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IEEE Standard for Ethernet

Amendment 10: Media Access Control Parameters, Physical Layers, and Management Parameters for 200 Gb/s and 400 Gb/s Operation

IEEE Computer Society

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LAN/MAN Standards Committee

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New York, NY 10016-5997
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(Amendment to
IEEE Std 802.3™-2015
as amended by
IEEE Std 802.3bw™-2015,
IEEE Std 802.3by™-2016,
IEEE Std 802.3bq™-2016,
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IEEE Std 802.3bz™-2016,
IEEE Std 802.3bu™-2016,
IEEE Std 802.3bv™-2017, and
IEEE Std 802.3-2015/Cor 1-2017)

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IEEE Standard for Ethernet

Amendment 10: Media Access Control Parameters, Physical Layers, and Management Parameters for 200 Gb/s and 400 Gb/s Operation

LAN/MAN Standards Committee
of the
IEEE Computer Society

Approved 6 December 2017
of the
IEEE-SA Standards Board

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Abstract: Clause 116 through Clause 124 and Annex 119A through Annex 120E are added by this amendment to IEEE Std 802.3-2015. This amendment includes IEEE 802.3 Media Access Control (MAC) parameters, Physical Layer specifications, and management parameters for the transfer of IEEE 802.3 format frames at 200 Gb/s and 400 Gb/s.

Keywords: 200 Gb/s Ethernet, 200GAUI-4, 200GAUI-8, 200GBASE-DR4, 200GBASE-FR4, 200GBASE-LR4, 200GBASE-R, 200GMII, 200GXS, 400 Gb/s Ethernet, 400GAUI-8, 400GAUI-16, 400GBASE-DR4, 400GBASE-FR8, 400GBASE-LR8, 400GBASE-SR16, 400GBASE-R, 400GMII, 400GXS, EEE, Energy Efficient Ethernet, Ethernet, FEC, forward error correction, IEEE 802.3™, IEEE 802.3bs™, MMF, PAM4, Physical Medium Dependent sublayer, PMD, SMF

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Introduction

This introduction is not part of IEEE Std 802.3bs-2017, IEEE Standard for Ethernet—Amendment 10: Media Access Control Parameters, Physical Layers, and Management Parameters for 200 Gb/s and 400 Gb/s Operation.

IEEE Std 802.3™ was first published in 1985. Since the initial publication, many projects have added functionality or provided maintenance updates to the specifications and text included in the standard. Each IEEE 802.3 project/amendment is identified with a suffix (e.g., IEEE Std 802.3ba™-2010).

The half duplex Media Access Control (MAC) protocol specified in IEEE Std 802.3-1985 is Carrier Sense Multiple Access with Collision Detection (CSMA/CD). This MAC protocol was key to the experimental Ethernet developed at Xerox Palo Alto Research Center, which had a 2.94 Mb/s data rate. Ethernet at 10 Mb/s was jointly released as a public specification by Digital Equipment Corporation (DEC), Intel and Xerox in 1980. Ethernet at 10 Mb/s was approved as an IEEE standard by the IEEE Standards Board in 1983 and subsequently published in 1985 as IEEE Std 802.3-1985. Since 1985, new media options, new speeds of operation, and new capabilities have been added to IEEE Std 802.3. A full duplex MAC protocol was added in 1997.

Some of the major additions to IEEE Std 802.3 are identified in the marketplace with their project number. This is most common for projects adding higher speeds of operation or new protocols. For example, IEEE Std 802.3u™ added 100 Mb/s operation (also called Fast Ethernet), IEEE Std 802.3z added 1000 Mb/s operation (also called Gigabit Ethernet), IEEE Std 802.3ae added 10 Gb/s operation (also called 10 Gigabit Ethernet), IEEE Std 802.3ah™ specified access network Ethernet (also called Ethernet in the First Mile) and IEEE Std 802.3ba added 40 Gb/s operation (also called 40 Gigabit Ethernet) and 100 Gb/s operation (also called 100 Gigabit Ethernet). These major additions are all now included in and are superseded by IEEE Std 802.3-2015 and are not maintained as separate documents.

At the publication date of IEEE Std 802.3bs-2017, IEEE Std 802.3 is composed of the following documents:

IEEE Std 802.3-2015

Section One—Includes Clause 1 through Clause 20 and Annex A through Annex H and Annex 4A. Section One includes the specifications for 10 Mb/s operation and the MAC, frame formats and service interfaces used for all speeds of operation.

Section Two—Includes Clause 21 through Clause 33 and Annex 22A through Annex 33E. Section Two includes management attributes for multiple protocols and speed of operation as well as specifications for providing power over twisted pair cabling for multiple operational speeds. It also includes general information on 100 Mb/s operation as well as most of the 100 Mb/s Physical Layer specifications.

Section Three—Includes Clause 34 through Clause 43 and Annex 36A through Annex 43C. Section Three includes general information on 1000 Mb/s operation as well as most of the 1000 Mb/s Physical Layer specifications.

Section Four—Includes Clause 44 through Clause 55 and Annex 44A through Annex 55B. Section Four includes general information on 10 Gb/s operation as well as most of the 10 Gb/s Physical Layer specifications.

Section Five—Includes Clause 56 through Clause 77 and Annex 57A through Annex 76A. Clause 56 through Clause 67 and Clause 75 through Clause 77, as well as associated annexes, specify subscriber access and other Physical Layers and sublayers for operation from 512 kb/s to 10 Gb/s, and defines services and protocol elements that enable the exchange of IEEE Std 802.3 format frames between

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stations in a subscriber access network. Clause 68 specifies a 10 Gb/s Physical Layer specification. Clause 69 through Clause 74 and associated annexes specify Ethernet operation over electrical backplanes at speeds of 1000 Mb/s and 10 Gb/s.

Section Six—Includes Clause 78 through Clause 95 and Annex 83A through Annex 93C. Clause 78 specifies Energy-Efficient Ethernet. Clause 79 specifies IEEE 802.3 Organizationally Specific Link Layer Discovery Protocol (LLDP) type, length, and value (TLV) information elements. Clause 80 through Clause 95 and associated annexes includes general information on 40 Gb/s and 100 Gb/s operation as well the 40 Gb/s and 100 Gb/s Physical Layer specifications. Clause 90 specifies Ethernet support for time synchronization protocols.

IEEE Std 802.3bw™-2015

Amendment 1—This amendment includes changes to IEEE Std 802.3-2015 and adds Clause 96. This amendment adds 100 Mb/s Physical Layer (PHY) specifications and management parameters for operation on a single balanced twisted-pair copper cable.

IEEE Std 802.3by™-2016

Amendment 2—This amendment includes changes to IEEE Std 802.3-2015 and adds Clause 105 through Clause 112, Annex 109A, Annex 109B, Annex 109C, Annex 110A, Annex 110B, and Annex 110C. This amendment adds MAC parameters, Physical Layers, and management parameters for the transfer of IEEE 802.3 format frames at 25 Gb/s.

IEEE Std 802.3bq™-2016

Amendment 3—This amendment includes changes to IEEE Std 802.3-2015 and adds Clause 113 and Annex 113A. This amendment adds new Physical Layers for 25 Gb/s and 40 Gb/s operation over balanced twisted-pair structured cabling systems.

IEEE Std 802.3bp™-2016

Amendment 4—This amendment includes changes to IEEE Std 802.3-2015 and adds Clause 97 and Clause 98. This amendment adds point-to-point 1 Gb/s Physical Layer (PHY) specifications and management parameters for operation on a single balanced twisted-pair copper cable in automotive and other applications not utilizing the structured wiring plant.

IEEE Std 802.3br™-2016

Amendment 5—This amendment includes changes to IEEE Std 802.3-2015 and adds Clause 99. This amendment adds a MAC Merge sublayer and a MAC Merge Service Interface to support for Interspersing Express Traffic over a single link.

IEEE Std 802.3bn™-2016

Amendment 6—This amendment adds the Physical Layer specifications and management parameters for symmetric and/or asymmetric operation of up to 10 Gb/s on point-to-multipoint Radio Frequency (RF) distribution plants comprising either amplified or passive coaxial media. It also extends the operation of Ethernet Passive Optical Networks (EPON) protocols, such as Multipoint Control Protocol (MPCP) and Operation Administration and Management (OAM).

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IEEE Std 802.3bz™-2016

Amendment 7—This amendment includes changes to IEEE Std 802.3-2015 and adds Clause 125 and Clause 126. This amendment adds new rates of 2.5 Gb/s and 5 Gb/s and new Physical Layers for operation at 2.5 Gb/s and 5 Gb/s over balanced twisted-pair structured cabling systems.

IEEE Std 802.3bu™-2016

Amendment 8—This amendment includes changes to IEEE Std 802.3-2015 to define a methodology for the provision of power via a single twisted pair to connected Data Terminal Equipment (DTE) with IEEE 802.3 single twisted-pair interfaces.

IEEE Std 802.3bv™-2017

Amendment 9—This amendment includes changes to IEEE Std 802.3-2015 and adds Clause 115 and Annex 115A. This amendment adds point-to-point 1000 Mb/s Physical Layer (PHY) specifications and management parameters for operation on duplex plastic optical fiber (POF) targeting use in automotive, industrial, home-network, and other applications.

IEEE Std 802.3™-2015/Cor 1-2017

This corrigendum clarifies which lane of the media dependent interface (MDI) of a multi-lane Physical Layer entity (PHY) is used as the timestamping reference point.

IEEE Std 802.3bs™-2017

Amendment 10—This amendment includes changes to IEEE Std 802.3-2015 and adds Clause 116 through Clause 124 and Annex 119A through Annex 120E. This amendment adds MAC parameters, Physical Layers, and management parameters for the transfer of IEEE 802.3 format frames at 200 Gb/s and 400 Gb/s.

A companion document IEEE Std 802.3.1 describes Ethernet management information base (MIB) modules for use with the Simple Network Management Protocol (SNMP). IEEE Std 802.3.1 is updated to add management capability for enhancements to IEEE Std 802.3 after approval of the enhancements.

IEEE Std 802.3 will continue to evolve. New Ethernet capabilities are anticipated to be added within the next few years as amendments to this standard.

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IEEE Standard for Ethernet

Amendment 10: Media Access Control Parameters, Physical Layers, and Management Parameters for 200 Gb/s and 400 Gb/s Operation

NOTE—The editing instructions contained in this amendment define how to merge the material contained therein into the existing base standard and its amendments to form the comprehensive standard.

The editing instructions are shown in ***bold italic***. Four editing instructions are used: change, delete, insert, and replace. ***Change*** is used to make corrections in existing text or tables. The editing instruction specifies the location of the change and describes what is being changed by using ~~strikethrough~~ (to remove old material) and underscore (to add new material). ***Delete*** removes existing material. ***Insert*** adds new material without disturbing the existing material. Deletions and insertions may require renumbering. If so, renumbering instructions are given in the editing instruction. ***Replace*** is used to make changes in figures or equations by removing the existing figure or equation and replacing it with a new one. Editing instructions, change markings, and this NOTE will not be carried over into future editions because the changes will be incorporated into the base standard.

Cross references that refer to clauses, tables, equations, or figures not covered by this amendment are highlighted in green.

1. Introduction

1.1 Overview

1.1.3 Architectural perspectives

1.1.3.2 Compatibility interfaces

Insert the following compatibility interfaces after "100 Gb/s Parallel Physical Interface (CPPI)":

q) *200 Gb/s Media Independent Interface (200GMII).* The 200GMII is designed to connect a 200 Gb/s capable MAC to a 200 Gb/s PHY. While conformance with implementation of this interface is not necessary to ensure communication, it allows flexibility in intermixing PHYs and DTEs at 200 Gb/s speeds. The 200GMII is a logical interconnection intended for use as an intra-chip interface. No mechanical connector is specified for use with the 200GMII. The 200GMII is optional.

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r) *200 Gb/s Attachment Unit Interface (200GAUI-n)*. The 200GAUI-n is a physical instantiation of the PMA service interface to extend the connection between 200 Gb/s capable PMAs. While conformance with implementation of this interface is not necessary to ensure communication, it is recommended, since it allows maximum flexibility in intermixing PHYs and DTEs at 200 Gb/s speeds. The 200GAUI-n is intended for use as a chip-to-chip or a chip-to-module interface. Two widths of 200GAUI-n are defined: an eight-lane version (200GAUI-8) in Annex 120B and Annex 120C, and a four-lane version (200GAUI-4) in Annex 120D and Annex 120E. No mechanical connector is specified for use with the 200GAUI-n. The 200GAUI-n is optional.

s) *400 Gb/s Media Independent Interface (400GMII)*. The 400GMII is designed to connect a 400 Gb/s capable MAC to a 400 Gb/s PHY. While conformance with implementation of this interface is not necessary to ensure communication, it allows flexibility in intermixing PHYs and DTEs at 400 Gb/s speeds. The 400GMII is a logical interconnection intended for use as an intra-chip interface. No mechanical connector is specified for use with the 400GMII. The 400GMII is optional.

t) *400 Gb/s Attachment Unit Interface (400GAUI-n)*. The 400GAUI-n is a physical instantiation of the PMA service interface to extend the connection between 400 Gb/s capable PMAs. While conformance with implementation of this interface is not necessary to ensure communication, it is recommended, since it allows maximum flexibility in intermixing PHYs and DTEs at 400 Gb/s speeds. The 400GAUI-n is intended for use as a chip-to-chip or a chip-to-module interface. Two widths of 400GAUI-n are defined: a sixteen-lane version (400GAUI-16) in Annex 120B and Annex 120C, and an eight-lane version (400GAUI-8) in Annex 120D and Annex 120E. No mechanical connector is specified for use with the 400GAUI-n. The 400GAUI-n is optional.

1.3 Normative references

Insert the following references in alphanumeric order:

ANSI/TIA-604-18:2015, FOCIS 18—Fiber Optic Connector Intermateability Standard—Type MPO-16

IEC 61753-021-2:2007, Fibre optic interconnecting devices and passive components performance standard—Part 021-2: Grade C/3 single-mode fibre optic connectors for category C—Controlled environment.

Change the reference for IEC 60793-2-10 as follows:

IEC 60793-2-10~~:2011~~, Optical fibres—Part 2-10: Product specifications—Sectional specification for category A1 multimode fibres.

1.4 Definitions

Insert the following 17 new definitions into the list after 1.4.72a 40GBASE-T (as inserted by IEEE Std 802.3bq-2016):

1.4.72b 200GBASE-DR4: IEEE 802.3 Physical Layer specification for 200 Gb/s using 200GBASE-R encoding and 4-level pulse amplitude modulation over four lanes of single-mode fiber, with reach up to at least 500 m. (See IEEE Std 802.3, Clause 121.)

1.4.72c 200GBASE-FR4: IEEE 802.3 Physical Layer specification for 200 Gb/s using 200GBASE-R encoding and 4-level pulse amplitude modulation over four WDM lanes on single-mode fiber, with reach up to at least 2 km. (See IEEE Std 802.3, Clause 122.)

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1.4.72d 200GBASE-LR4: IEEE 802.3 Physical Layer specification for 200 Gb/s using 200GBASE-R encoding and 4-level pulse amplitude modulation over four WDM lanes on single-mode fiber, with reach up to at least 10 km. (See IEEE Std 802.3, Clause 122.)

1.4.72e 200GBASE-R: An IEEE 802.3 family of Physical Layer devices using the physical coding sublayer defined in Clause 119 for 200 Gb/s operation. (See IEEE Std 802.3, Clause 119.)

1.4.72f 200 Gb/s Attachment Unit Interface (200GAUI-n): A physical instantiation of the PMA service interface to extend the connection between 200 Gb/s capable PMAs over n lanes, used for chip-to-chip or chip-to-module interconnections. Two widths of 200GAUI-n are defined: an eight-lane version (200GAUI-8), and a four-lane version (200GAUI-4). (See IEEE Std 802.3, Annex 120B and Annex 120C for 200GAUI-8, or Annex 120D and Annex 120E for 200GAUI-4.)

1.4.72g 200 Gb/s Media Independent Interface (200GMII): The interface between the Reconciliation Sublayer (RS) and the Physical Coding Sublayer (PCS) for 200 Gb/s operation. (See IEEE Std 802.3, Clause 117.)

1.4.72h 200GMII Extender: The 200 Gb/s Media Independent Interface Extender extends the reach of the 200GMII and consists of two 200GXS sublayers with a 200GAUI-n between them. (See IEEE Std 802.3, Clause 118.)

1.4.72i 200GXS: The 200 Gb/s Extender Sublayer (200GXS) is part of the 200GMII Extender. In functionality, it is almost identical to the 200GBASE-R PCS Sublayer defined in Clause 119. Two types of 200GXS are defined: the DTE 200GXS adjacent to the RS sublayer and the PHY 200GXS adjacent to the PHY. (See IEEE Std 802.3, Clause 118.)

1.4.72j 400GBASE-DR4: IEEE 802.3 Physical Layer specification for 400 Gb/s using 400GBASE-R encoding and 4-level pulse amplitude modulation over four lanes of single-mode fiber, with reach up to at least 500 m. (See IEEE Std 802.3, Clause 124.)

1.4.72k 400GBASE-FR8: IEEE 802.3 Physical Layer specification for 400 Gb/s using 400GBASE-R encoding and 4-level pulse amplitude modulation over eight WDM lanes on single-mode fiber, with reach up to at least 2 km. (See IEEE Std 802.3, Clause 122.)

1.4.72l 400GBASE-LR8: IEEE 802.3 Physical Layer specification for 400 Gb/s using 400GBASE-R encoding and 4-level pulse amplitude modulation over eight WDM lanes on single-mode fiber, with reach up to at least 10 km. (See IEEE Std 802.3, Clause 122.)

1.4.72m 400GBASE-R: An IEEE 802.3 family of Physical Layer devices using the physical coding sublayer defined in Clause 119 for 400 Gb/s operation. (See IEEE Std 802.3, Clause 119.)

1.4.72n 400GBASE-SR16: IEEE 802.3 Physical Layer specification for 400 Gb/s using 400GBASE-R encoding over sixteen lanes of multimode fiber, with reach up to at least 100 m. (See IEEE Std 802.3, Clause 123.)

1.4.72o 400 Gb/s Attachment Unit Interface (400GAUI-n): A physical instantiation of the PMA service interface to extend the connection between 400 Gb/s capable PMAs over n lanes, used for chip-to-chip or chip-to-module interconnections. Two widths of 400GAUI-n are defined: a sixteen-lane version (400GAUI-16), and an eight-lane version (400GAUI-8). (See IEEE Std 802.3, Annex 120B and Annex 120C for 400GAUI-16, or Annex 120D and Annex 120E for 400GAUI-8.)

1.4.72p 400 Gb/s Media Independent Interface (400GMII): The interface between the Reconciliation Sublayer (RS) and the Physical Coding Sublayer (PCS) for 400 Gb/s operation. (See IEEE Std 802.3, Clause 117.)

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1.4.72q 400GMII Extender: The 400 Gb/s Media Independent Interface Extender extends the reach of the 400GMII and consists of two 400GXS sublayers with a 400GAUI-n between them. (See IEEE Std 802.3, Clause 118.)

1.4.72r 400GXS: The 400 Gb/s Extender Sublayer (400GXS) is part of the 400GMII Extender. In functionality, it is almost identical to the 400GBASE-R PCS Sublayer defined in Clause 119. Two types of 400GXS are defined: the DTE 400GXS adjacent to the RS sublayer and the PHY 400GXS adjacent to the PHY. (See IEEE Std 802.3, Clause 118.)

Change 1.4.107 (as modified by IEEE Std 802.3by-2016) as follows:

1.4.107 BASE-R: An IEEE 802.3 family of Physical Layer devices using the 64B/66B encoding defined in Clause 49, Clause 82, ~~or~~ Clause 107, or Clause 119. (See IEEE Std 802.3, Clause 49, Clause 82, ~~or~~ Clause 107, or Clause 119.)

Change 1.4.325 as follows:

1.4.325 PCS lane (PCSL): In 40GBASE-R, ~~and~~ 100GBASE-R, 200GBASE-R, and 400GBASE-R, the PCS distributes encoded data to multiple logical lanes, these logical lanes are called PCS lanes. One or more PCS lanes can be multiplexed and carried on a physical lane together at the PMA service interface. (See IEEE Std 802.3, Clause 83 and Clause 120.)

1.5 Abbreviations

Insert the following new abbreviations into the list, in alphanumeric order:

200GAUI-n	200 Gb/s Attachment Unit Interface over n lanes
200GMII	200 Gb/s Media Independent Interface
200GXS	200GMII Extender Sublayer
400GAUI-n	400 Gb/s Attachment Unit Interface over n lanes
400GMII	400 Gb/s Media Independent Interface
400GXS	400GMII Extender Sublayer
SER	symbol error ratio

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4. Media Access Control

4.4 Specific implementations

4.4.2 MAC parameters

Change Table 4-2 (as modified by IEEE Std 802.3by-2016 and IEEE Std 802.3bz-2016) as follows:

Table 4–2—MAC parameters

Parameters	MAC data rate			
	Up to and including 100 Mb/s	**1 Gb/s**	**2.5 Gb/s, 5 Gb/s, 25 Gb/s, 40 Gb/s, ~~and~~ 100 Gb/s, 200 Gb/s, and 400 Gb/s**	**10 Gb/s**
slotTime	512 bit times	4096 bit times	not applicable	not applicable
interPacketGap[a]	96 bits	96 bits	96 bits	96 bits
attemptLimit	16	16	not applicable	not applicable
backoffLimit	10	10	not applicable	not applicable
jamSize	32 bits	32 bits	not applicable	not applicable
maxBasicFrameSize	1518 octets	1518 octets	1518 octets	1518 octets
maxEnvelopeFrameSize	2000 octets	2000 octets	2000 octets	2000 octets
minFrameSize	512 bits (64 octets)	512 bits (64 octets)	512 bits (64 octets)	512 bits (64 octets)
burstLimit	not applicable	65 536 bits	not applicable	not applicable
ipgStretchRatio	not applicable	not applicable	not applicable	104 bits

[a] References to interFrameGap or interFrameSpacing in other clauses (e.g., 13, 35, and 42) shall be interpreted as interPacketGap.

Change Note 7 below Table 4-2 as follows:

NOTE 7—For 40 Gb/s, ~~and~~ 100 Gb/s, 200 Gb/s, and 400 Gb/s operation, the received interpacket gap (the spacing between two packets, from the last bit of the FCS field of the first packet to the first bit of the Preamble of the second packet) can have a minimum value of 8 BT (bit times), as measured at the XLGMII, ~~or~~ CGMII, 200GMII, or 400GMII receive signals at the DTE due to clock tolerance and lane alignment requirements.

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30. Management

30.2 Managed objects

30.2.5 Capabilities

Insert two new rows between the rows for aRSFECIndicationEnable and aSNROpMarginChnlA in Table 30-1e (as modified by IEEE Std 802.3bq-2016) as follows (unchanged rows not shown):

Table 30–1e—Capabilities

			DTE								Repeater					MAU									
			Basic Package (mandatory)	Mandatory Package (mandatory)	Recommended Package (optional)	Optional Package (optional)	Array Package (optional)	Excessive Deferral Package (optional)	Multiple PHY Package (optional)	PHY Error Monitor Capability (optional)	Basic Control Capability (mandatory)	Performance Monitor Capability (optional)	Address Tracking Capability (optional)	100/1000 Mb/s Monitor Capability (optional)	1000 Mb/s Burst Monitor Capability (optional)	Basic Package (mandatory)	MAU Control Package (optional)	Media Loss Tracking Package (conditional)	Broadband DTE MAU Package (conditional)	MII Capability (conditional)	PHY Error Monitor Capability (optional)	MultiGBASE-T Operating Margin package (conditional)	Forward Error Correction Package (conditional)	Energy-Efficient Ethernet (optional)	Auto-Negotiation Package (mandatory)
…																									
	aPCSFECIndicationAbility	ATTRIBUTE	GET																				X		
	aPCSFECIndicationEnable	ATTRIBUTE	GET-SET																				X		
…																									

30.3 Layer management for DTEs

30.3.2 PHY device managed object class

30.3.2.1 PHY device attributes

30.3.2.1.2 aPhyType

Insert 200GBASE-R and 400GBASE-R PHY types at the end of "APPROPRIATE SYNTAX" section of 30.3.2.1.2 as follows:

APPROPRIATE SYNTAX:

...

200GBASE-R	Clause 119 200 Gb/s multi-PCS lane 64B/66B
400GBASE-R	Clause 119 400 Gb/s multi-PCS lane 64B/66B

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30.3.2.1.3 aPhyTypeList

Insert 200GBASE-R and 400GBASE-R PHY types at the end of "APPROPRIATE SYNTAX" section of 30.3.2.1.3 as follows:

APPROPRIATE SYNTAX:

...

200GBASE-R	Clause 119 200 Gb/s multi-PCS lane 64B/66B
400GBASE-R	Clause 119 400 Gb/s multi-PCS lane 64B/66B

30.3.2.1.5 aSymbolErrorDuringCarrier

Change the fourth paragraph of the "BEHAVIOUR DEFINED AS" section of 30.3.2.1.5 (as modified by IEEE Std 802.3by-2016) as follows:

BEHAVIOUR DEFINED AS:

For operation at 10 Gb/s, 25 Gb/s, 40 Gb/s, ~~and~~ 100 Gb/s, 200 Gb/s, and 400 Gb/s, it is a count of the number of times the receiving media is non-idle (the time between the Start of Packet Delimiter and the End of Packet Delimiter as defined by 46.2.5 and 81.2.5) for a period of time equal to or greater than minFrameSize, and during which there was at least one occurrence of an event that causes the PHY to indicate "Receive Error" on the media independent interface (see Table 46–4 and Table 81–3).

30.5 Layer management for medium attachment units (MAUs)

30.5.1 MAU managed object class

30.5.1.1 MAU attributes

30.5.1.1.2 aMAUType

Insert new MAU types immediately above 802.9a in "APPROPRIATE SYNTAX" section of 30.5.1.1.2 as follows:

APPROPRIATE SYNTAX:

...

200GBASE-R	Multi-lane PCS as specified in Clause 119 over undefined PMA/PMD
200GBASE-DR4	200GBASE-R PCS/PMA over 4-lane single-mode fiber PMD as specified in Clause 121
200GBASE-FR4	200GBASE-R PCS/PMA over 4 WDM lane single-mode fiber PMD with reach up to at least 2 km as specified in Clause 122
200GBASE-LR4	200GBASE-R PCS/PMA over 4 WDM lane single-mode fiber PMD with reach up to at least 10 km as specified in Clause 122
400GBASE-R	Multi-lane PCS as specified in Clause 119 over undefined PMA/PMD
400GBASE-SR16	400GBASE-R PCS/PMA over 16-lane multimode fiber PMD as specified in Clause 123
400GBASE-DR4	400GBASE-R PCS/PMA over 4-lane single-mode fiber PMD as specified in Clause 124
400GBASE-FR8	400GBASE-R PCS/PMA over 8 WDM lane single-mode fiber PMD with reach up to at least 2 km as specified in Clause 122
400GBASE-LR8	400GBASE-R PCS/PMA over 8 WDM lane single-mode fiber PMD with reach up to at least 10 km as specified in Clause 122

...

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Change the last paragraph of the "BEHAVIOUR DEFINED AS" section of 30.5.1.1.2 (as modified by IEEE Std 802.3by-2016) as follows:

The enumerations 1000BASE-X, 1000BASE-XHD, 1000BASE-XFD, 10GBASE-X, 10GBASE-R, 10GBASE-W, 25GBASE-R, 40GBASE-R, ~~and~~ 100GBASE-R, 200GBASE-R, and 400GBASE-R shall only be returned if the underlying PMD type is unknown.;

30.5.1.1.4 aMediaAvailable

Change the sixth paragraph of the "BEHAVIOUR DEFINED AS" section of 30.5.1.1.4 (as modified by IEEE Std 802.3bq-2016) as follows:

For 40 Gb/s, ~~and~~ 100 Gb/s, 200 Gb/s, and 400 Gb/s, the enumerations map to value of the link_fault variable (see 81.3.4) within the Link Fault Signaling state diagram (see 81.3.4.1 and Figure 46–11) as follows: the values OK and Link Interruption map to the enumeration "available", the value Local Fault maps to the enumeration "not available" and the value Remote Fault maps to the enumeration "remote fault."

30.5.1.1.12 aLaneMapping

Change 30.5.1.1.12 as follows:

ATTRIBUTE

APPROPRIATE SYNTAX:
A SEQUENCE of INTEGERs.

BEHAVIOUR DEFINED AS:
For 40/100/200/400GBASE-R PHYs and 100GBASE-P PHYs, an array of PCS lane identifiers. The indices of this array (0 to n – 1) denote the service interface lane number where n is the number of PCS lanes in use. Each element of this array contains the PCS lane number for the PCS lane that has been detected in the corresponding service interface lane.
If a Clause 45 MDIO Interface to the PCS is present, then this attribute will map to the Lane mapping registers (see 45.2.3.46 and 45.2.3.47).;

30.5.1.1.15 aFECAbility

Change 30.5.1.1.15 (as modified by IEEE Std 802.3by-2016) as follows:

ATTRIBUTE

APPROPRIATE SYNTAX:
An ENUMERATED VALUE that meets the requirement of the description below

unknown	initializing, true state not yet known
supported	FEC supported
not supported	FEC not supported

BEHAVIOUR DEFINED AS:
A read-only value that indicates if the PHY supports ~~a FEC sublayer for~~ forward error correction (see 65.2, Clause 74, Clause 91, ~~and~~ Clause 108, and Clause 119).

If a Clause 45 MDIO Interface is present and support for FEC is optional, then this attribute maps to the FEC capability register (see 45.2.8.2 or 45.2.1.92).;

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30.5.1.1.17 aFECCorrectedBlocks

Change 30.5.1.1.17 (as modified by IEEE Std 802.3by-2016) as follows:

ATTRIBUTE

APPROPRIATE SYNTAX:

A SEQUENCE of generalized nonresettable counters. Each counter has a maximum increment rate of 1 200 000 counts per second for 1000 Mb/s implementations, 5 000 000 counts per second for 10 Gb/s, 25 Gb/s, and 40 Gb/s implementations, ~~and~~ 2 500 000 counts per second for 100 Gb/s implementations, 40 000 000 counts per second for 200 Gb/s implementations, and 80 000 000 counts per second for 400 Gb/s implementations.

BEHAVIOUR DEFINED AS:

For 1000BASE-PX, 10/25/40/100/200/400GBASE-R, 100GBASE-P, 10GBASE-PR, or 10/1GBASE-PRX PHYs, an array of corrected FEC block counters. The counters do not increment for other PHY types. The indices of this array (0 to N – 1) denote the FEC sublayer instance number where N is the number of FEC sublayer instances in use. The number of FEC sublayer instances in use is set to one for PHYs that do not use PCS lanes or use a single FEC instance for all ~~multiple FEC~~ lanes. Each element of this array contains a count of corrected FEC blocks for that FEC sublayer instance.

Increment the counter by one for each received block that is corrected by the FEC function in the PHY for the corresponding lane or FEC sublayer instance.

If a Clause 45 MDIO Interface is present, then this attribute maps to the FEC corrected blocks counter(s) (see 45.2.8.5 and 45.2.1.94 for 10GBASE-R, 45.2.3.39 for 10GBASE-PR and 10/1GBASE-PRX, 45.2.1.116 for BASE-R, ~~and~~ 45.2.1.103 for RS-FEC, and 45.2.3.47l for PCS FEC).;

30.5.1.1.18 aFECUncorrectableBlocks

Change 30.5.1.1.18 (as modified by IEEE Std 802.3by-2016) as follows:

ATTRIBUTE

APPROPRIATE SYNTAX:

A SEQUENCE of generalized nonresettable counters. Each counter has a maximum increment rate of 1 200 000 counts per second for 1000 Mb/s implementations, and 5 000 000 counts per second for 10 Gb/s, 25 Gb/s, and 40 Gb/s implementations, ~~and~~ 2 500 000 counts per second for 100 Gb/s implementations, 40 000 000 counts per second for 200 Gb/s implementations, and 80 000 000 counts per second for 400 Gb/s implementations.

BEHAVIOUR DEFINED AS:

For 1000BASE-PX, 10/25/40/100/200/400GBASE-R, 100GBASE-P, 10GBASE-PR, or 10/1GBASE-PRX PHYs, an array of uncorrectable FEC block counters. The counters do not increment for other PHY types. The indices of this array (0 to N – 1) denote the FEC sublayer instance number where N is the number of FEC sublayer instances in use. The number of FEC sublayer instances in use is set to one for PHYs that do not use PCS lanes or use a single FEC instance for all ~~multiple FEC~~ lanes. Each element of this array contains a count of uncorrectable ~~corrected~~ FEC blocks for that FEC sublayer instance.

Increment the counter by one for each FEC block that is determined to be uncorrectable by the FEC function in the PHY for the corresponding lane or FEC sublayer instance.

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If a Clause 45 MDIO Interface is present, then this attribute maps to the FEC uncorrectable blocks counter(s) (see 45.2.8.6 and 45.2.1.95 for 10GBASE-R, 45.2.3.40 for 10GBASE-PR and 10/1GBASE-PRX, 45.2.1.117 for BASE-R, ~~and~~ 45.2.1.104 for RS-FEC, and 45.2.3.47m for PCS FEC).;

Insert 30.5.1.1.32 and 30.5.1.1.33 after 30.5.1.1.31 as follows:

30.5.1.1.32 aPCSFECIndicationAbility

ATTRIBUTE

APPROPRIATE SYNTAX:

An ENUMERATED VALUE that meets the requirement of the following description:

unknown	initializing, true state not yet known
supported	PCS FEC error indication bypass ability supported
not supported	PCS FEC error indication bypass ability not supported

BEHAVIOUR DEFINED AS:

A read-only value that indicates if the PHY supports an optional PCS FEC error indication bypass ability (see 119.2.5.3).

If a Clause 45 MDIO Interface is present, then this attribute maps to the PCS FEC status register (see 45.2.3.47k).;

30.5.1.1.33 aPCSFECIndicationEnable

ATTRIBUTE

APPROPRIATE SYNTAX:

An ENUMERATED VALUE that meets the requirement of the following description:

unknown	initializing, true state not yet known
disabled	PCS FEC error indication bypass disabled
enabled	PCS FEC error indication bypass enabled

BEHAVIOUR DEFINED AS:

A read-write value that indicates the mode of operation of the PCS FEC error indication bypass function (see 119.2.5.3).

A GET operation returns the current mode of operation of the PCS FEC. A SET operation changes the mode of operation of the PCS FEC to the indicated value.

If a Clause 45 MDIO Interface is present, then this attribute maps to the RS-FEC control register (see 45.2.3.47j).;

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45. Management Data Input/Output (MDIO) Interface

45.2 MDIO Interface Registers

45.2.1 PMA/PMD registers

Change the first sentence of 45.2.1 (as modified by IEEE Std 802.3by-2016) as follows:

For devices operating at 25 Gb/s or higher speeds, the PMA may be instantiated as multiple sublayers (see 83.1.4, ~~and~~ 109.1.4, and 120.1.4 for how MMD addresses are allocated to multiple PMA sublayers for the respective speeds).

