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Hou

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(45) **Date of Patent: Dec. 26, 2006**

(54) **ADAPTIVE MICROPHONE MATCHING IN MULTI-MICROPHONE DIRECTIONAL SYSTEM**

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(*) Notice: Subject to any disclaimer, the term of this patent is extended or adjusted under 35 U.S.C. 154(b) by 1008 days.

(21) Appl. No.: **09/808,694**

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(65) **Prior Publication Data**

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(51) **Int. Cl.**
H04R 3/00 (2006.01)

(52) **U.S. Cl.** **381/92**; 381/313

(58) **Field of Classification Search** 381/92, 381/94.1, 94.9, 313, 71.7, 71.6, 71.11
See application file for complete search history.

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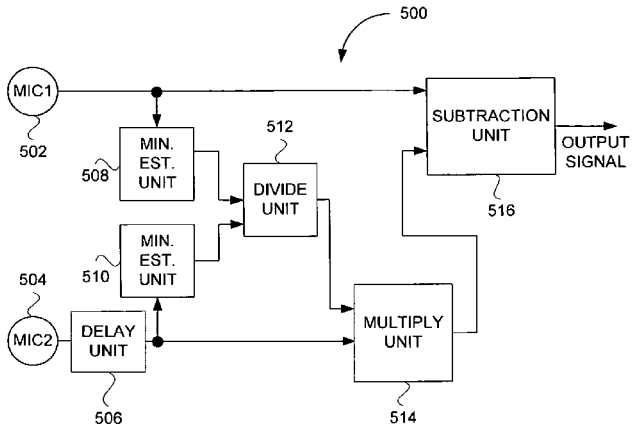
(Continued)

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(57) **ABSTRACT**

Improved approaches to matching sensitivities of microphones in multi-microphone directional processing systems. These approaches operate to adaptively match microphone sensitivities so that directional noise suppression is robust. As a result, microphone sensitivities remain matched not only over time but also while in actual use. These approaches are particularly useful for hearing aid applications in which directional noise suppression is important.

14 Claims, 14 Drawing Sheets



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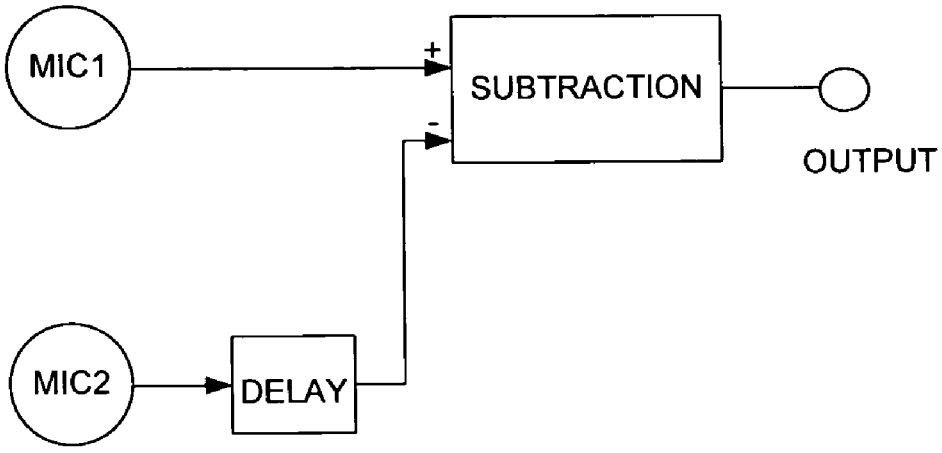

MIC1
+
SUBTRACTION
-
OUTPUT
MIC2
DELAY

FIG. 1

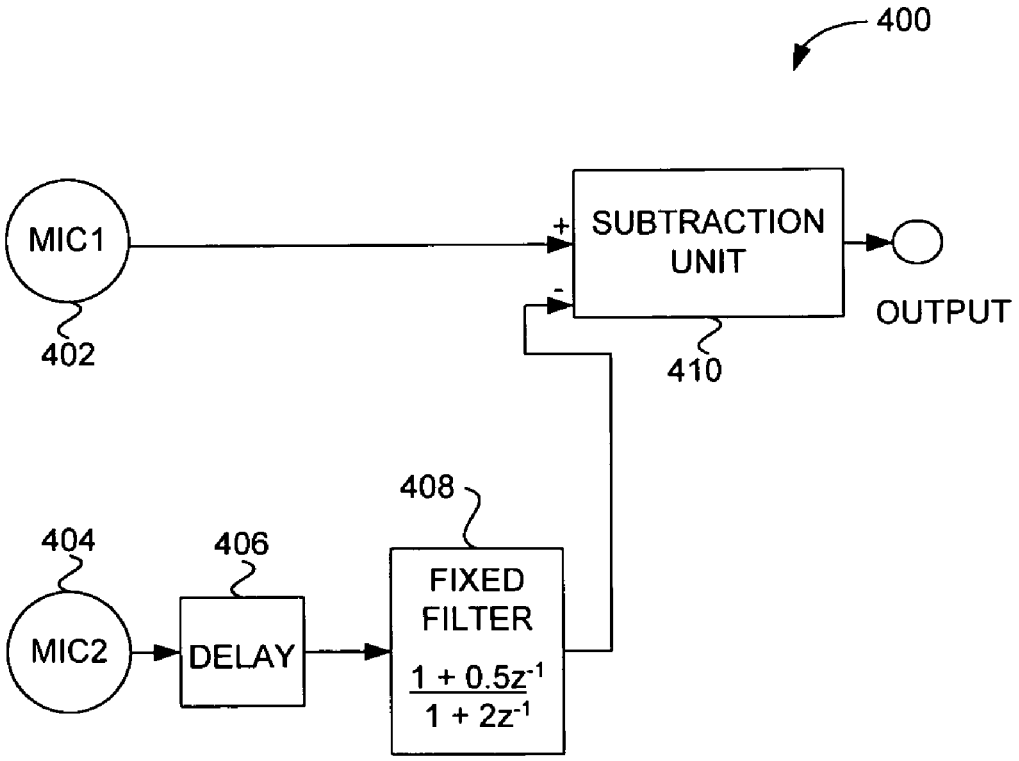

400
MIC1
+
SUBTRACTION UNIT
-
OUTPUT
402
410
408
404
406
MIC2
DELAY
FIXED FILTER
$\frac{1 + 0.5z^{-1}}{1 + 2z^{-1}}$

FIG. 4

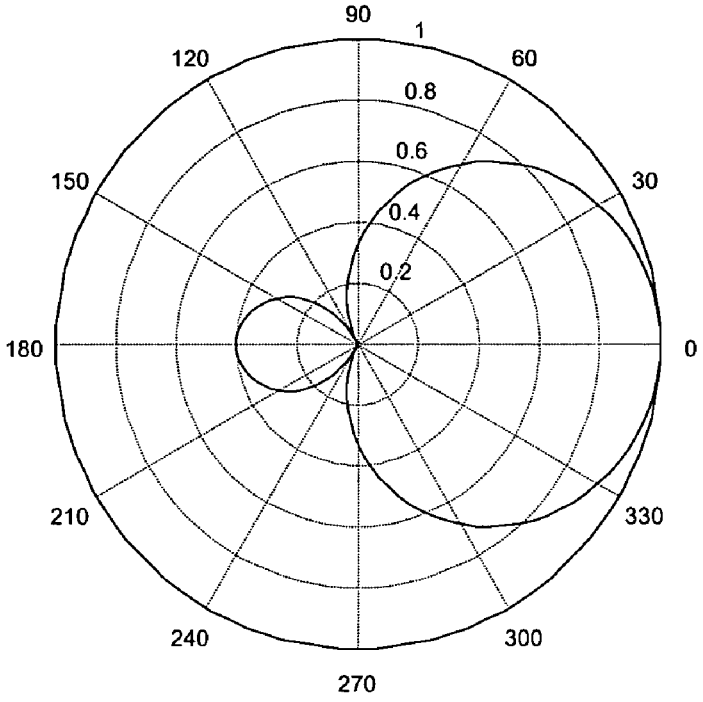

90
1
120
60
0.8
0.6
150
30
0.4
0.2
180
0
210
330
240
300
270

FIG. 2(a)

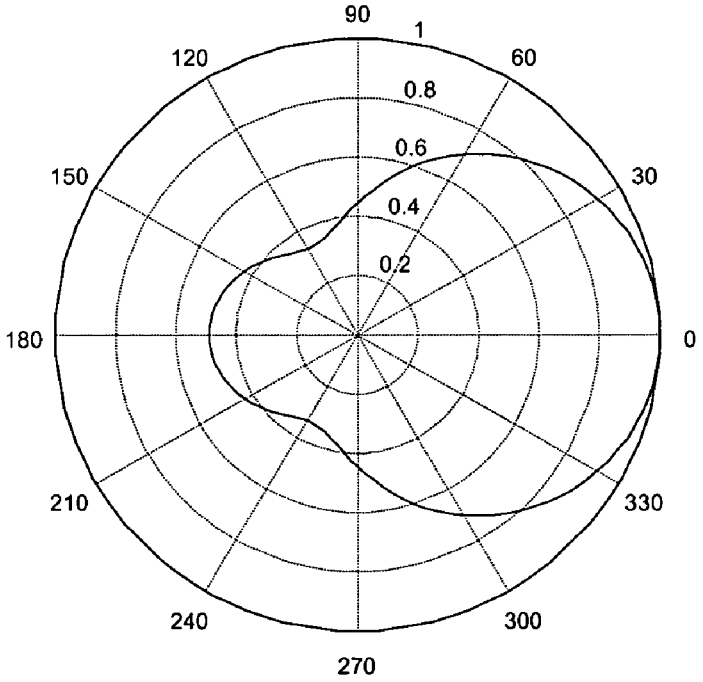

90
1
120
60
0.8
0.6
150
30
0.4
0.2
180
0
210
330
240
300
270

FIG. 2(b)