Change the reserved row for 1.23 through 1.29 in Table 45-3 (as modified by IEEE Std 802.3bv-2017) as follows (unchanged rows not shown):

Table 45–3—PMA/PMD registers

Register address	Register name	Subclause
…		
1.23	200G PMA/PMD extended ability	45.2.1.14e
1.24	400G PMA/PMD extended ability	45.2.1.14f
1.25, 1.26	Reserved	
1.27	PMD transmit disable extension	45.2.1.14g
1.28	PMD receive signal detect extension	45.2.1.14h
~~1.23 through~~ 1.29	Reserved	
…		

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Change the reserved row for 1.340 through 1.699 in Table 45-3 as follows (unchanged rows not shown):

Table 45–3—PMA/PMD registers

Register address	Register name	Subclause
…		
1.340 through 1.399~~699~~	Reserved	
1.400 through 1.407	200GAUI-8 and 400GAUI-16 chip-to-module recommended CTLE, lane 0 through lane 7	45.2.1.116a, 45.2.1.116b
1.408 through 1.415	400GAUI-16 chip-to-module recommended CTLE, lane 8 through lane 15	45.2.1.116c
1.416 through 1.499	Reserved	
1.500 through 1.515	200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, receive direction, lane 0 through lane 15	45.2.1.116d, 45.2.1.116e
1.516 through 1.531	200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, transmit direction, lane 0 through lane 15	45.2.1.116f, 45.2.1.116g
1.532 through 1.699	Reserved	
…		

Change the reserved row for 1.1511 through 1.1599 in Table 45-3 as follows (unchanged rows not shown):

Table 45–3—PMA/PMD registers

Register address	Register name	Subclause
…		
1.1511~~ through 1.1599~~	Reserved	
1.1512	PRBS13Q testing control	45.2.1.125a
1.1513 through 1.1599	Reserved	
…		

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Change the rows for 1.1600 through 1.1799 in Table 45-3 as follows (unchanged rows not shown):

Table 45–3—PMA/PMD registers

Register address	Register name	Subclause
…		
1.1600 through 1.16150~~9~~	PRBS Tx error counters, lane 0 through lane 15~~9~~	45.2.1.126
1.1616~~0~~ through 1.1699	Reserved	
1.1700 through 1.17150~~9~~	PRBS Rx error counters, lane 0 through lane 15~~9~~	45.2.1.127
1.1716~~0~~ through 1.1799	Reserved	
…		

45.2.1.1 PMA/PMD control 1 register (Register 1.0)

Change the row for bits 1.0.5:2 in Table 45-4 (as modified by IEEE Std 802.3by-2016 and IEEE Std 802.3bz-2016) as follows (unchanged rows not shown):

Table 45–4—PMA/PMD control 1 register bit definitions

Bit(s)	Name	Description	R/W[a]
…			
1.0.5:2	Speed selection	5 4 3 2 ~~1 x x x = Reserved~~ 1 1 x x = Reserved 1 0 1 x = Reserved 1 0 0 1 = 400 Gb/s 1 0 0 0 = 200 Gb/s 0 1 1 1 = 5 Gb/s 0 1 1 0 = 2.5 Gb/s 0 1 0 1 = Reserved 0 1 0 0 = 25 Gb/s 0 0 1 1 = 100 Gb/s 0 0 1 0 = 40 Gb/s 0 0 0 1 = 10PASS-TS/2BASE-TL 0 0 0 0 = 10 Gb/s	R/W
…			

[a] R/W = Read/Write, SC = Self-clearing, RO = Read only

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45.2.1.1.3 Speed selection (1.0.13, 1.0.6, 1.0.5:2)

Change the last paragraph of 45.2.1.1.3 (as modified by IEEE Std 802.3by-2016 and IEEE Std 802.3bz-2016) as follows:

When bits 5 through 2 are set to 0010 the use of a 40G PMA/PMD is selected; when set to 0011 the use of a 100G PMA/PMD is selected; when set to 0100 the use of a 25G PMA/PMD is selected; when set to 0110 the use of a 2.5G PMA/PMD is selected; when set to 0111 the use of a 5G PMA/PMD is selected; when set to 1000 the use of a 200G PMA/PMD is selected; when set to 1001 the use of a 400G PMA/PMD is selected. More specific selection is performed using the PMA/PMD control 2 register (Register 1.7) (see 45.2.1.6.3).

45.2.1.1.4 PMA remote loopback (1.0.1)

Change the last two sentences of the second paragraph of 45.2.1.1.4 as follows:

~~For 40/100 Gb/s operation, the remote loopback functionality is detailed in 83.5.9.~~ For 25/40/100 Gb/s operation, the remote loopback ability bit is specified in register 1.13~~ the 40G/100G PMA/PMD extended ability register~~. For 200 Gb/s and 400 Gb/s operation, the remote loopback ability bit is specified in registers 1.23 and 1.24, respectively.

45.2.1.1.5 PMA local loopback (1.0.0)

Change the second paragraph of 45.2.1.1.5 as follows:

The local loopback function is mandatory for the 1000BASE-KX, 10GBASE-KR, 10GBASE-X, 40GBASE-KR4, 40GBASE-CR4, and 100GBASE-CR10 port type and optional for all other port types, except 2BASE-TL, 10PASS-TS, and 10/1GBASE-PRX, which do not support loopback. ~~A device's ability to perform the local loopback function is advertised in the local loopback ability bit of the related speed-dependent status register.~~ A PMA that is unable to perform the local loopback function shall ignore writes to this bit and shall return a value of zero when read. ~~For 10 Gb/s operation, the~~ The local loopback functionality is detailed in the relevant PMA clause~~ 48.3.3 and 51.8~~. ~~For 40/100 Gb/s operation, the local loopback functionality is detailed in 83.5.8.~~ For 10/25/40/100/200/400 Gb/s operation, the local loopback ability bit is specified in the PMA/PMD status 2 register.

45.2.1.2 PMA/PMD status 1 register (Register 1.1)

45.2.1.2.3 Fault (1.1.7)

Change the fourth sentence of 45.2.1.2.3 (as modified by IEEE Std 802.3by-2016) as follows:

For 10/25/40/100/200/400 Gb/s operation, bit 1.1.7 is set to a one when either of the fault bits (1.8.11, 1.8.10) located in register 1.8 are set to a one.

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45.2.1.4 PMA/PMD speed ability (Register 1.4)

Change the reserved rows of Table 45–6 (as modified by IEEE Std 802.3bz-2016) as follows (unchanged rows not shown):

Table 45–6—PMA/PMD speed ability register bit definitions

Bit(s)	Name	Description	R/W[a]
1.4.15	~~Reserved for future speeds~~ 400G capable	~~Value always 0~~ 1 = PMA/PMD is capable of operating at 400 Gb/s 0 = PMA/PMD is not capable of operating at 400 Gb/s	RO
…			
1.4.12	~~Reserved for future speeds~~ 200G capable	~~Value always 0~~ 1 = PMA/PMD is capable of operating at 200 Gb/s 0 = PMA/PMD is not capable of operating at 200 Gb/s	RO
…			

[a] RO = Read only

Insert 45.2.1.4.aaa before 45.2.1.4.aa (as inserted by IEEE Std 802.3bz-2016) as follows:

45.2.1.4.aaa 400G capable (1.4.15)

When read as a one, bit 1.4.15 indicates that the PMA/PMD is able to operate at a data rate of 400 Gb/s. When read as a zero, bit 1.4.15 indicates that the PMA/PMD is not able to operate at a data rate of 400 Gb/s.

Insert 45.2.1.4.ac after 45.2.1.4.ab (as inserted by IEEE Std 802.3bz-2016) and before 45.2.1.4.a (as inserted by IEEE Std 802.3by-2016) as follows:

45.2.1.4.ac 200G capable (1.4.12)

When read as a one, bit 1.4.12 indicates that the PMA/PMD is able to operate at a data rate of 200 Gb/s. When read as a zero, bit 1.4.12 indicates that the PMA/PMD is not able to operate at a data rate of 200 Gb/s.

45.2.1.6 PMA/PMD control 2 register (Register 1.7)

Change Table 45-7 (as modified by IEEE Std 802.3bw-2015, IEEE Std 802.3by-2016, IEEE Std 802.3bq-2016, IEEE Std 802.3bp-2016, IEEE Std 802.3bn-2016, IEEE Std 802.3bz-2016, and IEEE Std 802.3bv-2017) as follows:

Table 45–7—PMA/PMD control 2 register bit definitions

Bit(s)	Name	Description	R/W[a]
1.7.15:10	Reserved	Value always 0	RO
1.7.9	PIASE	PMA ingress AUI stop enable	R/W
1.7.8	PEASE	PMA egress AUI stop enable	R/W

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Table 45–7—PMA/PMD control 2 register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
~~1.7.7:6~~ 1.7.7	Reserved	Value always 0	RO
~~1.7.5:0~~ 1.7.6:0	PMA/PMD type selection	6 5 4 3 2 1 0 1 1 x x x x x = reserved 1 0 1 1 1 1 x = reserved 1 0 1 1 1 0 1 = reserved 1 0 1 1 1 0 0 = 400GBASE-LR8 PMA/PMD 1 0 1 1 0 1 1 = 400GBASE-FR8 PMA/PMD 1 0 1 1 0 1 0 = 400GBASE-DR4 PMA/PMD 1 0 1 1 0 0 1 = 400GBASE-SR16 PMA/PMD 1 0 1 1 0 0 0 = reserved 1 0 1 0 1 1 x = reserved 1 0 1 0 1 0 1 = 200GBASE-LR4 PMA/PMD 1 0 1 0 1 0 0 = 200GBASE-FR4 PMA/PMD 1 0 1 0 0 1 1 = 200GBASE-DR4 PMA/PMD 1 0 1 0 0 1 0 = reserved 1 0 1 0 0 0 x = reserved 1 0 0 x x x x = reserved 0 1 1 1 1 1 x = reserved 0 1 1 1 1 0 1 = BASE-T1 PMA/PMD[b] 0 1 1 1 1 0 0 = reserved 0 1 1 1 0 1 1 = reserved 0 1 1 1 0 1 0 = 25GBASE-SR PMA/PMD 0 1 1 1 0 0 1 = 25GBASE-KR or 25GBASE-KR-S PMA/PMD 0 1 1 1 0 0 0 = 25GBASE-CR or 25GBASE-CR-S PMA/PMD 0 1 1 0 1 1 1 = 25GBASE-T PMA 0 1 1 0 1 1 0 = reserved 0 1 1 0 1 0 1 = reserved 0 1 1 0 1 0 0 = BASE-H PMA/PMD[c] 0 1 1 0 0 1 1 = 10GPASS-XR-U PMA/PMD 0 1 1 0 0 1 0 = 10GPASS-XR-D PMA/PMD 0 1 1 0 0 0 1 = 5GBASE-T PMA 0 1 1 0 0 0 0 = 2.5GBASE-T PMA 0 1 0 1 1 1 1 = 100GBASE-SR4 PMA/PMD 0 1 0 1 1 1 0 = 100GBASE-CR4 PMA/PMD 0 1 0 1 1 0 1 = 100GBASE-KR4 PMA/PMD 0 1 0 1 1 0 0 = 100GBASE-KP4 PMA/PMD 0 1 0 1 0 1 1 = 100GBASE-ER4 PMA/PMD 0 1 0 1 0 1 0 = 100GBASE-LR4 PMA/PMD 0 1 0 1 0 0 1 = 100GBASE-SR10 PMA/PMD 0 1 0 1 0 0 0 = 100GBASE-CR10 PMA/PMD 0 1 0 0 1 1 1 = reserved	R/W

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Table 45–7—PMA/PMD control 2 register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
1.7.6:0 (*continued*)	PMA/PMD type selection (*continued*)	0 1 0 0 1 1 0 = 40GBASE-T PMA 0 1 0 0 1 0 1 = 40GBASE-ER4 PMA/PMD 0 1 0 0 1 0 0 = 40GBASE-FR PMA/PMD 0 1 0 0 0 1 1 = 40GBASE-LR4 PMA/PMD 0 1 0 0 0 1 0 = 40GBASE-SR4 PMA/PMD 0 1 0 0 0 0 1 = 40GBASE-CR4 PMA/PMD 0 1 0 0 0 0 0 = 40GBASE-KR4 PMA/PMD 0 0 1 1 1 1 1 = 10/1GBASE-PRX-U4 0 0 1 1 1 1 0 = 10GBASE-PR-U4 0 0 1 1 1 0 1 = 10/1GBASE-PRX-D4 0 0 1 1 1 0 0 = 10GBASE-PR-D4 0 0 1 1 0 1 1 = reserved 0 0 1 1 0 1 0 = 10GBASE-PR-U3 0 0 1 1 0 0 1 = 10GBASE-PR-U1 0 0 1 1 0 0 0 = 10/1GBASE-PRX-U3 0 0 1 0 1 1 1 = 10/1GBASE-PRX-U2 0 0 1 0 1 1 0 = 10/1GBASE-PRX-U1 0 0 1 0 1 0 1 = 10GBASE-PR-D3 0 0 1 0 1 0 0 = 10GBASE-PR-D2 0 0 1 0 0 1 1 = 10GBASE-PR-D1 0 0 1 0 0 1 0 = 10/1GBASE-PRX-D3 0 0 1 0 0 0 1 = 10/1GBASE-PRX-D2 0 0 1 0 0 0 0 = 10/1GBASE-PRX-D1 0 0 0 1 1 1 1 = 10BASE-T PMA/PMD 0 0 0 1 1 1 0 = 100BASE-TX PMA/PMD 0 0 0 1 1 0 1 = 1000BASE-KX PMA/PMD 0 0 0 1 1 0 0 = 1000BASE-T PMA/PMD 0 0 0 1 0 1 1 = 10GBASE-KR PMA/PMD 0 0 0 1 0 1 0 = 10GBASE-KX4 PMA/PMD 0 0 0 1 0 0 1 = 10GBASE-T PMA 0 0 0 1 0 0 0 = 10GBASE-LRM PMA/PMD 0 0 0 0 1 1 1 = 10GBASE-SR PMA/PMD 0 0 0 0 1 1 0 = 10GBASE-LR PMA/PMD 0 0 0 0 1 0 1 = 10GBASE-ER PMA/PMD 0 0 0 0 1 0 0 = 10GBASE-LX4 PMA/PMD 0 0 0 0 0 1 1 = 10GBASE-SW PMA/PMD 0 0 0 0 0 1 0 = 10GBASE-LW PMA/PMD 0 0 0 0 0 0 1 = 10GBASE-EW PMA/PMD 0 0 0 0 0 0 0 = 10GBASE-CX4 PMA/PMD	

[a] R/W = Read/Write, RO = Read only
[b] If BASE-T1 is selected, bits 1.2100.3:0 are used to differentiate which BASE-T1 PMA/PMD is selected.
[c] If BASE-H PMA/PMD is selected, register 1.900 is used to differentiate which BASE-H PMA/PMD is selected.

Change the title and text of 45.2.1.6.3 as follows:

45.2.1.6.3 PMA/PMD type selection (1.7.6~~5~~:0)

The PMA/PMD type of the PMA/PMD shall be selected using bits ~~5~~6 to 0. The PMA/PMD type abilities of the PMA/PMD are advertised in bits 9 and 7 through 0 of the PMA/PMD status 2 register; the PMA/PMD extended ability register; ~~and~~ the 40G/100G PMA/PMD extended ability register; the 200G PMA/PMD extended ability register, and the 400G PMA/PMD extended ability register. A PMA/PMD shall ignore writes to the PMA/PMD type selection bits that select PMA/PMD types it has not advertised. It is the responsibility of the STA entity to ensure that mutually acceptable MMD types are applied consistently across all the MMDs on a particular PHY.

The PMA/PMD type selection defaults to a supported ability.

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45.2.1.7 PMA/PMD status 2 register (Register 1.8)

45.2.1.7.4 Transmit fault (1.8.11)

Insert the following rows at the bottom of Table 45-9 as follows (unchanged rows not shown):

Table 45–9—Transmit fault description location

PMA/PMD	Description location
…	
200GBASE-DR4	121.5.10
200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, 400GBASE-LR8	122.5.10
400GBASE-SR16	123.5.10
400GBASE-DR4	124.5.10

45.2.1.7.5 Receive fault (1.8.10)

Insert the following rows at the bottom of Table 45-10 as follows (unchanged rows not shown):

Table 45–10—Receive fault description location

PMA/PMD	Description location
…	
200GBASE-DR4	121.5.11
200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, 400GBASE-LR8	122.5.11
400GBASE-SR16	123.5.11
400GBASE-DR4	124.5.11

45.2.1.8 PMD transmit disable register (Register 1.9)

Change the first paragraph of 45.2.1.8 as follows:

The assignment of bits in the PMD transmit disable register is shown in Table 45–11. The transmit disable functionality is optional and a PMD's ability to perform the transmit disable functionality is advertised in the PMD transmit disable ability bit 1.8.8. A PMD that does not implement the transmit disable functionality shall ignore writes to the PMD transmit disable register and may return a value of zero for all bits. A PMD device that operates using a single lane and has implemented the transmit disable function shall use bit 1.9.0 to control the function. Such devices shall ignore writes to bits 1.9.~~1~~0<u>15</u>:1 and return a value of zero for those bits when they are read. The description of the transmit disable function for the various PMA/PMDs is given in Table 45–12.

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Change the reserved row and footnote a of Table 45–11 and insert new rows at the top of the table as follows (unchanged rows not shown):

Table 45–11—PMD transmit disable register bit definitions

Bit(s)	Name	Description	R/W[a]
~~1.9.15:11~~	~~Reserved~~	~~Value always 0~~	~~RO~~
1.9.15	PMD transmit disable 14	1 = Disable output on transmit lane 14 0 = Enable output on transmit lane 14	R/W
1.9.14	PMD transmit disable 13	1 = Disable output on transmit lane 13 0 = Enable output on transmit lane 13	R/W
1.9.13	PMD transmit disable 12	1 = Disable output on transmit lane 12 0 = Enable output on transmit lane 12	R/W
1.9.12	PMD transmit disable 11	1 = Disable output on transmit lane 11 0 = Enable output on transmit lane 11	R/W
1.9.11	PMD transmit disable 10	1 = Disable output on transmit lane 10 0 = Enable output on transmit lane 10	R/W
…			

[a] ~~RO = Read only,~~ R/W = Read/Write

Insert the following rows at the bottom of Table 45-12 as follows (unchanged rows not shown):

Table 45–12—Transmit disable description location

PMA/PMD	Description location
…	
200GBASE-DR4	121.5.7
200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8	122.5.7
400GBASE-SR16	123.5.7
400GBASE-DR4	124.5.7

Change the title and text of 45.2.1.8.1 as follows:

45.2.1.8.1 PMD transmit disable ~~9~~14 (1.9.~~10~~15)

When bit 1.9.~~10~~15 is set to a one, the PMD shall disable output on lane ~~9~~14 of the transmit path. When bit 1.9.~~10~~15 is set to zero, the PMD shall enable output on lane ~~9~~14 of the transmit path.

The default value for bit 1.9.~~10~~15 is zero.

NOTE—Transmission will not be enabled when this bit is set to zero unless the global PMD transmit disable bit is also zero.

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Change the title and text of 45.2.1.8.2 as follows:

45.2.1.8.2 PMD transmit disable ~~4, 5, 6, 7, 8~~ through 13 (1.9.5~~, 1.9.6, 1.9.7, 1.9.8, 1.9.9~~ through 1.9.14)

~~These b~~Bits 1.9.5 through 1.9.14 are defined similarly to bit 1.9.~~10~~15 for lanes 4 through 13, ~~5, 6, 7, and 8,~~ respectively.

45.2.1.9 PMD receive signal detect register (Register 1.10)

Change the text of 45.2.1.9 as follows:

The assignment of bits in the PMD receive signal detect register is shown in Table 45–13. The ~~10G~~ PMD receive signal detect register is mandatory. PMD types that use only a single lane indicate the status of the receive signal detect using bit 1.10.0 and return a value of zero for bits 1.10.~~10~~15:1. PMD types that use multiple wavelengths or lanes indicate the status of each lane in bits 1.10.~~10~~15:1 and the logical AND of those bits in bit 1.10.0.

Change the reserved row of Table 45–13 and insert new rows at the top of the table as follows (unchanged rows not shown)

Table 45–13—PMD receive signal detect register bit definitions

Bit(s)	Name	Description	R/W[a]
~~1.10.15:11~~	~~Reserved~~	~~Value always 0~~	~~RO~~
1.10.15	PMD receive signal detect 14	1 = Signal detected on receive lane 14 0 = Signal not detected on receive lane 14	RO
1.10.14	PMD receive signal detect 13	1 = Signal detected on receive lane 13 0 = Signal not detected on receive lane 13	RO
1.10.13	PMD receive signal detect 12	1 = Signal detected on receive lane 12 0 = Signal not detected on receive lane 12	RO
1.10.12	PMD receive signal detect 11	1 = Signal detected on receive lane 11 0 = Signal not detected on receive lane 11	RO
1.10.11	PMD receive signal detect 10	1 = Signal detected on receive lane 10 0 = Signal not detected on receive lane 10	RO
…			

[a] RO = Read only

Change the title and text of 45.2.1.9.1 as follows:

45.2.1.9.1 PMD receive signal detect ~~9~~14 (1.10.~~10~~15)

When bit 1.10.~~10~~15 is read as a one, a signal has been detected on lane ~~9~~14 of the PMD receive path. When bit 1.10.~~10~~15 is read as a zero, a signal has not been detected on lane ~~9~~14 of the PMD receive path.

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Change the title and text of 45.2.1.9.2 as follows:

45.2.1.9.2 PMD receive signal detect ~~4, 5, 6, 7, 8~~ through 13 (1.10.5~~, 1.10.6, 1.10.7, 1.10.8, 1.10.9~~ through 1.10.14)

~~These b~~Bits 1.10.5 through 1.10.14 are defined similarly to bit 1.10.~~10~~15 for lanes 4 through 13, ~~5, 6, 7, and 8,~~ respectively.

45.2.1.10 PMA/PMD extended ability register (Register 1.11)

Change the row for 1.11.13 in Table 45-14 (as modified by IEEE Std 802.3bz-2016) as follows (unchanged rows not shown):

Table 45–14—PMA/PMD Extended Ability register bit definitions

Bit(s)	Name	Description	R/W[a]
…			
1.11.13	~~Reserved~~ 200G/400G extended abilities	~~Value always zero~~ 1 = PMA/PMD has 200G/400G extended abilities listed in register 1.23 or register 1.24 0 = PMA/PMD does not have 200G/400G extended abilities	RO
…			

[a] RO = Read only

Insert 45.2.1.10.aab after 45.2.1.10.aaa (as inserted by IEEE Std 802.3bz-2016) and before 45.2.1.10.aa (as inserted by IEEE Std 802.3by-2016) as follows:

45.2.1.10.aab 200G/400G extended abilities (1.11.13)

When read as a one, bit 1.11.13 indicates that the PMA/PMD has 200G extended abilities listed in register 1.23 or 400G extended abilities listed in register 1.24. When read as a zero, bit 1.11.13 indicates that the PMA/PMD does not have 200G or 400G extended abilities.

Insert 45.2.1.14e through 45.2.1.14h after 45.2.1.14d (as inserted by IEEE Std 802.3bv-2017) as follows:

45.2.1.14e 200G PMA/PMD extended ability register (Register 1.23)

The assignment of bits in the 200G PMA/PMD extended ability register is shown in Table 45–17e.

Table 45–17e—200G PMA/PMD extended ability register bit definitions

Bit(s)	Name	Description	R/W[a]
1.23.15	200G PMA remote loopback ability	1 = 200G PMA has the ability to perform a remote loopback function 0 = 200G PMA does not have the ability to perform a remote loopback function	RO
1.23.14:6	Reserved	Value always 0	RO

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Table 45–17e—200G PMA/PMD extended ability register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
1.23.5	200GBASE-LR4 ability	1 = PMA/PMD is able to perform 200GBASE-LR4 0 = PMA/PMD is not able to perform 200GBASE-LR4	RO
1.23.4	200GBASE-FR4 ability	1 = PMA/PMD is able to perform 200GBASE-FR4 0 = PMA/PMD is not able to perform 200GBASE-FR4	RO
1.23.3	200GBASE-DR4 ability	1 = PMA/PMD is able to perform 200GBASE-DR4 0 = PMA/PMD is not able to perform 200GBASE-DR4	RO
1.23.2:0	Reserved	Value always 0	RO

[a] RO = Read only

45.2.1.14e.1 200G PMA remote loopback ability (1.23.15)

When read as a one, bit 1.23.15 indicates that the 200G PMA is able to perform the remote loopback function. When read as a zero, bit 1.23.15 indicates that the 200G PMA is not able to perform the remote loopback function. If a PMA is able to perform the remote loopback function, then it is controlled using the PMA remote loopback bit 1.0.1 (see 45.2.1.1.4).

45.2.1.14e.2 200GBASE-LR4 ability (1.23.5)

When read as a one, bit 1.23.5 indicates that the PMA/PMD is able to operate as a 200GBASE-LR4 PMA/PMD type. When read as a zero, bit 1.23.5 indicates that the PMA/PMD is not able to operate as a 200GBASE-LR4 PMA/PMD type.

45.2.1.14e.3 200GBASE-FR4 ability (1.23.4)

When read as a one, bit 1.23.4 indicates that the PMA/PMD is able to operate as a 200GBASE-FR4 PMA/PMD type. When read as a zero, bit 1.23.4 indicates that the PMA/PMD is not able to operate as a 200GBASE-FR4 PMA/PMD type.

45.2.1.14e.4 200GBASE-DR4 ability (1.23.3)

When read as a one, bit 1.23.3 indicates that the PMA/PMD is able to operate as a 200GBASE-DR4 PMA/PMD type. When read as a zero, bit 1.23.3 indicates that the PMA/PMD is not able to operate as a 200GBASE-DR4 PMA/PMD type.

45.2.1.14f 400G PMA/PMD extended ability register (Register 1.24)

The assignment of bits in the 400G PMA/PMD extended ability register is shown in Table 45–17f.

Table 45–17f—400G PMA/PMD extended ability register bit definitions

Bit(s)	Name	Description	R/W[a]
1.24.15	400G PMA remote loopback ability	1 = 400G PMA has the ability to perform a remote loopback function 0 = 400G PMA does not have the ability to perform a remote loopback function	RO
1.24.14:6	Reserved	Value always 0	RO

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Table 45–17f—400G PMA/PMD extended ability register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
1.24.5	400GBASE-LR8 ability	1 = PMA/PMD is able to perform 400GBASE-LR8 0 = PMA/PMD is not able to perform 400GBASE-LR8	RO
1.24.4	400GBASE-FR8 ability	1 = PMA/PMD is able to perform 400GBASE-FR8 0 = PMA/PMD is not able to perform 400GBASE-FR8	RO
1.24.3	400GBASE-DR4 ability	1 = PMA/PMD is able to perform 400GBASE-DR4 0 = PMA/PMD is not able to perform 400GBASE-DR4	RO
1.24.2	400GBASE-SR16 ability	1 = PMA/PMD is able to perform 400GBASE-SR16 0 = PMA/PMD is not able to perform 400GBASE-SR16	RO
1.24.1:0	Reserved	Value always 0	RO

[a] RO = Read only

45.2.1.14f.1 400G PMA remote loopback ability (1.24.15)

When read as a one, bit 1.24.15 indicates that the 400G PMA is able to perform the remote loopback function. When read as a zero, bit 1.24.15 indicates that the 400G PMA is not able to perform the remote loopback function. If a PMA is able to perform the remote loopback function, then it is controlled using the PMA remote loopback bit 1.0.1 (see 45.2.1.1.4).

45.2.1.14f.2 400GBASE-LR8 ability (1.24.5)

When read as a one, bit 1.24.5 indicates that the PMA/PMD is able to operate as a 400GBASE-LR8 PMA/PMD type. When read as a zero, bit 1.24.5 indicates that the PMA/PMD is not able to operate as a 400GBASE-LR8 PMA/PMD type.

45.2.1.14f.3 400GBASE-FR8 ability (1.24.4)

When read as a one, bit 1.24.4 indicates that the PMA/PMD is able to operate as a 400GBASE-FR8 PMA/PMD type. When read as a zero, bit 1.24.4 indicates that the PMA/PMD is not able to operate as a 400GBASE-FR8 PMA/PMD type.

45.2.1.14f.4 400GBASE-DR4 ability (1.24.3)

When read as a one, bit 1.24.3 indicates that the PMA/PMD is able to operate as a 400GBASE-DR4 PMA/PMD type. When read as a zero, bit 1.24.3 indicates that the PMA/PMD is not able to operate as a 400GBASE-DR4 PMA/PMD type.

45.2.1.14f.5 400GBASE-SR16 ability (1.24.2)

When read as a one, bit 1.24.2 indicates that the PMA/PMD is able to operate as a 400GBASE-SR16 PMA/PMD type. When read as a zero, bit 1.24.5 indicates that the PMA/PMD is not able to operate as a 400GBASE-SR16 PMA/PMD type.

45.2.1.14g PMD transmit disable extension register (Register 1.27)

The assignment of bits in the PMD transmit disable extension register is shown in Table 45–17g. The transmit disable functionality is optional and a PMD's ability to perform the transmit disable functionality is advertised in the PMD transmit disable ability bit 1.8.8. A PMD that does not implement the transmit disable functionality shall ignore writes to the PMD transmit disable extension register and may return a value of

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zero for all bits. A PMD device that operates using a single lane and has implemented the transmit disable function shall use bit 1.9.0 to control the function. Such devices shall ignore writes to bit 1.27.0 and return a value of zero for that bit when read.

NOTE—Disabling the transmitter on one or more lanes stops the entire link from carrying data.

Table 45–17g—PMD transmit disable extension register bit definitions

Bit(s)	Name	Description	R/W[a]
1.27.15:1	Reserved	Value always 0	RO
1.27.0	PMD transmit disable 15	1 = Disable output on transmit lane 15 0 = Enable output on transmit lane 15	R/W

[a] RO = Read only, R/W = Read/Write

45.2.1.14g.1 PMD transmit disable 15 (1.27.0)

When bit 1.27.0 is set to a one, the PMD shall disable output on lane 15 of the transmit path. When bit 1.27.0 is set to a zero, the PMD shall enable output on lane 15 of the transmit path.

The default value for bit 1.27.0 is zero.

NOTE—Transmission will not be enabled when this bit is set to a zero unless the global PMD transmit disable bit is also zero.

45.2.1.14h PMD receive signal detect extension register (Register 1.28)

The assignment of bits in the PMD receive signal detect extension register is shown in Table 45–17h. PMD types that use only a single lane indicate the status of the receive signal detect using bit 1.10.0 and return a value of zero for bit 1.28.0. PMD types that use multiple wavelengths or lanes indicate the status of each lane in bits 1.10.15:1 and 1.28.0 (if needed) and the logical AND of those bits in bit 1.10.0.

Table 45–17h—PMD receive signal detect extension register bit definitions

Bit(s)	Name	Description	R/W[a]
1.28.15:1	Reserved	Value always 0	RO
1.28.0	PMD receive signal detect 15	1 = Signal detected on receive lane 15 0 = Signal not detected on receive lane 15	RO

[a] RO = Read only

45.2.1.14h.1 PMD receive signal detect 15 (1.28.0)

When bit 1.28.0 is read as a one, a signal has been detected on lane 15 of the PMD receive path. When bit 1.28.0 is read as a zero, a signal has not been detected on lane 15 of the PMD receive path.

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Insert 45.2.1.116a through 45.2.1.116g after 45.2.1.116 as follows:

45.2.1.116a 200GAUI-8 and 400GAUI-16 chip-to-module recommended CTLE, lane 0 register (Register 1.400)

The assignment of bits in the 200GAUI-8 and 400GAUI-16 chip-to-module recommended CTLE, lane 0 register is shown in Table 45–90aa.

Table 45–90aa—200GAUI-8 and 400GAUI-16 chip-to-module recommended CTLE, lane 0 register bit definitions

Bit(s)	Name	Description	R/W[a]
1.400.15:5	Reserved	Value always 0	RO
1.400.4:1	Recommended CTLE peaking	4 3 2 1 1 1 x x = Reserved 1 0 1 x = Reserved 1 0 0 1 = 9 dB 1 0 0 0 = 8 dB 0 1 1 1 = 7 dB 0 1 1 0 = 6 dB 0 1 0 1 = 5 dB 0 1 0 0 = 4 dB 0 0 1 1 = 3 dB 0 0 1 0 = 2 dB 0 0 0 1 = 1 dB 0 0 0 0 = Reserved	R/W
1.400.0	Reserved	Value always 0	RO

[a] R/W = Read/Write, RO = Read only

45.2.1.116a.1 Recommended CTLE peaking (1.400.4:1)

The value of these bits sets the CTLE peaking value recommended for lane 0 by a host that implements the optional 200GAUI-8 or 400GAUI-16 chip-to-module interface defined in Annex 120C (see 120C.3.1). The module may optionally use this information to adjust its CTLE setting.

45.2.1.116b 200GAUI-8 and 400GAUI-16 chip-to-module recommended CTLE, lane 1 through lane 7 registers (Registers 1.401 through 1.407)

The 200GAUI-8 and 400GAUI-16 chip-to-module recommended CTLE, lane 1 through lane 7 registers are defined similarly to register 1.400 (which is used for lane 0, see 45.2.1.116a) but for lanes 1 through 7, respectively.

45.2.1.116c 400GAUI-16 chip-to-module recommended CTLE, lane 8 through lane 15 registers (Registers 1.408 through 1.415)

The 400GAUI-16 chip-to-module recommended CTLE, lane 8 through lane 15 registers are defined similarly to register 1.400 (which is used for lane 0, see 45.2.1.116a) but for lanes 8 through 15, respectively.

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45.2.1.116d 200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, receive direction, lane 0 register (Register 1.500)

The assignment of bits in the 200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, receive direction, lane 0 register is shown in Table 45–90ab. The transmitter, receive direction, is the transmitter that sends data towards the MAC.

Table 45–90ab—200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, receive direction, lane 0 register bit definitions

Bit(s)	Name	Description	R/W[a]
1.500.15	Request flag	1 = Change in equalization is requested 0 = No change in equalization is requested	RO
1.500.14:12	Post-cursor request	14 13 12 1 1 1 Reserved 1 1 0 Reserved 1 0 1 *Requested_eq_c1* = 5 (c(1) ratio –0.25) 1 0 0 *Requested_eq_c1* = 4 (c(1) ratio –0.2) 0 1 1 *Requested_eq_c1* = 3 (c(1) ratio –0.15) 0 1 0 *Requested_eq_c1* = 2 (c(1) ratio –0.1) 0 0 1 *Requested_eq_c1* = 1 (c(1) ratio –0.05) 0 0 0 *Requested_eq_c1* = 0 (c(1) ratio 0)	RO
1.500.11:10	Pre-cursor request	11 10 1 1 *Requested_eq_cm1* = 3 (c(–1) ratio –0.15) 1 0 *Requested_eq_cm1* = 2 (c(–1) ratio –0.1) 0 1 *Requested_eq_cm1* = 1 (c(–1) ratio –0.05) 0 0 *Requested_eq_cm1* = 0 (c(–1) ratio 0)	RO
1.500.9:7	Post-cursor remote setting	9 8 7 1 1 1 Reserved 1 1 0 Reserved 1 0 1 *Remote_eq_c1* = 5 (c(1) ratio –0.25) 1 0 0 *Remote_eq_c1* = 4 (c(1) ratio –0.2) 0 1 1 *Remote_eq_c1* = 3 (c(1) ratio –0.15) 0 1 0 *Remote_eq_c1* = 2 (c(1) ratio –0.1) 0 0 1 *Remote_eq_c1* = 1 (c(1) ratio –0.05) 0 0 0 *Remote_eq_c1* = 0 (c(1) ratio 0)	R/W
1.500.6:5	Pre-cursor remote setting	6 5 1 1 *Remote_eq_cm1* = 3 (c(–1) ratio –0.15) 1 0 *Remote_eq_cm1* = 2 (c(–1) ratio –0.1) 0 1 *Remote_eq_cm1* = 1 (c(–1) ratio –0.05) 0 0 *Remote_eq_cm1* = 0 (c(–1) ratio 0)	R/W
1.500.4:2	Post-cursor local setting	4 3 2 1 1 1 Reserved 1 1 0 Reserved 1 0 1 *Local_eq_c1* = 5 (c(1) ratio –0.25) 1 0 0 *Local_eq_c1* = 4 (c(1) ratio –0.2) 0 1 1 *Local_eq_c1* = 3 (c(1) ratio –0.15) 0 1 0 *Local_eq_c1* = 2 (c(1) ratio –0.1) 0 0 1 *Local_eq_c1* = 1 (c(1) ratio –0.05) 0 0 0 *Local_eq_c1* = 0 (c(1) ratio 0)	R/W
1.500.1:0	Pre-cursor local setting	1 0 1 1 *Local_eq_cm1* = 3 (c(–1) ratio –0.15) 1 0 *Local_eq_cm1* = 2 (c(–1) ratio –0.1) 0 1 *Local_eq_cm1* = 1 (c(–1) ratio –0.05) 0 0 *Local_eq_cm1* = 0 (c(–1) ratio 0)	R/W

[a] R/W = Read/Write, RO = Read only

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45.2.1.116d.1 Request flag (1.500.15)

The value of this bit indicates the value of the variable *Request_flag* in the lane 0 200GAUI-n or 400GAUI-n receiver in the receive direction (see 120B.3.2 and 120D.3.2.3). This indicates whether the 200GAUI-n or 400GAUI-n chip-to-chip device is issuing a request to change the remote transmitter equalization in the 200GAUI-n or 400GAUI-n chip-to-chip lane 0 transmitter in the receive direction. If a lane 0 200GAUI-n or 400GAUI-n receiver in the receive direction is not present in the package, then the value returned for this bit should be zero.

45.2.1.116d.2 Post-cursor request (1.500.14:12)

The value of these bits indicates the value of the variable *Requested_eq_c1* in the lane 0 200GAUI-n or 400GAUI-n receiver in the receive direction (see 120B.3.2 and 120D.3.2.3). When *Request_flag* is equal to 1, this value indicates the ratio of the post-cursor coefficient c(1), which is requested for the transmitter equalization in the 200GAUI-n or 400GAUI-n chip-to-chip lane 0 transmitter in the receive direction.

45.2.1.116d.3 Pre-cursor request (1.500.11:10)

The value of these bits indicates the value of the variable *Requested_eq_cm1* in the lane 0 200GAUI-n or 400GAUI-n receiver in the receive direction (see 120B.3.2 and 120D.3.2.3). When *Request_flag* is equal to 1, this value indicates the ratio of the pre-cursor coefficient c(–1), which is requested for the transmitter equalization in the 200GAUI-n or 400GAUI-n chip-to-chip lane 0 transmitter in the receive direction.