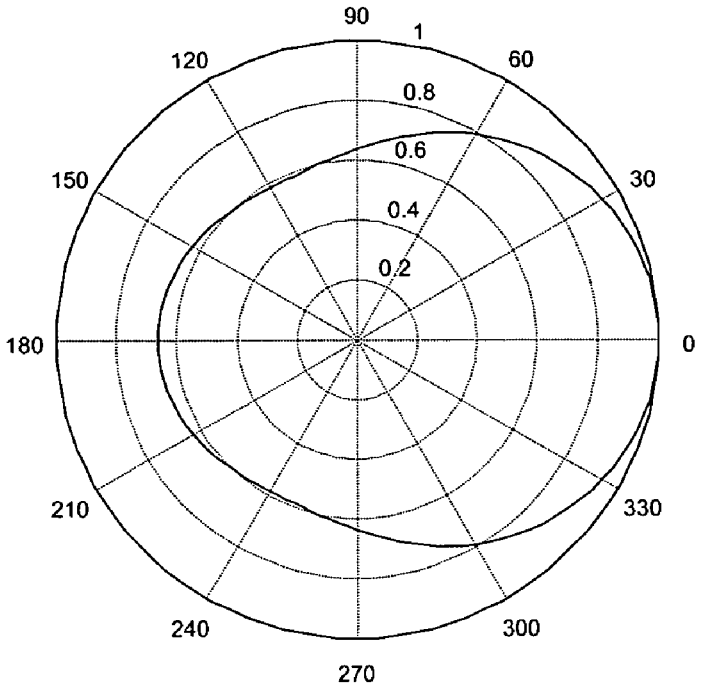

90
1
120
60
0.8
0.6
150
30
0.4
0.2
180
0
210
330
240
300
270

FIG. 2(c)

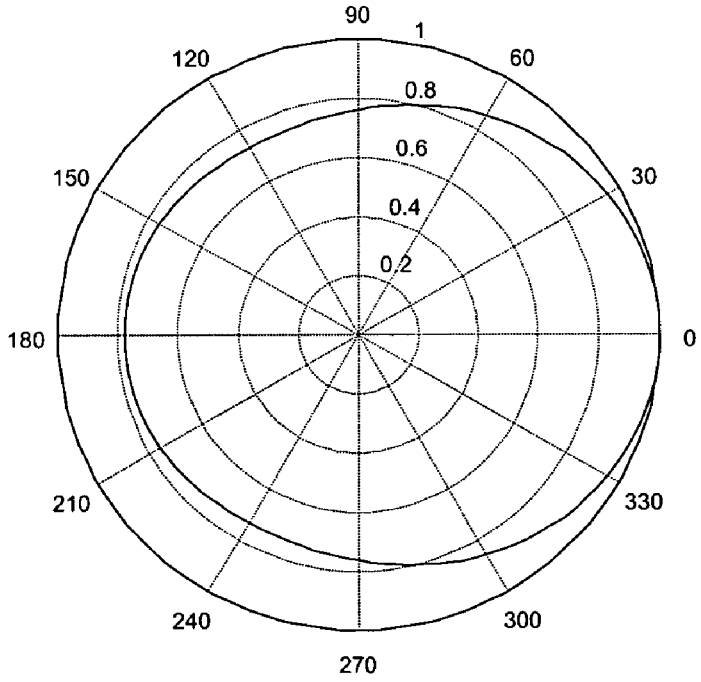

90
1
120
60
0.8
0.6
150
30
0.4
0.2
180
0
210
330
240
300
270

FIG. 2(d)

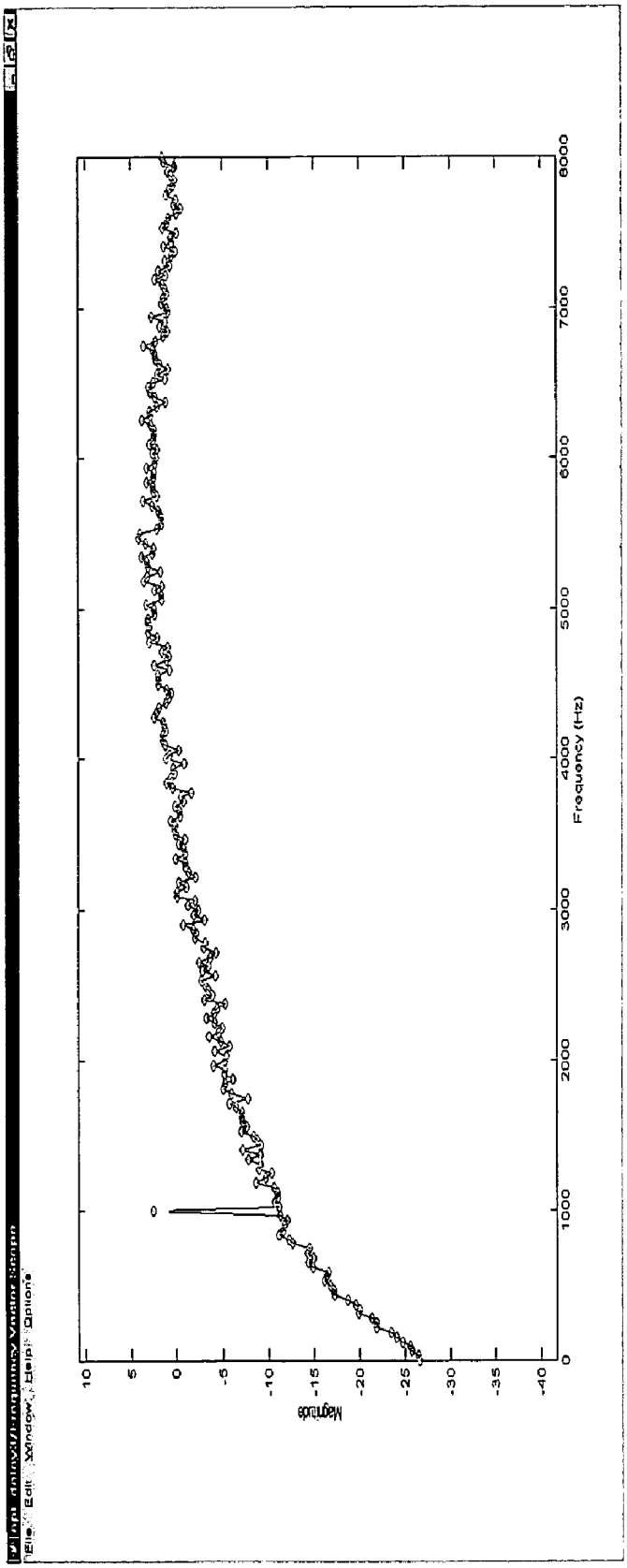

Magnitude
10
5
0
-5
-10
-15
-20
-25
-30
-35
-40
0
1000
2000
3000
4000
5000
6000
7000
8000
Frequency (Hz)

FIG. 3(a)

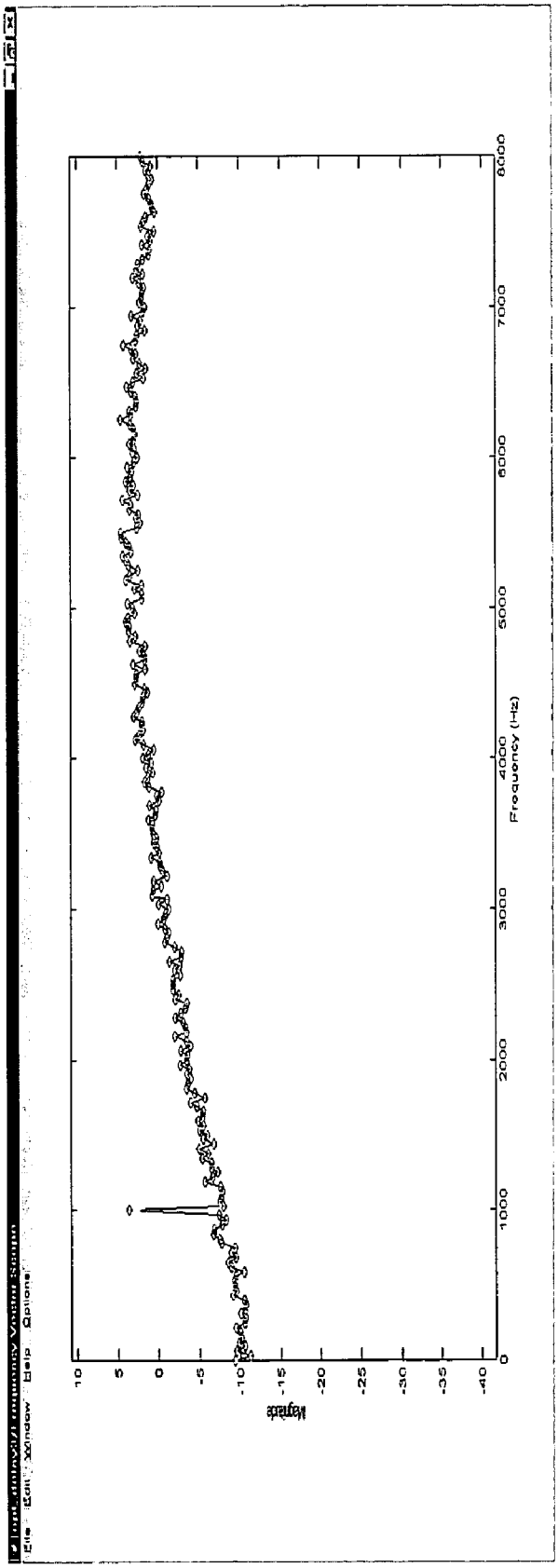

File Edit Window Help Options
Magnitude
10
5
0
-5
-10
-15
-20
-25
-30
-35
-40
0
1000
2000
3000
4000
5000
6000
7000
8000
Frequency (Hz)

FIG. 3(b)

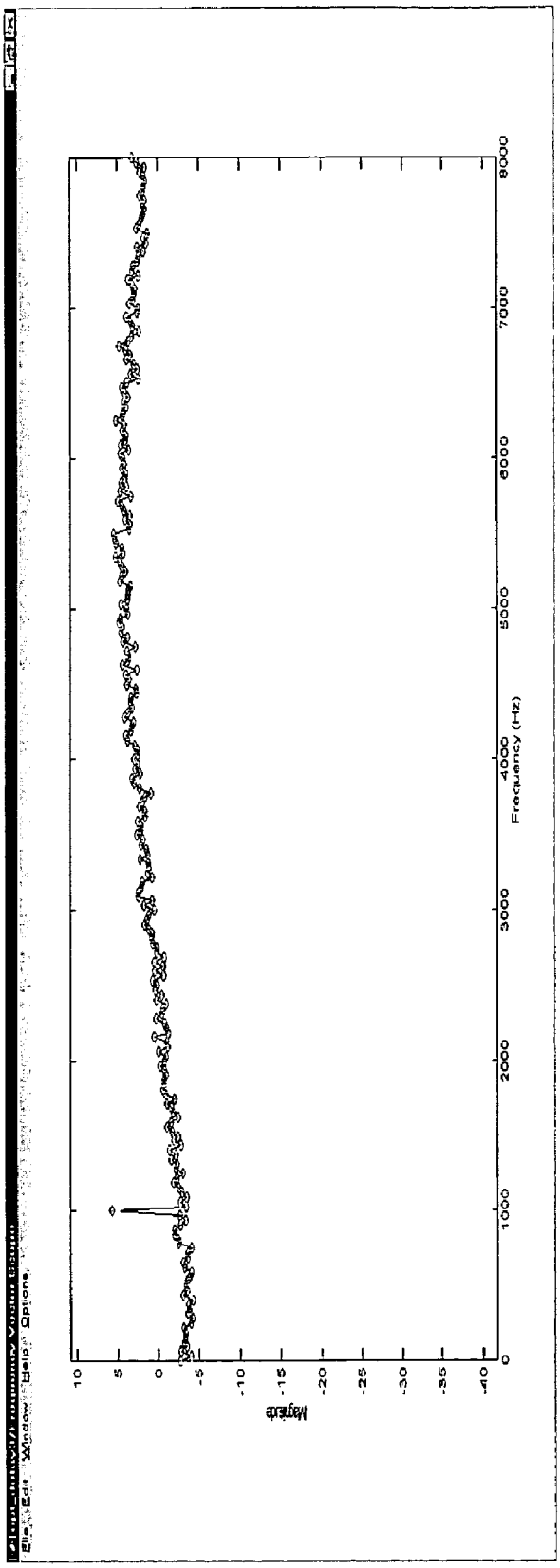

Magnitude
Frequency (Hz)
10
5
0
-5
-10
-15
-20
-25
-30
-35
-40
0
1000
2000
3000
4000
5000
6000
7000
8000

FIG. 3(c)

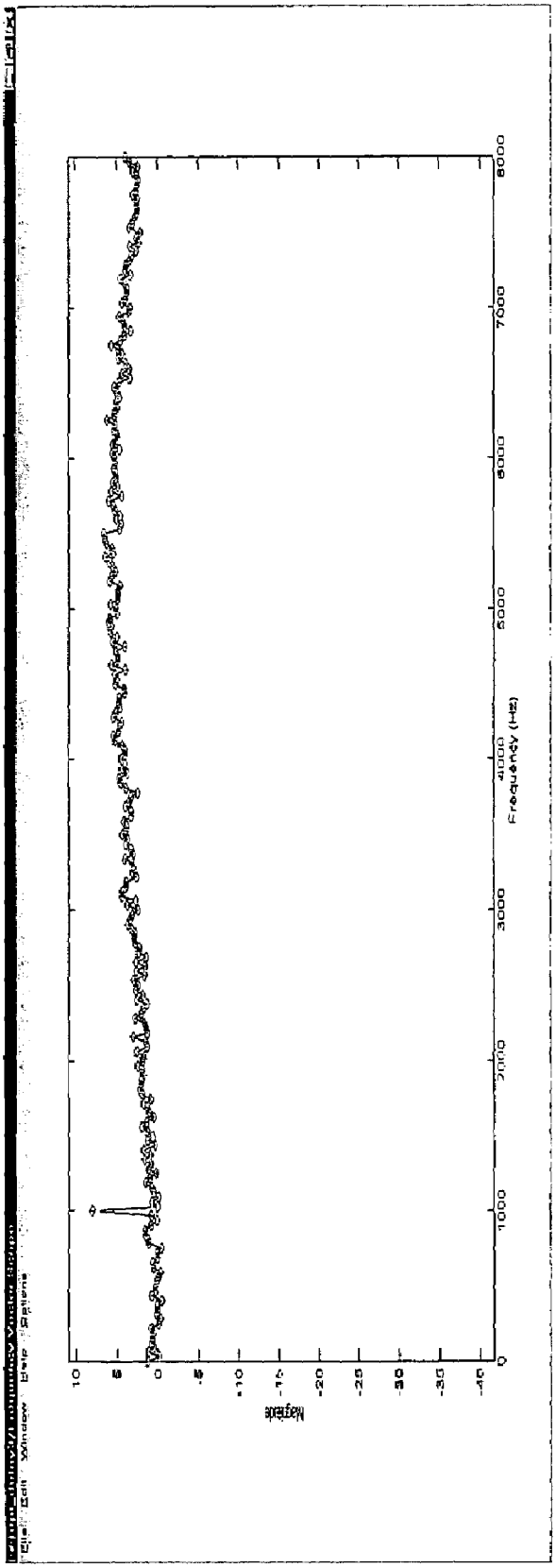

Magnitude
Frequency (Hz)
10
5
0
-5
-10
-15
-20
-25
-30
-35
-40
0
1000
2000
3000
4000
5000
6000
7000

FIG. 3(d)

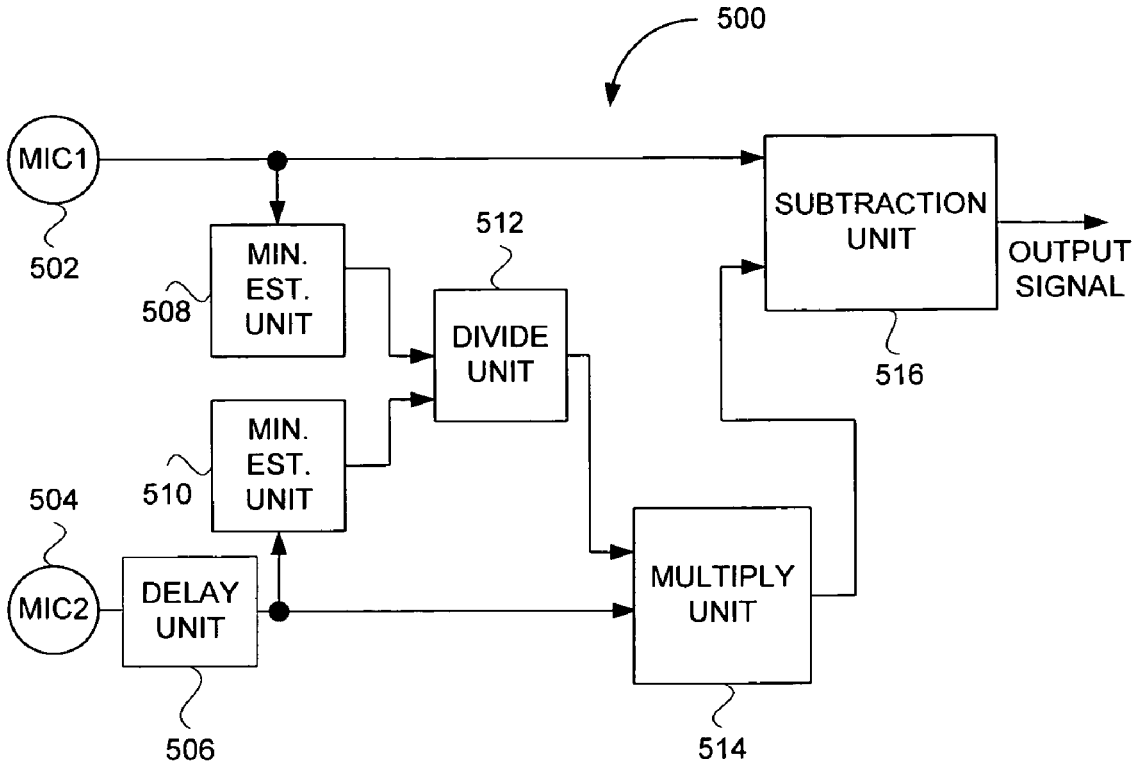

500
MIC1
502
MIN. EST. UNIT
508
512
DIVIDE UNIT
SUBTRACTION UNIT
OUTPUT SIGNAL
516
MIN. EST. UNIT
510
504
MIC2
DELAY UNIT
506
MULTIPLY UNIT
514