45.2.1.116d.4 Post-cursor remote setting (1.500.9:7)

The value of these bits sets the variable *Remote_eq_c1* for the lane 0 200GAUI-n or 400GAUI-n receiver in the receive direction (see 120B.3.2 and 120D.3.2.3). This is used by a 200GAUI-n or 400GAUI-n receiver that implements the optional transmitter equalization feedback as an indication of the ratio of the post-cursor coefficient c(1) being used in lane 0 of the 200GAUI-n or 400GAUI-n transmitter in the receive direction (see 120B.3.1 and 120D.3.1.5). It may be used to generate values for the request flag and the request bits. If a lane 0 200GAUI-n or 400GAUI-n receiver in the receive direction is not present in the package, then these bits have no effect.

45.2.1.116d.5 Pre-cursor remote setting (1.500.6:5)

The value of these bits sets the variable *Remote_eq_cm1* for the lane 0 200GAUI-n or 400GAUI-n receiver in the receive direction (see 120B.3.2 and 120D.3.2.3). This is used by a 200GAUI-n or 400GAUI-n receiver that implements the optional transmitter equalization feedback as an indication of the ratio of the pre-cursor coefficient c(–1) being used in lane 0 of the 200GAUI-n or 400GAUI-n transmitter in the receive direction (see 120B.3.1 and 120D.3.1.5). It may be used to generate values for the request flag and the request bits. If a lane 0 200GAUI-n or 400GAUI-n receiver in the receive direction is not present in the package, then these bits have no effect.

45.2.1.116d.6 Post-cursor local setting (1.500.4:2)

The value of these bits sets the variable *Local_eq_c1* for the lane 0 200GAUI-n or 400GAUI-n transmitter in the receive direction (see 120B.3.1 and 120D.3.1.5), which controls the weight of the transmitter equalization post-cursor coefficient c(1). If a lane 0 200GAUI-n or 400GAUI-n transmitter in the receive direction is not present in the package, then these bits have no effect.

45.2.1.116d.7 Pre-cursor local setting (1.500.1:0)

The value of these bits sets the variable *Local_eq_cm1* for the lane 0 200GAUI-n or 400GAUI-n transmitter in the receive direction (see 120B.3.1 and 120D.3.1.5), which controls the weight of the transmitter

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equalization pre-cursor coefficient c(–1). If a lane 0 200GAUI-n or 400GAUI-n transmitter in the receive direction is not present in the package, then these bits have no effect.

45.2.1.116e 200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, receive direction, lane 1 through lane 15 registers (Registers 1.501 through 1.515)

The 200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, receive direction, lane 1 through lane 15 registers are defined similarly to register 1.500 (which is used for lane 0, see 45.2.1.116d) but for lanes 1 through 15, respectively. The transmitter, receive direction, is the transmitter that sends data towards the MAC.

45.2.1.116f 200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, transmit direction, lane 0 register (Register 1.516)

The assignment of bits in the 200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, transmit direction, lane 0 register is shown in Table 45–90ac. The transmitter, transmit direction, is the transmitter that sends data towards the PMD.

Table 45–90ac—200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, transmit direction, lane 0 register bit definitions

Bit(s)	Name	Description	R/W[a]
1.516.15	Request flag	1 = Change in equalization is requested 0 = No change in equalization is requested	RO
1.516.14:12	Post-cursor request	14 13 12 1 1 1 Reserved 1 1 0 Reserved 1 0 1 *Requested_eq_c1* = 5 (c(1) ratio –0.25) 1 0 0 *Requested_eq_c1* = 4 (c(1) ratio –0.2) 0 1 1 *Requested_eq_c1* = 3 (c(1) ratio –0.15) 0 1 0 *Requested_eq_c1* = 2 (c(1) ratio –0.1) 0 0 1 *Requested_eq_c1* = 1 (c(1) ratio –0.05) 0 0 0 *Requested_eq_c1* = 0 (c(1) ratio 0)	RO
1.516.11:10	Pre-cursor request	11 10 1 1 *Requested_eq_cm1* = 3 (c(–1) ratio –0.15) 1 0 *Requested_eq_cm1* = 2 (c(–1) ratio –0.1) 0 1 *Requested_eq_cm1* = 1 (c(–1) ratio –0.05) 0 0 *Requested_eq_cm1* = 0 (c(–1) ratio 0)	RO
1.516.9:7	Post-cursor remote setting	9 8 7 1 1 1 Reserved 1 1 0 Reserved 1 0 1 *Remote_eq_c1* = 5 (c(1) ratio –0.25) 1 0 0 *Remote_eq_c1* = 4 (c(1) ratio –0.2) 0 1 1 *Remote_eq_c1* = 3 (c(1) ratio –0.15) 0 1 0 *Remote_eq_c1* = 2 (c(1) ratio –0.1) 0 0 1 *Remote_eq_c1* = 1 (c(1) ratio –0.05) 0 0 0 *Remote_eq_c1* = 0 (c(1) ratio 0)	R/W
1.516.6:5	Pre-cursor remote setting	6 5 1 1 *Remote_eq_cm1* = 3 (c(–1) ratio –0.15) 1 0 *Remote_eq_cm1* = 2 (c(–1) ratio –0.1) 0 1 *Remote_eq_cm1* = 1 (c(–1) ratio –0.05) 0 0 *Remote_eq_cm1* = 0 (c(–1) ratio 0)	R/W

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Table 45–90ac—200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, transmit direction, lane 0 register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
1.516.4:2	Post-cursor local setting	4 3 2 1 1 1 Reserved 1 1 0 Reserved 1 0 1 *Local_eq_c1* = 5 (c(1) ratio –0.25) 1 0 0 *Local_eq_c1* = 4 (c(1) ratio –0.2) 0 1 1 *Local_eq_c1* = 3 (c(1) ratio –0.15) 0 1 0 *Local_eq_c1* = 2 (c(1) ratio –0.1) 0 0 1 *Local_eq_c1* = 1 (c(1) ratio–0.05) 0 0 0 *Local_eq_c1* = 0 (c(1) ratio 0)	R/W
1.516.1:0	Pre-cursor local setting	1 0 1 1 *Local_eq_cm1* = 3 (c(–1) ratio –0.15) 1 0 *Local_eq_cm1* = 2 (c(–1) ratio –0.1) 0 1 *Local_eq_cm1* = 1 (c(–1) ratio –0.05) 0 0 *Local_eq_cm1* = 0 (c(–1) ratio 0)	R/W

[a] R/W = Read/Write, RO = Read only

45.2.1.116f.1 Request flag (1.516.15)

The value of this bit indicates the value of the variable *Request_flag* in the lane 0 200GAUI-n or 400GAUI-n receiver in the transmit direction (see 120B.3.2 and 120D.3.2.3). This indicates whether the 200GAUI-n or 400GAUI-n chip-to-chip device is issuing a request to change the remote transmitter equalization in the 200GAUI-n or 400GAUI-n chip-to-chip lane 0 transmitter in the transmit direction. If a lane 0 200GAUI-n or 400GAUI-n receiver in the transmit direction is not present in the package, then the value returned for this bit should be zero.

45.2.1.116f.2 Post-cursor request (1.516.14:12)

The value of these bits indicates the value of the variable *Requested_eq_c1* in the lane 0 200GAUI-n or 400GAUI-n receiver in the transmit direction (see 120B.3.2 and 120D.3.2.3). When *Request_flag* is equal to 1, this value indicates the ratio of the post-cursor coefficient c(1), which is requested for the transmitter equalization in the 200GAUI-n or 400GAUI-n chip-to-chip lane 0 transmitter in the transmit direction.

45.2.1.116f.3 Pre-cursor request (1.516.11:10)

The value of these bits indicates the value of the variable *Requested_eq_cm1* in the lane 0 200GAUI-n or 400GAUI-n receiver in the transmit direction (see 120B.3.2 and 120D.3.2.3). When *Request_flag* is equal to 1, this value indicates the ratio of the pre-cursor coefficient c(–1), which is requested for the transmitter equalization in the 200GAUI-n or 400GAUI-n chip-to-chip lane 0 transmitter in the transmit direction.

45.2.1.116f.4 Post-cursor remote setting (1.516.9:7)

The value of these bits sets the variable *Remote_eq_c1* for the lane 0 200GAUI-n or 400GAUI-n receiver in the transmit direction (see 120B.3.2 and 120D.3.2.3). This is used by a 200GAUI-n or 400GAUI-n receiver that implements the optional transmitter equalization feedback as an indication of the ratio of the post-cursor coefficient c(1) being used in lane 0 of the 200GAUI-n or 400GAUI-n transmitter in the transmit direction (see 120B.3.1 and 120D.3.1.5). It may be used to generate values for the request flag and the request bits. If a lane 0 200GAUI-n or 400GAUI-n receiver in the transmit direction is not present in the package, then these bits have no effect.

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45.2.1.116f.5 Pre-cursor remote setting (1.516.6:5)

The value of these bits sets the variable *Remote_eq_cm1* for the lane 0 200GAUI-n or 400GAUI-n receiver in the transmit direction (see 120B.3.2 and 120D.3.2.3). This is used by a 200GAUI-n or 400GAUI-n receiver that implements the optional transmitter equalization feedback as an indication of the ratio of the pre-cursor coefficient c(–1) being used in lane 0 of the 200GAUI-n or 400GAUI-n transmitter in the transmit direction (see 120B.3.1 and 120D.3.1.5). It may be used to generate values for the request flag and the request bits. If a lane 0 200GAUI-n or 400GAUI-n receiver in the transmit direction is not present in the package, then these bits have no effect.

45.2.1.116f.6 Post-cursor local setting (1.516.4:2)

The value of these bits sets the variable *Local_eq_c1* for the lane 0 200GAUI-n or 400GAUI-n transmitter in the transmit direction (see 120B.3.1 and 120D.3.1.5), which controls the weight of the transmitter equalization post-cursor coefficient c(1). If a lane 0 200GAUI-n or 400GAUI-n transmitter in the transmit direction is not present in the package, then these bits have no effect.

45.2.1.116f.7 Pre-cursor local setting (1.516.1:0)

The value of these bits sets the variable *Local_eq_cm1* for the lane 0 200GAUI-n or 400GAUI-n transmitter in the transmit direction (see 120B.3.1 and 120D.3.1.5), which controls the weight of the transmitter equalization pre-cursor coefficient c(–1). If a lane 0 200GAUI-n or 400GAUI-n transmitter in the transmit direction is not present in the package, then these bits have no effect.

45.2.1.116g 200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, transmit direction, lane 1 through lane 15 registers (Registers 1.517 through 1.531)

The 200GAUI-n and 400GAUI-n chip-to-chip transmitter equalization, transmit direction, lane 1 through lane 15 registers are defined similarly to register 1.516 (which is used for lane 0, see 45.2.1.116f) but for lanes 1 through 15, respectively. The transmitter, transmit direction, is the transmitter that sends data towards the PMD.

45.2.1.123 Test-pattern ability (Register 1.1500)

Change 45.2.1.123 as follows:

The test-pattern ability register is used for PHY types that implement SSPRQ, square wave, ~~testing~~ and PRBS testing in the PMA. These functions are described in 83.5.10 and 120.5.11. The assignment of bits in the test-pattern ability register is shown in Table 45–92.

Table 45–92—Test-pattern ability register bit definitions

Bit(s)	Name	Description	R/W[a]
~~1.1500.15:13~~ 1.1500.15:14	Reserved	Value always 0	RO
1.1500.13	SSPRQ Tx generator ability	1 = SSPRQ transmit direction pattern generator supported 0 = SSPRQ transmit direction pattern generator not supported	RO
1.1500.12	Square wave test ability	1 = Square wave testing supported 0 = Square wave testing not supported	RO
~~1.1500.11:6~~	~~Reserved~~	~~Value always 0~~	~~RO~~

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Table 45–92—Test-pattern ability register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
1.1500.11	PRBS13Q Tx generator ability	1 = PRBS13Q transmit direction pattern generator supported 0 = PRBS13Q transmit direction pattern generator not supported	RO
1.1500.10	PRBS13Q Rx generator ability	1 = PRBS13Q receive direction pattern generator supported 0 = PRBS13Q receive direction pattern generator not supported	RO
1.1500.9	PRBS31Q Tx generator ability	1 = PRBS31Q transmit direction pattern generator supported 0 = PRBS31Q transmit direction pattern generator not supported	RO
1.1500.8	PRBS31Q Tx checker ability	1 = PRBS31Q transmit direction pattern checker supported 0 = PRBS31Q transmit direction pattern checker not supported	RO
1.1500.7	PRBS31Q Rx generator ability	1 = PRBS31Q receive direction pattern generator supported 0 = PRBS31Q receive direction pattern generator not supported	RO
1.1500.6	PRBS31Q Rx checker ability	1 = PRBS31Q receive direction pattern checker supported 0 = PRBS31Q receive direction pattern checker not supported	RO
1.1500.5	PRBS9 Tx generator ability	1 = PRBS9 transmit direction pattern generator supported 0 = PRBS9 transmit direction pattern generator not supported	RO
1.1500.4	PRBS9 Rx generator ability	1 = PRBS9 receive direction pattern generator supported 0 = PRBS9 receive direction pattern generator not supported	RO
1.1500.3	PRBS31 Tx generator ability	1 = PRBS31 ~~T~~transmit direction pattern generator supported 0 = PRBS31 ~~Tx~~transmit direction pattern generator not supported	RO
1.1500.2	PRBS31 Tx checker ability	1 = PRBS31 ~~T~~transmit direction pattern checker supported 0 = PRBS31 ~~Tx~~transmit direction pattern checker not supported	RO
1.1500.1	PRBS31 Rx generator ability	1 = PRBS31 ~~R~~receive direction pattern generator supported 0 = PRBS31 ~~Rx~~receive direction pattern generator not supported	RO
1.1500.0	PRBS31 Rx checker ability	1 = PRBS31 ~~R~~receive direction pattern checker supported 0 = PRBS31 ~~Rx~~receive direction pattern checker not supported	RO

[a] RO = Read only

If SSPRQ pattern testing is supported and register 1.1500 is implemented then bit 1.1500.13 shall indicate the generation ability in the transmit direction. The SSPRQ pattern test is controlled by register 1.1501 (see 45.2.1.124).

If square wave testing is supported and ~~this~~ register 1.1500 is implemented then bit 1.1500.12 shall be set to 1. The square wave test is controlled by register 1.1510 (see 45.2.1.125).

If PRBS13Q pattern testing is supported and register 1.1500 is implemented then bit 1.1500.11 shall indicate the generation ability in the transmit direction and bit 1.1500.10 shall indicate the generation ability in the receive direction. The PRBS13Q pattern test is controlled by register 1.1512 (see 45.2.1.125a).

If PRBS31Q pattern testing is supported and register 1.1500 is implemented then bit 1.1500.9 shall indicate the generation ability in the transmit direction and bit 1.1500.7 shall indicate the generation ability in the receive direction. Bit 1.1500.8 shall indicate the checker ability in the transmit direction and bit 1.1500.6 shall indicate the checker ability in the receive direction. The PRBS31Q pattern test is controlled by register 1.1501 (see 45.2.1.124).

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If PRBS9 pattern testing is supported and ~~this~~ register 1.1500 is implemented then bit 1.1500.5 shall indicate the generation ability in the transmit direction and bit 1.1500.4 shall indicate the generation ability in the receive direction. The PRBS pattern test is controlled by register 1.1501 (see 45.2.1.124).

If PRBS31 pattern testing is supported and ~~this~~ register 1.1500 is implemented then bit 1.1500.3 shall indicate the generation ability in the transmit direction and bit 1.1500.1 shall indicate the generation ability in the receive direction. Bit 1.1500.2 shall indicate the checker ability in the transmit direction and bit 1.1500.0 shall indicate the checker ability in the receive direction. The PRBS pattern test is controlled by register 1.1501 (see 45.2.1.124).

45.2.1.124 PRBS pattern testing control (Register 1.1501)

Change the first paragraph of 45.2.1.124 and Table 45–93 as follows:

The PRBS pattern testing control register is used for PHY types that implement PRBS pattern testing in the PMA. This function is described in 83.5.10 and 120.5.11. The assignment of bits in the PRBS pattern testing control register is shown in Table 45–93.

Table 45–93—PRBS pattern testing control register bit definitions

Bit(s)	Name	Description	R/W[a]
1.1501.15:~~12~~	Reserved	Value always 0	RO
1.1501.14	SSPRQ pattern enable	1 = Enable SSPRQ test-pattern 0 = Disable SSPRQ test-pattern	R/W
1.1501.13	PRBS31Q pattern enable	1 = Enable PRBS31Q test-pattern 0 = Disable PRBS31Q test-pattern	R/W
1.1501.12	Reserved	Value always 0	RO
1.1501.11	Transmitter linearity test pattern enable	1 = Enable transmitter linearity test-pattern 0 = Disable transmitter linearity test-pattern	R/W
1.1501.10	QPRBS13 pattern enable	1 = Enable QPRBS13 test-pattern 0 = Disable QPRBS13 test-pattern	R/W
1.1501.9	JP03B pattern enable	1 = Enable JP03B test-pattern 0 = Disable JP03B test-pattern	R/W
1.1501.8	JP03A pattern enable	1 = Enable JP03A test-pattern 0 = Disable JP03A test-pattern	R/W
1.1501.7	PRBS31 pattern enable	1 = Enable PRBS31 test-pattern 0 = Disable PRBS31 test-pattern	R/W
1.1501.6	PRBS9 pattern enable	1 = Enable PRBS9 test-pattern 0 = Disable PRBS9 test-pattern	R/W
1.1501.5:4	Reserved	Value always 0	RO
1.1501.3	Tx generator enable	1 = Enable transmit direction test-pattern generator 0 = Disable transmit direction test-pattern generator	R/W
1.1501.2	Tx checker enable	1 = Enable transmit direction test-pattern checker 0 = Disable transmit direction test-pattern checker	R/W
1.1501.1	Rx generator enable	1 = Enable receive direction test-pattern generator 0 = Disable receive direction test-pattern generator	R/W
1.1501.0	Rx checker enable	1 = Enable receive direction test-pattern checker 0 = Disable receive direction test-pattern checker	R/W

[a] R/W = Read/Write, RO = Read only

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Change the last two paragraphs of 45.2.1.124 as follows:

Register 1.1501, bit 3 enables pattern ~~PRBS~~ generation in the transmit direction. Register 1.1501, bit 2 enables pattern ~~PRBS~~ checking in the transmit direction. Register 1.1501, bit 1 enables pattern ~~PRBS~~ generation in the receive direction. Register 1.1501, bit 0 enables pattern ~~PRBS~~ checking in the receive direction. If none ~~neither~~ of the bits 3:0 ~~7 and 6~~ are asserted then bits 7 and 6 ~~3:0~~ have no effect.

Register 1.1501 bit 8 enables testing with the JP03A pattern defined in 94.2.9.1 for 100GBASE-KP4 PMA/PMD. Register 1.1501 bit 9 enables testing with the JP03B pattern defined in 94.2.9.2 for 100GBASE-KP4 PMA/PMD. Register ~~field~~ 1.1501 bit 10 enables testing with the QPRBS13 pattern defined in 94.2.9.3 for 100GBASE-KP4 PMA/PMD. Register ~~field~~ 1.1501 bit 11 enables the transmitter linearity test pattern defined in 94.2.9.4 for 100GBASE-KP4 PMA/PMD. Register 1.1501 bit 13 enables the PRBS31Q test pattern defined in 120.5.11.2.2. Register 1.1501 bit 14 enables the SSPRQ test pattern defined in 120.5.11.2.3. The assertion of bits 1.1501.8, 1.1501.9, 1.1501.10, 1.1501.11, 1.1501.13, and 1.1501.14 are mutually exclusive. If more than one bit is asserted, the behavior is undefined. The assertion of 1.1501.8, 1.1501.9, 1.501.10, and 1.1501.11 operates in conjunction with register 1.1501 bit 3 for 100GBASE-KP4 PMA/PMD. For other PMA/PMD types or if bit 1.1501.3 is not asserted, then 1.1501.8, 1.1501.9, 1.501.10, and 1.1501.11 have no effect. For the 200GBASE-R and 400GBASE-R PMAs, the assertion of register 1.1501 bit 13 operates in conjunction with register 1.1501 bits 3 and 1. If neither bit is asserted, then register 1.1501 bit 13 has no effect. Also for the 200GBASE-R and 400GBASE-R PMAs, the assertion of register 1.1501 bit 14 operates in conjunction with register 1.1501 bit 3. If bit 1.1501.3 is not asserted, then register 1.1501 bit 14 has no effect. For other PMA/PMD types register 1.1501 bits 13 and 14 have no effect.

45.2.1.125 Square wave testing control (Register 1.1510)

Change 45.2.1.125 as follows:

The square wave testing control register is used for PHY types that implement transmit square wave testing in the PMA. This function is described in 83.5.10 and 120.5.11. The assignment of bits in the square wave testing control register is shown in Table 45–94.

Table 45–94—Square wave testing control register bit definitions

Bit(s)	Name	Description	R/W[a]
~~1.1510.15:10~~	~~Reserved~~	~~Value always 0~~	~~RO~~
1.1510.15	Lane 15 SW enable	1 = Enable square wave on lane 15 0 = Disable square wave on lane 15	R/W
1.1510.14	Lane 14 SW enable	1 = Enable square wave on lane 14 0 = Disable square wave on lane 14	R/W
1.1510.13	Lane 13 SW enable	1 = Enable square wave on lane 13 0 = Disable square wave on lane 13	R/W
1.1510.12	Lane 12 SW enable	1 = Enable square wave on lane 12 0 = Disable square wave on lane 12	R/W
1.1510.11	Lane 11 SW enable	1 = Enable square wave on lane 11 0 = Disable square wave on lane 11	R/W
1.1510.10	Lane 10 SW enable	1 = Enable square wave on lane 10 0 = Disable square wave on lane 10	R/W

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Table 45–94—Square wave testing control register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
1.1510.9	Lane 9 SW enable	1 = Enable square wave on lane 9 0 = Disable square wave on lane 9	R/W
1.1510.8	Lane 8 SW enable	1 = Enable square wave on lane 8 0 = Disable square wave on lane 8	R/W
1.1510.7	Lane 7 SW enable	1 = Enable square wave on lane 7 0 = Disable square wave on lane 7	R/W
1.1510.6	Lane 6 SW enable	1 = Enable square wave on lane 6 0 = Disable square wave on lane 6	R/W
1.1510.5	Lane 5 SW enable	1 = Enable square wave on lane 5 0 = Disable square wave on lane 5	R/W
1.1510.4	Lane 4 SW enable	1 = Enable square wave on lane 4 0 = Disable square wave on lane 4	R/W
1.1510.3	Lane 3 SW enable	1 = Enable square wave on lane 3 0 = Disable square wave on lane 3	R/W
1.1510.2	Lane 2 SW enable	1 = Enable square wave on lane 2 0 = Disable square wave on lane 2	R/W
1.1510.1	Lane 1 SW enable	1 = Enable square wave on lane 1 0 = Disable square wave on lane 1	R/W
1.1510.0	Lane 0 SW enable	1 = Enable square wave on lane 0 0 = Disable square wave on lane 0	R/W

[a] R/W = Read/Write~~, RO = Read only~~

Register 1.1510, bits 0 through 15~~9~~ enable square wave output on PMA lanes 0 through 15~~9~~ respectively. Lanes for which a square wave pattern is not enabled act as determined by other registers.

Insert 45.2.1.125a after 45.2.1.125 as follows:

45.2.1.125a PRBS13Q testing control (Register 1.1512)

The PRBS13Q testing control register is used for PHY types that implement PRBS13Q testing in the PMA. This function is described in 120.5.11.2.1. The assignment of bits in the PRBS13Q testing control register is shown in Table 45–94a.

Table 45–94a—PRBS13Q testing control register bit definitions

Bit(s)	Name	Description	R/W[a]
1.1512.15:8	Reserved	Value always 0	RO
1.1512.7	Lane 7 PRBS13Q enable	1 = Enable PRBS13Q on lane 7 0 = Disable PRBS13Q on lane 7	R/W
1.1512.6	Lane 6 PRBS13Q enable	1 = Enable PRBS13Q on lane 6 0 = Disable PRBS13Q on lane 6	R/W

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Table 45–94a—PRBS13Q testing control register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
1.1512.5	Lane 5 PRBS13Q enable	1 = Enable PRBS13Q on lane 5 0 = Disable PRBS13Q on lane 5	R/W
1.1512.4	Lane 4 PRBS13Q enable	1 = Enable PRBS13Q on lane 4 0 = Disable PRBS13Q on lane 4	R/W
1.1512.3	Lane 3 PRBS13Q enable	1 = Enable PRBS13Q on lane 3 0 = Disable PRBS13Q on lane 3	R/W
1.1512.2	Lane 2 PRBS13Q enable	1 = Enable PRBS13Q on lane 2 0 = Disable PRBS13Q on lane 2	R/W
1.1512.1	Lane 1 PRBS13Q enable	1 = Enable PRBS13Q on lane 1 0 = Disable PRBS13Q on lane 1	R/W
1.1512.0	Lane 0 PRBS13Q enable	1 = Enable PRBS13Q on lane 0 0 = Disable PRBS13Q on lane 0	R/W

[a] R/W = Read/Write, RO = Read only

Register 1.1512, bits 0 through 7 enable PRBS13Q output on PMA lanes 0 through 7 respectively. The assertion of register 1.1512 bits 0 through 7 operates in conjunction with register 1.1501 bits 3 and 1. If neither bit is asserted, then register 1.1512 bits 0 through 7 have no effect. Lanes for which a PRBS13Q pattern is not enabled act as determined by other registers.

Change the title and text of 45.2.1.126 as follows:

45.2.1.126 PRBS Tx pattern testing error counter (Register 1.1600 through 1.1615~~, 1.1601, 1.1602, 1.1603, 1.1604, 1.1605, 1.1606, 1.1607, 1.1608, 1.1609~~)

The PRBS Tx pattern testing error counter registers are used for PHY types that implement PRBS Tx pattern testing in the PMA. This function is described in 83.5.10 and 120.5.11. The assignment of bits in the PRBS Tx pattern testing error counter registers are shown in Table 45–95. Register 1.1600 contains the PRBS Tx pattern testing error counter for lane 0, register 1.1601 contains the PRBS Tx pattern testing error counter for lane 1, and registers 1.1602 through 1.1615~~09~~ contain the PRBS Tx pattern testing error counters for lanes 2 through 15~~9~~ respectively. Counters corresponding to lanes that are not implemented in a PMA shall read all zeros.

Table 45–95—PRBS Tx pattern testing error counter

Bit(s)	Name	Description	R/W[a]
1.1600[b].15:0	Error counter		RO, NR

[a] RO = Read only, NR = Non Roll-over.
[b] All instances of address 1.1600 also apply to addresses 1.1601 through 1.1615~~09~~.

The PRBS Tx pattern testing error counter is a 16-bit counter as defined in 83.5.10 and 120.5.11. These bits shall be reset to all zeros when the register is read by the management function or upon execution of the PMA reset. These bits shall be held at all ones in the case of overflow.

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Change the title and content of 45.2.1.127 as follows:

45.2.1.127 PRBS Rx pattern testing error counter (Register 1.1700 through 1.1715~~, 1.1701, 1.1702, 1.1703, 1.1704, 1.1705, 1.1706, 1.1707, 1.1708, 1.1709~~)

The PRBS Rx pattern testing error counter registers are used for PHY types that implement PRBS Rx pattern testing in the PMA. This function is described in 83.5.10 and 120.5.11. The assignment of bits in the PRBS Rx pattern testing error counter registers is identical to the PRBS Tx pattern testing error counter as shown in Table 45–95. Register 1.1700 contains the PRBS Rx pattern testing error counter for lane 0, register 1.1701 contains the PRBS Rx pattern testing error counter for lane 1, and registers 1.1702 through 1.1715~~09~~ contain the PRBS Rx pattern testing error counters for lanes 2 through 15~~9~~ respectively. Counters corresponding to lanes that are not implemented in a PMA shall read all zeros.

The PRBS Rx pattern testing error counter is a 16-bit counter as defined in 83.5.10 and 120.5.11. These bits shall be reset to all zeros when the register is read by the management function or upon execution of the PMA reset. These bits shall be held at all ones in the case of overflow.

45.2.3 PCS registers

Change the reserved row for 3.9 through 3.13 in Table 45-119 as follows (unchanged rows not shown):

Table 45–119—PCS registers

Register address	Register name	Subclause
…		
3.9	PCS status 3	45.2.3.7a
3.10~~9~~ through 3.13	Reserved	
…		

Change the reserved row for 3.523 through 3.1799 in Table 45-119 (as modified by IEEE Std 802.3bv-2017) as follows (unchanged rows not shown):

Table 45–119—PCS registers

Register address	Register name	Subclause
…		
3.523 through 3.~~1799~~599	Reserved	
3.600 through 3.631	PCS FEC symbol error counter, lane 0 to 15	45.2.3.47h, 45.2.3.47i
3.632 through 3.799	Reserved	
3.800	PCS FEC control	45.2.3.47j
3.801	PCS FEC status	45.2.3.47k
3.802, 3.803	PCS FEC corrected codewords counter	45.2.3.47l
3.804, 3.805	PCS FEC uncorrected codewords counter	45.2.3.47m
3.806, 3.807	PCS FEC degraded SER activate threshold	45.2.3.47n

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Table 45–119—PCS registers *(continued)*

Register address	Register name	Subclause
3.808, 3.809	PCS FEC degraded SER deactivate threshold	45.2.3.47o
3.810, 3.811	PCS FEC degraded SER interval	45.2.3.47p
3.812 through 3.1799	Reserved	
…		

45.2.3.1 PCS control 1 register (Register 3.0)

Change the row for bits 3.0.5:2 in Table 45-120 (as modified by IEEE Std 802.3bz-2016) as follows (unchanged rows not shown):

Table 45–120—PCS control 1 register bit definitions

Bit(s)	Name	Description	R/W[a]
…			
3.0.5:2	Speed selection	5 4 3 2 1 1 x x = Reserved ~~1 0 1 x = Reserved~~ 1 0 1 1 = Reserved 1 0 1 0 = 400 Gb/s 1 0 0 1 = 200 Gb/s ~~Reserved~~ 1 0 0 0 = 5 Gb/s 0 1 1 1 = 2.5 Gb/s 0 1 1 0 = Reserved 0 1 0 1 = 25 Gb/s 0 1 0 0 = 100 Gb/s 0 0 1 1 = 40 Gb/s 0 0 1 0 = 10/1 Gb/s 0 0 0 1 = 10PASS-TS/2BASE-TL 0 0 0 0 = 10 Gb/s	R/W
…			

[a] RO = Read only, R/W = Read/Write, SC = Self-clearing

45.2.3.2 PCS status 1 register (Register 3.1)

45.2.3.2.7 PCS receive link status (3.1.2)

Change the third sentence of 45.2.3.2.7 (as modified by IEEE Std 802.3by-2016 and IEEE Std 802.3bq-2016) as follows:

When a 10/25/40/100/200/400GBASE-R, 10GBASE-W, or any MultiGBASE-T mode of operation is selected for the PCS using the PCS type selection field (3.7.3:0), this bit is a latching low version of bit 3.32.12.

45.2.3.4 PCS speed ability (Register 3.4)

Change the reserved row of Table 45-122 (as modified by IEEE Std 802.3bz-2016) as follows (unchanged rows not shown):

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Table 45–122—PCS speed ability register bit definitions

Bit(s)	Name	Description	R/W[a]
3.4.15:~~8~~10	Reserved for future speeds	Value always 0	RO
3.4.9	400G capable	1 = PCS is capable of operating at 400 Gb/s 0 = PCS is not capable of operating at 400 Gb/s	RO
3.4.8	200G capable	1 = PCS is capable of operating at 200 Gb/s 0 = PCS is not capable of operating at 200 Gb/s	RO
…			

[a] RO = Read only

Insert 45.2.3.4.8 and 45.2.3.4.9 after 45.2.3.4.7 (as inserted by IEEE Std 802.3bz-2016) as follows:

45.2.3.4.8 200G capable (3.4.8)

When read as a one, bit 3.4.8 indicates that the PCS is able to operate at a data rate of 200 Gb/s. When read as a zero, bit 3.4.8 indicates that the PCS is not able to operate at a data rate of 200 Gb/s.

45.2.3.4.9 400G capable (3.4.9)

When read as a one, bit 3.4.9 indicates that the PCS is able to operate at a data rate of 400 Gb/s. When read as a zero, bit 3.4.9 indicates that the PCS is not able to operate at a data rate of 400 Gb/s.

45.2.3.6 PCS control 2 register (Register 3.7)

Change Table 45–123 (as modified by IEEE Std 802.3by-2016, IEEE Std 802.3bq-2016, and IEEE Std 802.3bz-2016) as follows:

Table 45–123—PCS control 2 register bit definitions

Bit(s)	Name	Description	R/W[a]
3.7.15:4	Reserved	Value always 0	RO
3.7.3:0	PCS type selection	3 2 1 0 1 1 1~~0~~ x = reserved 1 1 0 1 = Select 400GBASE-R PCS type 1 1 0 0 = Select 200GBASE-R PCS type 1 0 1 1 = Select 5GBASE-T PCS type 1 0 1 0 = Select 2.5GBASE-T PCS type 1 0 0 1 = Select 25GBASE-T PCS type 1 0 0 0 = reserved 0 1 1 1 = Select 25GBASE-R PCS type 0 1 1 0 = Select 40GBASE-T PCS type 0 1 0 1 = Select 100GBASE-R PCS type 0 1 0 0 = Select 40GBASE-R PCS type 0 0 1 1 = Select 10GBASE-T PCS type 0 0 1 0 = Select 10GBASE-W PCS type 0 0 0 1 = Select 10GBASE-X PCS type 0 0 0 0 = Select 10GBASE-R PCS type	R/W

[a] RO = Read only, R/W = Read/Write

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45.2.3.6.1 PCS type selection (3.7.3:0)

Change the second sentence of 45.2.3.6.1 (as modified by IEEE Std 802.3bq-2016) as follows:

The PCS type abilities of the PCS are advertised in bits 3.8.9, ~~and~~ 3.8.7:0, and 3.9.1:0.

Insert 45.2.3.7a after 45.2.3.7 as follows:

45.2.3.7a PCS status 3 register (Register 3.9)

The assignment of bits in the PCS status 3 register is shown in Table 45–124a. All the bits in the PCS status 3 register are read only; a write to the PCS status 3 register shall have no effect.

Table 45–124a—PCS status 3 register bit definitions

Bit(s)	Name	Description	R/W[a]
3.9.15:2	Reserved	Value always 0	RO
3.9.1	400GBASE-R capable	1 = PCS is able to support 400GBASE-R PCS type 0 = PCS is not able to support 400GBASE-R PCS type	RO
3.9.0	200GBASE-R capable	1 = PCS is able to support 200GBASE-R PCS type 0 = PCS is not able to support 200GBASE-R PCS type	RO

[a] RO = Read only

45.2.3.7a.1 400GBASE-R capable (3.9.1)

When read as a one, bit 3.9.1 indicates that the PCS is able to support the 400GBASE-R PCS type. When read as a zero, bit 3.9.1 indicates that the PCS is not able to support the 400GBASE-R PCS type.

45.2.3.7a.2 200GBASE-R capable (3.9.0)

When read as a one, bit 3.9.0 indicates that the PCS is able to support the 200GBASE-R PCS type. When read as a zero, bit 3.9.0 indicates that the PCS is not able to support the 200GBASE-R PCS type.

45.2.3.9a EEE control and capability 2 (Register 3.21)

Change the reserved row for 3.21.15:3 in Table 45-125a (as inserted by IEEE Std 802.3bq-2016) as follows (unchanged rows not shown):

Table 45–125a—EEE control and capability 2 register bit definitions

Bit(s)	Name	Description	R/W[a]
~~3.21.15:3~~ 3.21.15:6	Reserved	Value always 0	RO
3.21.5	400GBASE-R fast wake	1 = EEE fast wake is supported for 400GBASE-R 0 = EEE fast wake is not supported for 400GBASE-R	RO
3.21.4	Reserved	Value always 0	RO

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Table 45–125a—EEE control and capability 2 register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
3.21.3	200GBASE-R fast wake	1 = EEE fast wake is supported for 200GBASE-R 0 = EEE fast wake is not supported for 200GBASE-R	RO
…			

[a] RO = Read only

Insert 45.2.3.9a.a and 45.2.3.9a.b before 45.2.3.9a.1 (as inserted by IEEE Std 802.3bq-2016) as follows:

45.2.3.9a.a 400GBASE-R EEE fast wake supported (3.21.5)

If the PCS supports EEE fast wake operation for 400GBASE-R, this bit shall be set to a one; otherwise this bit shall be set to a zero.

45.2.3.9a.b 200GBASE-R EEE fast wake supported (3.21.3)

If the PCS supports EEE fast wake operation for 200GBASE-R, this bit shall be set to a one; otherwise this bit shall be set to a zero.

45.2.3.13 BASE-R and MultiGBASE-T PCS status 1 register (Register 3.32)

45.2.3.13.1 BASE-R and MultiGBASE-T receive link status (3.32.12)

Change the last sentence of 45.2.3.13.1 (as modified by IEEE Std 802.3by-2016, IEEE Std 802.3bq-2016, and IEEE Std 802.3bz-2016) as follows:

This bit is a reflection of the PCS_status variable defined in 49.2.14.1 for 10/25GBASE-R, in 126.3.7.1 for 2.5GBASE-T and 5GBASE-T, in 55.3.7.1 for 10GBASE-T, in 113.3.7.1 for 25GBASE-T and 40GBASE-T, ~~and~~ in 82.3.1 for 40/100GBASE-R, and in 119.3 for 200/400GBASE-R.