FIG. 5

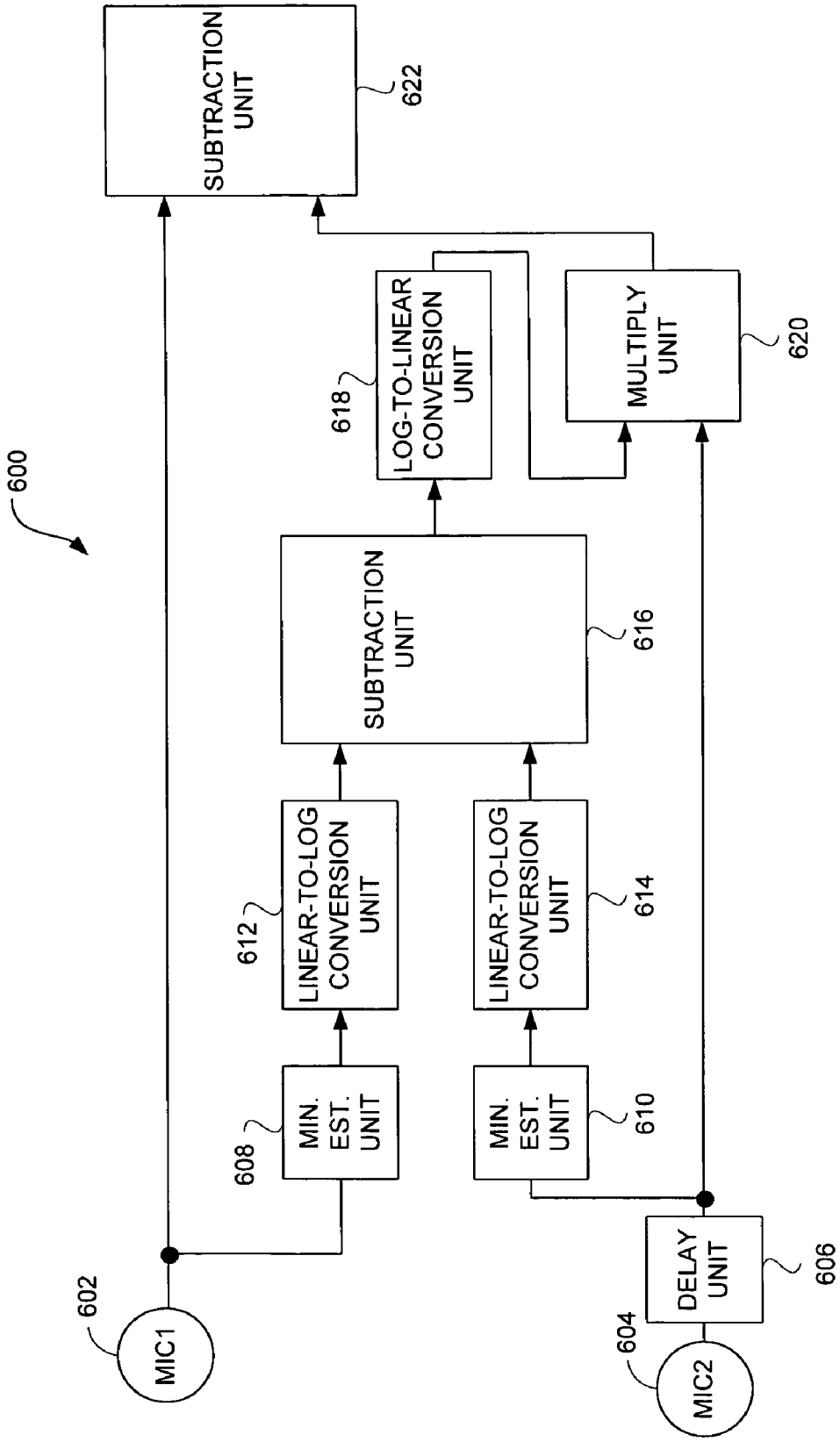

600
602
MIC1
604
MIC2
606
DELAY UNIT
608
MIN. EST. UNIT
610
MIN. EST. UNIT
612
LINEAR-TO-LOG CONVERSION UNIT
614
LINEAR-TO-LOG CONVERSION UNIT
616
SUBTRACTION UNIT
618
LOG-TO-LINEAR CONVERSION UNIT
620
MULTIPLY UNIT
622
SUBTRACTION UNIT

FIG. 6

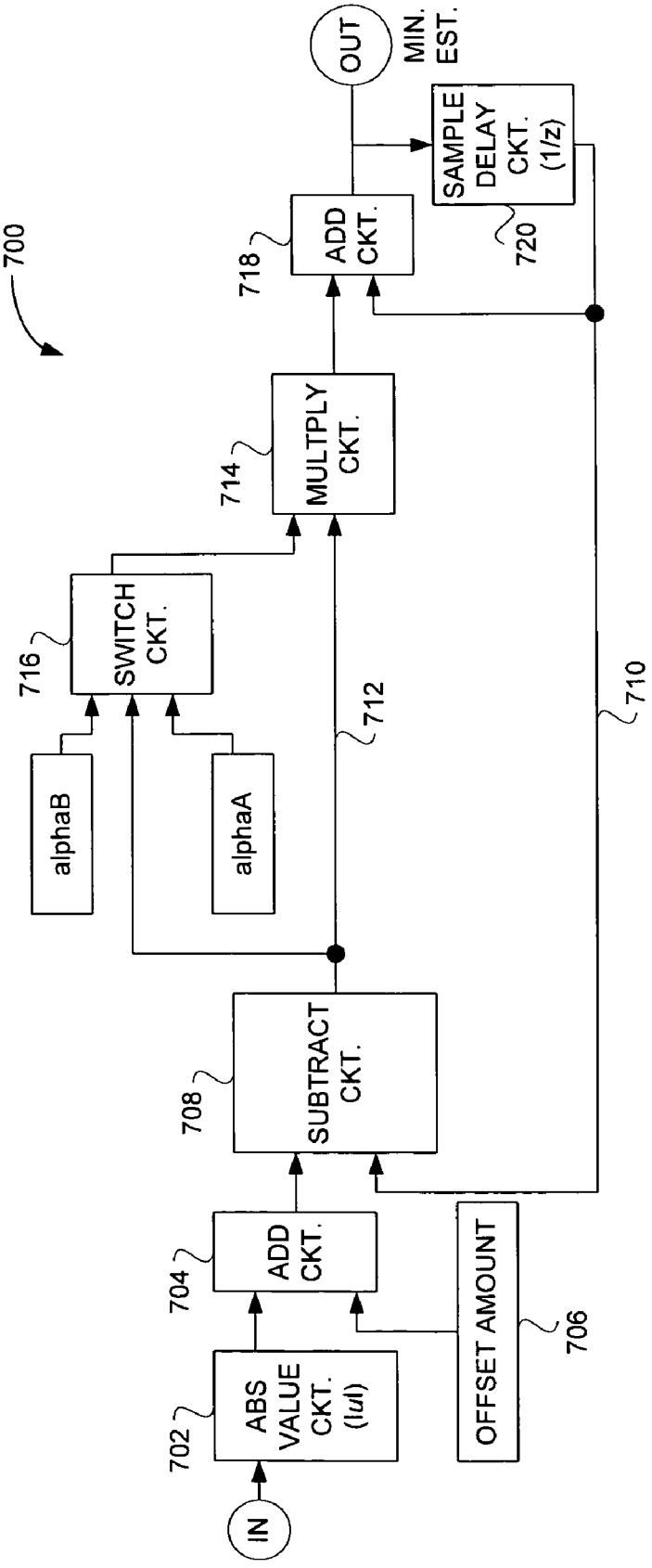

700
IN
ABS VALUE CKT. (|u|)
702
ADD CKT.
704
OFFSET AMOUNT
706
SUBTRACT CKT.
708
710
712
alphaB
alphaA
SWITCH CKT.
716
MULTPLY CKT.
714
ADD CKT.
718
SAMPLE DELAY CKT. (1/z)
720
OUT
MIN. EST.

FIG. 7

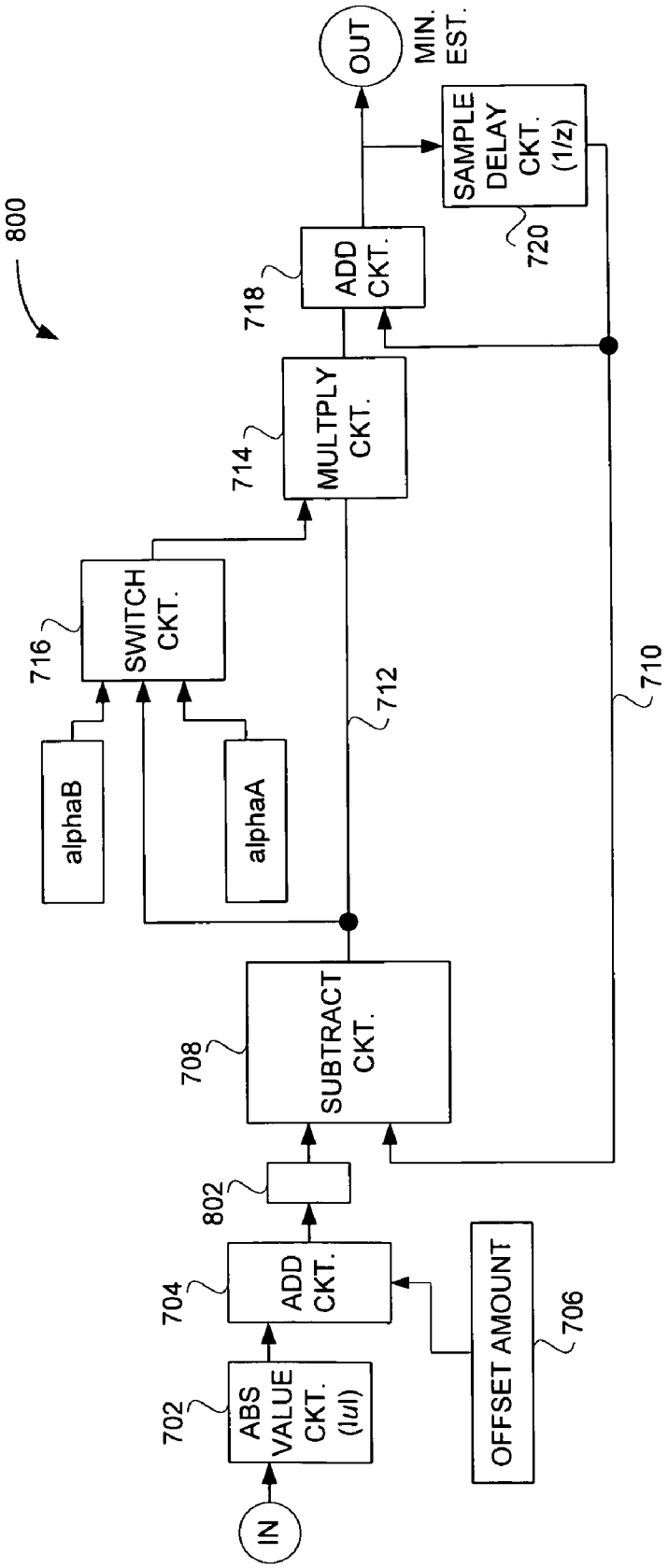

800
IN
ABS VALUE CKT. (|ul|)
702
ADD CKT.
704
OFFSET AMOUNT
706
802
SUBTRACT CKT.
708
710
712
alphaB
alphaA
SWITCH CKT.
716
MULTPLY CKT.
714
ADD CKT.
718
SAMPLE DELAY CKT. (1/z)
720
OUT
MIN. EST.

FIG. 8

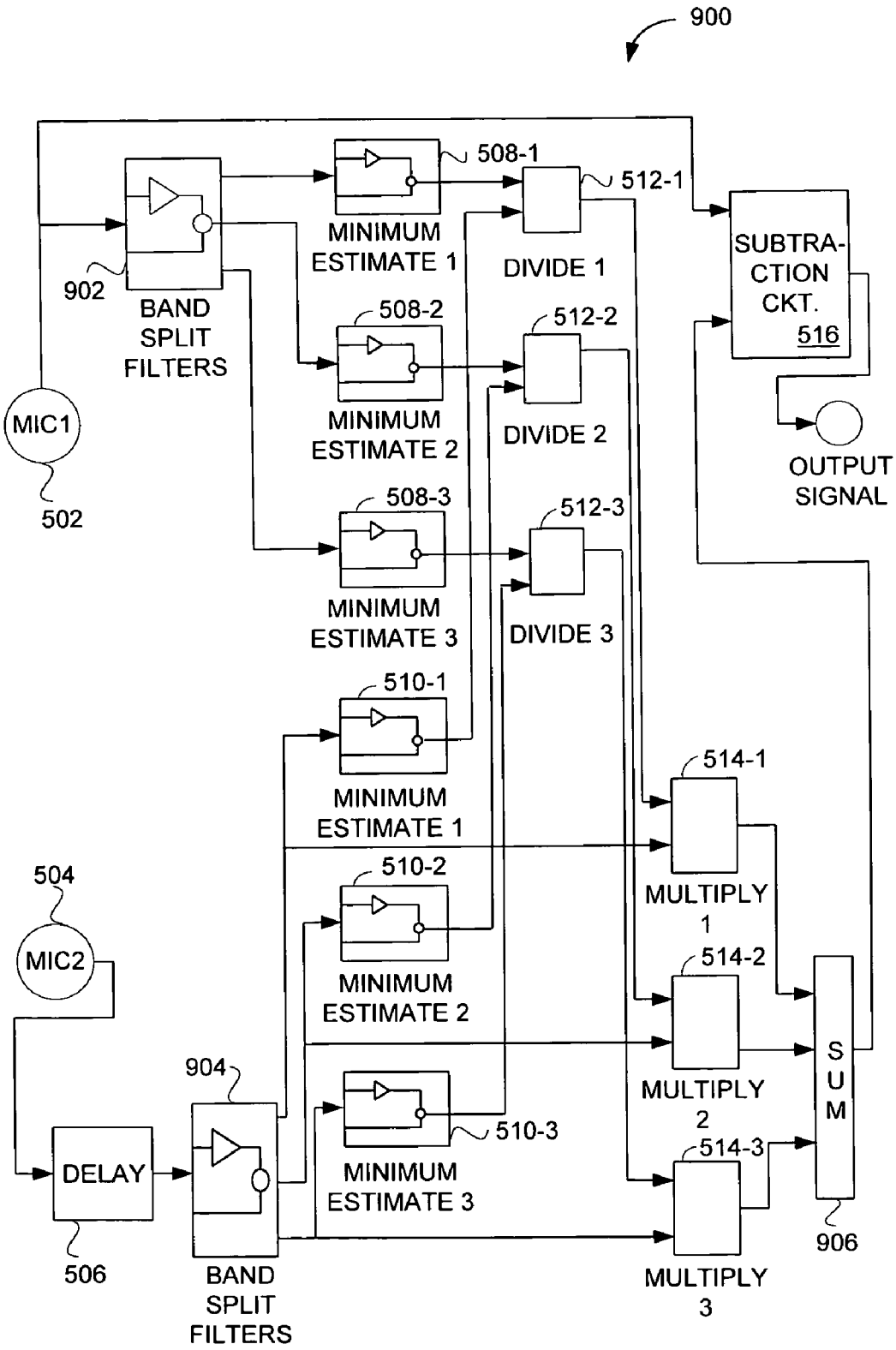

900
508-1
512-1
MINIMUM ESTIMATE 1
DIVIDE 1
SUBTRA-CTION CKT. 516
902
BAND SPLIT FILTERS
508-2
512-2
MINIMUM ESTIMATE 2
DIVIDE 2
MIC1
502
OUTPUT SIGNAL
508-3
512-3
MINIMUM ESTIMATE 3
DIVIDE 3
510-1
514-1
MINIMUM ESTIMATE 1
MULTIPLY 1
504
510-2
MIC2
MINIMUM ESTIMATE 2
514-2
MULTIPLY 2
S U M
904
510-3
514-3
DELAY
MINIMUM ESTIMATE 3
906
506
BAND SPLIT FILTERS
MULTIPLY 3

FIG. 9

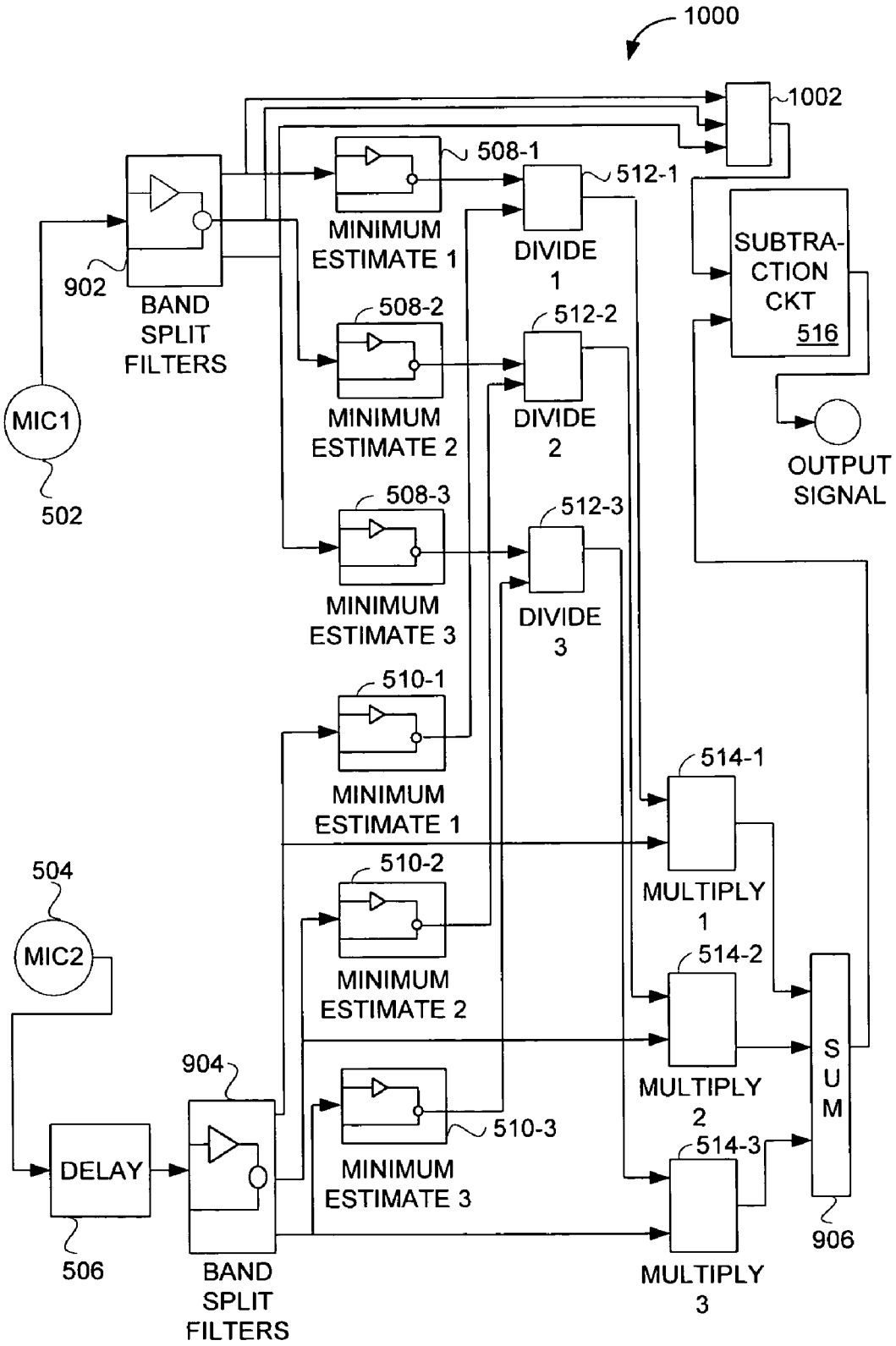

1000
1002
508-1
512-1
MINIMUM ESTIMATE 1
DIVIDE 1
SUBTRA-CTION CKT 516
902
BAND SPLIT FILTERS
508-2
512-2
MINIMUM ESTIMATE 2
DIVIDE 2
MIC1
502
OUTPUT SIGNAL
508-3
512-3
MINIMUM ESTIMATE 3
DIVIDE 3
510-1
514-1
MINIMUM ESTIMATE 1
MULTIPLY 1
504
510-2
MIC2
MINIMUM ESTIMATE 2
514-2
MULTIPLY 2
S U M
904
510-3
514-3
DELAY
MINIMUM ESTIMATE 3
906
506
BAND SPLIT FILTERS
MULTIPLY 3

FIG. 10

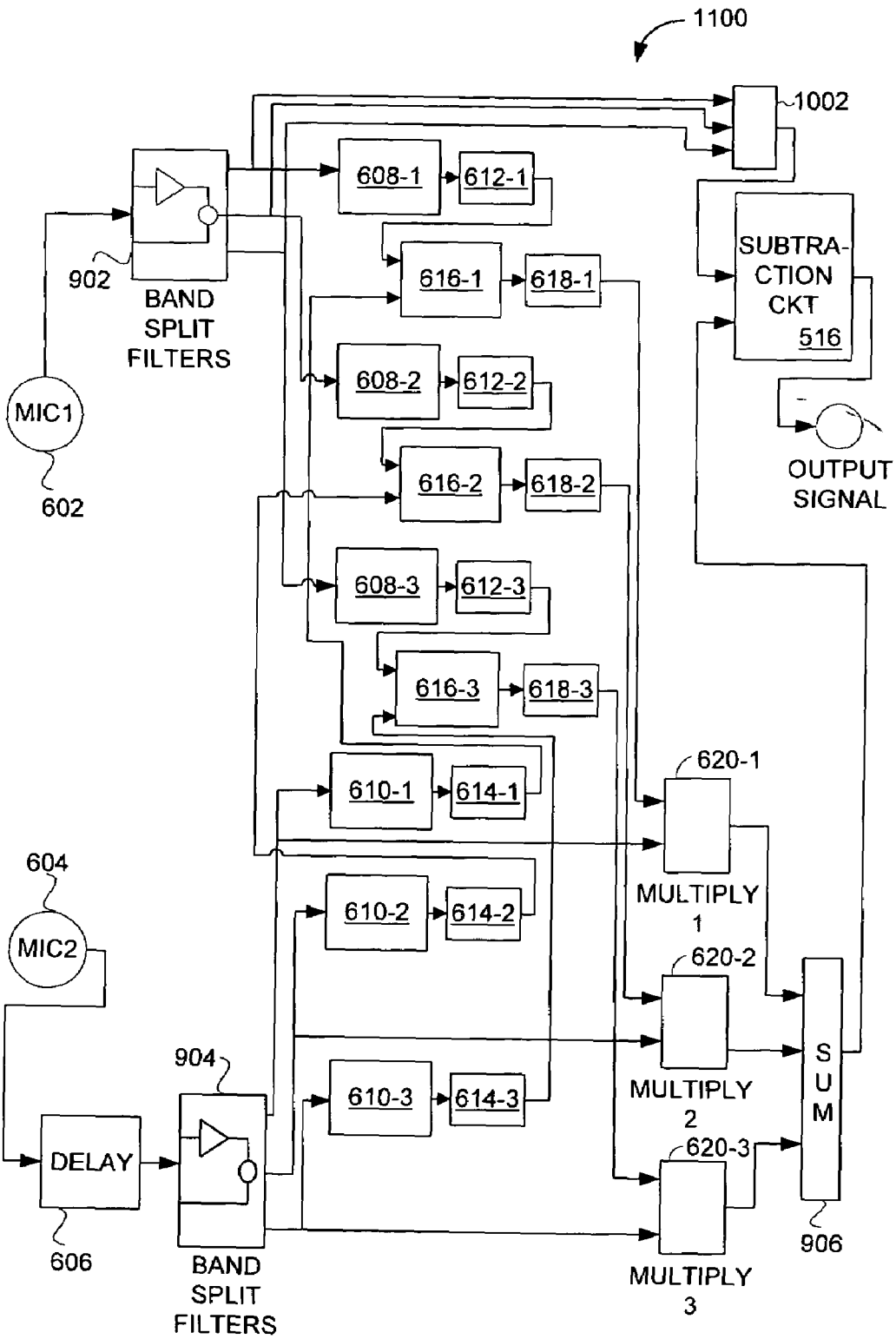

1100
1002
608-1
612-1
616-1
618-1
608-2
612-2
616-2
618-2
608-3
612-3
616-3
618-3
610-1
614-1
610-2
614-2
610-3
614-3
902
BAND SPLIT FILTERS
MIC1
602
SUBTRA-CTION CKT 516
OUTPUT SIGNAL
620-1
MULTIPLY 1
620-2
MULTIPLY 2
620-3
MULTIPLY 3
SUM
906
604
MIC2
DELAY
606
904
BAND SPLIT FILTERS