45.2.3.17 BASE-R PCS test-pattern control register (Register 3.42)

Change the fifth sentence of 45.2.3.17 (as modified by IEEE Std 802.3by-2016) as follows:

Scrambled idle test patterns are defined for 25/40/100/200/400GBASE-R PCS only.

45.2.3.46 Lane 0 mapping register (Register 3.400)

Change the text of 45.2.3.46 as follows:

The assignment of bits in the Lane 0 mapping register is shown in Table 45–160. When the multi-lane PCS described in Clause 82 or Clause 119 detects and locks the alignment marker for service interface lane 0, the detected PCS lane number is recorded in this register. The contents of the Lane 0 mapping register is valid when Lane 0 aligned bit (3.52.0) is set to one and is invalid otherwise.

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Insert 45.2.3.47h through 45.2.3.47p after 45.2.3.47g (as inserted by IEEE Std 802.3bv-2017) as follows:

45.2.3.47h PCS FEC symbol error counter lane 0 (Register 3.600, 3.601)

The assignment of bits in the PCS FEC symbol error counter lane 0 register is shown in Table 45–160h. Symbol errors detected in PCS lane 0 are counted and shown in register 3.600.15:0 and 3.601.15:0. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 3.600, 3.601 are used to read the value of a 32-bit counter. When registers 3.600 and 3.601 are used to read the 32-bit counter value, the register 3.600 is read first, the value of the register 3.601 is latched when (and only when) register 3.600 is read, and reads of register 3.601 return the latched value rather than the current value of the counter.

Table 45–160h—PCS FEC symbol error counter lane 0 register bit definitions

Bit(s)	Name	Description	R/W[a]
3.600.15:0	PCS FEC symbol errors, lane 0 lower	FEC_symbol_error_counter_0[15:0]	RO, NR
3.601.15:0	PCS FEC symbol errors, lane 0 upper	FEC_symbol_error_counter_0[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

45.2.3.47i PCS FEC symbol error counter lane 1 through 15 (Registers 3.602 through 3.631)

The behavior of the PCS FEC symbol error counters, lane 1 through 15 is identical to that described for PCS lane 0 in 45.2.3.47h. Errors detected in each PCS lane are counted and shown in the corresponding register. PCS lane 1, lower 16 bits are shown in register 3.602; PCS lane 1, upper 16 bits are shown in register 3.603; PCS lane 2, lower 16 bits are shown in register 3.604; through register 3.631 for PCS lane 15, upper 16 bits.

45.2.3.47j PCS FEC control register (Register 3.800)

The assignment of bits in the PCS FEC control register is shown in Table 45–160i.

Table 45–160i—PCS FEC control register bit definitions

Bit(s)	Name	Description	R/W[a]
3.800.15:3	Reserved	Value always 0	RO
3.800.2	PCS FEC degraded SER enable	1 = FEC decoder signals degraded SER 0 = FEC decoder does not signal degraded SER	R/W
3.800.1	PCS FEC bypass indication enable	1 = FEC decoder does not indicate errors 0 = FEC decoder indicates errors	R/W
3.800.0	Reserved	Value always 0	RO

[a] R/W = Read/Write, RO = Read only

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45.2.3.47j.1 PCS FEC degraded SER enable (3.800.2)

This bit enables the PCS FEC decoder to signal the presence of a degraded SER. When set to a one, this bit enables degraded SER signaling. When set to a zero, degraded SER signaling is disabled. Writes to this bit are ignored and reads return a zero if the PCS FEC does not have the ability to signal the presence of a degraded SER (see 119.2.5.3).

45.2.3.47j.2 PCS FEC bypass indication enable (3.800.1)

This bit enables the PCS FEC decoder to bypass error indication to the upper layers through the sync bits. When set to a one, this bit enables bypass of the error indication. When set to a zero, errors are indicated to the upper layers through the sync bits. Writes to this bit are ignored and reads return a zero if the PCS FEC does not have the ability to bypass indicating decoding errors (see 119.2.5.3).

45.2.3.47k PCS FEC status register (Register 3.801)

The assignment of bits in the PCS FEC status register is shown in Table 45–160j.

Table 45–160j—PCS FEC status register bit definitions

Bit(s)	Name	Description	R/W[a]
3.801.15:7	Reserved	Value always 0	RO
3.801.6	Local degraded SER received	1 = local degraded SER received 0 = no local degraded SER received	RO
3.801.5	Remote degraded SER received	1 = remote degraded SER received 0 = no remote degraded SER received	RO
3.801.4	PCS FEC degraded SER	1 = SER is degraded 0 = SER is not degraded	RO
3.801.3	PCS FEC degraded SER ability	1 = FEC decoder has the ability to signal the presence of a degraded SER 0 = FEC decoder does not have the ability to signal the presence of a degraded SER	RO
3.801.2	PCS FEC high SER	1 = FEC errors have exceeded threshold 0 = FEC errors have not exceeded threshold	RO/LH
3.801.1	PCS FEC bypass indication ability	1 = FEC decoder has the ability to bypass error indication 0 = FEC decoder does not have the ability to bypass error indication	RO
3.801.0	Reserved	Value always 0	RO

[a] RO = Read only, LH = Latching high

45.2.3.47k.1 Local degraded SER received (3.801.6)

When read as a one, bit 3.801.6 indicates that the local degraded SER signal has been received. This bit reflects the state of rx_local_degraded (see 119.2.6.2.2).

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45.2.3.47k.2 Remote degraded SER received (3.801.5)

When read as a one, bit 3.801.5 indicates that the remote degraded SER signal has been received. This bit reflects the state of rx_rm_degraded (see 119.2.6.2.2).

45.2.3.47k.3 PCS FEC degraded SER (3.801.4)

When read as a one, bit 3.801.4 indicates that the local PCS has detected a degradation of the received signal. This bit reflects the state of the variable FEC_degraded_SER (see 119.2.5.3). The value of bit 3.801.4 is unspecified if the value of the PCS FEC degraded SER activate threshold (registers 3.806 and 3.807) is less than the value of the PCS FEC degraded SER deactivate threshold (registers 3.808 and 3.809).

45.2.3.47k.4 PCS FEC degraded SER ability (3.801.3)

The PCS FEC decoder may have the option to signal the presence of a degraded SER (see 119.2.5.3). This bit is set to one to indicate that the decoder has the ability to signal the presence of a degraded SER. The bit is set to zero if this ability is not supported.

45.2.3.47k.5 PCS FEC high SER (3.801.2)

When PCS FEC bypass indication enable (bit 3.800.1) is set to one, this bit is set to one if the number of FEC symbol errors in a window of 8192 codewords exceeds the threshold (see 119.2.5.3) and is set to zero otherwise. The bit is set to zero if PCS FEC bypass indication enable (bit 3.800.1) is set to zero. This bit shall be implemented with latching high behavior.

45.2.3.47k.6 PCS FEC bypass indication ability (3.801.1)

The PCS FEC decoder may have the option to perform error detection without error indication (see 119.2.5.3) to reduce the delay contributed by the FEC decoder. This bit is set to one to indicate that the decoder has this ability to bypass error indication. The bit is set to zero if this ability is not supported.

45.2.3.47l PCS FEC corrected codewords counter (Register 3.802, 3.803)

The assignment of bits in the PCS FEC corrected codewords counter register is shown in Table 45–160k. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 3.802, 3.803 are used to read the value of a 32-bit counter. When registers 3.802 and 3.803 are used to read the 32-bit counter value, the register 3.802 is read first, the value of the register 3.803 is latched when (and only when) register 3.802 is read, and reads of register 3.803 return the latched value rather than the current value of the counter.

Table 45–160k—PCS FEC corrected codewords counter register bit definitions

Bit(s)	Name	Description	R/W[a]
3.802.15:0	FEC corrected codewords lower	FEC_corrected_cw_counter[15:0]	RO, NR
3.803.15:0	FEC corrected codewords upper	FEC_corrected_cw_counter[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

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45.2.3.47m PCS FEC uncorrected codewords counter (Register 3.804, 3.805)

The assignment of bits in the PCS FEC uncorrected codewords counter register is shown in Table 45–160l. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 3.804, 3.805 are used to read the value of a 32-bit counter. When registers 3.804 and 3.805 are used to read the 32-bit counter value, the register 3.804 is read first, the value of the register 3.805 is latched when (and only when) register 3.804 is read, and reads of register 3.805 return the latched value rather than the current value of the counter.

Table 45–160l—PCS FEC uncorrected codewords counter register bit definitions

Bit(s)	Name	Description	R/W[a]
3.804.15:0	FEC uncorrected codewords lower	FEC_uncorrected_cw_counter[15:0]	RO, NR
3.805.15:0	FEC uncorrected codewords upper	FEC_uncorrected_cw_counter[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

45.2.3.47n PCS FEC degraded SER activate threshold register (Register 3.806, 3.807)

The assignment of bits in the PCS FEC degraded SER activate threshold register is shown in Table 45–160m. The value controls the threshold used to set the PCS FEC degraded SER bit (3.801.4) as defined in 119.2.5.3.

Table 45–160m—PCS FEC degraded SER activate threshold register bit definitions

Bit(s)	Name	Description	R/W[a]
3.806.15:0	PCS FEC degraded SER activate threshold lower	FEC_degraded_SER_activate_threshold[15:0]	R/W
3.807.15:0	PCS FEC degraded SER activate threshold upper	FEC_degraded_SER_activate_threshold[31:16]	R/W

[a] R/W = Read/Write

45.2.3.47o PCS FEC degraded SER deactivate threshold register (Register 3.808, 3.809)

The assignment of bits in the PCS FEC degraded SER deactivate threshold register is shown in Table 45–160n. The value controls the threshold used to clear the PCS FEC degraded SER bit (3.801.4) as defined in 119.2.5.3.

Table 45–160n—PCS FEC degraded SER deactivate threshold register bit definitions

Bit(s)	Name	Description	R/W[a]
3.808.15:0	PCS FEC degraded SER deactivate threshold lower	FEC_degraded_SER_deactivate_threshold[15:0]	R/W
3.809.15:0	PCS FEC degraded SER deactivate threshold upper	FEC_degraded_SER_deactivate_threshold[31:16]	R/W

[a] R/W = Read/Write

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45.2.3.47p PCS FEC degraded SER interval register (Register 3.810, 3.811)

The assignment of bits in the PCS FEC degraded SER interval register is shown in Table 45–160o. The value controls the interval used to set and clear the PCS FEC degraded SER bit (3.801.4) as defined in 119.2.5.3.

Table 45–160o—PCS FEC degraded SER interval register bit definitions

Bit(s)	Name	Description	R/W[a]
3.810.15:0	PCS FEC degraded SER interval lower	FEC_degraded_SER_interval[15:0]	R/W
3.811.15:0	PCS FEC degraded SER interval upper	FEC_degraded_SER_interval[31:16]	R/W

[a] R/W = Read/Write

45.2.4 PHY XS registers

Change the reserved row for 4.26 through 4.1799 in Table 45-164 as follows (unchanged rows not shown):

Table 45–164—PHY XS registers

Register address	Register name	Subclause
…		
4.26 through 4.31 ~~4.1799~~	Reserved	
4.32	BASE-R PHY XS status 1	45.2.4.11a
4.33 through 4.41	Reserved	
4.42	BASE-R PHY XS test pattern control	45.2.4.11b
4.43 through 4.49	Reserved	
4.50	Multi-lane BASE-R PHY XS alignment status 1	45.2.4.11c
4.51	Reserved	
4.52	Multi-lane BASE-R PHY XS alignment status 3	45.2.4.11d
4.53	Multi-lane BASE-R PHY XS alignment status 4	45.2.4.11e
4.54	Reserved	
4.400 through 4.415	PHY XS lane mapping, lane 0 through 15	45.2.4.11f, 45.2.4.11g
4.415 through 4.599	Reserved	
4.600 through 4.631	PHY XS FEC symbol error counter, lane 0 to 15	45.2.4.11h, 45.2.4.11i
4.632 through 4.799	Reserved	
4.800	PHY XS FEC control	45.2.4.11j
4.801	PHY XS FEC status	45.2.4.11k
4.802, 4.803	PHY XS FEC corrected codewords counter	45.2.4.11l

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Table 45–164—PHY XS registers *(continued)*

Register address	Register name	Subclause
4.804, 4.805	PHY XS FEC uncorrected codewords counter	45.2.4.11m
4.806, 4.807	PHY XS FEC degraded SER activate threshold	45.2.4.11n
4.808, 4.809	PHY XS FEC degraded SER deactivate threshold	45.2.4.11o
4.810, 4.811	PHY XS FEC degraded SER interval	45.2.4.11p
4.812 through 4.1799	Reserved	
…		

45.2.4.1 PHY XS control 1 register (Register 4.0)

Change the row for 4.0.5:2 in Table 45-165 as follows (unchanged rows not shown):

Table 45–165—PHY XS control 1 register bit definitions

Bit(s)	Name	Description	R/W[a]
…			
4.0.5:2	Speed selection	5 4 3 2 ~~1 x x x = Reserved~~ ~~x 1 x x = Reserved~~ ~~x x 1 x = Reserved~~ 1 1 x x = Reserved 1 0 1 1 = Reserved 1 0 1 0 = 400 Gb/s 1 0 0 1 = 200 Gb/s 1 0 0 0 = Reserved 0 1 x x = Reserved 0 0 1 x = Reserved 0 0 0 1 = Reserved 0 0 0 0 = 10 Gb/s	R/W
…			

[a] R/W = Read/Write, SC = Self-clearing, RO = Read only

45.2.4.4 PHY XS speed ability (Register 4.4)

Change Table 45-167 as follows:

Table 45–167—PHY XS speed ability register bit definitions

Bit(s)	Name	Description	R/W[a]
4.4.15:10	Reserved~~ for future speeds~~	Value always 0	RO
4.4.9	400G capable	1 = PHY XS is capable of operating at 400 Gb/s 0 = PHY XS is not capable of operating at 400 Gb/s	RO
4.4.8	200G capable	1 = PHY XS is capable of operating at 200 Gb/s 0 = PHY XS is not capable of operating at 200 Gb/s	RO

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Table 45–167—PHY XS speed ability register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
4.4.7:1	Reserved	Value always 0	RO
4.4.0	10G capable	1 = PHY XS is capable of operating at 10 Gb/s 0 = PHY XS is not capable of operating at 10 Gb/s	RO

[a] RO = Read only

Insert 45.2.4.4.a and 45.2.4.4.b before 45.2.4.4.1 as follows:

45.2.4.4.a 400G capable (4.4.9)

When read as a one, bit 4.4.9 indicates that the PHY XS is able to operate at a data rate of 400 Gb/s. When read as a zero, bit 4.4.9 indicates that the PHY-XS is not able to operate at a data rate of 400 Gb/s.

45.2.4.4.b 200G capable (4.4.8)

When read as a one, bit 4.4.8 indicates that the PHY XS is able to operate at a data rate of 200 Gb/s. When read as a zero, bit 4.4.8 indicates that the PHY-XS is not able to operate at a data rate of 200 Gb/s.

Insert 45.2.4.11a through 45.2.4.11p after 45.2.4.11 as follows:

45.2.4.11a BASE-R PHY XS status 1 register (Register 4.32)

The assignment of bits in the BASE-R PHY XS status 1 register is shown in Table 45–171a. A PHY XS device that does not implement BASE-R shall return a zero for all bits in the BASE-R PHY XS status 1 register. It is the responsibility of the STA management entity to ensure that a port type is supported by all MMDs before interrogating any of its status bits.

Table 45–171a—BASE-R PHY XS status 1 register bit definitions

Bit(s)	Name	Description	R/W[a]
4.32.15:13	Reserved	Value always 0	RO
4.32.12	BASE-R PHY XS receive link status	1 = BASE-R PHY XS receive link up 0 = BASE-R PHY XS receive link down	RO
4.32.11:0	Reserved	Value always 0	RO

[a] RO = Read only

45.2.4.11a.1 BASE-R PHY XS receive link status (4.32.12)

When read as a one, bit 4.32.12 indicates that the PHY XS is in a fully operational state. When read as a zero, bit 4.32.12 indicates that the PHY XS is not fully operational. This bit is a reflection of the PHY XS equivalent of the PCS_status variable defined in 119.3 for 200/400GBASE-R.

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45.2.4.11b BASE-R PHY XS test-pattern control register (Register 4.42)

The assignment of bits in the BASE-R PHY XS test-pattern control register is shown in Table 45–171b. This register is only required when the 200/400GBASE-R capability is supported. The test-pattern methodology is described in 119.2.4.9.

Table 45–171b—BASE-R PHY XS test-pattern control register bit definitions

Bit(s)	Name	Description	R/W[a]
4.42.15:4	Reserved	Value always 0	RO
4.42.3	Transmit test-pattern enable	1 = Enable transmit test pattern 0 = Disable transmit test pattern	R/W
4.42.2:0	Reserved	Value always 0	RO

[a] RO = Read only, R/W = Read/Write

45.2.4.11b.1 Transmit test-pattern enable (4.42.3)

When bit 4.42.3 is set to a one, pattern testing is enabled on the transmit path. When bit 4.42.3 is set to a zero, pattern testing is disabled on the transmit path.

The default value for bit 4.42.3 is zero.

45.2.4.11c Multi-lane BASE-R PHY XS alignment status 1 register (Register 4.50)

The assignment of bits in the multi-lane BASE-R PHY XS alignment status 1 register is shown in Table 45–171c. A PHY XS device that does not implement multi-lane BASE-R PHY XS shall return a zero for all bits in the multi-lane BASE-R PHY XS alignment status 1 register. It is the responsibility of the STA management entity to ensure that a port type is supported by all MMDs before interrogating any of its status bits.

Table 45–171c—Multi-lane BASE-R PHY XS alignment status 1 register bit definitions

Bit(s)	Name	Description	R/W[a]
4.50.15:13	Reserved	Value always 0	RO
4.50.12	PHY XS lane alignment status	1 = PHY XS receive lanes locked and aligned 0 = PHY XS receive lanes not locked and aligned	RO
4.50.11:0	Reserved	Value always 0	RO

[a] RO = Read only

45.2.4.11c.1 PHY XS lane alignment status (4.50.12)

When read as a one, bit 4.50.12 indicates that the PHY XS has locked and aligned all receive lanes. When read as a zero, bit 4.50.12 indicates that the PHY XS has not locked and aligned all receive lanes.

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45.2.4.11d Multi-lane BASE-R PHY XS alignment status 3 register (Register 4.52)

The assignment of bits in the multi-lane BASE-R PHY XS alignment status 3 register is shown in Table 45–171d. A PHY XS device that does not implement multi-lane BASE-R PHY XS shall return a zero for all bits in the multi-lane BASE-R PHY XS alignment status 3 register. A device that implements multi-lane BASE-R PHY XS shall return a zero for all bits in the multi-lane BASE-R PHY XS alignment status 3 register that are not required for the PHY XS configuration. It is the responsibility of the STA management entity to ensure that a port type is supported by all MMDs before interrogating any of its status bits.

Table 45–171d—Multi-lane BASE-R PHY XS alignment status 3 register bit definitions

Bit(s)	Name	Description	R/W[a]
4.52.15:8	Reserved	Value always 0	RO
4.52.7	Lane 7 aligned	1 = Lane 7 alignment marker is locked 0 = Lane 7 alignment marker is not locked	RO
4.52.6	Lane 6 aligned	1 = Lane 6 alignment marker is locked 0 = Lane 6 alignment marker is not locked	RO
4.52.5	Lane 5 aligned	1 = Lane 5 alignment marker is locked 0 = Lane 5 alignment marker is not locked	RO
4.52.4	Lane 4 aligned	1 = Lane 4 alignment marker is locked 0 = Lane 4 alignment marker is not locked	RO
4.52.3	Lane 3 aligned	1 = Lane 3 alignment marker is locked 0 = Lane 3 alignment marker is not locked	RO
4.52.2	Lane 2 aligned	1 = Lane 2 alignment marker is locked 0 = Lane 2 alignment marker is not locked	RO
4.52.1	Lane 1 aligned	1 = Lane 1 alignment marker is locked 0 = Lane 1 alignment marker is not locked	RO
4.52.0	Lane 0 aligned	1 = Lane 0 alignment marker is locked 0 = Lane 0 alignment marker is not locked	RO

[a] RO = Read only

45.2.4.11d.1 Lane 7 aligned (4.52.7)

When read as a one, bit 4.52.7 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 7. When read as a zero, bit 4.52.7 indicates that the PHY XS receiver lane 7 has not achieved alignment marker lock. This bit reflects the state of amps_lock[7] (see 119.2.6.2.2).

45.2.4.11d.2 Lane 6 aligned (4.52.6)

When read as a one, bit 4.52.6 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 6. When read as a zero, bit 4.52.6 indicates that the PHY XS receiver lane 6 has not achieved alignment marker lock. This bit reflects the state of amps_lock[6] (see 119.2.6.2.2).

45.2.4.11d.3 Lane 5 aligned (4.52.5)

When read as a one, bit 4.52.5 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 5. When read as a zero, bit 4.52.5 indicates that the PHY XS receiver lane 5 has not achieved alignment marker lock. This bit reflects the state of amps_lock[5] (see 119.2.6.2.2).

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45.2.4.11d.4 Lane 4 aligned (4.52.4)

When read as a one, bit 4.52.4 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 4. When read as a zero, bit 4.52.4 indicates that the PHY XS receiver lane 4 has not achieved alignment marker lock. This bit reflects the state of amps_lock[4] (see 119.2.6.2.2).

45.2.4.11d.5 Lane 3 aligned (4.52.3)

When read as a one, bit 4.52.3 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 3. When read as a zero, bit 4.52.3 indicates that the PHY XS receiver lane 3 has not achieved alignment marker lock. This bit reflects the state of amps_lock[3] (see 119.2.6.2.2).

45.2.4.11d.6 Lane 2 aligned (4.52.2)

When read as a one, bit 4.52.2 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 2. When read as a zero, bit 4.52.2 indicates that the PHY XS receiver lane 2 has not achieved alignment marker lock. This bit reflects the state of amps_lock[2] (see 119.2.6.2.2).

45.2.4.11d.7 Lane 1 aligned (4.52.1)

When read as a one, bit 4.52.1 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 1. When read as a zero, bit 4.52.1 indicates that the PHY XS receiver lane 1 has not achieved alignment marker lock. This bit reflects the state of amps_lock[1] (see 119.2.6.2.2).

45.2.4.11d.8 Lane 0 aligned (4.52.0)

When read as a one, bit 4.52.0 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 0. When read as a zero, bit 4.52.0 indicates that the PHY XS receiver lane 0 has not achieved alignment marker lock. This bit reflects the state of amps_lock[0] (see 119.2.6.2.2).

45.2.4.11e Multi-lane BASE-R PHY XS alignment status 4 register (Register 4.53)

The assignment of bits in the multi-lane BASE-R PHY XS alignment status 4 register is shown in Table 45–171e. A PHY XS device that does not implement multi-lane BASE-R PHY XS shall return a zero for all bits in the multi-lane BASE-R PHY XS alignment status 4 register. A device that implements multi-lane BASE-R PHY XS shall return a zero for all bits in the multi-lane BASE-R PHY XS alignment status 4 register that are not required for the PHY XS configuration. It is the responsibility of the STA management entity to ensure that a port type is supported by all MMDs before interrogating any of its status bits.

Table 45–171e—Multi-lane BASE-R PHY XS alignment status 4 register bit definitions

Bit(s)	Name	Description	R/W[a]
4.53.15:8	Reserved	Value always 0	RO
4.53.7	Lane 15 aligned	1 = Lane 15 alignment marker is locked 0 = Lane 15 alignment marker is not locked	RO
4.53.6	Lane 14 aligned	1 = Lane 14 alignment marker is locked 0 = Lane 14 alignment marker is not locked	RO
4.53.5	Lane 13 aligned	1 = Lane 13 alignment marker is locked 0 = Lane 13 alignment marker is not locked	RO

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Table 45–171e—Multi-lane BASE-R PHY XS alignment status 4 register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
4.53.4	Lane 12 aligned	1 = Lane 12 alignment marker is locked 0 = Lane 12 alignment marker is not locked	RO
4.53.3	Lane 11 aligned	1 = Lane 11 alignment marker is locked 0 = Lane 11 alignment marker is not locked	RO
4.53.2	Lane 10 aligned	1 = Lane 10 alignment marker is locked 0 = Lane 10 alignment marker is not locked	RO
4.53.1	Lane 9 aligned	1 = Lane 9 alignment marker is locked 0 = Lane 9 alignment marker is not locked	RO
4.53.0	Lane 8 aligned	1 = Lane 8 alignment marker is locked 0 = Lane 8 alignment marker is not locked	RO

[a] RO = Read only

45.2.4.11e.1 Lane 15 aligned (4.53.7)

When read as a one, bit 4.53.7 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 15. When read as a zero, bit 4.53.7 indicates that the PHY XS receiver lane 15 has not achieved alignment marker lock. This bit reflects the state of amps_lock[15] (see 119.2.6.2.2).

45.2.4.11e.2 Lane 14 aligned (4.53.6)

When read as a one, bit 4.53.6 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 14. When read as a zero, bit 4.53.6 indicates that the PHY XS receiver lane 14 has not achieved alignment marker lock. This bit reflects the state of amps_lock[14] (see 119.2.6.2.2).

45.2.4.11e.3 Lane 13 aligned (4.53.5)

When read as a one, bit 4.53.5 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 13. When read as a zero, bit 4.53.5 indicates that the PHY XS receiver lane 13 has not achieved alignment marker lock. This bit reflects the state of amps_lock[13] (see 119.2.6.2.2).

45.2.4.11e.4 Lane 12 aligned (4.53.4)

When read as a one, bit 4.53.4 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 12. When read as a zero, bit 4.53.4 indicates that the PHY XS receiver lane 12 has not achieved alignment marker lock. This bit reflects the state of amps_lock[12] (see 119.2.6.2.2).

45.2.4.11e.5 Lane 11 aligned (4.53.3)

When read as a one, bit 4.53.3 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 11. When read as a zero, bit 4.53.3 indicates that the PHY XS receiver lane 11 has not achieved alignment marker lock. This bit reflects the state of amps_lock[11] (see 119.2.6.2.2).

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45.2.4.11e.6 Lane 10 aligned (4.53.2)

When read as a one, bit 4.53.2 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 10. When read as a zero, bit 4.53.2 indicates that the PHY XS receiver lane 10 has not achieved alignment marker lock. This bit reflects the state of amps_lock[10] (see 119.2.6.2.2).

45.2.4.11e.7 Lane 9 aligned (4.53.1)

When read as a one, bit 4.53.1 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 9. When read as a zero, bit 4.53.1 indicates that the PHY XS receiver lane 9 has not achieved alignment marker lock. This bit reflects the state of amps_lock[9] (see 119.2.6.2.2).

45.2.4.11e.8 Lane 8 aligned (4.53.0)

When read as a one, bit 4.53.0 indicates that the PHY XS receiver has achieved alignment marker lock for service interface lane 8. When read as a zero, bit 4.53.0 indicates that the PHY XS receiver lane 8 has not achieved alignment marker lock. This bit reflects the state of amps_lock[8] (see 119.2.6.2.2).

45.2.4.11f PHY XS lane mapping, lane 0 register (Register 4.400)

The assignment of bits in the PHY XS lane mapping, lane 0 register is shown in Table 45–171f. When the multi-lane PHY XS described in Clause 118 detects and locks the alignment marker for service interface lane 0, the detected PCS lane number is recorded in this register. The contents of the PHY XS lane mapping, lane 0 register is valid when Lane 0 aligned bit (4.52.0) is set to one and is invalid otherwise.

Table 45–171f—PHY XS lane mapping, lane 0 register bit definitions

Bit(s)	Name	Description	R/W[a]
4.400.15:5	Reserved	Value always 0	RO
4.400.4:0	Lane 0 mapping	PHY XS lane received in service interface lane 0	RO

[a] RO = Read only

45.2.4.11g PHY XS lane mapping, lane 1 through lane 15 registers (Registers 4.401 through 4.415)

The definition of the PHY XS lane mapping, lane 1 through 15 registers is identical to that described for lane 0 in 45.2.4.11f. The lane mapping for lane 1 is in register 4.401; lane 2 is in register 4.402; etc.

45.2.4.11h PHY XS FEC symbol error counter lane 0 (Register 4.600, 4.601)

The assignment of bits in the PHY XS FEC symbol error counter lane 0 register is shown in Table 45–171g. Symbol errors detected in PHY XS FEC lane 0 are counted and shown in register 4.600.15:0 and 4.601.15:0. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 4.600, 4.601 are used to read the value of a 32-bit counter. When registers 4.600 and 4.601 are used to read the 32-bit counter value, the register 4.600 is read first, the value of the register 4.601 is latched when (and only when) register 4.600 is read, and reads of register 4.601 return the latched value rather than the current value of the counter.

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Table 45–171g—PHY XS FEC symbol error counter lane 0 register bit definitions

Bit(s)	Name	Description	R/W[a]
4.600.15:0	PHY XS FEC symbol errors, lane 0 lower	FEC_symbol_error_counter_0[15:0]	RO, NR
4.601.15:0	PHY XS FEC symbol errors, lane 0 upper	FEC_symbol_error_counter_0[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

45.2.4.11i PHY XS FEC symbol error counter lane 1 through 15 (Registers 4.602 through 4.631)

The behavior of the PHY XS FEC symbol error counters, lane 1 through 15 is identical to that described for PHY XS FEC lane 0 in 45.2.4.11h. Errors detected in each PHY XS FEC lane are counted and shown in the corresponding register. PHY XS FEC lane 1, lower 16 bits are shown in register 4.602; PHY XS FEC lane 1, upper 16 bits are shown in register 4.603; PHY XS FEC lane 2, lower 16 bits are shown in register 4.604; through register 4.631 for PHY XS FEC lane 15, upper 16 bits.

45.2.4.11j PHY XS FEC control register (Register 4.800)

The assignment of bits in the PHY XS FEC control register is shown in Table 45–171h.

Table 45–171h—PHY XS FEC control register bit definitions

Bit(s)	Name	Description	R/W[a]
4.800.15:3	Reserved	Value always 0	RO
4.800.2	PHY XS FEC degraded SER enable	1 = FEC decoder signals degraded SER 0 = FEC decoder does not signal degraded SER	R/W
4.800.1	PHY XS FEC bypass indication enable	1 = FEC decoder does not indicate errors 0 = FEC decoder indicates errors	R/W
4.800.0	Reserved	Value always 0	RO

[a] R/W = Read/Write, RO = Read only

45.2.4.11j.1 PHY XS FEC degraded SER enable (4.800.2)

This bit enables the PHY XS FEC decoder to signal the presence of a degraded SER. When set to a one, this bit enables degraded SER signaling. When set to a zero, degraded SER signaling is disabled. Writes to this bit are ignored and reads return a zero if the PHY XS FEC does not have the ability to signal the presence of a degraded SER (see 119.2.5.3 for equivalent PCS behavior).

45.2.4.11j.2 PHY XS FEC bypass indication enable (4.800.1)

This bit enables the PHY XS FEC decoder to bypass error indication to the upper layers through the sync bits. When set to a one, this bit enables bypass of the error indication. When set to a zero, errors are indicated to the upper layers through the sync bits. Writes to this bit are ignored and reads return a zero if the PHY XS FEC does not have the ability to bypass indicating decoding errors (see 119.2.5.3 for equivalent PCS behavior).

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45.2.4.11k PHY XS FEC status register (Register 4.801)

The assignment of bits in the PHY XS FEC status register is shown in Table 45–171i.

Table 45–171i—PHY XS FEC status register bit definitions

Bit(s)	Name	Description	R/W[a]
4.801.15:6	Reserved	Value always 0	RO
4.801.5	Remote degraded SER received	1 = remote degraded SER received 0 = no remote degraded SER received	RO
4.801.4	PHY XS FEC degraded SER	1 = SER is degraded 0 = SER is not degraded	RO
4.801.3	PHY XS FEC degraded SER ability	1 = FEC decoder has the ability to signal the presence of a degraded SER 0 = FEC decoder does not have the ability to signal the presence of a degraded SER	RO
4.801.2	PHY XS FEC high SER	1 = FEC errors have exceeded threshold 0 = FEC errors have not exceeded threshold	RO/LH
4.801.1	PHY XS FEC bypass indication ability	1 = FEC decoder has the ability to bypass error indication 0 = FEC decoder does not have the ability to bypass error indication	RO
4.801.0	Reserved	Value always 0	RO

[a] RO = Read only, LH = Latching high

45.2.4.11k.1 Remote degraded SER received (4.801.5)

When read as a one, bit 4.801.5 indicates that the remote degraded SER signal has been received. This bit reflects the state of rx_rm_degraded for the PHY XS (see 119.2.6.2.2 for equivalent PCS behavior).

45.2.4.11k.2 PHY XS FEC degraded SER (4.801.4)

When PHY XS FEC degraded SER enable (bit 4.800.2) is set to one, bit 4.801.4 is set to one if the number of FEC symbol errors in a window of PHY XS FEC degraded SER interval (registers 4.810 and 4.811) codewords exceeds the PHY XS FEC degraded SER activate threshold (registers 4.806 and 4.807) and is cleared if the number of FEC symbol errors in the same window is below the PHY XS FEC degraded SER deactivate threshold (registers 4.808 and 4.809). If the number of FEC symbol errors in the window is between the two thresholds, then bit 4.801.4 remains in its previous state. The bit is set to zero if PHY XS FEC degraded SER enable (bit 4.800.2) is set to zero (see 119.2.5.3 for equivalent PCS behavior). The value of bit 4.801.4 is undefined if the value of the PHY XS FEC degraded SER activate threshold is less than the value of the PHY XS FEC degraded SER deactivate threshold.

45.2.4.11k.3 PHY XS FEC degraded SER ability (4.801.3)

The PHY XS FEC decoder may have the option to signal the presence of a degraded SER (see 119.2.5.3 for equivalent PCS behavior). This bit is set to one to indicate that the decoder has the ability to signal the presence of a degraded SER. The bit is set to zero if this ability is not supported.

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45.2.4.11k.4 PHY XS FEC high SER (4.801.2)

When PHY XS FEC bypass indication enable (bit 4.800.1) is set to one, this bit is set to one if the number of FEC symbol errors in a window of 8192 codewords exceeds the threshold (see 119.2.5.3 for equivalent PCS behavior) and is set to zero otherwise. The bit is set to zero if PHY XS FEC bypass indication enable (bit 4.800.1) is set to zero. This bit shall be implemented with latching high behavior.

45.2.4.11k.5 PHY XS FEC bypass indication ability (4.801.1)

The PHY XS FEC decoder may have the option to perform error detection without error indication (see 119.2.5.3 for equivalent PCS behavior) to reduce the delay contributed by the FEC decoder. This bit is set to one to indicate that the decoder has this ability to bypass error indication. The bit is set to zero if this ability is not supported.

45.2.4.11l PHY XS FEC corrected codewords counter (Register 4.802, 4.803)

The assignment of bits in the PHY XS FEC corrected codewords counter register is shown in Table 45–171j. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 4.802, 4.803 are used to read the value of a 32-bit counter. When registers 4.802 and 4.803 are used to read the 32-bit counter value, the register 4.802 is read first, the value of the register 4.803 is latched when (and only when) register 4.802 is read, and reads of register 4.803 return the latched value rather than the current value of the counter.

Table 45–171j—PHY XS FEC corrected codewords counter register bit definitions

Bit(s)	Name	Description	R/W[a]
4.802.15:0	FEC corrected codewords lower	FEC_corrected_cw_counter[15:0]	RO, NR
4.803.15:0	FEC corrected codewords upper	FEC_corrected_cw_counter[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

45.2.4.11m PHY XS FEC uncorrected codewords counter (Register 4.804, 4.805)

The assignment of bits in the PHY XS FEC uncorrected codewords counter register is shown in Table 45–171k. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 4.804, 4.805 are used to read the value of a 32-bit counter. When registers 4.804 and 4.805 are used to read the 32-bit counter value, the register 4.804 is read first, the value of the register 4.805 is latched when (and only when) register 4.804 is read, and reads of register 4.805 return the latched value rather than the current value of the counter.

Table 45–171k—PHY XS FEC uncorrected codewords counter register bit definitions

Bit(s)	Name	Description	R/W[a]
4.804.15:0	FEC uncorrected codewords lower	FEC_uncorrected_cw_counter[15:0]	RO, NR
4.805.15:0	FEC uncorrected codewords upper	FEC_uncorrected_cw_counter[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

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45.2.4.11n PHY XS FEC degraded SER activate threshold register (Register 4.806, 4.807)

The assignment of bits in the PHY XS FEC degraded SER activate threshold register is shown in Table 45–171l. The value controls the threshold used to set the PHY XS FEC degraded SER bit (4.801.4) in an equivalent manner to that defined for the FEC_degraded_SER bit in 119.2.5.3.