FIG. 11

ADAPTIVE MICROPHONE MATCHING IN MULTI-MICROPHONE DIRECTIONAL SYSTEM

CROSS-REFERENCE TO RELATED APPLICATIONS

This application claims the benefit of U.S. Provisional Application No. 60/189,282, filed Mar. 14, 2000, and entitled "METHODS FOR ADAPTIVE MICROPHONE MATCHING IN MULTI-MICROPHONE DIRECTIONAL SYSTEM", the contents of which is hereby incorporated by reference. This application is also related to U.S. application Ser. No. 09/788,271, filed Feb. 16, 2001, and entitled "NULL ADAPTATION IN MULTI-MICROPHONE DIRECTIONAL SYSTEM", the contents of which is hereby incorporated by reference.

BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates to multi-microphone sound pick-up systems and, more particularly, to matching microphone sensitivity in multi-microphone sound pick-up systems.

2. Description of the Related Art

Suppressing interfering noise is still a major challenge for most communication devices involving a sound pick up system such as a microphone or a multi-microphone array. The multi-microphone array can selectively enhance sounds coming from certain directions while suppressing interference coming from other directions.

FIG. **1** shows a typical direction processing system in a two-microphone hearing aid. The two microphones pick-up sounds and convert them into electronic or digital signals. The output signal form the second microphone is delayed and subtracted from the output signal of the first microphone. The result is a signal with interference from certain directions being suppressed. In other words, the output signal is dependent on which directions the input signals come from. Therefore, the system is directional. The physical distance between the two microphones and the delay are two variables that control the characteristics of the directionality. For hearing aid applications, the physical distance is limited by the physical dimension of the hearing aid. The delay can be set in a delta-sigma analog-to-digital converter (A/D) or by use of an all-pass filter.

The sensitivity of the microphones of the sound pick up system must be matched in order to achieve good directionality. When the sensitivities of the microphones are not properly matched, then the directionality is substantially degraded and thus the ability to suppress interference coming from a particular direction is poor. FIGS. **2**(*a*), **2**(*b*), **2**(*c*) and **2**(*d*) illustrate representative polar patterns for microphone sensitivity discrepancies of 0, 1, 2, and 3 dB, respectively. Note that the representative polar pattern shown in FIG. **2**(*a*) is the desired polar pattern which offers maximized directionality. The representative polar patterns shown in FIGS. **2**(*b*)–**2**(*d*) are distorted polar patterns that respectively illustrate directionality becoming progressively worse as the sensitivity discrepancy increases respectively from 1, 2 and 3 dB. FIGS. **3**(*a*), **3**(*b*), **3**(*c*) and **3**(*d*) illustrate representative spectrum response for microphone sensitivity discrepancies of 0, 1, 2, and 3 dB, respectively, with reference to a 1 kHz pure tone in white noise. Note that the Signal-to-Noise Ratio of the spectrum shown in FIGS. **3**(*a*)–**3**(*d*) is 14, 11, 9 and 7 dB, respectively. Accordingly, a good match of sensitivity between microphones is very important to good directionality.

Conventionally, manufacturers manually match the microphone for their multi-microphone directional processing systems. While manual matching of the microphones provides for improved directionality, the operational or manufacturing costs are substantial. Besides cost-effectiveness, manual matching has other problems that compromise manual matching. One problem is that microphone sensitivity tends to drift over time. Hence, once matched microphones can become mismatched over time. Another problem is that the sensitivity difference can depend on how the multi-microphone directional processing systems is used. For example, in hearing aid applications, a microphone pair that is perfectly matched as determined by measurements at manufacture may become mismatched when the hearing aid is put on a patient. This can occur because at manufacture the microphones are measured in a field where sound pressure level is the same everywhere (free field), while in real life situation (in situ) sound pressure may not distribute uniformly at microphone locations. Hence, when such pressure differences result, the microphones are in effect mismatched. In another word, because the microphones are matched in free field, not in situ, the microphones can actually be mismatched when used in real life, which degrades directionality.

Some manufacturers have used a fixed filter in their designs of multi-microphone directional processing systems. FIG. **4** illustrates a conventional two-microphone directional processing system **400** having a first microphone **402**, a second microphone **404**, a delay **406**, a fixed filter **408**, and a subtraction unit **410**. The fixed filter **408** can serve to compensate for a mismatch in microphone sensitivity. The fixed filter approach is more cost-effective that the manual matching. However, the other problems (e.g., drift over time and in-situ mismatch) of manual matching are still present with the fixed filter approach.

Thus, there is a need for improved approaches to match sensitivities of microphones in multi-microphone directional processing systems.

SUMMARY OF THE INVENTION

Broadly speaking, the invention relates to improved approaches to matching sensitivities of microphones in multi-microphone directional processing systems. These approaches operate to adaptively match microphone sensitivities so that directional noise suppression is robust. As a result, microphone sensitivities remain matched not only over time but also while in actual use. These approaches are particularly useful for hearing aid applications in which directional noise suppression is important.

The invention can be implemented in numerous ways including as a method, system, apparatus, device, and computer readable medium. Several embodiments of the invention are discussed below.

As an adaptive directional sound processing system, one embodiment of the invention includes at least: at least first and second microphones spaced apart by a distance, the first microphones producing a first electronic sound signal and the second microphone producing a second electronic sound signal; means for processing the second electronic sound signal to adaptively produce a compensation scaling amount that compensates for sensitivity differences between the first and second microphones; a scaling circuit operatively connected to the means for scaling and the second microphone, the scaling circuit operates to scale the second electronic

sound signal in accordance with the compensation scaling amount; and a subtraction circuit operatively connected to the scaling circuit and the first microphone, the subtraction circuit producing an output difference signal by subtracting the scaled second electronic sound signal from the first electronic sound signal.

As an adaptive directional sound processing system, another embodiment of the invention includes at least: at least first and second microphones spaced apart by a predetermined distance, the first microphones producing a first electronic sound signal and the second microphone producing a second electronic sound signal; a first minimum estimate circuit operatively coupled to the first microphone, the first minimum estimate circuit produces a first minimum estimate for the first electronic sound signal from the first microphone; a second minimum estimate circuit operatively coupled to the second microphone, the second minimum estimate circuit produces a second minimum estimate for the second electronic sound signal from the second microphone; a divide circuit operatively connected to the first and second minimum estimate circuits, the divide circuit operates to produce a scaling signal from the first and second minimum estimates; a multiply circuit operatively connected to the divide circuit and the second microphone, the multiply circuit operates to multiply the second electronic sound signal by the scaling signal to produce a scaled second electronic sound signal; and a subtraction circuit operatively connected to the multiply circuit and the first microphone, the subtraction circuit producing an output difference signal by subtracting the scaled second electronic sound signal from the first electronic sound signal.

As a hearing aid device having an adaptive directional sound processing, one embodiment of the invention includes at least: at least first and second microphones spaced apart by a distance, the first microphones producing a first electronic sound signal and the second microphone producing a second electronic sound signal; sensitivity difference detection circuitry operatively connected to the first and second microphones, the sensitivity difference detection circuitry adaptively produces a compensation scaling amount corresponding to sensitivity differences between the first and second microphones; a scaling circuit operatively connected to the sensitivity difference detection circuitry and the second microphone, the scaling circuit operates to scale the second electronic sound signal in accordance with the compensation scaling amount; and a subtraction circuit operatively connected to the scaling circuit and the first microphone, the subtraction circuit producing an output difference signal by subtracting the scaled second electronic sound signal from the first electronic sound signal.

As a method for adaptively measuring and compensating for acoustical differences between sound signals picked up by microphones, one embodiment of the invention includes at least the acts of: receiving first and second electronic sound signals from first and second microphones, respectively; determining a compensation scaling amount that compensates for acoustic differences with respect to the first and second microphones; scaling the second electronic sound signal in accordance with the compensation scaling amount; and producing a differential electronic sound signal by subtracting the scaled second electronic sound signal from the first electronic sound signal.

Other aspects and advantages of the invention will become apparent from the following detailed description taken in conjunction with the accompanying drawings which illustrate, by way of example, the principles of the invention.

BRIEF DESCRIPTION OF THE DRAWINGS

The invention will be readily understood by the following detailed description in conjunction with the accompanying drawings, wherein like reference numerals designate like structural elements, and in which:

FIG. **1** shows a typical direction processing system in a two-microphone hearing aid;

FIGS. **2**(*a*)–**2**(*d*) illustrate representative polar patterns for various microphone sensitivity discrepancies;

FIGS. **3**(*a*)–**3**(*d*) illustrate representative Signal-to-Noise Ratio spectrums respectively corresponding to the representative polar patterns shown in FIGS. **2**(*a*)–**2**(*d*);

FIG. **4** illustrates a conventional two-microphone directional processing system;

FIG. **5** is a block diagram of a two-microphone directional processing system according to one embodiment of the invention;

FIG. **6** is a block diagram of a two-microphone directional processing system according to another embodiment of the invention;

FIG. **7** is a block diagram of a minimum estimate unit according to one embodiment of the invention;

FIG. **8** is a block diagram of a minimum estimate unit according to another embodiment of the invention;

FIG. **9** is a block diagram of a multi-microphone directional processing system that operates to perform multi-band adaptive compensation for microphone mismatch;

FIG. **10** is a block diagram of a multi-microphone directional processing system according to one embodiment of the invention; and

FIG. **11** is a block diagram of a multi-microphone directional processing system according to another embodiment of the invention.