Table 45–171l—PHY XS FEC degraded SER activate threshold register bit definitions

Bit(s)	Name	Description	R/W[a]
4.806.15:0	PHY XS FEC degraded SER activate threshold lower	FEC_degraded_SER_activate_threshold[15:0]	R/W
4.807.15:0	PHY XS FEC degraded SER activate threshold upper	FEC_degraded_SER_activate_threshold[31:16]	R/W

[a] R/W = Read/Write

45.2.4.11o PHY XS FEC degraded SER deactivate threshold register (Register 4.808, 4.809)

The assignment of bits in the PHY XS FEC degraded SER deactivate threshold register is shown in Table 45–171m. The value controls the threshold used to clear the PHY XS FEC degraded SER bit (4.801.4) in an equivalent manner to that defined for the FEC_degraded_SER bit in 119.2.5.3.

Table 45–171m—PHY XS FEC degraded SER deactivate threshold register bit definitions

Bit(s)	Name	Description	R/W[a]
4.808.15:0	PHY XS FEC degraded SER deactivate threshold lower	FEC_degraded_SER_deactivate_threshold[15:0]	R/W
4.809.15:0	PHY XS FEC degraded SER deactivate threshold upper	FEC_degraded_SER_deactivate_threshold[31:16]	R/W

[a] R/W = Read/Write

45.2.4.11p PHY XS FEC degraded SER interval register (Register 4.810, 4.811)

The assignment of bits in the PHY XS FEC degraded SER interval register is shown in Table 45–171n. The value controls the interval used to set and clear the PHY XS FEC degraded SER bit (4.801.4) in an equivalent manner to that defined for the FEC_degraded_SER bit in 119.2.5.3.

Table 45–171n—PHY XS FEC degraded SER interval register bit definitions

Bit(s)	Name	Description	R/W[a]
4.810.15:0	PHY XS FEC degraded SER interval lower	FEC_degraded_SER_interval[15:0]	R/W
4.811.15:0	PHY XS FEC degraded SER interval upper	FEC_degraded_SER_interval[31:16]	R/W

[a] R/W = Read/Write

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45.2.5 DTE XS registers

Change the reserved row for 5.26 through 5.1799 in Table 45-175 as follows (unchanged rows not shown):

Table 45–175—DTE XS registers

Register address	Register name	Subclause
…		
5.26 through 5.31 ~~5.1799~~	Reserved	
5.32	BASE-R DTE XS status 1	45.2.5.11a
5.33 through 5.41	Reserved	
5.42	BASE-R DTE XS test pattern control	45.2.5.11b
5.43 through 5.49	Reserved	
5.50	Multi-lane BASE-R DTE XS alignment status 1	45.2.5.11c
5.51	Reserved	
5.52	Multi-lane BASE-R DTE XS alignment status 3	45.2.5.11d
5.53	Multi-lane BASE-R DTE XS alignment status 4	45.2.5.11e
5.54	Reserved	
5.400 through 5.415	DTE XS lane mapping, lane 0 through 15	45.2.5.11f, 45.2.5.11g
5.415 through 5.599	Reserved	
5.600 through 5.631	DTE XS FEC symbol error counter, lane 0 to 15	45.2.5.11h, 45.2.5.11i
5.632 through 5.799	Reserved	
5.800	DTE XS FEC control	45.2.5.11j
5.801	DTE XS FEC status	45.2.5.11k
5.802, 5.803	DTE XS FEC corrected codewords counter	45.2.5.11l
5.804, 5.805	DTE XS FEC uncorrected codewords counter	45.2.5.11m
5.806, 5.807	DTE XS FEC degraded SER activate threshold	45.2.5.11n
5.808, 5.809	DTE XS FEC degraded SER deactivate threshold	45.2.5.11o
5.810, 5.811	DTE XS FEC degraded SER interval	45.2.5.11p
5.812 through 5.1799	Reserved	
…		

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45.2.5.1 DTE XS control 1 register (Register 5.0)

Change the row for 5.0.5:2 in Table 45-176 as follows (unchanged rows not shown):

Table 45–176—DTE XS control 1 register bit definitions

Bit(s)	Name	Description	R/W[a]
…			
5.0.5:2	Speed selection	5 4 3 2 ~~1 x x x = Reserved~~ ~~x 1 x x = Reserved~~ ~~x x 1 x = Reserved~~ 1 1 x x = Reserved 1 0 1 1 = Reserved 1 0 1 0 = 400 Gb/s 1 0 0 1 = 200 Gb/s 1 0 0 0 = Reserved 0 1 x x = Reserved 0 0 1 x = Reserved 0 0 0 1 = Reserved 0 0 0 0 = 10 Gb/s	R/W
…			

[a] R/W = Read/Write, SC = Self-clearing, RO = Read only

45.2.5.4 DTE XS speed ability (Register 5.4)

Change Table 45-178 as follows:

Table 45–178— DTE XS speed ability register bit definitions

Bit(s)	Name	Description	R/W[a]
5.4.15:10	Reserved~~ for future speeds~~	Value always 0	RO
5.4.9	400G capable	1 = DTE XS is capable of operating at 400 Gb/s 0 = DTE XS is not capable of operating at 400 Gb/s	RO
5.4.8	200G capable	1 = DTE XS is capable of operating at 200 Gb/s 0 = DTE XS is not capable of operating at 200 Gb/s	RO
5.4.7:1	Reserved	Value always 0	RO
5.4.0	10G capable	1 = DTE XS is capable of operating at 10 Gb/s 0 = DTE XS is not capable of operating at 10 Gb/s	RO

[a] RO = Read only

Insert 45.2.5.4.a and 45.2.5.4.b before 45.2.5.4.1 as follows:

45.2.5.4.a 400G capable (5.4.9)

When read as a one, bit 5.4.9 indicates that the DTE XS is able to operate at a data rate of 400 Gb/s. When read as a zero, bit 5.4.9 indicates that the DTE XS is not able to operate at a data rate of 400 Gb/s.

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45.2.5.4.b 200G capable (5.4.8)

When read as a one, bit 5.4.8 indicates that the DTE XS is able to operate at a data rate of 200 Gb/s. When read as a zero, bit 5.4.8 indicates that the DTE XS is not able to operate at a data rate of 200 Gb/s.

Insert 45.2.5.11a through 45.2.5.11p after 45.2.5.11 as follows:

45.2.5.11a BASE-R DTE XS status 1 register (Register 5.32)

The assignment of bits in the BASE-R DTE XS status 1 register is shown in Table 45–182a. A DTE XS device that does not implement BASE-R shall return a zero for all bits in the BASE-R DTE XS status 1 register. It is the responsibility of the STA management entity to ensure that a port type is supported by all MMDs before interrogating any of its status bits.

Table 45–182a—BASE-R DTE XS status 1 register bit definitions

Bit(s)	Name	Description	R/W[a]
5.32.15:13	Reserved	Value always 0	RO
5.32.12	BASE-R DTE XS receive link status	1 = BASE-R DTE XS receive link up 0 = BASE-R DTE XS receive link down	RO
5.32.11:0	Reserved	Value always 0	RO

[a] RO = Read only

45.2.5.11a.1 BASE-R DTE XS receive link status (5.32.12)

When read as a one, bit 5.32.12 indicates that the DTE XS is in a fully operational state. When read as a zero, bit 5.32.12 indicates that the DTE XS is not fully operational. This bit is a reflection of the DTE XS equivalent of the PCS_status variable defined in 119.3 for 200/400GBASE-R.

45.2.5.11b BASE-R DTE XS test-pattern control register (Register 5.42)

The assignment of bits in the BASE-R DTE XS test-pattern control register is shown in Table 45–182b. This register is only required when the 200/400GBASE-R capability is supported. The test-pattern methodology is described in 119.2.4.9.

Table 45–182b—BASE-R DTE XS test-pattern control register bit definitions

Bit(s)	Name	Description	R/W[a]
5.42.15:4	Reserved	Value always 0	RO
5.42.3	Transmit test-pattern enable	1 = Enable transmit test pattern 0 = Disable transmit test pattern	R/W
5.42.2:0	Reserved	Value always 0	RO

[a] RO = Read only, R/W = Read/Write

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45.2.5.11b.1 Transmit test-pattern enable (5.42.3)

When bit 5.42.3 is set to a one, pattern testing is enabled on the transmit path. When bit 5.42.3 is set to a zero, pattern testing is disabled on the transmit path.

The default value for bit 5.42.3 is zero.

45.2.5.11c Multi-lane BASE-R DTE XS alignment status 1 register (Register 5.50)

The assignment of bits in the multi-lane BASE-R DTE XS alignment status 1 register is shown in Table 45–182c. A DTE XS device that does not implement multi-lane BASE-R DTE XS shall return a zero for all bits in the multi-lane BASE-R DTE XS alignment status 1 register. It is the responsibility of the STA management entity to ensure that a port type is supported by all MMDs before interrogating any of its status bits.

Table 45–182c—Multi-lane BASE-R DTE XS alignment status 1 register bit definitions

Bit(s)	Name	Description	R/W[a]
5.50.15:13	Reserved	Value always 0	RO
5.50.12	DTE XS lane alignment status	1 = DTE XS receive lanes locked and aligned 0 = DTE XS receive lanes not locked and aligned	RO
5.50.11:0	Reserved	Value always 0	RO

[a] RO = Read only

45.2.5.11c.1 DTE XS lane alignment status (5.50.12)

When read as a one, bit 5.50.12 indicates that the DTE XS has locked and aligned all receive lanes. When read as a zero, bit 5.50.12 indicates that the DTE XS has not locked and aligned all receive lanes.

45.2.5.11d Multi-lane BASE-R DTE XS alignment status 3 register (Register 5.52)

The assignment of bits in the multi-lane BASE-R DTE XS alignment status 3 register is shown in Table 45–182d. A DTE XS device that does not implement multi-lane BASE-R DTE XS shall return a zero for all bits in the multi-lane BASE-R DTE XS alignment status 3 register. A device that implements multi-lane BASE-R DTE XS shall return a zero for all bits in the multi-lane BASE-R DTE XS alignment status 3 register that are not required for the DTE XS configuration. It is the responsibility of the STA management entity to ensure that a port type is supported by all MMDs before interrogating any of its status bits.

Table 45–182d—Multi-lane BASE-R DTE XS alignment status 3 register bit definitions

Bit(s)	Name	Description	R/W[a]
5.52.15:8	Reserved	Value always 0	RO
5.52.7	Lane 7 aligned	1 = Lane 7 alignment marker is locked 0 = Lane 7 alignment marker is not locked	RO
5.52.6	Lane 6 aligned	1 = Lane 6 alignment marker is locked 0 = Lane 6 alignment marker is not locked	RO

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Table 45–182d—Multi-lane BASE-R DTE XS alignment status 3 register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
5.52.5	Lane 5 aligned	1 = Lane 5 alignment marker is locked 0 = Lane 5 alignment marker is not locked	RO
5.52.4	Lane 4 aligned	1 = Lane 4 alignment marker is locked 0 = Lane 4 alignment marker is not locked	RO
5.52.3	Lane 3 aligned	1 = Lane 3 alignment marker is locked 0 = Lane 3 alignment marker is not locked	RO
5.52.2	Lane 2 aligned	1 = Lane 2 alignment marker is locked 0 = Lane 2 alignment marker is not locked	RO
5.52.1	Lane 1 aligned	1 = Lane 1 alignment marker is locked 0 = Lane 1 alignment marker is not locked	RO
5.52.0	Lane 0 aligned	1 = Lane 0 alignment marker is locked 0 = Lane 0 alignment marker is not locked	RO

[a] RO = Read only

45.2.5.11d.1 Lane 7 aligned (5.52.7)

When read as a one, bit 5.52.7 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 7. When read as a zero, bit 5.52.7 indicates that the DTE XS receiver lane 7 has not achieved alignment marker lock. This bit reflects the state of amps_lock[7] (see 119.2.6.2.2).

45.2.5.11d.2 Lane 6 aligned (5.52.6)

When read as a one, bit 5.52.6 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 6. When read as a zero, bit 5.52.6 indicates that the DTE XS receiver lane 6 has not achieved alignment marker lock. This bit reflects the state of amps_lock[6] (see 119.2.6.2.2).

45.2.5.11d.3 Lane 5 aligned (5.52.5)

When read as a one, bit 5.52.5 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 5. When read as a zero, bit 5.52.5 indicates that the DTE XS receiver lane 5 has not achieved alignment marker lock. This bit reflects the state of amps_lock[5] (see 119.2.6.2.2).

45.2.5.11d.4 Lane 4 aligned (5.52.4)

When read as a one, bit 5.52.4 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 4. When read as a zero, bit 5.52.4 indicates that the DTE XS receiver lane 4 has not achieved alignment marker lock. This bit reflects the state of amps_lock[4] (see 119.2.6.2.2).

45.2.5.11d.5 Lane 3 aligned (5.52.3)

When read as a one, bit 5.52.3 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 3. When read as a zero, bit 5.52.3 indicates that the DTE XS receiver lane 3 has not achieved alignment marker lock. This bit reflects the state of amps_lock[3] (see 119.2.6.2.2).

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45.2.5.11d.6 Lane 2 aligned (5.52.2)

When read as a one, bit 5.52.2 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 2. When read as a zero, bit 5.52.2 indicates that the DTE XS receiver lane 2 has not achieved alignment marker lock. This bit reflects the state of amps_lock[2] (see 119.2.6.2.2).

45.2.5.11d.7 Lane 1 aligned (5.52.1)

When read as a one, bit 5.52.1 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 1. When read as a zero, bit 5.52.1 indicates that the DTE XS receiver lane 1 has not achieved alignment marker lock. This bit reflects the state of amps_lock[1] (see 119.2.6.2.2).

45.2.5.11d.8 Lane 0 aligned (5.52.0)

When read as a one, bit 5.52.0 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 0. When read as a zero, bit 5.52.0 indicates that the DTE XS receiver lane 0 has not achieved alignment marker lock. This bit reflects the state of amps_lock[0] (see 119.2.6.2.2).

45.2.5.11e Multi-lane BASE-R DTE XS alignment status 4 register (Register 5.53)

The assignment of bits in the multi-lane BASE-R DTE XS alignment status 4 register is shown in Table 45–182e. A DTE XS device that does not implement multi-lane BASE-R DTE XS shall return a zero for all bits in the multi-lane BASE-R DTE XS alignment status 4 register. A device that implements multi-lane BASE-R DTE XS shall return a zero for all bits in the multi-lane BASE-R DTE XS alignment status 4 register that are not required for the DTE XS configuration. It is the responsibility of the STA management entity to ensure that a port type is supported by all MMDs before interrogating any of its status bits.

Table 45–182e—Multi-lane BASE-R DTE XS alignment status 4 register bit definitions

Bit(s)	Name	Description	R/W[a]
5.53.15:8	Reserved	Value always 0	RO
5.53.7	Lane 15 aligned	1 = Lane 15 alignment marker is locked 0 = Lane 15 alignment marker is not locked	RO
5.53.6	Lane 14 aligned	1 = Lane 14 alignment marker is locked 0 = Lane 14 alignment marker is not locked	RO
5.53.5	Lane 13 aligned	1 = Lane 13 alignment marker is locked 0 = Lane 13 alignment marker is not locked	RO
5.53.4	Lane 12 aligned	1 = Lane 12 alignment marker is locked 0 = Lane 12 alignment marker is not locked	RO
5.53.3	Lane 11 aligned	1 = Lane 11 alignment marker is locked 0 = Lane 11 alignment marker is not locked	RO
5.53.2	Lane 10 aligned	1 = Lane 10 alignment marker is locked 0 = Lane 10 alignment marker is not locked	RO
5.53.1	Lane 9 aligned	1 = Lane 9 alignment marker is locked 0 = Lane 9 alignment marker is not locked	RO
5.53.0	Lane 8 aligned	1 = Lane 8 alignment marker is locked 0 = Lane 8 alignment marker is not locked	RO

[a] RO = Read only

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45.2.5.11e.1 Lane 15 aligned (5.53.7)

When read as a one, bit 5.53.7 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 15. When read as a zero, bit 5.53.7 indicates that the DTE XS receiver lane 15 has not achieved alignment marker lock. This bit reflects the state of amps_lock[15] (see 119.2.6.2.2).

45.2.5.11e.2 Lane 14 aligned (5.53.6)

When read as a one, bit 5.53.6 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 14. When read as a zero, bit 5.53.6 indicates that the DTE XS receiver lane 14 has not achieved alignment marker lock. This bit reflects the state of amps_lock[14] (see 119.2.6.2.2).

45.2.5.11e.3 Lane 13 aligned (5.53.5)

When read as a one, bit 5.53.5 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 13. When read as a zero, bit 5.53.5 indicates that the DTE XS receiver lane 13 has not achieved alignment marker lock. This bit reflects the state of amps_lock[13] (see 119.2.6.2.2).

45.2.5.11e.4 Lane 12 aligned (5.53.4)

When read as a one, bit 5.53.4 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 12. When read as a zero, bit 5.53.4 indicates that the DTE XS receiver lane 12 has not achieved alignment marker lock. This bit reflects the state of amps_lock[12] (see 119.2.6.2.2).

45.2.5.11e.5 Lane 11 aligned (5.53.3)

When read as a one, bit 5.53.3 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 11. When read as a zero, bit 5.53.3 indicates that the DTE XS receiver lane 11 has not achieved alignment marker lock. This bit reflects the state of amps_lock[11] (see 119.2.6.2.2).

45.2.5.11e.6 Lane 10 aligned (5.53.2)

When read as a one, bit 5.53.2 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 10. When read as a zero, bit 5.53.2 indicates that the DTE XS receiver lane 10 has not achieved alignment marker lock. This bit reflects the state of amps_lock[10] (see 119.2.6.2.2).

45.2.5.11e.7 Lane 9 aligned (5.53.1)

When read as a one, bit 5.53.1 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 9. When read as a zero, bit 5.53.1 indicates that the DTE XS receiver lane 9 has not achieved alignment marker lock. This bit reflects the state of amps_lock[9] (see 119.2.6.2.2).

45.2.5.11e.8 Lane 8 aligned (5.53.0)

When read as a one, bit 5.53.0 indicates that the DTE XS receiver has achieved alignment marker lock for service interface lane 8. When read as a zero, bit 5.53.0 indicates that the DTE XS receiver lane 8 has not achieved alignment marker lock. This bit reflects the state of amps_lock[8] (see 119.2.6.2.2).

45.2.5.11f DTE XS lane mapping, lane 0 register (Register 5.400)

The assignment of bits in the DTE XS lane mapping, lane 0 register is shown in Table 45–182f. When the multi-lane DTE XS described in Clause 118 detects and locks the alignment marker for service interface lane 0, the detected PCS lane number is recorded in this register. The contents of the DTE XS lane mapping, lane 0 register is valid when Lane 0 aligned bit (5.52.0) is set to one and is invalid otherwise.

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Table 45–182f—DTE XS lane mapping, lane 0 register bit definitions

Bit(s)	Name	Description	R/W[a]
5.400.15:5	Reserved	Value always 0	RO
5.400.4:0	Lane 0 mapping	DTE XS lane received in service interface lane 0	RO

[a] RO = Read only

45.2.5.11g DTE XS lane mapping, lane 1 through lane 15 registers (Registers 5.401 through 5.415)

The definition of the DTE XS lane mapping, lane 1 through 15 registers is identical to that described for lane 0 in 45.2.5.11f. The lane mapping for lane 1 is in register 5.401; lane 2 is in register 5.402; etc.

45.2.5.11h DTE XS FEC symbol error counter lane 0 (Register 5.600, 5.601)

The assignment of bits in the DTE XS FEC symbol error counter lane 0 register is shown in Table 45–182g. Symbol errors detected in DTE XS FEC lane 0 are counted and shown in register 5.600.15:0 and 5.601.15:0. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 5.600, 5.601 are used to read the value of a 32-bit counter. When registers 5.600 and 5.601 are used to read the 32-bit counter value, the register 5.600 is read first, the value of the register 5.601 is latched when (and only when) register 5.600 is read, and reads of register 5.601 return the latched value rather than the current value of the counter.

Table 45–182g—DTE XS FEC symbol error counter lane 0 register bit definitions

Bit(s)	Name	Description	R/W[a]
5.600.15:0	DTE XS FEC symbol errors, lane 0 lower	FEC_symbol_error_counter_0[15:0]	RO, NR
5.601.15:0	DTE XS FEC symbol errors, lane 0 upper	FEC_symbol_error_counter_0[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

45.2.5.11i DTE XS FEC symbol error counter lane 1 through 15 (Registers 5.602 through 5.631)

The behavior of the DTE XS FEC symbol error counters, lane 1 through 15 is identical to that described for DTE XS FEC lane 0 in 45.2.5.11h. Errors detected in each DTE XS FEC lane are counted and shown in the corresponding register. DTE XS FEC lane 1, lower 16 bits are shown in register 5.602; DTE XS FEC lane 1, upper 16 bits are shown in register 5.603; DTE XS FEC lane 2, lower 16 bits are shown in register 5.604; through register 5.631 for DTE XS FEC lane 15, upper 16 bits.

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45.2.5.11j DTE XS FEC control register (Register 5.800)

The assignment of bits in the DTE XS FEC control register is shown in Table 45–182h.

Table 45–182h—DTE XS FEC control register bit definitions

Bit(s)	Name	Description	R/W[a]
5.800.15:2	Reserved	Value always 0	RO
5.800.2	DTE XS FEC degraded SER enable	1 = FEC decoder signals degraded SER 0 = FEC decoder does not signal degraded SER	R/W
5.800.1	DTE XS FEC bypass indication enable	1 = FEC decoder does not indicate errors 0 = FEC decoder indicates errors	R/W
5.800.0	Reserved	Value always 0	RO

[a] R/W = Read/Write, RO = Read only

45.2.5.11j.1 DTE XS FEC degraded SER enable (5.800.2)

This bit enables the DTE XS FEC decoder to signal the presence of a degraded SER. When set to a one, this bit enables degraded SER signaling. When set to a zero, degraded SER signaling is disabled. Writes to this bit are ignored and reads return a zero if the DTE XS FEC does not have the ability to signal the presence of a degraded SER (see 119.2.5.3 for equivalent PCS behavior).

45.2.5.11j.2 DTE XS FEC bypass indication enable (5.800.1)

This bit enables the DTE XS FEC decoder to bypass error indication to the upper layers through the sync bits. When set to a one, this bit enables bypass of the error indication. When set to a zero, errors are indicated to the upper layers through the sync bits. Writes to this bit are ignored and reads return a zero if the DTE XS FEC does not have the ability to bypass indicating decoding errors (see 119.2.5.3 for equivalent PCS behavior).

45.2.5.11k DTE XS FEC status register (Register 5.801)

The assignment of bits in the DTE XS FEC status register is shown in Table 45–182i.

Table 45–182i—DTE XS FEC status register bit definitions

Bit(s)	Name	Description	R/W[a]
5.801.15:7	Reserved	Value always 0	RO
5.801.6	Local degraded SER received	1 = local degraded SER received 0 = no local degraded SER received	RO
5.801.5	Remote degraded SER received	1 = remote degraded SER received 0 = no remote degraded SER received	RO
5.801.4	DTE XS FEC degraded SER	1 = SER is degraded 0 = SER is not degraded	RO

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Table 45–182i—DTE XS FEC status register bit definitions *(continued)*

Bit(s)	Name	Description	R/W[a]
5.801.3	DTE XS FEC degraded SER ability	1 = FEC decoder has the ability to signal the presence of a degraded SER 0 = FEC decoder does not have the ability to signal the presence of a degraded SER	RO
5.801.2	DTE XS FEC high SER	1 = FEC errors have exceeded threshold 0 = FEC errors have not exceeded threshold	RO/LH
5.801.1	DTE XS FEC bypass indication ability	1 = FEC decoder has the ability to bypass error indication 0 = FEC decoder does not have the ability to bypass error indication	RO
5.801.0	Reserved	Value always 0	RO

[a] RO = Read only, LH = Latching high

45.2.5.11k.1 Local degraded SER received (5.801.6)

When read as a one, bit 5.801.6 indicates that the local degraded SER signal has been received. This bit reflects the state of rx_local_degraded (see 118.2.1) for the DTE XS.

45.2.5.11k.2 Remote degraded SER received (5.801.5)

When read as a one, bit 5.801.5 indicates that the remote degraded SER signal has been received. This bit reflects the state of rx_rm_degraded for the DTE XS (see 119.2.6.2.2 for equivalent PCS behavior).

45.2.5.11k.3 DTE XS FEC degraded SER (5.801.4)

When DTE XS FEC degraded SER enable (bit 5.800.2) is set to one, bit 5.801.4 is set to one if the number of FEC symbol errors in a window of DTE XS FEC degraded SER interval (registers 5.810 and 5.811) codewords exceeds the DTE XS FEC degraded SER activate threshold (registers 5.806 and 5.807) and is cleared if the number of FEC symbol errors in the same window is below the DTE XS FEC degraded SER deactivate threshold (registers 5.808 and 5.809). If the number of FEC symbol errors in the window is between the two thresholds, then bit 5.801.4 remains in its previous state. The bit is set to zero if DTE XS FEC degraded SER enable (bit 5.800.2) is set to zero (see 119.2.5.3 for equivalent PCS behavior). The value of bit 5.801.4 is undefined if the value of the DTE XS FEC degraded SER activate threshold is less than the value of the DTE XS FEC degraded SER deactivate threshold.

45.2.5.11k.4 DTE XS FEC degraded SER ability (5.801.3)

The DTE XS FEC decoder may have the option to signal the presence of a degraded SER (see 119.2.5.3 for equivalent PCS behavior). This bit is set to one to indicate that the decoder has the ability to signal the presence of a degraded SER. The bit is set to zero if this ability is not supported.

45.2.5.11k.5 DTE XS FEC high SER (5.801.2)

When DTE XS FEC bypass indication enable (bit 5.800.1) is set to one, this bit is set to one if the number of FEC symbol errors in a window of 8192 codewords exceeds the threshold (see 119.2.5.3) and is set to zero otherwise. The bit is set to zero if DTE XS FEC bypass indication enable (bit 5.800.1) is set to zero. This bit shall be implemented with latching high behavior.

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45.2.5.11k.6 DTE XS FEC bypass indication ability (5.801.1)

The DTE XS FEC decoder may have the option to perform error detection without error indication (see 119.2.5.3) to reduce the delay contributed by the FEC decoder. This bit is set to one to indicate that the decoder has this ability to bypass error indication. The bit is set to zero if this ability is not supported.

45.2.5.11l DTE XS FEC corrected codewords counter (Register 5.802, 5.803)

The assignment of bits in the DTE XS FEC corrected codewords counter register is shown in Table 45–182j. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 5.802, 5.803 are used to read the value of a 32-bit counter. When registers 5.802 and 5.803 are used to read the 32-bit counter value, the register 5.802 is read first, the value of the register 5.803 is latched when (and only when) register 5.802 is read, and reads of register 5.803 return the latched value rather than the current value of the counter.

Table 45–182j—DTE XS FEC corrected codewords counter register bit definitions

Bit(s)	Name	Description	R/W[a]
5.802.15:0	FEC corrected codewords lower	FEC_corrected_cw_counter[15:0]	RO, NR
5.803.15:0	FEC corrected codewords upper	FEC_corrected_cw_counter[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

45.2.5.11m DTE XS FEC uncorrected codewords counter (Register 5.804, 5.805)

The assignment of bits in the DTE XS FEC uncorrected codewords counter register is shown in Table 45–182k. These bits shall be reset to all zeros when the register is read by the management function or upon PHY reset. These bits shall be held at all ones in the case of overflow. Registers 5.804, 5.805 are used to read the value of a 32-bit counter. When registers 5.804 and 5.805 are used to read the 32-bit counter value, the register 5.804 is read first, the value of the register 5.805 is latched when (and only when) register 5.804 is read, and reads of register 5.805 return the latched value rather than the current value of the counter.

Table 45–182k—DTE XS FEC uncorrected codewords counter register bit definitions

Bit(s)	Name	Description	R/W[a]
5.804.15:0	FEC uncorrected codewords lower	FEC_uncorrected_cw_counter[15:0]	RO, NR
5.805.15:0	FEC uncorrected codewords upper	FEC_uncorrected_cw_counter[31:16]	RO, NR

[a] RO = Read only, NR = Non Roll-over

45.2.5.11n DTE XS FEC degraded SER activate threshold register (Register 5.806, 5.807)

The assignment of bits in the DTE XS FEC degraded SER activate threshold register is shown in Table 45–182l. The value controls the threshold used to set the DTE XS FEC degraded SER bit (5.801.4) in an equivalent manner to that defined for the FEC_degraded_SER bit in 119.2.5.3.

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Table 45–182l—DTE XS FEC degraded SER activate threshold register bit definitions

Bit(s)	Name	Description	R/W[a]
5.806.15:0	DTE XS FEC degraded SER activate threshold lower	FEC_degraded_SER_activate_threshold[15:0]	R/W
5.807.15:0	DTE XS FEC degraded SER activate threshold upper	FEC_degraded_SER_activate_threshold[31:16]	R/W

[a] R/W = Read/Write

45.2.5.11o DTE XS FEC degraded SER deactivate threshold register (Register 5.808, 5.809)

The assignment of bits in the DTE XS FEC degraded SER deactivate threshold register is shown in Table 45–182m. The value controls the threshold used to clear the DTE XS FEC degraded SER bit (5.801.4) in an equivalent manner to that defined for the FEC_degraded_SER bit in 119.2.5.3.

Table 45–182m—DTE XS FEC degraded SER deactivate threshold register bit definitions

Bit(s)	Name	Description	R/W[a]
5.808.15:0	DTE XS FEC degraded SER deactivate threshold lower	FEC_degraded_SER_deactivate_threshold[15:0]	R/W
5.809.15:0	DTE XS FEC degraded SER deactivate threshold upper	FEC_degraded_SER_deactivate_threshold[31:16]	R/W

[a] R/W = Read/Write

45.2.5.11p DTE XS FEC degraded SER interval register (Register 5.810, 5.811)

The assignment of bits in the DTE XS FEC degraded SER interval register is shown in Table 45–182n. The value controls the interval used to set and clear the DTE XS FEC degraded SER bit (5.801.4) in an equivalent manner to that defined for the FEC_degraded_SER bit in 119.2.5.3.

Table 45–182n—DTE XS FEC degraded SER interval register bit definitions

Bit(s)	Name	Description	R/W[a]
5.810.15:0	DTE XS FEC degraded SER interval lower	FEC_degraded_SER_interval[15:0]	R/W
5.811.15:0	DTE XS FEC degraded SER interval upper	FEC_degraded_SER_interval[31:16]	R/W

[a] R/W = Read/Write

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78. Energy-Efficient Ethernet (EEE)

78.1 Overview

Change the third paragraph of 78.1 (as modified by IEEE Std 802.3by-2016) as follows:

EEE supports operation over twisted-pair cabling systems, twinax cable, electrical backplanes, optical fiber, the XGXS for 10 Gb/s PHYs, the 25GAUI for 25 Gb/s PHYs, the XLAUI for 40 Gb/s PHYs, ~~and~~ the CAUI-10 or CAUI-4 for 100 Gb/s PHYs, the 200GAUI-n and 200GXS for 200 Gb/s PHYs, and the 400GAUI-n and 400GXS for 400 Gb/s PHYs. Table 78–1 lists the supported PHYs and interfaces and their associated clauses.

78.1.4 PHY types optionally supporting EEE

Insert new rows at the end of Table 78-1 (as modified by IEEE Std 802.3by-2016) as follows (unchanged rows not shown):

Table 78–1—Clauses associated with each PHY or interface type

PHY or interface type	Clause
…	
200GBASE-DR4[b]	119, 120, 121
200GBASE-FR4[b]	119, 120, 122
200GBASE-LR4[b]	119, 120, 122
400GBASE-SR16[b]	119, 120, 123
400GBASE-DR4[b]	119, 120, 124
400GBASE-FR8[b]	119, 120, 122
400GBASE-LR8[b]	119, 120, 122

[b] The deep sleep mode of EEE is not supported for this PHY.

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78.5 Communication link access latency

Insert three new rows (including footnote c) at the end of Table 78-4 as follows (unchanged rows not shown):

Table 78–4—Summary of the LPI timing parameters for supported PHYs or interfaces

PHY or interface type	Case	$T_{w_sys_tx}$ (min) (μs)	T_{w_phy} (min) (μs)	$T_{phy_shrink_tx}$ (max) (μs)	$T_{phy_shrink_rx}$ (max) (μs)	$T_{w_sys_rx}$ (min) (μs)
…						
200GXS/400GXS [c]		0.34				
200GBASE-R fast wake		0.34	0.3	0	0	0.25
400GBASE-R fast wake		0.34	0.3	0	0	0.25

[c] The minimum $T_{w_sys_tx}$ of a PHY is increased by the indicated period if there is a 200GXS/400GXS in the transmit path. A PHY that includes a 200GXS/400GXS in the receive path may require an increase of $T_{w_sys_tx}$ on the link partner; this may be negotiated using LLDP (see 79.3.5).

Change the title of 78.5.1 as follows:

78.5.1 ~~10 Gb/s~~ PHY extension using extender sublayers ~~XGXS~~

Insert a new paragraph at the end of 78.5.1 as follows:

The 200GXS or 400GXS (see Clause 118) can be inserted between the RS and a 200 Gb/s or 400 Gb/s PHY, respectively, to transparently extend the physical reach of the 200GMII or 400GMII. The LPI signaling can operate through the 200GXS or 400GXS with the PHY timing parameters modified as described in Table 78–4.

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90. Ethernet support for time synchronization protocols

90.1 Introduction

Change the second paragraph of 90.1 (as modified by IEEE Std 802.3by-2016) as follows:

The TSSI is defined for the full-duplex mode of operation only. It supports MAC operation at various data rates. The MII (Clause 22), GMII (Clause 35), XGMII (Clause 46), 25GMII (Clause 106), XLGMII (Clause 81), ~~and~~ CGMII (Clause 81), 200GMII (Clause 117), and 400GMII (Clause 117) specifications are all compatible with the gRS sublayer defined in 90.5.

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Insert Clause 116 to Clause 124 as follows:

116. Introduction to 200 Gb/s and 400 Gb/s networks

116.1 Overview

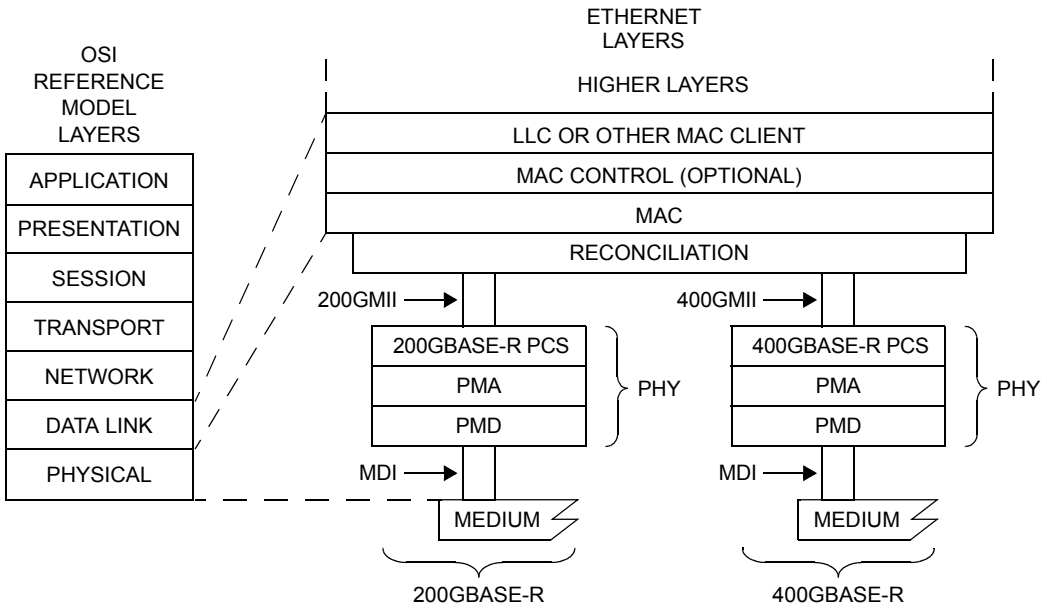
116.1.1 Scope

This clause describes the general requirements for 200 Gigabit and 400 Gigabit Ethernet.

200 Gigabit Ethernet uses the IEEE 802.3 MAC sublayer operating at a data rate of 200 Gb/s, coupled with any IEEE 802.3 200GBASE Physical Layer implementation and is defined for full duplex operation only. 400 Gigabit Ethernet uses the IEEE 802.3 MAC sublayer operating at a data rate of 400 Gb/s, coupled with any IEEE 802.3 400GBASE Physical Layer implementation and is defined for full duplex operation only.

116.1.2 Relationship of 200 Gigabit and 400 Gigabit Ethernet to the ISO OSI reference model

200 Gigabit and 400 Gigabit Ethernet couples the IEEE 802.3 MAC to a family of 200 Gb/s and 400 Gb/s Physical Layers. The relationships among 200 Gigabit and 400 Gigabit Ethernet, the IEEE 802.3 MAC, and the ISO Open System Interconnection (OSI) reference model are shown in Figure 116–1.



200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
LLC = LOGICAL LINK CONTROL
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 116–1—Architectural positioning of 200 Gigabit and 400 Gigabit Ethernet

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While this specification defines interfaces in terms of bits, octets, and frames, implementations may choose other data-path widths for implementation convenience. The only exceptions are as follows:

a) The 200GMII and 400GMII, which, when implemented as logical interconnection points between the MAC sublayer and the Physical Layer (PHY), use a 64-bit wide data path as specified in Clause 117. Physical instantiations of this interface may use other data-path widths.

b) The management interface, which, when physically implemented as the MDIO/MDC (Management Data Input/Output and Management Data Clock) at an observable interconnection port, uses a bit-wide data path as specified in Clause 45.

c) The PMA service interface, which, when physically implemented as 400GAUI-16 (400 Gb/s sixteen-lane Attachment Unit Interface) at an observable interconnection port, uses a 16-lane data path as specified in Annex 120B or Annex 120C.

d) The PMA service interface, which, when physically implemented as 200GAUI-8 (200 Gb/s eight-lane Attachment Unit Interface) or 400GAUI-8 (400 Gb/s eight-lane Attachment Unit Interface) at an observable interconnection port, uses an 8-lane data path as specified in Annex 120B, Annex 120C, Annex 120D, or Annex 120E.

e) The PMA service interface, which, when physically implemented as 200GAUI-4 (200 Gb/s four-lane Attachment Unit Interface) at an observable interconnection port, uses a 4-lane data path as specified in Annex 120D or Annex 120E.

f) The MDI as specified in Clause 123 for 400GBASE-SR16 uses a 16-lane data path.

g) The MDI as specified in Clause 122 for 400GBASE-FR8 and 400GBASE-LR8 uses an 8-lane data path.

h) The MDIs as specified in Clause 121 for 200GBASE-DR4, in Clause 122 for 200GBASE-FR4 and 200GBASE-LR4, and in Clause 124 for 400GBASE-DR4, all use a 4-lane data path.

116.1.3 Nomenclature

The nomenclature employed by the 200 Gb/s and 400 Gb/s Physical Layer is explained as follows.

The alpha-numeric prefix 200GBASE in the port type (e.g., 200GBASE-R) represents a family of Physical Layer devices operating at a speed of 200 Gb/s. The alpha-numeric prefix 400GBASE in the port type (e.g., 400GBASE-R) represents a family of Physical Layer devices operating at a speed of 400 Gb/s.

200GBASE-R represents a family of Physical Layer devices using the Physical Coding Sublayer for 200 Gb/s operation over multiple PCS lanes (see Clause 119). Physical Layer devices listed in Table 116–1 are defined for operation at 200 Gb/s.

Table 116–1—200 Gb/s PHYs

Name	Description
200GBASE-DR4	200 Gb/s PHY using 200GBASE-R encoding over four lanes of single-mode fiber, with reach up to at least 500 m (see Clause 121)
200GBASE-FR4	200 Gb/s PHY using 200GBASE-R encoding over four WDM lanes on single-mode fiber, with reach up to at least 2 km (see Clause 122)
200GBASE-LR4	200 Gb/s PHY using 200GBASE-R encoding over four WDM lanes on single-mode fiber, with reach up to at least 10 km (see Clause 122)

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400GBASE-R represents a family of Physical Layer devices using the Physical Coding Sublayer for 400 Gb/s operation over multiple PCS lanes (see Clause 119). Physical Layer devices listed in Table 116–2 are defined for operation at 400 Gb/s.

Table 116–2—400 Gb/s PHYs

Name	Description
400GBASE-SR16	400 Gb/s PHY using 400GBASE-R encoding over sixteen lanes of multimode fiber, with reach up to at least 100 m (see Clause 123)
400GBASE-DR4	400 Gb/s PHY using 400GBASE-R encoding over four lanes of single-mode fiber, with reach up to at least 500 m (see Clause 124)
400GBASE-FR8	400 Gb/s PHY using 400GBASE-R encoding over eight WDM lanes on single-mode fiber, with reach up to at least 2 km (see Clause 122)
400GBASE-LR8	400 Gb/s PHY using 400GBASE-R encoding over eight WDM lanes on single-mode fiber, with reach up to at least 10 km (see Clause 122)

116.1.4 Physical Layer signaling systems

This standard specifies a family of Physical Layer implementations. Table 116–3 and Table 116–4 specify the correlation between PHY types and clauses. Implementations conforming to one or more PHY types must meet the requirements of the corresponding clauses.

Table 116–3—PHY type and clause correlation (200GBASE optical)

PHY type	Clause[a]												
	78	117		118	119	120	120B	120C	120D	120E	121	122	122
	EEE	RS	200GMII	200GMII Extender	200GBASE-R PCS	200GBASE-R PMA	200GAUI-8 C2C	200GAUI-8 C2M	200GAUI-4 C2C	200GAUI-4 C2M	200GBASE-DR4 PMD	200GBASE-FR4 PMD	200GBASE-LR4 PMD
200GBASE-DR4	O	M	O	O	M	M	O	O	O	O	M		
200GBASE-FR4	O	M	O	O	M	M	O	O	O	O		M	
200GBASE-LR4	O	M	O	O	M	M	O	O	O	O			M

[a] O = Optional, M = Mandatory.

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Table 116–4—PHY type and clause correlation (400GBASE optical)

PHY type	Clause[a]													
	78	117		118	119	120	120B	120C	120D	120E	123	124	122	122
	EEE	RS	400GMII	400GMII Extender	400GBASE-R PCS	400GBASE-R PMA	400GAUI-16 C2C	400GAUI-16 C2M	400GAUI-8 C2C	400GAUI-8 C2M	400GBASE-SR16 PMD	400GBASE-DR4 PMD	400GBASE-FR8 PMD	400GBASE-LR8 PMD
400GBASE-SR16	O	M	O	O	M	M	O	O	O	O	M			
400GBASE-DR4	O	M	O	O	M	M	O	O	O	O		M		
400GBASE-FR8	O	M	O	O	M	M	O	O	O	O			M	
400GBASE-LR8	O	M	O	O	M	M	O	O	O	O				M

[a] O = Optional, M = Mandatory.

116.2 Summary of 200 Gigabit and 400 Gigabit Ethernet sublayers

116.2.1 Reconciliation Sublayer (RS) and Media Independent Interface

The Media Independent Interface (Clause 117) provides a logical interconnection between the MAC sublayer and Physical Layer entities (PHY). The Media Independent Interface is not intended to be physically instantiated, rather it can logically connect layers within a device.

The 200GMII supports 200 Gb/s operation and the 400GMII supports 400 Gb/s operation through its 64-bit-wide transmit and receive data paths. The Reconciliation Sublayer (RS) provides a mapping between the signals provided at the Media Independent Interface (200GMII and 400GMII) and the MAC/PLS service definition.

While the 200GMII and 400GMII are optional interfaces, they are used extensively in this standard as a basis for functional specification and provides a common service interface for the physical coding sublayer defined in Clause 119.

116.2.2 200GMII and 400GMII Extender Sublayers (200GXS and 400GXS)

The 200 Gigabit 200GMII Extender Sublayer (200GXS) is part of the 200GMII Extender (Clause 118). It is identical in function to the 200GBASE-R PCS in Clause 119 with the exceptions defined in Clause 118. The 400 Gigabit 400GMII Extender Sublayer (400GXS) is part of the 400GMII Extender (Clause 118). It is identical in function to the 400GBASE-R PCS in Clause 119 with the exceptions defined in Clause 118.

The optional 200GMII Extender (Clause 118) can be inserted between the Reconciliation Sublayer and the PHY to transparently extend the reach of the 200GMII. The optional 400GMII Extender (Clause 118) can be inserted between the Reconciliation Sublayer and the PHY to transparently extend the reach of the 400GMII.

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116.2.3 Physical Coding Sublayer (PCS)

The terms 200GBASE-R and 400GBASE-R refer to a specific family of Physical Layer implementations based upon the 64B/66B data coding method specified in Clause 119 and the PMA specifications defined in Clause 120. Clause 119 PCSs perform encoding (decoding) of data from (to) the 200GMII or 400GMII to 256B/257B code blocks, apply FEC, distribute the data to multiple lanes, and transfer the encoded data to the PMA.

The 200GBASE-R PCS has almost the same functionality as the 200GXS, and therefore may be configured as a 200GXS in order to implement part of the optional 200GMII Extender (see Clause 118). The 400GBASE-R PCS has almost the same functionality as the 400GXS, and therefore may be configured as a 400GXS in order to implement part of the optional 400GMII Extender (see Clause 118).

116.2.4 Physical Medium Attachment (PMA) sublayer

The PMA provides a medium-independent means for the PCS to support the use of a range of physical media. The 200GBASE-R and 400GBASE-R PMAs perform the mapping of transmit and receive data streams between the PCS and PMA via the PMA service interface, and the mapping and multiplexing of transmit and receive data streams between the PMA and PMD via the PMD service interface. In addition, the PMA performs retiming of the received data stream when appropriate, optionally provides data loopback at the PMA or PMD service interface, and optionally provides test pattern generation and checking.

The 200GBASE-R and 400GBASE-R PMAs are specified in Clause 120.

116.2.5 Physical Medium Dependent (PMD) sublayer

The Physical Medium Dependent sublayer is responsible for interfacing to the transmission medium. The PMD is located just above the Medium Dependent Interface (MDI). The MDI, logically subsumed within each PMD subclause, is the actual medium attachment for the various supported media.

The 200GBASE-R PMDs and their corresponding media are specified in Clause 121 and Clause 122. The 400GBASE-R PMDs and their corresponding media are specified in Clause 122 through Clause 124.

116.2.6 Management interface (MDIO/MDC)

The optional MDIO/MDC management interface (Clause 45) provides an interconnection between MDIO Manageable Devices (MMDs) and Station Management (STA) entities.

116.2.7 Management

Managed objects, attributes, and actions are defined for all 200 Gigabit and 400 Gigabit Ethernet components. These items are defined in Clause 30.

116.3 Service interface specification method and notation

The service interface specification for the 200GBASE-R and 400GBASE-R Physical Layers is as per the definition in 1.2.2. Note that the 200GBASE-R and 400GBASE-R inter-sublayer service interfaces use multiple scalar REQUEST and INDICATION primitives, to indicate the transfer of multiple independent streams of data units, as defined in 116.3.1 through 116.3.3.

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116.3.1 Inter-sublayer service interface

The inter-sublayer service interface is described in an abstract manner and does not imply any particular implementation. The inter-sublayer service interface primitives are defined as follows:

IS_UNITDATA_*i*.request
IS_UNITDATA_*i*.indication
IS_SIGNAL.indication

The IS_UNITDATA_*i*.request (where *i* = 0 to n – 1, and n is the number of streams of data units) primitive is used to define the transfer of multiple streams of data units from a sublayer to the next lower (closer to the medium) sublayer. The IS_UNITDATA_*i*.indication (where *i* = 0 to n – 1, and n is the number of streams of data units) primitive is used to define the transfer of multiple streams of data units from a sublayer to the next higher (closer to the MAC) sublayer. The IS_SIGNAL.indication primitive is used to define the transfer of signal status from a sublayer to the next higher sublayer.

116.3.2 Instances of the Inter-sublayer service interface

The inter-sublayer interface can be instantiated between different sublayers, hence a prefix notation is defined to identify a specific instance of an inter-sublayer service interface. The following prefixes are defined:

a) PMD:—for primitives issued on the interface between the PMD sublayer and the PMA sublayer called the PMD service interface.
b) PMA:—for primitives issued on the interface between the PMA sublayer and the PCS, DTE 200GXS, DTE 400GXS, or PMA sublayer above called the PMA service interface.
c) PHY XS:—for primitives issued on the interface between the PMA sublayer and the PHY 200GXS sublayer or PHY 400GXS sublayer called the PHY XS service interface.

Examples of inter-sublayer service interfaces for 200GBASE-R and 400GBASE-R with their corresponding instance names are illustrated in Figure 116–2 and Figure 116–3, respectively. For example, the primitives for one instance of the inter-sublayer service interface, named the PMD service interface, are identified as follows:

PMD:IS_UNITDATA_*i*.request
PMD:IS_UNITDATA_*i*.indication
PMD:IS_SIGNAL.indication.

Primitives for other instances, of inter-sublayer interfaces, are represented in a similar manner as described above.

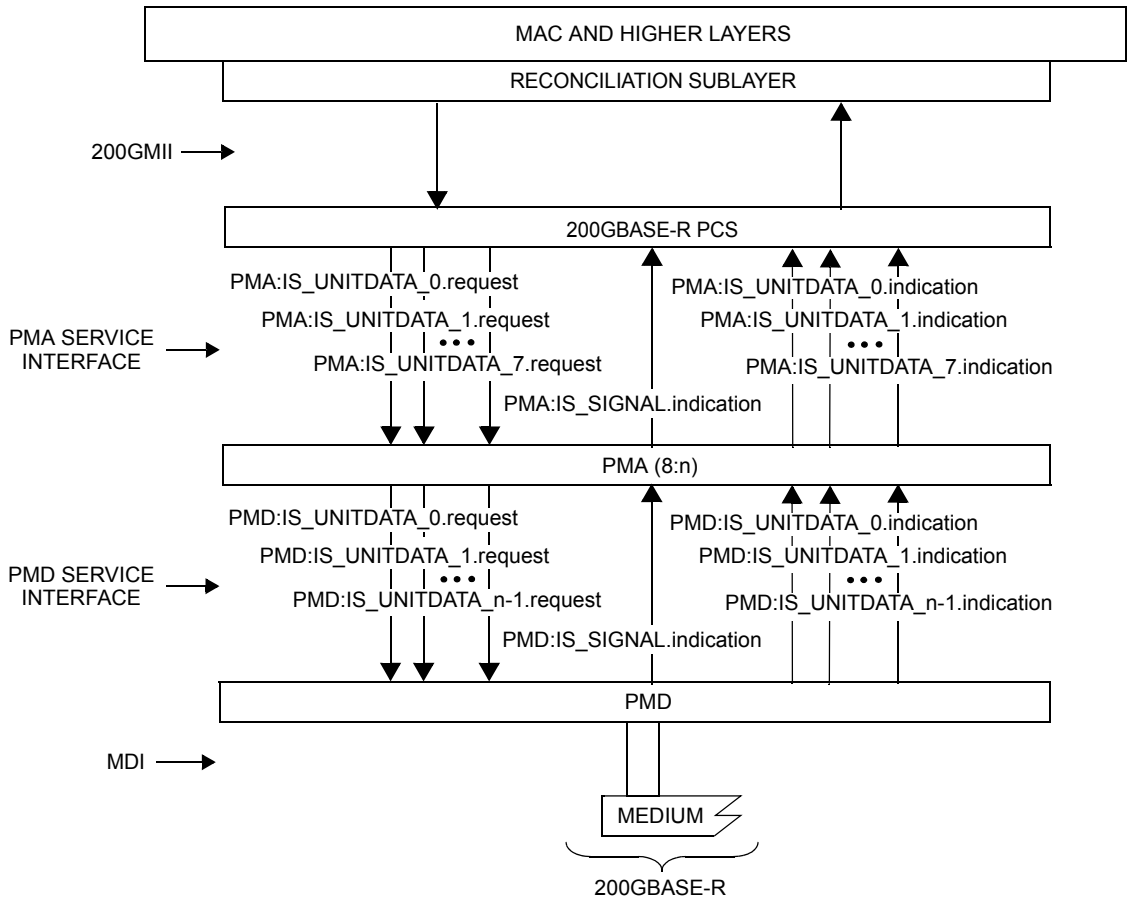
116.3.3 Semantics of inter-sublayer service interface primitives

The semantics of the inter-sublayer service interface primitives for the 200GBASE-R and 400GBASE-R sublayers are described in 116.3.3.1 through 116.3.3.3.

116.3.3.1 IS_UNITDATA_*i*.request

The IS_UNITDATA_*i*.request (where *i* = 0 to n – 1) primitive is used to define the transfer of multiple streams of data units from a sublayer to the next lower sublayer, where n is the number of parallel streams of data units.

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200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT
n = NUMBER OF PARALLEL STREAMS OF DATA UNITS

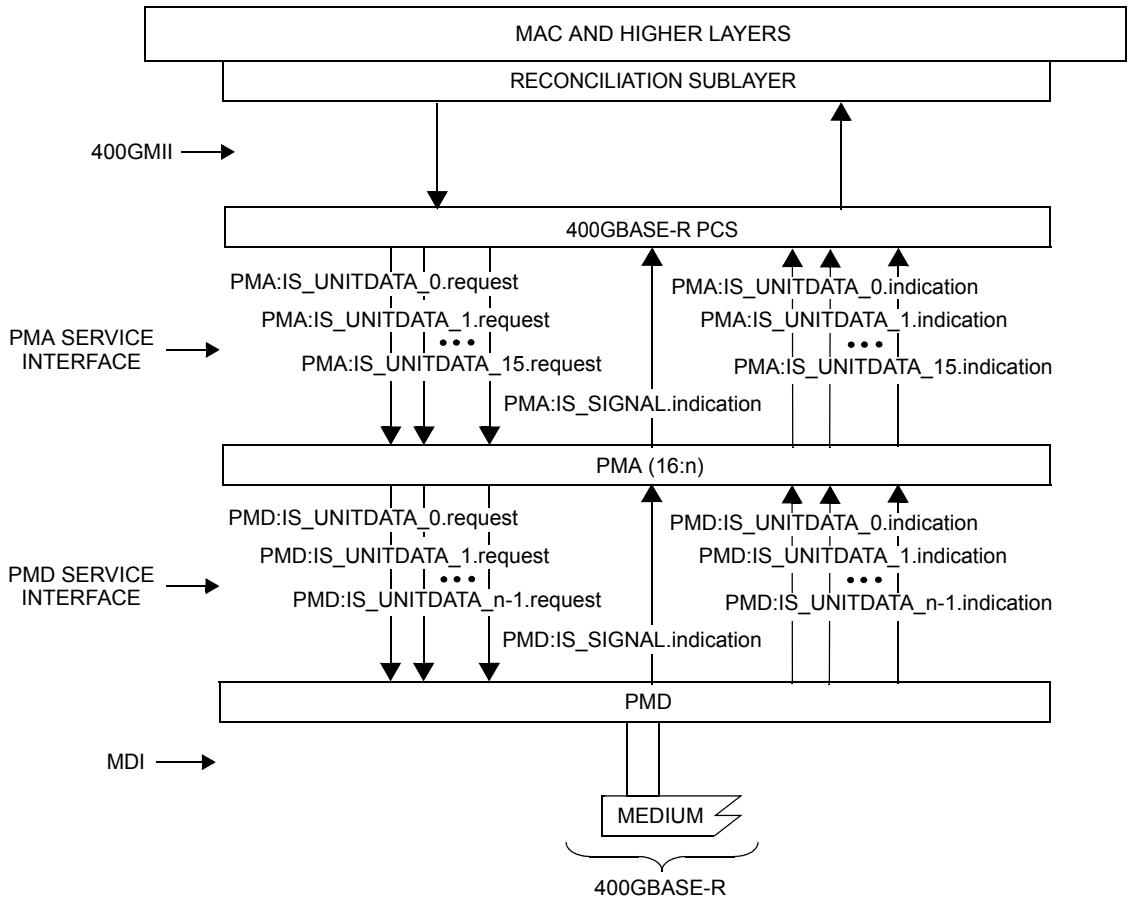
Figure 116–2—200GBASE-R inter-sublayer service interfaces

116.3.3.1.1 Semantics of the service primitive

IS_UNITDATA_0.request(tx_symbol)
IS_UNITDATA_1.request(tx_symbol)
...
IS_UNITDATA_n–1.request(tx_symbol)

The data conveyed by IS_UNITDATA_0.request to IS_UNITDATA_n–1.request consists of n parallel continuous streams of encoded symbols, one stream for each lane. Depending on the specific instance of the inter-sublayer service interface each of the tx_symbol parameters can either take one of two values: zero or one; or take one of four values: zero, one, two, or three.

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400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT
n = NUMBER OF PARALLEL STREAMS OF DATA UNITS

Figure 116–3—400GBASE-R inter-sublayer service interfaces

116.3.3.1.2 When generated

The sublayer continuously sends n parallel symbol streams IS_UNITDATA_*i*.request(tx_symbol) to the next lower sublayer, each at a nominal signaling rate defined by a specific instance of the inter-sublayer service interface.

116.3.3.1.3 Effect of receipt

The effect of receipt of this primitive is defined by the sublayer that receives this primitive.

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116.3.3.2 IS_UNITDATA_*i*.indication

The IS_UNITDATA_*i*.indication (where *i* = 0 to n – 1) primitive is used to define the transfer of multiple streams of data units from the sublayer to the next higher sublayer, where n is the number of parallel streams of data units.

116.3.3.2.1 Semantics of the service primitive

IS_UNITDATA_0.indication(rx_symbol)
IS_UNITDATA_1.indication(rx_symbol)
...
IS_UNITDATA_n–1.indication(rx_symbol)

The data conveyed by IS_UNITDATA_0.indication to IS_UNITDATA_n–1.indication consists of n parallel continuous streams of encoded symbols, one stream for each lane. Depending on the specific instance of the inter-sublayer service interface each of the rx_symbol parameters can either take one of two values: zero or one; or take one of four values: zero, one, two, or three.

116.3.3.2.2 When generated

The sublayer continuously sends n parallel symbol streams IS_UNITDATA_*i*.indication(rx_symbol) to the next higher sublayer, each at a nominal signaling rate defined by a specific instance of the inter-sublayer service interface.

116.3.3.2.3 Effect of receipt

The effect of receipt of this primitive is defined by the sublayer that receives this primitive.

116.3.3.3 IS_SIGNAL.indication

The IS_SIGNAL.indication primitive is generated by the sublayer to the next higher sublayer to indicate the status of the receive process. This primitive is generated by the receive process to propagate the detection of severe error conditions (e.g., no valid signal being received by the sublayer that generates this primitive) to the next higher sublayer.

116.3.3.3.1 Semantics of the service primitive

IS_SIGNAL.indication(SIGNAL_OK)

The SIGNAL_OK parameter can take on one of two values: OK or FAIL. A value of FAIL denotes that invalid data is being presented (rx_symbol parameters undefined) by the sublayer to the next higher sublayer. A value of OK does not guarantee valid data is being presented by the sublayer to the next higher sublayer.

116.3.3.3.2 When generated

The sublayer generates the IS_SIGNAL.indication primitive to the next higher sublayer whenever there is change in the value of the SIGNAL_OK parameter.

116.3.3.3.3 Effect of receipt

The effect of receipt of this primitive is defined by the sublayer that receives this primitive.

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116.4 Delay constraints

Predictable operation of the MAC Control PAUSE operation (Clause 31, Annex 31B) demands that there be an upper bound on the propagation delays through the network. This implies that MAC, MAC Control sublayer, and PHY implementers must conform to certain delay maxima, and that network planners and administrators conform to constraints regarding the cable topology and concatenation of devices. Table 116–5 and Table 116–6 contain the values of maximum sublayer delay (sum of transmit and receive delays at one end of the link) in bit times as specified in 1.4 and pause_quanta as specified in 31B.2 for 200 Gigabit and 400 Gigabit Ethernet, respectively. If a PHY contains an Auto-Negotiation sublayer, the delay of the Auto-Negotiation sublayer is included within the delay of the PMD and medium.

Table 116–5—Sublayer delay constraints (200GBASE)

Sublayer	Maximum (bit time)[a]	Maximum (pause_quanta)[b]	Maximum (ns)	Notes[c]
200G MAC, RS, and MAC Control	49 152	96	245.76	See 117.1.4.
200GBASE-R PCS or 200GXS[d]	160 256	313	801.28	See 119.5.
200GBASE-R PMA	18 432	36	92.16	See 120.5.4.
200GBASE-DR4 PMD	4 096	8	20.48	Includes 2 m of fiber. See 121.3.1.
200GBASE-FR4 PMD	4 096	8	20.48	Includes 2 m of fiber. See 122.3.1.
200GBASE-LR4 PMD	4 096	8	20.48	Includes 2 m of fiber. See 122.3.1.

[a] For 200GBASE-R, 1 bit time (BT) is equal to 5 ps. (See 1.4.117 for the definition of bit time.)
[b] For 200GBASE-R, 1 pause_quantum is equal to 2.56 ns. (See 31B.2 for the definition of pause_quanta.)
[c] Should there be a discrepancy between this table and the delay requirements of the relevant sublayer clause, the sublayer clause prevails.
[d] If an implementation includes the 200GMII extender, the delay associated with the 200GMII extender includes two 200GXS sublayers.

Table 116–6—Sublayer delay constraints (400GBASE)

Sublayer	Maximum (bit time)[a]	Maximum (pause_quanta)[b]	Maximum (ns)	Notes[c]
400G MAC, RS, and MAC Control	98 304	192	245.76	See 117.1.4.
400GBASE-R PCS or 400GXS[d]	320 000	625	800	See 119.5.
400GBASE-R PMA	36 864	72	92.16	See 120.5.4.
400GBASE-SR16 PMD	8 192	16	20.48	Includes 2 m of fiber. See 123.3.1.
400GBASE-DR4 PMD	8 192	16	20.48	Includes 2 m of fiber. See 124.3.1.
400GBASE-FR8 PMD	8 192	16	20.48	Includes 2 m of fiber. See 122.3.1.
400GBASE-LR8 PMD	8 192	16	20.48	Includes 2 m of fiber. See 122.3.1.

[a] For 400GBASE-R, 1 bit time (BT) is equal to 2.5 ps. (See 1.4.117 for the definition of bit time.)
[b] For 400GBASE-R, 1 pause_quantum is equal to 1.28 ns. (See 31B.2 for the definition of pause_quanta.)
[c] Should there be a discrepancy between this table and the delay requirements of the relevant sublayer clause, the sublayer clause prevails.
[d] If an implementation includes the 400GMII extender, the delay associated with the 400GMII extender includes two 400GXS sublayers.

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See 80.4 for the calculation of bit time per meter of fiber or electrical cable.

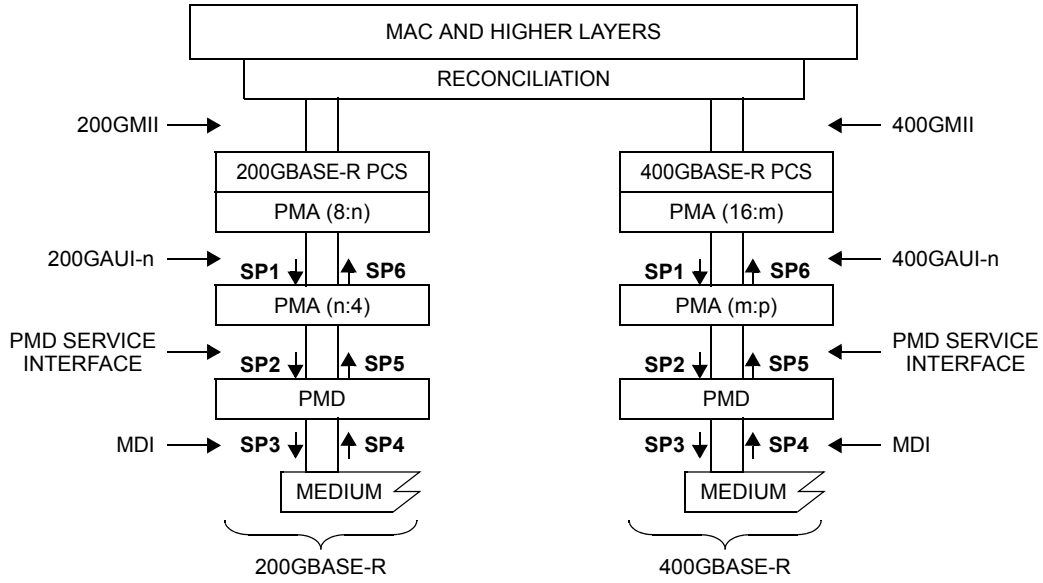
See 31B.3.7 for PAUSE reaction timing constraints for stations at operating speeds of 200 Gb/s and 400 Gb/s.

116.5 Skew constraints

Skew (or relative delay) can be introduced between lanes by both active and passive elements of a 200GBASE-R or 400GBASE-R link. Skew is defined as the difference between the times of the earliest PCS lane and latest PCS lane for the one to zero transition of the alignment marker sync bits. The PCS deskew function (see 119.2.5.1) compensates for all lane-to-lane Skew observed at the receiver. The Skew between the lanes must be kept within limits as shown in Table 116–7 so that the transmitted information on the lanes can be reassembled by the receive PCS.

Skew Variation may be introduced due to variations in electrical, thermal or environmental characteristics. Skew Variation is defined as the change in Skew between any PCS lane and any other PCS lane over the entire time that the link is in operation. From the time the link is brought up, Skew Variation is limited so that each PCS lane always traverses the same lane between any pair of adjacent sublayers while the link remains in operation.

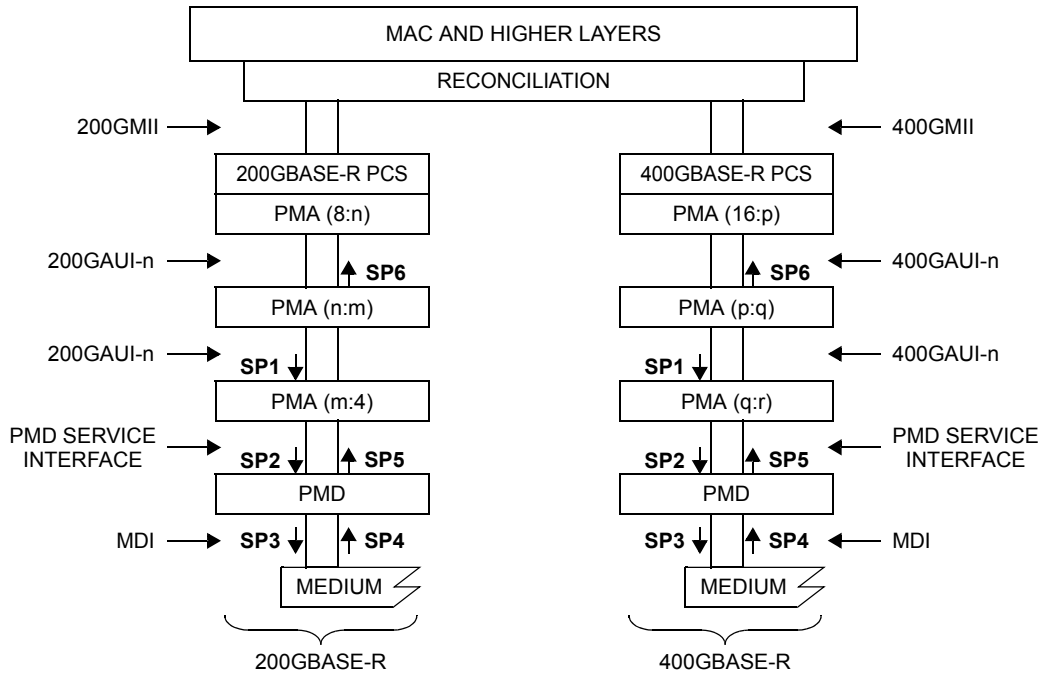
The maximum Skew and Skew Variation at physically instantiated interfaces is specified at Skew points SP1, SP2, and SP3 for the transmit direction and SP4, SP5, and SP6 for the receive direction as illustrated in Figure 116–4 (single 200GAUI-n or 400GAUI-n interface) and Figure 116–5 (multiple 200GAUI-n or 400GAUI-n interfaces).



200GAUI-n = 200 Gb/s ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GAUI-n = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT
n = 8 or 4
m = 16 or 8
p = 16, 8, or 4

Figure 116–4—200GBASE-R and 400GBASE-R Skew points for single 200GAUI-n or 400GAUI-n

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200GAUI-n = 200 Gb/s ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GAUI-n = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT
n = 8 or 4
m = 8 or 4
p = 16 or 8
q = 16 or 8
r = 16, 8, or 4

Figure 116–5—200GBASE-R and 400GBASE-R Skew points for multiple 200GAUI-n or 400GAUI-n

In the transmit direction, the Skew points are defined in the following locations:

— SP1 on the 200GAUI-n/400GAUI-n interface, at the input of the PMA closest to the PMD;
— SP2 on the PMD service interface, at the input of the PMD;
— SP3 at the output of the PMD, at the MDI.

In the receive direction, the Skew points are defined in the following locations:

— SP4 at the MDI, at the input of the PMD;
— SP5 on the PMD service interface, at the output of the PMD;
— SP6 on the 200GAUI-n/400GAUI-n interface, at the output of the PMA closest to the 200GBASE-R/400GBASE-R PCS or DTE 200GXS/400GXS.

The allowable limits for Skew are shown in Table 116–7 and the allowable limits for Skew Variation are shown in Table 116–8.

The Skew requirements for the PCS, PMA and PMD sublayers are specified in the respective clauses as noted in Table 116–7 and Table 116–8.

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Table 116–7—Summary of Skew constraints

Skew points	Maximum Skew (ns)[a]	Maximum Skew for 200GBASE-R or 400GBASE-R PCS lane (UI)[b]	Notes[c]
SP1	29	» 770	See 120.5.3.1
SP2	43	» 1142	See 120.5.3.3, 121.3.2, 122.3.2, 123.3.2, or 124.3.2
SP3	54	» 1434	See 121.3.2, 122.3.2, 123.3.2, or 124.3.2
SP4	134	» 3559	See 121.3.2, 122.3.2, 123.3.2, or 124.3.2
SP5	145	» 3852	See 121.3.2, 122.3.2, 123.3.2, or 124.3.2
SP6	160	» 4250	See 120.5.3.5
At PCS receive	180	» 4781	See 119.2.5.1

[a] The Skew limit includes 1 ns allowance for PCB traces that are associated with the Skew points.
[b] The symbol ≈ indicates approximate equivalent of maximum Skew in UI based on 1 UI equals 37.64706 ps at PCS lane signaling rate of 26.5625 GBd.
[c] Should there be a discrepancy between this table and the Skew requirements of the relevant sublayer clause, the sublayer clause prevails.

Table 116–8—Summary of Skew Variation constraints

Skew points	Maximum Skew Variation (ns)	Maximum Skew Variation for 26.5625 GBd PMD lane (UI)[a]	Maximum Skew Variation for 53.125 GBd PMD lane (UI)[b]	Notes[c]
SP1	0.2	» 5	N/A	See 120.5.3.1
SP2	0.4	»11	» 21	See 120.5.3.3, 121.3.2, 122.3.2, 123.3.2, or 124.3.2
SP3	0.6	» 16	» 32	See 121.3.2, 122.3.2, 123.3.2, or 124.3.2
SP4	3.4	» 90	» 181	See 121.3.2, 122.3.2, 123.3.2, or 124.3.2
SP5	3.6	» 96	» 191	See 121.3.2, 122.3.2, 123.3.2, or 124.3.2
SP6	3.8	» 101	N/A	See 120.5.3.5
At PCS receive	4	» 106	N/A	See 119.2.5.1

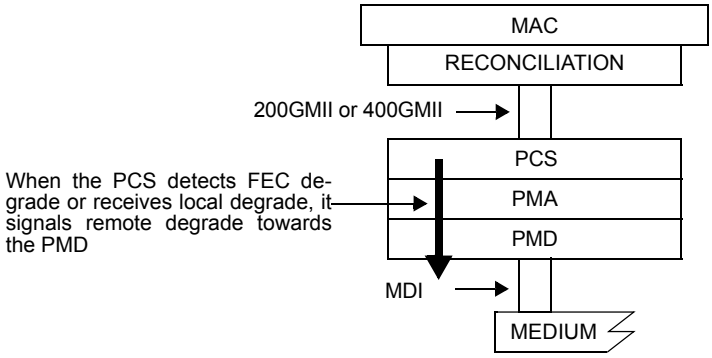
[a] The symbol ≈ indicates approximate equivalent of maximum Skew Variation in UI based on 1 UI equals 37.64706 ps at PMD lane signaling rate of 26.5625 GBd.
[b] The symbol ≈ indicates approximate equivalent of maximum Skew Variation in UI based on 1 UI equals 18.82353 ps at PMD lane signaling rate of 53.125 GBd.
[c] Should there be a discrepancy between this table and the Skew requirements of the relevant sublayer clause, the sublayer clause prevails.

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116.6 FEC Degrade

FEC degrade is an optional feature allowing for the detection of a non-service affecting link degradation condition based on exceeding a threshold for FEC corrected errors. If there are multiple FEC decoders in a given direction of transmission between the MAC sublayers at each end of the link, a Local Degrade condition is cascaded in that direction of transmission to convey the fact that one or more FEC decoders in the path have exceeded their threshold of FEC corrected errors. If any FEC decoder in a given direction of transmission exceeds its provisioned threshold for FEC corrected errors, a Remote Degrade condition is indicated in the opposite direction of transmission from the PCS or XS closest to the MAC.