DETAILED DESCRIPTION OF THE INVENTION

The invention relates to improved approaches to matching sensitivities of microphones in multi-microphone directional processing systems. These approaches operate to adaptively match microphone sensitivities so that directional noise suppression is robust. As a result, microphone sensitivities remain matched not only over time but also while in actual use. These approaches are particularly useful for hearing aid applications in which directional noise suppression is important.

According to one aspect, the invention operates to adaptively measure a sensitivity difference between microphones in a multi-microphone directional processing system, and then compensate (or correct) an electronic sound signal from one or more of the microphones. As a result of the adaptive processing, the microphones "effectively" become matched and remain matched over time and while in use.

Consequently, the invention enables multi-microphone directional processing systems to achieve superior directionality and consistent Signal-to-Noise Ratio (SNR) across all conditions. The invention is described below with respect to embodiments particularly well suited for use with hearing aid applications. However, it should be recognized that the invention is not limited to hearing aid applications, but is applicable to other sound pick-up systems.

Embodiments of this aspect of the invention are discussed below with reference to FIGS. **5**–**11**. However, those skilled in the art will readily appreciate that the detailed description

given herein with respect to these figures is for explanatory purposes as the invention extends beyond these limited embodiments.

As noted above, microphone matching is important for multi-microphone directional systems. Different and undesired responses will result when the sensitivities of the microphones are not matched. The acoustic delay between the microphones further complicates matching problems. For example, even if the microphones are perfectly matched, the instantaneous response of the microphones can be different because of the delay and/or fluctuation in the acoustic signals. Therefore, it is not enough to simply use the difference of the responses to correct the problem. More complex processing is necessary to eliminate the effects of acoustic delay between the microphones and/or the fluctuation in the acoustic signals.

According to one aspect of the invention, responses from each microphone are processed such that the resulting processed signals are not sensitive to the acoustic delay between the microphones and the fluctuation of acoustic conditions. A difference between the processed signals from the microphone channels can then be used to scale at least one microphone's response so as to compensate or correct for sensitivity differences between the microphones.

FIG. **5** is a block diagram of a two-microphone directional processing system **500** according to one embodiment of the invention. The two-microphone directional processing system **500** includes a first microphone **502** and a second microphone **504**. The first microphone **502** produces a first electronic sound signal and the second microphone **504** produces a second electronic sound signal. A delay unit **506** delays the second electronic sound signal. The two-microphone directional processing system **500** also includes a first minimum estimate unit **508**, a second minimum estimate unit **510** and a divide unit **512**. The first minimum estimate unit **508** estimates the minimum for the first electronic sound signal. The second minimum estimate unit **510** estimates the minimum of the second electronic sound signal. Typically, these minimums are measured over a time constant duration, such that the minimum is a relatively long-term minimum. The divide unit **512** produces a quotient by dividing the first minimum estimate by the second minimum estimate. The quotient represents a scaling amount that is sent to a multiplication unit **514**. The second electronic sound signal is then multiplied with the scaling amount to produce a compensated sound signal. The compensated sound signal is thus compensated (or corrected) for the relative difference in sensitivity between the mismatched first and second microphones **502** and **504**. A subtraction unit **516** then subtracts the compensated electronic sound signal from the first electronic sound signal to produce an output signal. At this point, the output signal has been processed by the two-microphone directional processing system **500** to have robust directionality despite a mismatch between the first and second microphones **502** and **504**.

The two-microphone directional processing system **500** uses a single-band adaptive compensation scheme to compensate for sensitivity differences between the microphones. In this embodiment, minimum estimates and division calculations are performed. The minimum estimates can, for example, be performed by minimum estimate units shown in more detail below with respect to FIGS. **7** and **8**. It should also be noted that the delay unit **506** can be positioned within the two-microphone directional processing system **500** anywhere in the channel associated with the second electronic sound signal prior to the subtraction unit **516**. Still further, it should be noted that a multiple-band adaptive compensation scheme could alternatively be utilized.

Moreover, although the two-microphone directional processing system **500** uses minimum estimates of the electronic sound signals produced by the first and second microphones **502** and **504**, other signal characteristics can alternatively be used. For example, Root-Mean-Square (RMS) average of the electronic sound signals produced by the microphones could be used. With such an approach, the RMS average could be measured over a time constant duration. The time constant can be set such that the average is relatively long-term so as to avoid impact of signal fluctuations. The time constant with an RMS approach is likely to be longer than the time constant for the minimum approach.

The two-microphone directional processing system **500** operates to scale the intensity of an electronic sound signal from one or more of the microphones. With respect to the two-microphone directional processing system **500**, the processing (including the scaling) is performed in a linear domain. However, the scaling or other processing can also be performed in a logarithm (or dB) domain.

FIG. **6** is a block diagram of a two-microphone directional processing system **600** according to another embodiment of the invention. The two-microphone directional processing system **600** includes a first microphone **602** and a second microphone **604**. The first microphone **602** produces a first electronic sound signal and the second microphone **604** produces a second electronic sound signal. A delay unit **606** delays the second electronic sound signal. The two-microphone directional processing system **600** also includes a first minimum estimate unit **608** and a second minimum estimate unit **610**. The first minimum estimate unit **608** estimates the minimum for the first electronic sound signal. The second minimum estimate unit **610** estimates the minimum of the second electronic sound signal. Typically, these minimums are measured over a time constant duration, such that the minimum is a relatively long-term minimum.

The two-microphone directional processing system **600** also includes a first linear-to-log conversion unit **612**, a second linear-to-log conversion unit **614**, a subtraction unit **616**, and a log-to-linear conversion unit **618**. The first minimum estimate is converted from the linear domain to the logarithm domain by the first linear-to-log conversion unit **612**, and the second minimum estimate is converted from the linear domain to the logarithm domain by the second linear-to-log conversion unit **614**. The subtraction unit **616** then subtracts the second minimum estimate from the first minimum estimate to produce a difference amount. The log-to-linear conversion unit **614** then converts the difference amount to the linear domain.

The converted difference amount produced by the log-to-linear conversion unit **614** represents a scaling amount that is sent to a multiplication unit **620**. The second electronic sound signal is then multiplied with the scaling amount to produce a compensated sound signal. The compensated sound signal is thus compensated (or corrected) for the relative difference in sensitivity between the mismatched first and second microphones **602** and **604**. A subtraction unit **622** then subtracts the compensated electronic sound signal from the first electronic sound signal to produce an output signal. The output signal has been processed by the two-microphone directional processing system **500** to have robust directionality despite a physical mismatch between the first and second microphones **602** and **604**.

It should be noted that the two-microphone directional processing system **600** is generally similar to the two-

microphone directional processing system **500** illustrated in FIG. **5**. Both use similar circuitry to produce a single-band adaptive compensation scheme for a multi-microphone directional processing system. However, the divide unit **512** shown in FIG. **5** is replaced by the linear-to-log conversion units **612** and **614**, the subtraction unit **616** and the log-to-linear conversion unit **618** shown in FIG. **6**. Mathematically, the divide unit **512** is equivalent to the combination of the linear-to-log conversion units **612** and **614**, the subtraction unit **616** and the log-to-linear conversion unit **618**. However, with certain approximations, the design shown in FIG. **6** may be able to perform a "divide" operation more efficiently. Also the delay unit **606** in FIG. **6** can be positioned anywhere in the channel associated with the second electronic sound signal prior to the subtraction unit **622**.

FIG. **7** is a block diagram of a minimum estimate unit **700** according to one embodiment of the invention. The minimum estimate unit **700** is, for example, suitable for use as the minimum estimate units discussed above with respect to FIGS. **5** and **6**. The minimum estimate unit **700** receives an input signal (e.g., electronic sound signal) that is to have its minimum estimated. The input signal is supplied to an absolute value circuit **702** that determines the absolute value of the input signal. An add circuit **704** adds the absolute value of the input signal together with an offset amount **706** and thus produces an offset absolute value signal. The addition of the offset amount, which is typically a small positive value, such as 0.000000000001, is used to avoid overflow in division or logarithm calculations performed in subsequent circuitry in the multi-microphone directional processing systems. The offset absolute value signal from the add circuit **704** is supplied to a subtract circuit **708**. The subtract circuit **708** subtracts a previous output **710** from the offset absolute value signal to produce a difference signal **712**. The difference signal **712** is supplied to a multiply circuit **714**. In addition, the difference signal **712** is supplied to a switch circuit **716**. The switch circuit **716** selects one of two constants that are supplied to the multiply circuit **714**. A first of the constants is referred to as alphaB and is supplied to the multiply circuit **714** when the difference signal **712** is greater than or equal to zero. Alternatively, a second constant, alphaA, is supplied to the multiply circuit **714** when the difference signal **712** is not greater than or equal to zero. The constants, alphaA and alphaB, are typically small positive values, with alphaA being greater than alphaB. In one implementation, alphaA is 0.00005 and alphaB is 0.000005. The multiply circuit **714** multiplies the difference signal **712** by the selected constant to produce an adjustment amount. The adjustment amount is supplied to an add circuit **718**. The add circuit **718** adds the adjustment amount to the previous output **710** to produce a minimum estimate for the input signal. A sample delay circuit **720** delays the minimum estimate by a delay (1/z) to yield the previous output **710** (where 1/z represents a delay operation).

FIG. **8** is a block diagram of a minimum estimate unit **800** according to another embodiment of the invention. The minimum estimate unit **800** is, for example, similar in design to the minimum estimate unit **700** illustrated in FIG. **7**. The minimum estimate unit **800**, however, further includes a linear-to logarithm conversion unit **802** that converts the offset absolute value signal into a logarithmic offset signal before being supplied to the subtract circuit **708**.

The minimum estimate unit **800** is, for example, suitable for use as the minimum estimate units discussed above with respect to FIG. **6**. Note that, however, the linear-to-logarithm conversion units **612** and **614** would not be needed when the minimum estimate unit **800** is used in the system because there is already a linear-to-logarithm conversion unit inside the minimum estimate unit **800**.