Figure 116–6 illustrates the signaling of the Remote Degraded condition in the case that there is no Clause 118 extender sublayer present between the MAC and the PCS. Note that the PCS will not initiate the signaling for local degrade in this configuration as there are no additional FEC decoders in the receive direction between the PCS and the MAC.



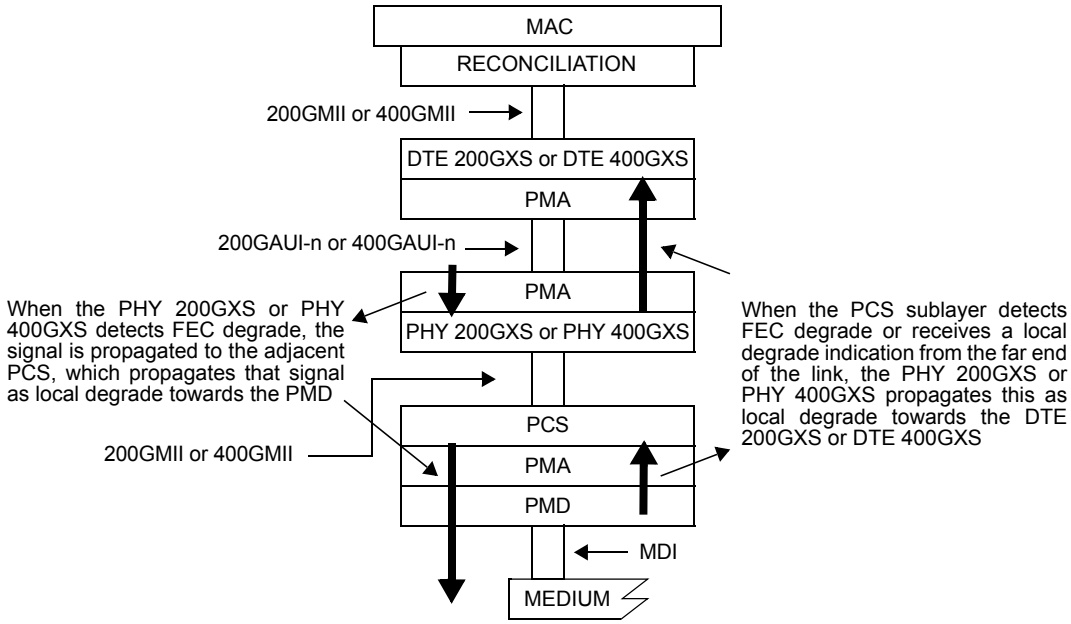
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 116–6—Remote Degrade signaling without Extender Sublayer

Figure 116–7 illustrates the signaling of the Local Degraded condition in the case that a Clause 118 extender sublayer is present between the MAC and the PCS.

Figure 116–8 illustrates the signaling of the Remote Degraded condition in the case that a Clause 118 extender sublayer is present between the MAC and the PCS.

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200GAUI-n = 200 Gb/s n-LANE ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
200GXS = 200GMII EXTENDER SUBLAYER
400GAUI-n = 400 Gb/s n-LANE ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
400GXS = 400GMII EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 116–7—Local Degrade Signaling with Extender Sublayer

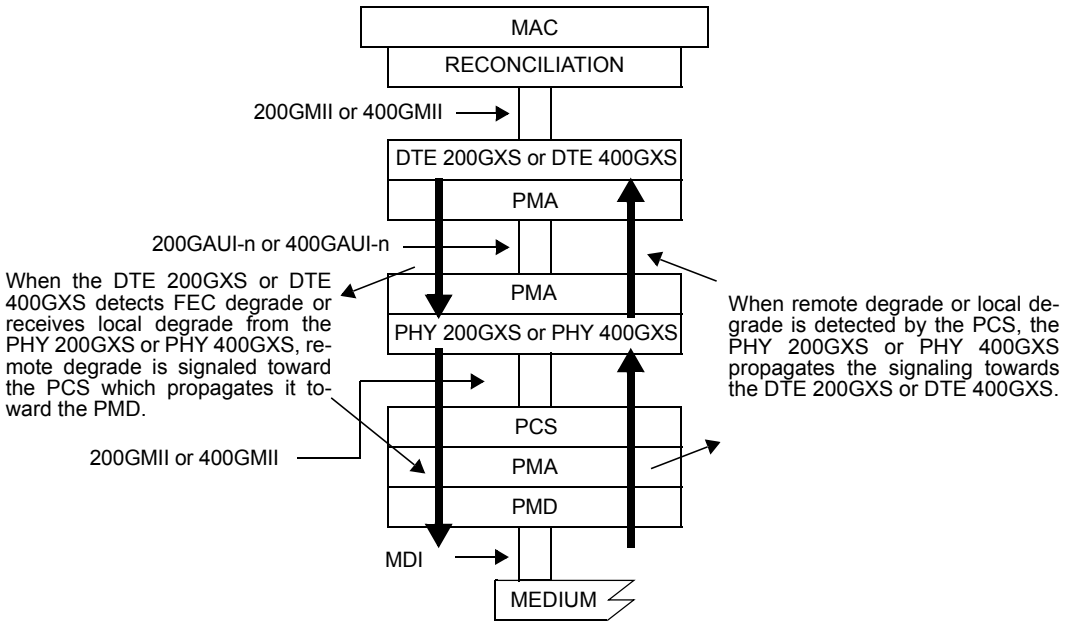
116.7 State diagrams

State diagrams take precedence over text.

The conventions of 1.2 are adopted, along with the extensions listed in 21.5.

Multiple states of a function that have a transition to a common state utilizing different qualifiers (for example, multiple exit conditions to an IDLE or WAIT state) may be indicated by a shared arrow. An exit transition arrow must connect to the shared arrow, and the qualifier must be met prior to termination of the transition arrow on the shared arrow. The shared arrow has no qualifier.

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200GAUI-n = 200 Gb/s n-LANE ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
200GXS = 200GMII EXTENDER SUBLAYER
400GAUI-n = 400 Gb/s n-LANE ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
400GXS = 400GMII EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 116–8—Remote Degrade signaling with Extender Sublayer

116.8 Protocol implementation conformance statement (PICS) proforma

The supplier of a protocol implementation that is claimed to conform to any part of IEEE Std 802.3, Clause 117 through Clause 124, and related annexes demonstrates compliance by completing a protocol implementation conformance statement (PICS) proforma.

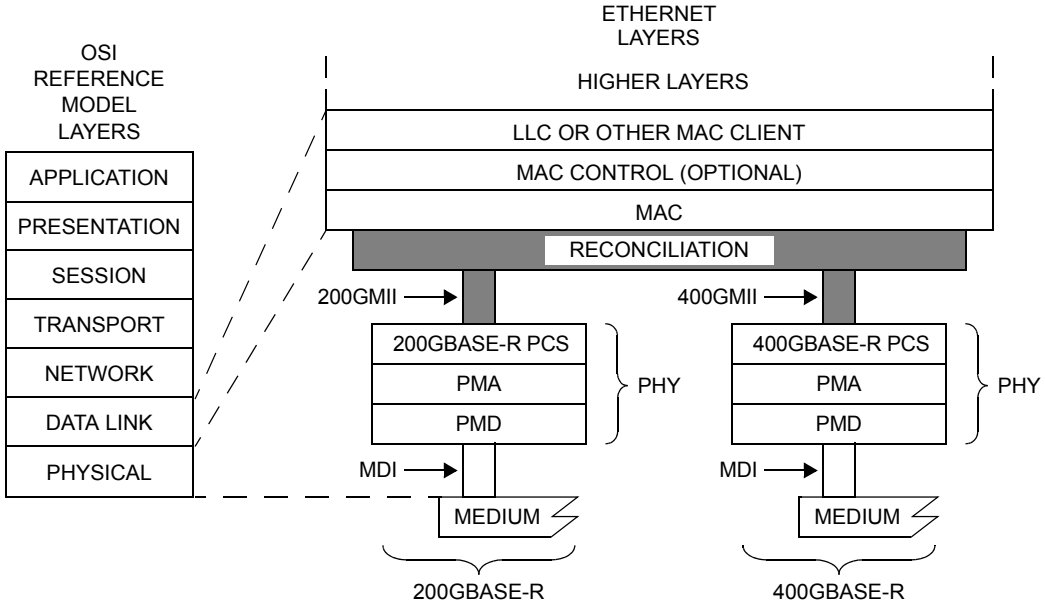
A completed PICS proforma is the PICS for the implementation in question. The PICS is a statement of which capabilities and options of the protocol have been implemented. A PICS is included at the end of each clause as appropriate. Each of the 200 Gigabit and 400 Gigabit Ethernet PICS conforms to the same notation and conventions used in 21.6.

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117. Reconciliation Sublayer (RS) and Media Independent Interface for 200 Gb/s and 400 Gb/s operation (200GMII and 400GMII)

117.1 Overview

This clause defines the characteristics of the Reconciliation Sublayer (RS) and the Media Independent Interface between Ethernet media access controllers and various PHYs. Figure 117–1 shows the relationship of the RS and Media Independent Interface to the ISO/IEC OSI reference model.



200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
LLC = LOGICAL LINK CONTROL
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 117–1—RS and MII relationship to the ISO/IEC Open Systems Interconnection (OSI) reference model and IEEE 802.3 Ethernet model

The 200GMII and the 400GMII are optional logical interfaces between the MAC sublayer and the Physical Layer (PHY). The 200GXS/400GXS sublayer in conjunction with the 200GAUI-n/400GAUI-n interface may be used to optionally extend the 200GMII/400GMII.

The RS adapts the bit serial protocols of the MAC to the parallel format of the PCS service interface. Though the 200GMII and 400GMII are optional interfaces, they are used in this standard as a basis for specification. The Physical Coding Sublayer (PCS) is specified to the 200GMII/400GMII, so if not implemented, a conforming implementation shall behave functionally as if the RS and 200GMII/400GMII were implemented.

The 200GMII/400GMII have the following characteristics:

a) The 200GMII supports a speed of 200 Gb/s.
b) The 400GMII supports a speed of 400 Gb/s.
c) Data and delimiters are synchronous to a clock reference.

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d) They provide independent 64-bit wide transmit and receive data paths.

e) They supports full duplex operation only.

117.1.1 Summary of major concepts

The following are the major concepts of the 200GMII/400GMII:

a) The 200GMII/400GMII are functionally similar to other media independent interfaces that have been defined for lower speeds, as they all define an interface allowing independent development of MAC and PHY logic.

b) The RS converts between the MAC serial data stream and the parallel data paths of the 200GMII/400GMII.

c) The RS maps the signal set provided at the 200GMII/400GMII to the PLS service primitives provided at the MAC.

d) Each direction of data transfer is independent and serviced by data, control, and clock signals.

e) The RS generates continuous data or control characters on the transmit path and expects continuous data or control characters on the receive path.

f) The RS participates in link fault detection and reporting by monitoring the receive path for status reports that indicate an unreliable link, and generating status reports on the transmit path to report detected link faults to the DTE on the remote end of the connecting link.

g) The 200GMII/400GMII may also support Low Power Idle (LPI) signaling for PHY types supporting Energy Efficient Ethernet (EEE) (see Clause 78).

h) The 200GMII/400GMII can be extended through the use of a pair of 200GXS/400GXS sublayers (DTE XS and PHY XS) with a 200GAUI-n/400GAUI-n between them.

117.1.2 Application

This clause applies to the interface between the MAC and PHY. This logical interface is used to provide media independence so that an identical media access controller may be used with supported PHY types.

117.1.3 Rate of operation

The 200GMII is specified to support 200 Gb/s operation. The 400GMII is specified to support 400 Gb/s operation.

117.1.4 Delay constraints

The maximum cumulative MAC Control, MAC, and RS delay (sum of transmit and receive delays at one end of the link) shall meet the values specified in Table 117–1. A description of overall system delay constraints and the definitions for bit times and pause_quanta can be found in 116.4 and its references.

Table 117–1—Delay constraints

Sublayer	Maximum (bit time)	Maximum (pause_quanta)	Maximum (ns)
200 Gb/s MAC, RS, and MAC Control	49 152	96	245.76
400 Gb/s MAC, RS, and MAC Control	98 304	192	245.76

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117.1.5 Allocation of functions

The allocation of functions at the 200GMII/400GMII balances the need for media independence with interface simplicity. The 200GMII/400GMII provides media independence by separating the Data Link and Physical Layers of the OSI seven-layer reference model.

117.1.6 200GMII/400GMII structure

The 200GMII/400GMII structure is identical to the CGMII structure specified in 81.1.6.

117.1.7 Mapping of 200GMII/400GMII signals to PLS service primitives

The Reconciliation Sublayer (RS) shall map the signals provided at the 200GMII/400GMII to the PLS service primitives defined in Clause 6. The PLS service primitives provided by the RS and described here behave in exactly the same manner as defined in Clause 6. Full duplex operation only is implemented at either 200 Gb/s or 400 Gb/s, therefore, PLS service primitives supporting CSMA/CD operation are not mapped through the RS to the 200GMII/400GMII. This behavior and restrictions are the same as described in 22.7, with the details of the signaling described in 117.3. LPI_REQUEST shall not be set to ASSERT unless the attached link has been operational for at least one second (i.e., link_status = OK, according to the underlying PCS/PMA).

EEE capability requires the use of the MAC defined in Annex 4A for simplified full duplex operation (with carrier sense deferral). This provides full duplex operation but uses the carrier sense signal to defer transmission until the system returns from its low power state.

Mappings for the following primitives are defined for 200 Gb/s and 400 Gb/s operation:

PLS_DATA.request
PLS_DATA.indication
PLS_CARRIER.indication
PLS_SIGNAL.indication
PLS_DATA_VALID.indication

The RS maps all primitives in an identical manner as the CGMII does and as specified in 81.1.7.

117.2 200GMII/400GMII data stream

The 200GMII/400GMII data stream has the same characteristics as the CGMII data stream described in 81.2.

117.3 200GMII/400GMII functional specifications

The 200GMII/400GMII functions identically to the CGMII specified in 81.3.

117.4 LPI Assertion and Detection

LPI assertion and detection function identically to the CGMII specified in 81.4, with the single exception that the PMA stop signaling described in 81.4.4 is not applicable.

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117.5 Protocol implementation conformance statement (PICS) proforma for Clause 117, Reconciliation Sublayer (RS) and Media Independent Interface for 200 Gb/s and 400 Gb/s operation (200GMII and 400GMII)[2]

117.5.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Clause 117, Reconciliation Sublayer (RS) and Media Independent Interface for 200 Gb/s and 400 Gb/s operation (200GMII and 400GMII), shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

117.5.2 Identification

117.5.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

117.5.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Clause 117, Reconciliation Sublayer (RS) and Media Independent Interface for 200 Gb/s and 400 Gb/s operation (200GMII and 400GMII)
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[2]*Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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117.5.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
*MII	PHY support of either 200GMII or 400GMII	117.2, 117.3		O	Yes [ ] No [ ]
*LPI	Implementation of LPI	117.1.7		O	Yes [ ] No [ ]

117.5.4 PICS proforma tables for Reconciliation Sublayer (RS) and Media Independent Interface for 200 Gb/s and 400 Gb/s operation (200GMII and 400GMII)

117.5.4.1 General

Item	Feature	Subclause	Value/Comment	Status	Support
G1	Cumulative MAC Control, MAC and RS delay	117.1.4	Per Table 117–1	M	Yes [ ]
G3	Lane structure	117.1.6	Per Table 81-2	M	Yes [ ]

117.5.4.2 Mapping of PLS service primitives

Item	Feature	Subclause	Value/Comment	Status	Support
PL1	Mapping to Clause 6	117.1.7	RS implements mapping to Clause 6 PLS service primitives	MII:M	Yes [ ] N/A [ ]
PL2	Mapping of PLS_DATA.requests	117.1.7	In sequence TXD<0> to TXD<63>	MII:M	Yes [ ] N/A [ ]
PL3	Start control character creation	117.1.7	First octet of preamble converted to Start control character	MII:M	Yes [ ] N/A [ ]
PL4	TXD and TXC generation	117.1.7	For each 64 PLS_DATA.requests	MII:M	Yes [ ] N/A [ ]
PL5	Terminate control character creation	117.1.7	DATA_COMPLETE causes creation of Terminate control character in next lane in sequence	MII:M	Yes [ ] N/A [ ]
PL6	Mapping RXD to PLS_DATA.indications	117.1.7	Create PLS_DATA.indications in sequence from RXD<0> to RXD<63>	MII:M	Yes [ ] N/A [ ]
PL7	PLS_DATA.indication generation	117.1.7	Generate 64 PLS_DATA.indications for each RXD<63:0> until Terminate then generating 0, 8, 16, 24, 32, 40, 48, or 56	MII:M	Yes [ ] N/A [ ]
PL8	Start control character conversion	117.1.7	Convert valid Start control character to preamble before generating PLS_DATA.indications	MII:M	Yes [ ] N/A [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
PL9	Terminate control character	117.1.7	No PLS_DATA.indications generated	MII:M	Yes [ ] N/A [ ]
PL10	PLS_DATA_VALID.indication generation	117.1.7	On change of value of DATA_VALID_STATUS	MII:M	Yes [ ] N/A [ ]
PL11	DATA_VALID_STATUS	117.1.7	Value of DATA_VALID on a lane 0 Start control character preceded by eight idles, a Sequence ordered set, or a Terminate character	MII:M	Yes [ ] N/A [ ]
PL12	DATA_VALID_STATUS	117.1.7	Value of DATA_NOT_VALID on any control character but Error	MII:M	Yes [ ] N/A [ ]
PL13	Frame not ending with Terminate control character	117.1.7	Ensure MAC detects CRC error	MII:M	Yes [ ] N/A [ ]

117.5.4.3 Data stream structure

Item	Feature	Subclause	Value/Comment	Status	Support
DS1	Frame transfer	117.2	Within 200GMII/400GMII data stream	MII:M	Yes [ ] N/A [ ]
DS2	Bit mapping	117.2	Per Figure 81–4	MII:M	Yes [ ] N/A [ ]
DS3	Content of <data>	117.2	Consist of data octets	MII:M	Yes [ ] N/A [ ]
DS4	Recognition of <efd>	117.2	Terminate recognized in any lane	MII:M	Yes [ ] N/A [ ]

117.5.4.4 200GMII/400GMII signal functional specifications

Item	Feature	Subclause	Value/Comment	Status	Support
FS1	TX_CLK active edge	117.3	TXD and TXC sampled on the rising edge of TX_CLK	MII:M	Yes [ ] N/A [ ]
FS2	TX_CLK frequency	117.3	One-sixty-fourth of the MAC transmit data rate	MII:M	Yes [ ] N/A [ ]
FS3	TXC assertion and de-assertion	117.3	De-asserted for data, asserted for control character	MII:M	Yes [ ] N/A [ ]
FS4	TXC clock	117.3	Synchronous to TX_CLK	MII:M	Yes [ ] N/A [ ]
FS5	TXD encoding	117.3	Per Table 81–3	MII:M	Yes [ ] N/A [ ]
FS6	TXD clock	117.3	Synchronous to TX_CLK	MII:M	Yes [ ] N/A [ ]
FS7	Start alignment	117.3	Start control character aligned to lane 0	MII:M	Yes [ ] N/A [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
FS8	RX_CLK active edge	117.3	RXD and RXC sampled on the rising edge of RX_CLK	MII:M	Yes [ ] N/A [ ]
FS9	RX_CLK frequency	117.3	One-sixty-fourth of the MAC receive data rate	MII:M	Yes [ ] N/A [ ]
FS10	Loss of receive signal	117.3	Source RX_CLK from nominal clock	MII:M	Yes [ ] N/A [ ]
FS11	RXC assertion and de-assertion	117.3	De-asserted for data, asserted for control character	MII:M	Yes [ ] N/A [ ]
FS12	RXC clock	117.3	Synchronous to RX_CLK	MII:M	Yes [ ] N/A [ ]
FS13	RXD decoding	117.3	Per Table 81–4	MII:M	Yes [ ] N/A [ ]
FS14	RXD clock	117.3	Synchronous to RX_CLK	MII:M	Yes [ ] N/A [ ]
FS15	Received Error control character	117.3	RS cause MAC FrameCheckError	MII:M	Yes [ ] N/A [ ]
FS16	DATA_VALID assertion	117.3	RS not assert DATA_VALID unless Start control character in lane 0	MII:M	Yes [ ] N/A [ ]

117.5.4.5 Link fault signaling state diagram

Item	Feature	Subclause	Value/Comment	Status	Support
LF1	Link fault signaling state diagram	117.3	Implement per Figure 81–11	MII:M	Yes [ ] N/A [ ]
LF2	link_fault = OK and MAC frames	117.3	RS services MAC frame transmission requests	MII:M	Yes [ ] N/A [ ]
LF3	link_fault = OK and no MAC frames	117.3	In absence of MAC frames, RS transmits Idle control characters	MII:M	Yes [ ] N/A [ ]
LF4	link_fault = Local Fault	117.3	RS transmits continuous Remote Fault Sequence ordered sets	MII:M	Yes [ ] N/A [ ]
LF5	link_fault = Remote Fault	117.3	RS transmits continuous Idle control characters	MII:M	Yes [ ] N/A [ ]

117.5.4.6 LPI functions

Item	Feature	Subclause	Value/Comment	Status	Support
L1	Assertion of LPI in Tx direction	117.3	As defined in Table 81–3	LPI:M	Yes [ ] N/A [ ]
L2	Assertion of LPI in Rx direction	117.3	As defined in Table 81–4	LPI:M	Yes [ ] N/A [ ]

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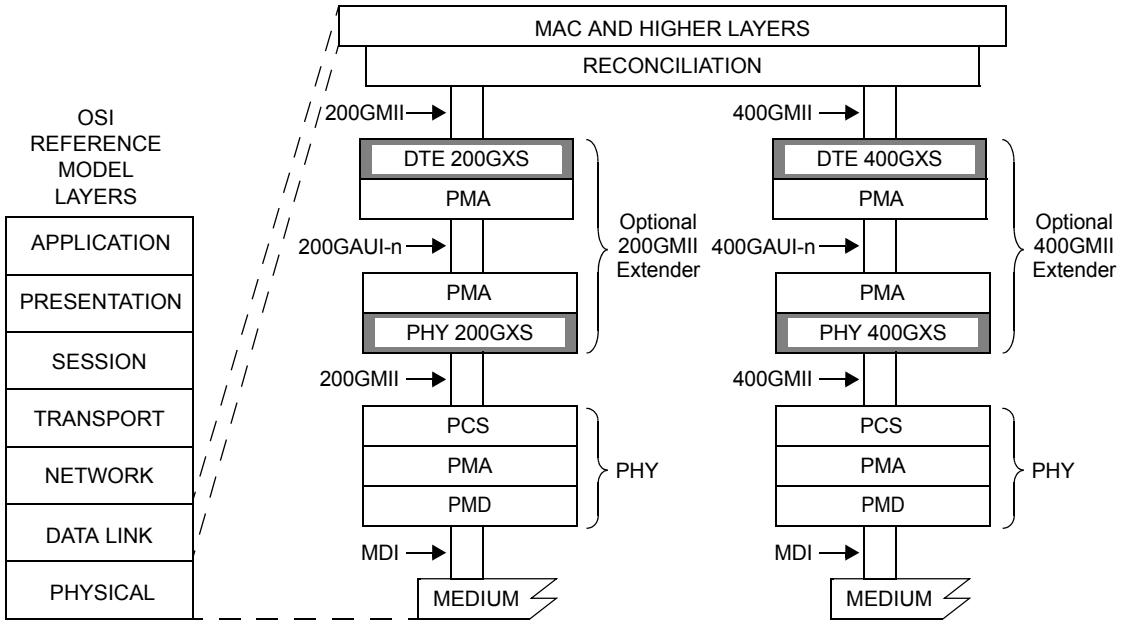
118. 200GMII Extender, 400GMII Extender, 200GMII Extender Sublayer (200GXS), and 400GMII Extender Sublayer (400GXS)

118.1 Overview

This clause defines the functional characteristics for the optional 200GMII Extender and 200GMII Extender Sublayer (200GXS), and also for the optional 400GMII Extender and 400GMII Extender Sublayer (400GXS). Figure 118–1 shows the relationship of the 200GMII/400GMII Extender and 200GXS/400GXS sublayer with other sublayers to the ISO Open System Interconnection (OSI) reference model.

The 200GMII/400GMII Extender allows the extension of the 200GMII/400GMII to the PCS via a physical instantiation. The 200GMII/400GMII Extender is composed of a 200GXS/400GXS at the RS end, a 200GXS/400GXS at the PHY end with a physical instantiation of 200GAUI-n/400GAUI-n between two adjacent PMA sublayers.

A 200GMII/400GMII Extender with the optional Energy-Efficient Ethernet (EEE) capability (see Clause 78) encodes and decodes Low Power Idle (LPI) signals. The assertion of LPI at the 200GMII/400GMII is encoded in the transmitted symbols. Detection of LPI encoding in the received symbols is indicated as LPI at the 200GMII/400GMII.



200GAUI-n = 200 Gb/s n-LANE ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
200GXS = 200GMII EXTENDER SUBLAYER
400GAUI-n = 400 Gb/s n-LANE ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
400GXS = 400GMII EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 118–1—200GXS and 400GXS relationship to the ISO/IEC Open System Interconnection (OSI) reference model and the IEEE 802.3 Ethernet model

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118.1.1 Summary of major concepts

The following is a list of the major concepts of the 200GMII/400GMII Extender:

a) Simple signal mapping to the 200GMII/400GMII
b) The optional 200GMII/400GMII Extender can be inserted between the Reconciliation Sublayer and the PHY to transparently extend the reach of the 200GMII/400GMII
c) Independent transmit and receive data paths
d) The 200GXS/400GXS leverages all functions in the Clause 119 PCS and supports physical instantiations of the 200GAUI-n/400GAUI-n
e) Optionally extends LPI signaling to the PHY for EEE

118.1.2 200GXS/400GXS Sublayer

The 200GXS, if implemented, shall be identical in function to the 200GBASE-R PCS in Clause 119 with the addition of the functions defined in 118.2. A single device may be configured as either a 200GXS or the 200GBASE-R PCS and may be managed through different optional management registers.

The 400GXS, if implemented, shall be identical in function to the 400GBASE-R PCS in Clause 119 with the addition of the functions defined in 118.2. A single device may be configured as either a 400GXS or the 400GBASE-R PCS and may be managed through different optional management registers.

118.1.3 200GAUI-n/400GAUI-n

A 200GMII Extender may use any of the following physical instantiations of the 200GAUI-n:

— 200GAUI-8 chip-to-chip (Annex 120B)
— 200GAUI-8 chip-to-module (Annex 120C)
— 200GAUI-4 chip-to-chip (Annex 120D)
— 200GAUI-4 chip-to-module (Annex 120E)

A 400GMII Extender may use any of the following physical instantiations of the 400GAUI-n:

— 400GAUI-16 chip-to-chip (Annex 120B)
— 400GAUI-16 chip-to-module (Annex 120C)
— 400GAUI-8 chip-to-chip (Annex 120D)
— 400GAUI-8 chip-to-module (Annex 120E)

118.2 FEC Degrade

The propagation of FEC degrade signaling across PCS and XS sublayers is described in 116.6 and is based on the optional FEC degrade signaling described in Clause 119 with the changes described for the DTE XS in 118.2.1 and for the PHY XS in 118.2.2.

118.2.1 DTE XS FEC Degrade signaling

The variable tx_am_sf is set as follows:

tx_am_sf<2:0> = {FEC_degraded_SER + rx_local_degraded,0,0}

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118.2.2 PHY XS FEC Degrade signaling

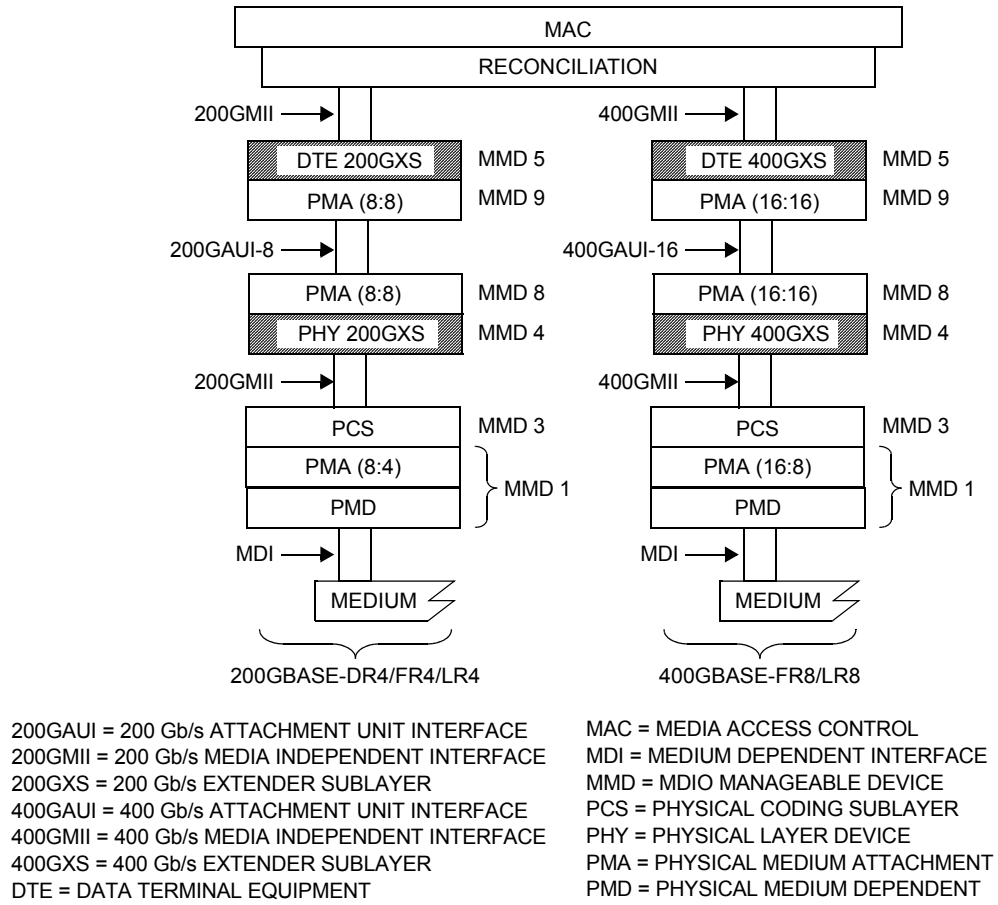
The variable tx_am_sf is set as follows:

tx_am_sf<2:0> = {PCS:rx_rm_degraded, PCS:FEC_degraded_SER + PCS:rx_local_degraded, 0}

Where PCS:rx_rm_degraded, PCS:FEC_degraded_SER, and PCS:rx_local_degraded are the rx_rm_degraded, FEC_degraded_SER, and rx_local_degraded variables from the adjacent PCS.

118.3 200GXS and 400GXS partitioning example

A partitioning example and MMD numbering using the 200GXS and 400GXS is shown in Figure 118–2.



200GAUI = 200 Gb/s ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
200GXS = 200 Gb/s EXTENDER SUBLAYER
400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
400GXS = 400 Gb/s EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 118–2—Example 200GBASE-DR4/FR4/LR4 and 400GBASE-FR8/LR8 PMA layering with 200GXS and 400GXS

118.4 200GXS and 400GXS MDIO function mapping

The optional MDIO capability described in Clause 45 defines several registers that provide control and status information for and about the 200GXS or 400GXS. If MDIO is implemented, it shall map MDIO PHY XS and DTE XS control bits to Clause 119 control variables as shown in Table 118–1 and Table 118–3, respectively. Similarly, if MDIO is implemented, it shall map MDIO PHY XS and DTE XS status bits to Clause 119 status variables as shown in Table 118–2 and Table 118–4, respectively.