The two constants, alphaA and alphaB, are used in the minimum estimate units **700**, **800** to determine how the minimum estimate changes with the input signal. Because the constant alphaA is greater than the constant alphaB, the minimum estimate tracks the value level (or minimum level) of the input signal. Since the value level is typically a good indicator of the noise level in the sound, the minimum estimate produced by the minimum estimate units **700**, **800** is a good indicator of background noise level.

As noted above, the present invention can also be implemented in circuits that utilize multi-band adaptive compensation for mismatch of microphone sensitivities. FIG. **9** is a block diagram of a multi-microphone directional processing system **900** that operates to perform multi-band adaptive compensation for microphone mismatch. Although any number of bands can be used, the multi-microphone directional processing system **900** uses three bands. The multi-microphone directional processing system **900** is generally similar in operation to the two-microphone directional processing system **500** illustrated in FIG. **5**. However, the multi-microphone directional processing system **900** further includes band split filters **902** and **904** that divide or separate the electronic sound signals from each of the microphones into different frequency ranges. Typically, the band split banks would be the same for each microphone. The band split filters **902** split the first electronic sound signal into first, second and third partial sound signals that are respectively delivered to minimum estimate circuits **508-1**, **508-2** and **508-3**. The minimum estimates produced by the minimum estimate circuits **508-1**, **508-2** and **508-3** are respectively supplied to the divide circuits **512-1**, **512-2** and **512-3**. The divide circuits **512-1**, **512-2** and **512-3** yield first, second and third scaling amounts. The first, second and third scaling amounts produced by the divide circuits **512-1**, **512-2** and **512-3** are respectively supplied to the multiply circuits **514-1**, **514-2** and **514-3**. The multiply circuits **514-1**, **514-2** and **514-3** respectively multiply the first, second and third partial sound signals for the second electronic sound signal by the corresponding first, second and third scaling amounts to produce first, second and third partial scaled second electronic sound signals. The first, second and third partial scaled second electronic sound signals output from the multiply circuits **514-1**, **514-2** and **514-3** are then summed by a sum circuit **906** to produce the compensated sound signal. The compensated sound signal is thus compensated (or corrected) for the relative difference in sensitivity between the mismatched first and second microphones **502** and **504**. The compensated sound signal is then subtracted from the first electronic sound signal by the subtraction circuit **516** to produce the output signal.

FIG. **10** is a block diagram of a multi-microphone directional processing system **1000** according to one embodiment of the invention. The multi-microphone directional processing system **1000** illustrated in FIG. **10** is generally similar to the multi-microphone directional processing system **900** illustrated in FIG. **9**. However, the multi-microphone directional processing system **1000** further includes a sum circuit **1002**. The sum circuit **1002** operates to sum each of the partial first electronic sound signals produced by the band split filters **902** prior to being supplied to the subtraction circuit **518**. The multi-microphone directional processing system **1000** thus compensates for delay induced by the

band split filters **902** and **904** by addition of the sum circuit **1002** to the multi-microphone directional processing system **1000**.

FIG. **11** is a block diagram of a multi-microphone directional processing system **1100** according to another embodiment of the invention. The multi-microphone directional processing system **1100** includes the band split filters **902** and **904** as discussed above with respect to FIG. **9**, and optionally includes the sum circuit **1002** as discussed above with respect to FIG. **10**. In addition, like FIG. **6**, the multi-microphone directional processing system **1100** utilizes the logarithm domain to effectively perform division operations in a multi-band adaptive manner. Hence, FIG. **11** represents a multi-band adaptive compensation scheme using the approach discussed above with respect to FIG. **6**.

The invention is preferably implemented in hardware, but can be implemented in software or a combination of hardware and software. The invention can also be embodied as computer readable code on a computer readable medium. The computer readable medium is any data storage device that can store data which can be thereafter be read by a computer system. Examples of the computer readable medium include read-only memory, random-access memory, CD-ROMs, magnetic tape, optical data storage devices, carrier waves. The computer readable medium can also be distributed over a network coupled computer systems so that the computer readable code is stored and executed in a distributed fashion.

The advantages of the invention are numerous. Different embodiments or implementations may yield one or more of the following advantages. One advantage of the invention is that directional noise suppression is not affected by microphone mismatch. Another advantage of the invention is that the directional noise suppression is not affected by the drift of microphone sensitivity over time. Still another advantage of the invention is that directional noise suppression is not affected by the non-uniform distribution of sound pressure in real-life application. Thus, the invention enables the multi-microphone system processing system to achieve superior directionality and consistent Signal-to-Noise Ratio (SNR) across all conditions.

The many features and advantages of the present invention are apparent from the written description and, thus, it is intended by the appended claims to cover all such features and advantages of the invention. Further, since numerous modifications and changes will readily occur to those skilled in the art, it is not desired to limit the invention to the exact construction and operation as illustrated and described. Hence, all suitable modifications and equivalents may be resorted to as falling within the scope of the invention.

What is claimed is:

1. An adaptive directional sound processing system, comprising:

at least first and second microphones spaced apart by a predetermined distance, said first microphone producing a first electronic sound signal and said second microphone producing a second electronic sound signal;

a first minimum estimate circuit operatively coupled to said first microphone, said first minimum estimate circuit produces a first minimum estimate for the first electronic sound signal from said first microphone;

a second minimum estimate circuit operatively coupled to said second microphone, said second minimum estimate circuit produces a second minimum estimate for the second electronic sound signal from said second microphone;

a divide circuit operatively connected to said first and second minimum estimate circuits, said divide circuit operates to produce a scaling signal from the first and second minimum estimates;

a multiply circuit operatively connected to said divide circuit and said second microphone, said multiply circuit operates to multiply the second electronic sound signal by the scaling signal to produce a scaled second electronic sound signal; and

a subtraction circuit operatively connected to said multiply circuit and said first microphone, said subtraction circuit producing an output difference signal by subtracting the scaled second electronic sound signal from the first electronic sound signal.

2. An adaptive directional sound processing system as recited in claim **1**, wherein said adaptive directional sound processing system further comprises:

a delay circuit that delays the second electronic sound signal or the scaled second electronic sound signal by a delay amount.

3. An adaptive directional sound processing system as recited in claim **1**, wherein said divide circuit operates in a linear domain.

4. An adaptive directional sound processing system as recited in claim **1**, wherein said divide circuit operates in a logarithm domain.

5. An adaptive directional sound processing system as recited in claim **1**, wherein said divide circuit comprises:

a first linear-to-logarithm conversion circuit operatively coupled to said first minimum estimate circuit to produce a converted first minimum estimate circuit;

a second linear-to-logarithm conversion circuit operatively coupled to said second minimum estimate circuit to produce a converted second minimum estimate circuit;

a subtraction circuit operatively connected to said a first linear-to-logarithm conversion circuit and said second linear-to-logarithm conversion circuit to produce a difference signal; and

a logarithm-to-linear conversion circuit operatively connected to said subtraction circuit to converted the difference signal to the scaling signal.

6. An adaptive directional sound processing system as recited in claim **1**, wherein at least one of said first minimum estimate circuit and said second minimum estimate circuit comprises:

a subtraction circuit that subtracts the first electronic sound signal from a previous minimum estimate in producing a difference signal;

a multiply circuit that multiplies the difference signal by a scale amount to produce an adjustment amount; and

an addition circuit that adds the adjustment amount to the previous minimum estimate in producing a current minimum estimate.

7. An adaptive directional sound processing system as recited in claim **1**, wherein, wherein said adaptive directional sound processing system resides within a hearing aid device.

8. A method for adaptively measuring and compensating for acoustical differences between sound signals picked up by microphones, said method comprising:

(a) receiving first and second electronic sound signals from first and second microphones, respectively;

(b) determining a compensation scaling amount that compensates for acoustic differences with respect to the first and second microphones;

(c) scaling the second electronic sound signal in accordance with the compensation scaling amount; and

(d) producing a differential electronic sound signal by subtracting the scaled second electronic sound signal from the first electronic sound signal,

wherein said determining (b) comprises:

(b1) determining a first minimum estimate of the first electronic sound signal;

(b2) determining a second minimum estimate of the second electronic sound signal; and

(b3) dividing the first minimum estimate by the second minimum estimate to produce the compensation scaling amount.

9. A method as recited in claim **8**, wherein the acoustic differences pertain to at least differences in microphone sensitivity.

10. A method as recited in claim **8**, wherein the microphones are provided within a hearing aid device, and wherein said method is performed by the hearing aid device.

11. A method for adaptively measuring and compensating for acoustical differences between sound signals picked up by microphones, said method comprising:

(a) receiving first and second electronic sound signals from first and second microphones, respectively;

(b) determining a compensation scaling amount that compensates for acoustic differences with respect to the first and second microphones;

(c) scaling the second electronic sound signal in accordance with the compensation scaling amount; and

(d) producing a differential electronic sound signal by subtracting the scaled second electronic sound signal from the first electronic sound signal,

wherein said determining (b) comprises: p**2** (b1) measuring a sensitivity difference between the first and second microphones while in use; and

(b2) producing the compensation scaling amount based on the sensitivity difference,

wherein the acoustic differences pertain to at least differences in microphone sensitivity, and

wherein said measuring (b1) of the sensitivity difference is performed using minimum estimates of the first and second sound signals or maximum estimates of the first and second sound signals.

12. A method as recited in claim **11**, wherein the microphones are provided within a hearing aid device, and wherein said method is performed by the hearing aid device.

13. A method for adaptively measuring and compensating for acoustical differences between sound signals picked up by microphones, said method comprising:

(a) receiving first and second electronic sound signals from first and second microphones, respectively;

(b) determining a compensation scaling amount that compensates for acoustic differences with respect to the first and second microphones;

(c) scaling the second electronic sound signal in accordance with the compensation scaling amount; and

(d) producing a differential electronic sound signal by subtracting the scaled second electronic sound signal from the first electronic sound signal,

wherein said determining (b) comprises:

(b1) determining a first minimum estimate of the first electronic sound signal;

(b2) determining a second minimum estimate of the second electronic sound signal;

(b3) converting the first minimum estimate to a logarithm scale first minimum estimate;

(b4) converting the second minimum estimate to a logarithm scale second minimum estimate;

(b5) subtracting the logarithm scale second minimum estimate from the logarithm scale first minimum estimate to produce a difference signal; and

(b6) converting the difference signal from the logarithm scale to a linear scale, the converted difference signal being the compensation scaling amount.

14. A method as recited in claim **13**, wherein the microphones are provided within a hearing aid device, and wherein said method is performed by the hearing aid device.

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