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Table 118–1—MDIO PHY XS to Clause 119 control variable mapping

MDIO control variable	PHY XS register name	Register/ bit number	Clause 119 control variable
Reset	PHY XS control 1 register	4.0.15	reset
Loopback	PHY XS control 1 register	4.0.14	Loopback
Transmit test-pattern enable	BASE-R PHY XS test-pattern control register	4.42.3	tx_test_mode
PHY XS FEC bypass indication enable	PHY XS FEC control register	4.800.1	FEC_bypass_indication_enable
PHY XS FEC degraded SER enable	PHY XS FEC control register	4.800.2	FEC_degraded_SER_enable
PHY XS FEC degraded SER activate threshold	PHY XS FEC degraded SER activate threshold register	4.806, 4.807	FEC_degraded_SER_activate_threshold
PHY XS FEC degraded SER deactivate threshold	PHY XS FEC degraded SER deactivate threshold register	4.808, 4.809	FEC_degraded_SER_deactivate_threshold
PHY XS FEC degraded SER interval	PHY XS FEC degraded SER interval	4.810, 4.811	FEC_degraded_SER_interval

Table 118–2—MDIO PHY XS to Clause 119 status variable mapping

MDIO status variable	PHY XS register name	Register/ bit number	Clause 119 status variable
BASE-R PHY XS receive link status	BASE-R PHY XS status 1	4.32.12	PCS_status
Lane *x* aligned	Multi-lane BASE-R PHY XS alignment status 3 and 4	4.52.7:0 4.53.7:0	am_lock<*x*>
PHY XS lane alignment status	Multi-lane BASE-R PHY XS alignment status 1	4.50.12	align_status
Lane *x* mapping	PHY XS lane mapping, lane 0 through lane 15	4.400 through 4.415	pcs_lane_mapping<*x*>
PHY XS FEC bypass indication ability	PHY XS FEC status	4.801.1	FEC_bypass_indication_ability
FEC corrected codewords	PHY XS FEC corrected codewords counter	4.802, 4.803	FEC_corrected_cw_counter
FEC uncorrected codewords	PHY XS FEC uncorrected codewords counter	4.804, 4.805	FEC_uncorrected_cw_counter
PHY XS FEC symbol errors, lane 0 to lane 15	PHY XS FEC symbol error counter, lane 0 to lane 15	4.600 to 4.631	FEC_symbol_error_counter_*i*
Tx LPI indication	PHY XS status 1	4.1.9	Tx LPI indication
Tx LPI received	PHY XS status 1	4.1.11	Tx LPI received
Rx LPI indication	PHY XS status 1	4.1.8	Rx LPI indication
Rx LPI received	PHY XS status 1	4.1.10	Rx LPI received
EEE wake error counter	EEE wake error counter	4.22	Wake_error_counter

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Table 118–2—MDIO PHY XS to Clause 119 status variable mapping *(continued)*

MDIO status variable	PHY XS register name	Register/ bit number	Clause 119 status variable
PHY XS FEC degraded SER ability	PHY XS FEC status register	4.801.3	FEC_degraded_SER_ability
PHY XS FEC degraded SER	PHY XS FEC status register	4.801.4	FEC_degraded_SER
Remote degraded SER received	PHY XS FEC status register	4.801.5	rx_rm_degraded

Table 118–3—MDIO DTE XS to Clause 119 control variable mapping

MDIO control variable	DTE XS register name	Register/ bit number	Clause 119 control variable
Reset	DTE XS control 1 register	5.0.15	reset
Loopback	DTE XS control 1 register	5.0.14	Loopback
Transmit test-pattern enable	BASE-R DTE XS test-pattern control register	5.42.3	tx_test_mode
DTE XS FEC bypass indication enable	DTE XS FEC control register	5.800.1	FEC_bypass_indication_enable
DTE XS FEC degraded SER enable	DTE XS FEC control register	5.800.2	FEC_degraded_SER_enable
DTE XS FEC degraded SER activate threshold	DTE XS FEC degraded SER activate threshold register	5.806, 5.807	FEC_degraded_SER_activate_threshold
DTE XS FEC degraded SER deactivate threshold	DTE XS FEC degraded SER deactivate threshold register	5.808, 5.809	FEC_degraded_SER_deactivate_threshold
DTE XS FEC degraded SER interval	DTE XS FEC degraded SER interval	5.810, 5.811	FEC_degraded_SER_interval

Table 118–4—MDIO DTE XS to Clause 119 status variable mapping

MDIO status variable	DTE XS register name	Register/ bit number	Clause 119 status variable
BASE-R DTE XS receive link status	BASE-R DTE XS status 1	5.32.12	PCS_status
Lane *x* aligned	Multi-lane BASE-R DTE XS alignment status 3 and 4	5.52.7:0 5.53.7:0	am_lock<*x*>
DTE XS lane alignment status	Multi-lane BASE-R DTE XS alignment status 1	5.50.12	align_status
Lane *x* mapping	DTE XS lane mapping, lane 0 through lane 15	5.400 through 5.415	pcs_lane_mapping<*x*>
DTE XS FEC bypass indication ability	DTE XS FEC status	5.801.1	FEC_bypass_indication_ability
FEC corrected codewords	DTE XS FEC corrected codewords counter	5.802, 5.803	FEC_corrected_cw_counter

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Table 118–4—MDIO DTE XS to Clause 119 status variable mapping *(continued)*

MDIO status variable	DTE XS register name	Register/ bit number	Clause 119 status variable
FEC uncorrected codewords	DTE XS FEC uncorrected codewords counter	5.804, 5.805	FEC_uncorrected_cw_counter
DTE XS FEC symbol errors, lane 0 to lane 15	DTE XS FEC symbol error counter, lane 0 to lane 15	5.600 to 5.631	FEC_symbol_error_counter_*i*
Tx LPI indication	DTE XS status 1	5.1.9	Tx LPI indication
Tx LPI received	DTE XS status 1	5.1.11	Tx LPI received
Rx LPI indication	DTE XS status 1	5.1.8	Rx LPI indication
Rx LPI received	DTE XS status 1	5.1.10	Rx LPI received
EEE wake error counter	EEE wake error counter	5.22	Wake_error_counter
DTE XS FEC degraded SER ability	DTE XS FEC status register	5.801.3	FEC_degraded_SER_ability
DTE XS FEC degraded SER	DTE XS FEC status register	5.801.4	FEC_degraded_SER
Remote degraded received	DTE XS FEC status register	5.801.5	rx_rm_degraded
Local degraded received	DTE XS FEC status register	5.801.6	rx_local_degraded

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118.5 Protocol implementation conformance statement (PICS) proforma for Clause 118, 200GMII Extender, 400GMII Extender, 200GMII Extender Sublayer (200GXS), and 400GMII Extender Sublayer (400GXS)[3]

118.5.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Clause 118, 200GMII Extender, 400GMII Extender, 200GMII Extender Sublayer (200GXS), and 400GMII Extender Sublayer (400GXS), shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

118.5.2 Identification

118.5.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

118.5.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Clause 118, 200GMII Extender, 400GMII Extender, 200GMII Extender Sublayer (200GXS), and 400GMII Extender Sublayer (400GXS)
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[3]*Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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118.5.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
MII	200GMII or 400GMII logical interface	117, 118.1	Logical interface is supported	O	Yes [ ] No [ ]
*PHYXS	PHY 200GXS or PHY 400GXS	118.1		O/2	Yes [ ] No [ ]
*DTEXS	DTE 200GXS or DTE 400GXS	118.1		O/2	Yes [ ] No [ ]
*200GXS	Using the XS for 200GBASE-R	118.1.2		O.1	Yes [ ] No [ ]
*400GXS	Using the XS for 400GBASE-R	118.1.2		O.1	Yes [ ] No [ ]
*MD	MDIO	45, 118.4	Registers and interface supported	O	Yes [ ] No [ ]
*BI	Bypass indication	119.2.5.3	Capability is supported	O	Yes [ ] No [ ]
DC	Delay constraints	119.5	Conforms to delay constraints specified in 119.5	M	Yes [ ]
EEE	EEE capability	119.2.3.3	Capability is supported	O	Yes [ ] No [ ]
JTM	Supports test-pattern mode	119.2.1, 119.2.4.9		M	Yes [ ]
FDD	Support for optional FEC degraded detection	119.2.5.3	In the FEC decoder can optionally detect FEC SER degraded at a programmable threshold	O	Yes [ ] No [ ]

118.5.4 PICS proforma tables for 200GMII Extender, 400GMII Extender, 200GMII Extender Sublayer (200GXS), and 400GMII Extender Sublayer (400GXS)

118.5.4.1 Transmit function

Item	Feature	Subclause	Value/Comment	Status	Support
TF1	64B/66B to 256B/257B transcoder	119.2.4.2	tx_xcoded<256:0> constructed per 119.2.4.2	M	Yes [ ]
TF2	Transmission bit ordering	119.2.4.8	First bit transmitted is bit 0	M	Yes [ ]
TF3	Pad value	119.2.4.4	PRBS9	M	Yes [ ]
TF4	Alignment marker insertion	119.2.4.4		M	Yes [ ]
TF5	Pre-FEC distribution	119.2.4.5	Distribute the data to two FEC codewords	M	Yes [ ]
TF6	Reed-Solomon encoder	119.2.4.6	RS(544,514)	M	Yes [ ]
TF7	Symbol distribution	119.2.4.7	Distribution is based on 10b symbols	M	Yes [ ]

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118.5.4.2 Receive function

Item	Feature	Subclause	Value/Comment	Status	Support
RF1	Skew tolerance	119.2.5.1	Maximum Skew of 180 ns between PCS lanes and a maximum Skew Variation of 4 ns	M	Yes [ ]
RF2	Lane reorder and de-interleave	119.2.5.2	Order the PCS lanes according to the PCS lane number and de-interleave the FEC codewords	M	Yes [ ]
RF3	Reed-Solomon decoder	119.2.5.3	Corrects any combination of up to *t*=15 symbol errors in a codeword	M	Yes [ ]
RF4	Reed-Solomon decoder	119.2.5.3	Capable of indicating when a codeword was not corrected.	M	Yes [ ]
RF5	Error monitoring while error indication is bypassed	119.2.5.3	When the number of symbol errors in a block of 8192 codewords exceeds 5560 assert hi_ser for 60 ms to 75 ms	BI:M	Yes [ ] N/A [ ]
RF6	256B/257B to 64B/66B transcoder	119.2.5.7	rx_coded_*j*<65:0>, *j*=0 to 3 constructed per 119.2.5.7	M	Yes [ ]

118.5.4.3 64B/66B coding rules

Item	Feature	Subclause	Value/Comment	Status	Support
C1	Encoder (and ENCODE function) implements the code as specified	119.2.3, 119.2.6.2.3		M	Yes [ ]
C2	Decoder (and DECODE function) implements the code as specified	119.2.3, 119.2.6.2.3		M	Yes [ ]
C3	Only valid block types are transmitted	119.2.3.2		M	Yes [ ]
C4	Invalid block types are treated as an error	119.2.3.2		M	Yes [ ]
C5	Only valid control characters are transmitted	119.2.3.3		M	Yes [ ]
C6	Invalid control characters are treated as an error	119.2.3.3		M	Yes [ ]
C7	Idles do not interrupt data	119.2.3.5		M	Yes [ ]
C8	IDLE control code insertion and deletion	119.2.3.5	Insertion or Deletion in groups of 8 /I/s	M	Yes [ ]
C9	Sequence ordered set deletion	119.2.3.8	Only one whole ordered set of two consecutive sequence ordered sets may be deleted	M	Yes [ ]

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118.5.4.4 Scrambler and descrambler

Item	Feature	Subclause	Value/Comment	Status	Support
S1	Scrambler	119.2.4.3	Performs as shown in Figure 49–8	M	Yes [ ]
S2	Descrambler	119.2.5.6	Performs as shown in Figure 49–10	M	Yes [ ]

118.5.4.5 Alignment markers

Item	Feature	Subclause	Value/Comment	Status	Support
AM1	Alignment marker insertion	119.2.4.4	Alignment markers are inserted periodically as in 119.2.4.4	M	Yes [ ]
AM2	Alignment marker form	119.2.4.4	Alignment markers are formed as described in 119.2.4.4	M	Yes [ ]
AM3	Lane mapping	119.2.4.4	PCS lane number is captured	MD:M	Yes [ ] N/A [ ]
AM4	Alignment marker removal	119.2.5.5	Alignment markers are removed prior to descrambling as described in 119.2.5.5	M	Yes [ ]

118.5.5 Test-pattern modes

Item	Feature	Subclause	Value/Comment	Status	Support
JT1	Scrambled idle transmit test-pattern generator is implemented	119.2.4.9		M	Yes [ ]

118.5.6 Bit order

Item	Feature	Subclause	Value/Comment	Status	Support
B1	Transmit bit order	119.2.4.8	Placement of bits into the PCS lanes as shown in Figure 119–10 or Figure 119–11	M	Yes [ ]

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118.5.7 Management

Item	Feature	Subclause	Value/Comment	Status	Support
M1	Alternate access to XS management objects is provided	119.3		O	Yes [ ] No [ ]
M2	Mapping of MDIO control bits and MDIO status bits for PHY 200GXS or PHY 400GXS	118.4	See Table 118–1 and Table 118–2	MD*PHY XS:M	Yes [ ] N/A [ ]
M3	Mapping of MDIO control bits and MDIO status bits for DTE 200GXS or DTE 400GXS	118.4	See Table 118–3 and Table 118–4	MD*DTEX S:M	Yes [ ] N/A [ ]

118.5.7.1 State diagrams

Item	Feature	Subclause	Value/Comment	Status	Support
SM1	Alignment marker lock	119.2.6	Implements 8 alignment marker lock processes as depicted in Figure 119–12	200GXS: M	Yes [ ] N/A [ ]
SM2	Alignment marker lock	119.2.6	Implements 16 alignment marker lock processes as depicted in Figure 119–12	400GXS: M	Yes [ ] N/A [ ]
SM3	The SLIP function evaluates all possible block positions	119.2.6.2.3		M	Yes [ ]
SM4	PCS synchronization state diagram	119.2.6	Meets the requirements of Figure 119–13	M	Yes [ ]
SM5	Transmit process	119.2.6	Meets the requirements of Figure 119–14	M	Yes [ ]
SM6	Receive process	119.2.6	Meets the requirements of Figure 119–15	M	Yes [ ]

118.5.7.2 Loopback

Item	Feature	Subclause	Value/Comment	Status	Support
L1	Supports loopback	119.4		M	Yes [ ]
L2	When in loopback, transmits what it receives from the 200GMII/400GMII	119.4		M	Yes [ ]
L3	When in loopback, ignore all data presented by the PMA sublayer	119.4		M	Yes [ ]

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118.5.7.3 Delay constraints

Item	Feature	Subclause	Value/Comment	Status	Support
TIM1	PCS delay constraint	119.5	No more than 160 256 BT for sum of transmit and receive path delays for 200GBASE-R.	200GXS :M	Yes [ ] N/A [ ]
TIM2	PCS delay constraint	119.5	No more than 320 000 BT for sum of transmit and receive path delays for 400GBASE-R.	400GXS :M	Yes [ ] N/A [ ]

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119. Physical Coding Sublayer (PCS) for 64B/66B, type 200GBASE-R and 400GBASE-R

119.1 Overview

119.1.1 Scope

This clause specifies the Physical Coding Sublayer (PCS) that is common to two families of Physical Layer implementation known as 200GBASE-R and 400GBASE-R. The 200GBASE-R PCS is a sublayer of the 200 Gb/s PHYs listed in Table 116–1. The 400GBASE-R PCS is a sublayer of the 400 Gb/s PHYs listed in Table 116–2. The terms 200GBASE-R and 400GBASE-R are used when referring generally to Physical Layers using the PCS defined in this clause. Both 200GBASE-R and 400GBASE-R are based on a 64B/66B code. The 64B/66B code supports transmission of data and control characters. The 64B/66B code is then transcoded to 256B/257B encoding to reduce the overhead and make room for Forward Error Correction (FEC). The 256B/257B encoded data is then FEC encoded before being transmitted. Data distribution is introduced to support multiple lanes in the Physical Layer. Part of the distribution includes the periodic insertion of an alignment marker, which allows the receive PCS to align data from multiple lanes.

119.1.2 Relationship of 200GBASE-R and 400GBASE-R to other standards

Figure 119–1 depicts the relationship of the 200GBASE-R and 400GBASE-R sublayers (shown shaded), the Ethernet MAC and Reconciliation Sublayers, and the higher layers.

This clause borrows heavily from Clause 82 and Clause 91. 64B/66B encoding is reused with the addition of transcoding and PCS FEC.

119.1.3 Physical Coding Sublayer (PCS)

The PCS service interface is the Media Independent Interface (200GMII/400GMII), which is defined in Clause 117. The 200GMII provides a uniform interface to the Reconciliation Sublayer for all 200 Gb/s PHY implementations. The 400GMII provides a uniform interface to the Reconciliation Sublayer for all 400 Gb/s PHY implementations.

The 200GBASE-R and 400GBASE-R PCSs provide all services required by the 200GMII/400GMII, including the following:

a) Encoding (decoding) of eight 200GMII/400GMII data octets to (from) 66-bit blocks (64B/66B).
b) Transcoding from 66-bit blocks to (from) 257-bit blocks.
c) Reed-Solomon encoding (decoding) the 257-bit blocks.
d) Transferring encoded data to (from) the PMA.
e) Compensation for any rate differences caused by the insertion or deletion of alignment markers or due to any rate difference between the 200GMII/400GMII and PMA through the insertion or deletion of idle control characters.
f) Determining when a functional link has been established and informing the management entity via the MDIO when the PHY is ready for use.

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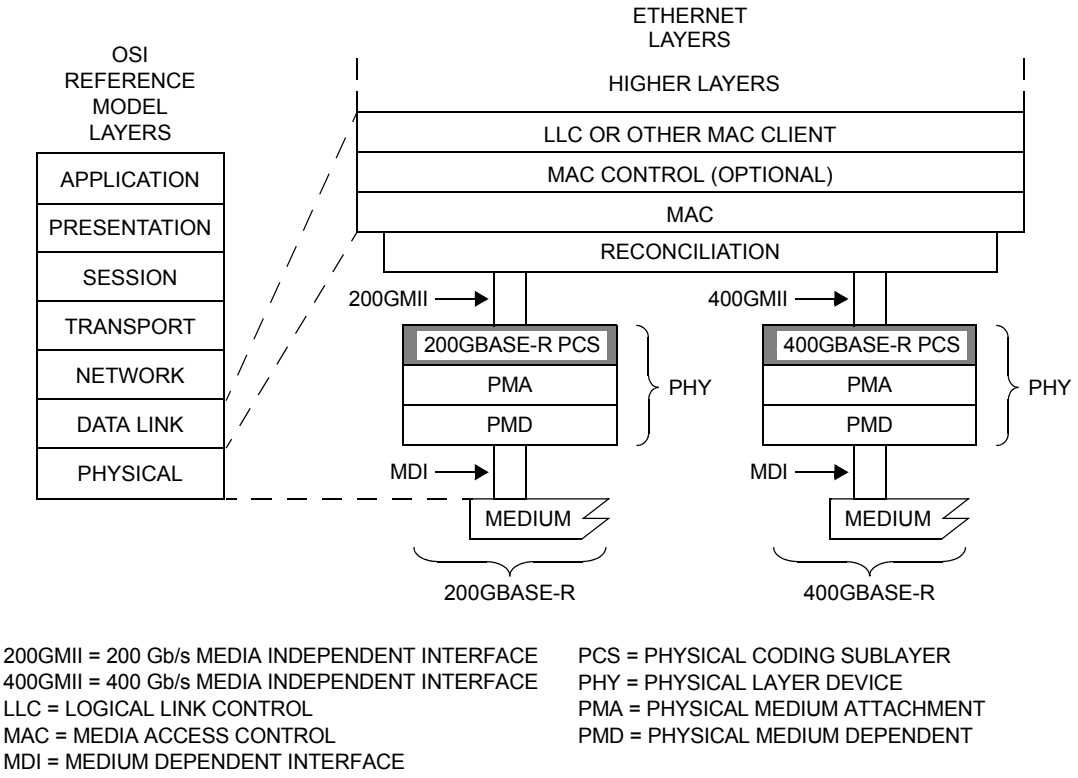


Figure 119–1—200GBASE-R and 400GBASE-R PCS relationship to the ISO/IEC Open Systems Interconnection (OSI) reference model and IEEE 802.3 Ethernet model

119.1.4 Inter-sublayer interfaces

The upper interface of the PCS may connect to the Reconciliation Sublayer through the 200GMII/400GMII. The lower interface of the PCS connects to the PMA sublayer to support a PMD. The 200GBASE-R PCS has a nominal rate at the PMA service interface of 26.5625 Gtransfers/s on each of 8 PCS lanes, which provides capacity for the MAC data rate of 200 Gb/s. The 400GBASE-R PCS has a nominal rate at the PMA service interface of 26.5625 Gtransfers/s on each of 16 PCS lanes, which provides capacity for the MAC data rate of 400 Gb/s.

It is important to note that, while this specification defines interfaces in terms of bits, octets, and frames, implementations may choose other data-path widths for implementation convenience.

119.1.4.1 PCS service interface (200GMII/400GMII)

The PCS service interface allows the 200GBASE-R or 400GBASE-R PCS to transfer information to and from a PCS client. The PCS client is the Reconciliation Sublayer. The PCS Service Interface is precisely defined as the Media Independent Interface (200GMII/400GMII) in Clause 117.

119.1.4.2 Physical Medium Attachment (PMA) service interface

The PMA service interface for the PCS is described in an abstract manner and does not imply any particular implementation. The PMA Service Interface supports the exchange of encoded data between the PCS and PMA sublayer. The PMA service interface is defined in 120.3 and is an instance of the inter-sublayer service interface definition in 116.3.

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119.1.5 Functional block diagram

Figure 119–2 provides a functional block diagram of the 200GBASE-R and 400GBASE-R PCSs.

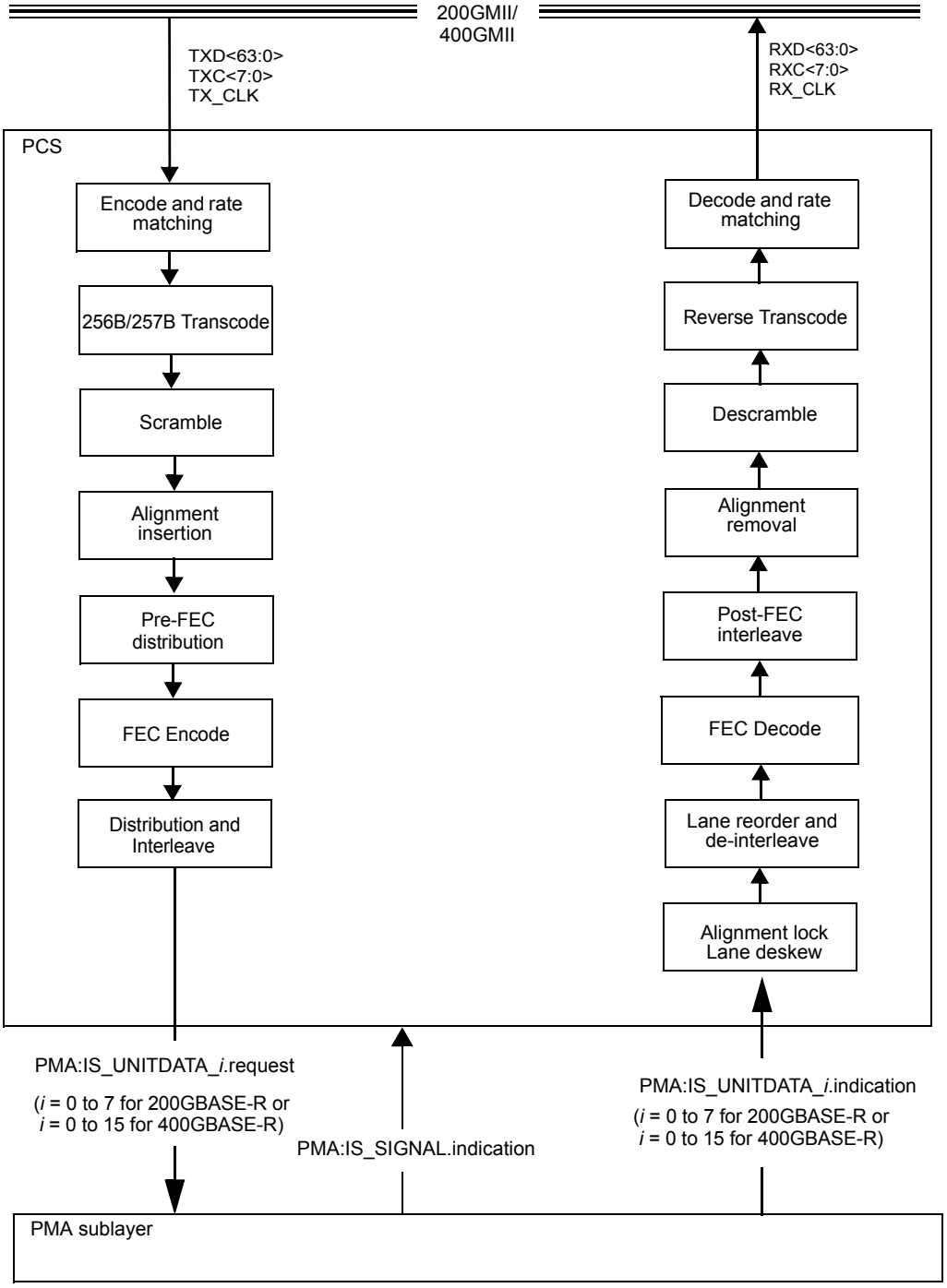


Figure 119–2—Functional block diagram

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119.2 Physical Coding Sublayer (PCS)

119.2.1 Functions within the PCS

The 200GBASE-R and 400GBASE-R PCS are composed of the PCS Transmit and PCS Receive processes. The PCS shields the Reconciliation Sublayer (and MAC) from the specific nature of the underlying channel. The PCS transmit channel can operate in normal mode or test-pattern mode.

When communicating with the 200GMII/400GMII, the PCS uses an eight octet-wide, synchronous data path, with packet delineation being provided by transmit control signals (TXC<n> = 1) and receive control signals (RXC<n> = 1). When communicating with the PMA, the 200GBASE-R PCS uses 8 encoded bit streams (also known as PCS lanes) and the 400GBASE-R PCS uses 16 encoded bit streams. Per direction (RX or TX), the PCS lanes originate from a common clock but may vary in phase and skew dynamically. The PMA sublayer operates independently of block and packet boundaries. The PCS provides the functions necessary to map packets between the 200GMII/400GMII format and the PMA service interface format.

When the transmit channel is in normal mode, the PCS Transmit process continuously generates 66-bit blocks based upon the TXD <63:0> and TXC <7:0> signals on the 200GMII/400GMII. The 66-bit blocks are transcoded to 257-bit blocks to reduce the line coding overhead. The transcoded blocks are then scrambled to control clock content and baseline wander. Alignment markers are periodically added to the data stream. The data stream is distributed to two 5140-bit blocks and then FEC encoded to control errors. The two FEC codewords are then interleaved before data is distributed to individual PCS lanes.

Transmit data-units are sent to the service interface via the PMA:IS_UNITDATA_*i*.request primitive.

When the transmit channel is in test-pattern mode, a test pattern is packed into the transmit data-units that are sent to the PMA service interface via the PMA:IS_UNITDATA_*i*.request primitive.

The PCS Synchronization process continuously monitors PMA:IS_SIGNAL.indication(SIGNAL_OK). When SIGNAL_OK indicates OK, then the PCS synchronization process accepts data-units via the PMA:IS_UNITDATA_*i*.indication primitive. It attains alignment marker lock based on the common marker (CM) portion that is periodically transmitted on every PCS lane. After alignment markers are found on all PCS lanes, the individual PCS lanes are identified using the unique marker portion (UM) and then re-ordered and deskewed. Note that a particular transmit PCS lane can be received on any receive lane of the service interface due to the skew and multiplexing that occurs in the path.

The PCS deskew process deskews, aligns and reorders the individual PCS lanes, forms a single stream, and sets the align_status flag to indicate whether the PCS has obtained alignment. The PCS then de-interleaves and processes the FEC codewords. Next the PCS re-interleaves the corrected FEC codewords on a 10-bit basis to form a single stream. The PCS then removes alignment markers, descrambles the data, transcodes the data back to 64B/66B and then decodes the 64B/66B encoded data.

The PCS shall provide transmit test-pattern mode for the scrambled idle pattern (see 119.2.4.9).

119.2.2 Use of blocks

The PCS maps 200GMII/400GMII signals into 66-bit blocks, and vice versa, using a 64B/66B coding scheme. The PCS functions ENCODE and DECODE generate, manipulate, and interpret blocks as defined in 119.2.3.

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119.2.3 64B/66B code

The PCS uses 64B/66B coding to support transmission of control and data characters. The 64B/66B codestream is then transcoded into a 256B/257B stream and FEC bits are added in this PCS before transmission.

119.2.3.1 Notation conventions

For values shown as binary, the leftmost bit is the first transmitted bit.

64B/66B encodes 8 data octets or control characters into a block. Blocks containing control characters also contain a block type field. Data octets are labeled D_0 to D_7. The control characters /I/ and /E/ are labeled C_0 to C_7. The control characters, /Q/ and /Fsig/, for ordered sets are labeled as O_0 since they are only valid on the first octet of the 200GMII/400GMII. The control character for start is labeled as S_0 for the same reason. The control character for terminate is labeled as T_0 to T_7. The four trailing zero data octets in ordered sets are labeled as Z_4 to Z_7.

One 200GMII/400GMII transfer provides eight characters that are encoded into one 66-bit transmission block. The subscript in the above labels indicates the position of the character in the eight characters from the 200GMII/400GMII transfer.

Contents of block type fields, data octets, and control characters are shown as hexadecimal values. The LSB of the hexadecimal value represents the first transmitted bit. For instance, the block type field 0x1E is sent from left to right as 01111000. The bits of a transmitted or received block are labeled TxB<65:0> and RxB<65:0>, respectively, where TxB<0> and RxB<0> represent the first transmitted bit. The value of the sync header is shown as a binary value. Binary values are shown with the first transmitted bit (the LSB) on the left.

119.2.3.2 64B/66B block structure

Blocks consist of 66 bits. The first two bits of a block are the synchronization header (sync header). Blocks are either data blocks or control blocks. The sync header is 01 for data blocks and 10 for control blocks. The remainder of the block contains the payload.

Data blocks contain eight data characters. Control blocks begin with an 8-bit block type field that indicates the format of the remainder of the block. For control blocks containing a Start, Terminate, or ordered set, that character is implied by the block type field. Other control characters are encoded in a 7-bit control code. Each control block encodes eight characters.

The format of the blocks is as shown in Figure 82-5. In the figure, the column labeled Input Data shows, in abbreviated form, the eight characters used to create the 66-bit block. These characters are either data characters or control characters and, when transferred across the 200GMII/400GMII, the corresponding TXC or RXC bit is set accordingly. Within the Input Data column, D_0 through D_7 are data octets and are transferred with the corresponding TXC or RXC bit set to zero. All other characters are control characters and are transferred with the corresponding TXC or RXC bit set to one. The single bit fields (thin rectangles with no label in the figure) are sent as zero and ignored upon receipt.

All values of block type field not listed in Figure 82-5 are invalid; they shall not be transmitted and shall be considered an error if received.

119.2.3.3 Control codes

The same set of control characters are supported by the 200GMII/400GMII and the PCS. The representations of the control characters are the control codes. The 200GMII/400GMII encodes a control

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character into an octet (an eight bit value). The PCS encodes the start and terminate control characters implicitly by the block type field. The PCS encodes the ordered set control codes using the block type field. The PCS encodes each of the other control characters into a 7-bit control code.

The control characters and their mappings to 200GBASE-R/400GBASE-R control codes and 200GMII/400GMII control codes are specified in Table 82-1. All 200GMII/400GMII and 200GBASE-R/400GBASE-R control code values that do not appear in the table shall not be transmitted and shall be considered an error if received. The ability to transmit or receive Low Power Idle (LPI) is required for PHYs that support EEE (see Clause 78). If EEE has not been negotiated, LPI shall not be transmitted and shall be treated as an error if received.

119.2.3.4 Valid and invalid blocks

The valid and invalid blocks are identical to those in 82.2.3.5 with the exception that it is a 200GMII/400GMII data stream instead of XLGMII/CGMII.

119.2.3.5 Idle (/I/)

Idle and LPI control characters are identical to those in 82.2.3.6.

119.2.3.6 Start (/S/)

Start control characters are identical to those in 82.2.3.7.

119.2.3.7 Terminate (/T/)

The terminate control characteristics are identical to those in 82.2.3.8.

119.2.3.8 Ordered set (/O/)

Ordered sets are specified identically as in 82.2.3.9.

119.2.3.9 Error (/E/)

In both the 64B/66B encoder and decoder, the /E/ is generated whenever an /E/ is detected. The /E/ is also generated when invalid blocks are detected. The /E/ allows the PCS to propagate detected errors. See R_TYPE and T_TYPE function definitions in 119.2.6.2.3 for further information.

119.2.4 Transmit

119.2.4.1 Encode and rate matching

The PCS generates 66-bit blocks based upon the TXD<63:0> and TXC<7:0> signals received from the 200GMII/400GMII. One 200GMII/400GMII data transfer is encoded into one 66-bit block. If the transmit PCS spans multiple clock domains, it may also perform clock rate compensation via the deletion of idle control characters or sequence ordered sets or the insertion of idle (or LPI) control characters.

Idle (or LPI) control characters or sequence ordered sets are removed, if necessary, to accommodate the insertion of the alignment markers. See 119.2.3.5 and 119.2.3.8 for the deletion and insertion rules, and 119.2.4.4 for more details on alignment markers.

The transmit PCS generates blocks as specified in the transmit state diagram as shown in Figure 119–14. The contents of each block are contained in a vector tx_coded<65:0>, which is passed to the transcoder. tx_coded<1:0> contains the sync header and the remainder of the bits contain the block payload.

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Note—The stream of 66-bit blocks generated by this process, together with the FEC_degraded_SER and rx_local_degraded bits should be used as the reference signal for mapping to OTN.

119.2.4.2 64B/66B to 256B/257B transcoder

The transcoder constructs a 257-bit block, tx_xcoded<256:0>, from a group of four 66-bit blocks, tx_coded_*j*<65:0> where *j*=0 to 3. For each group of four 66-bit blocks, *j*=3 corresponds to the most recently received block. Bit 0 in each 66-bit block is the first bit received and corresponds to the first bit of the synchronization header.

Note—This transcoder differs from that described in 91.5.2.5, there is no scrambling of the first 5 bits since this clause scrambles the complete 257-bit blocks after transcoding.

If for all *j*=0 to 3, tx_coded_*j*<0>=0 and tx_coded_*j*<1>=1, tx_xcoded<256:0> shall be constructed as follows:

a) tx_xcoded<0> = 1

b) tx_xcoded<(64*j*+64):(64*j*+1)> = tx_coded_*j*<65:2> for *j*=0 to 3

If for all *j*=0 to 3, tx_coded_*j*<0> ≠ tx_coded_*j*<1> (valid synchronization header) and for any *j*=0 to 3, tx_coded_*j*<0>=1 and tx_coded_*j*<1>=0, tx_xcoded<256:0> shall be constructed as follows:

a1) tx_xcoded<0> = 0

b1) tx_xcoded<*j*+1> = tx_coded_*j*<1> for *j*=0 to 3

c1) Let *c* be the smallest value of *j* such that tx_coded_*c*<0>=1. In other words, tx_coded_*c* is the first 66-bit control block that was received in the current group of four blocks.

d1) Let tx_payloads<(64*j*+63):64*j*> = tx_coded_*j*<65:2> for *j*=0 to 3

e1) Omit tx_coded_*c*<9:6>, which is the second nibble (based on transmission order) of the block type field for tx_coded_*c,* from tx_xcoded per the following expressions.
tx_xcoded<(64*c*+8):5> = tx_payloads<(64*c*+3):0>
tx_xcoded<256:(64*c*+9)> = tx_payloads<255:(64*c*+8)>

If for any *j*=0 to 3, tx_coded_*j*<0> = tx_coded_*j*<1> (invalid synchronization header), tx_xcoded<256:0> shall be constructed as follows:

a2) tx_xcoded<0> = 0

b2) tx_xcoded<*j*+1> = 1 for *j*=0 to 3

c2) Let tx_payloads<(64*j*+63):64*j*> = tx_coded_*j*<65:2> for *j*=0 to 3

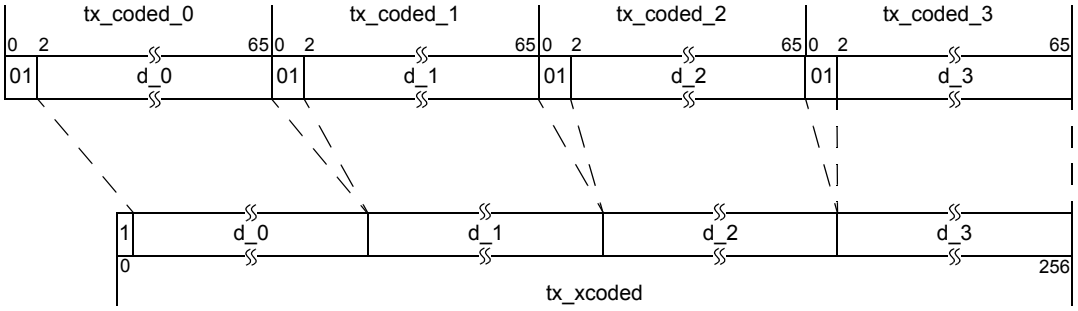
d2) Omit the second nibble (based on transmission order) of tx_coded_0 per the following expressions.
tx_xcoded<8:5> = tx_payloads<3:0>
tx_xcoded<256:9> = tx_payloads<255:8>

Several examples of the construction of tx_xcoded<256:0> are shown in Figure 119–3. In Figure 119–3, d_*j* indicates the *j*th 66-bit block contains only data octets, c_*j* indicates the *j*th 66-bit block contains one or more control characters, f_*j* denotes the first nibble of the block type field for 66-bit block *j*, and s_*j* denotes the second nibble of the block type field for 66-bit block *j*.

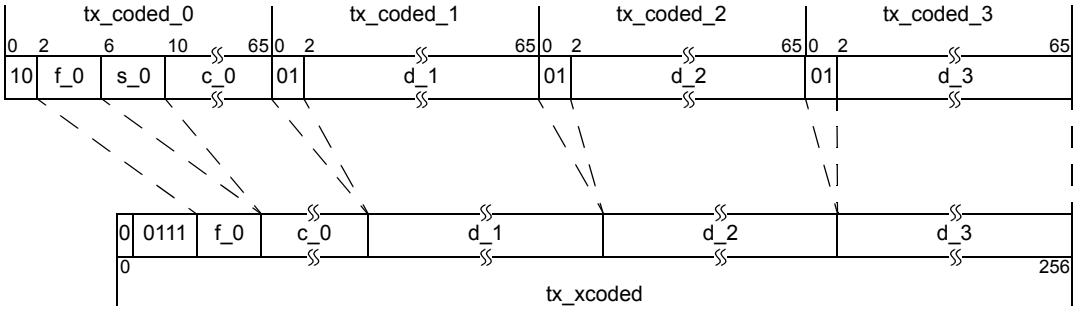
For each 257-bit block, bit 0 shall be the first bit transmitted.

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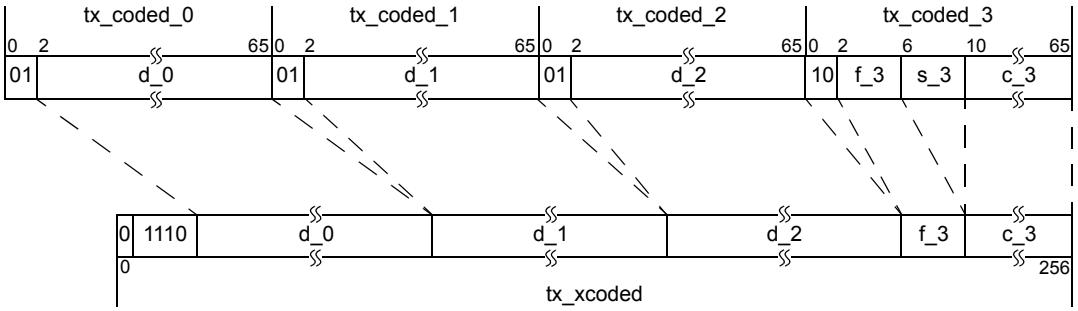
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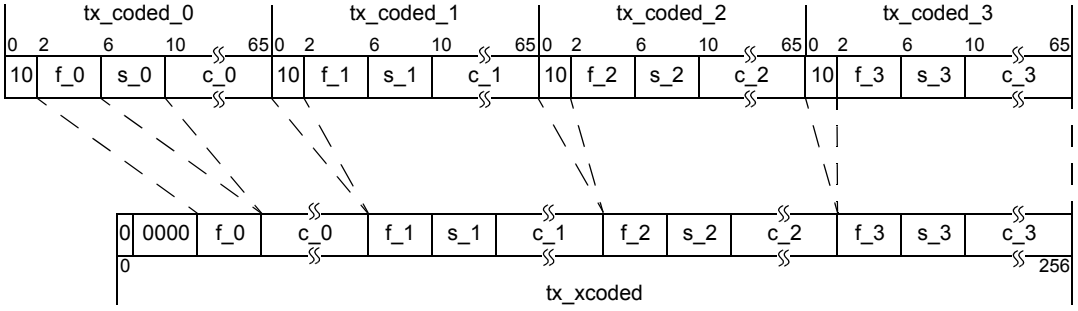
Example 1: All data blocks



Example 2: Control block followed by three data blocks



Example 3: Three data blocks followed by a control block



Example 4: All control blocks

Figure 119–3—Examples of the construction of tx_xcoded

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119.2.4.3 Scrambler

The transcoded 257-bit block, tx_xcoded<256:0>, shall be scrambled with a self-synchronizing scrambler to generate tx_scrambled<256:0>. The scrambler polynomial is identical to that in Clause 49, see Equation (49-1) for the definition of the polynomial.

119.2.4.4 Alignment marker mapping and insertion

In order to support deskew and reordering of the individual PCS lanes at the receive PCS, alignment markers corresponding to PCS lanes are periodically inserted after being processed by the alignment marker mapping function. The alignment marker mapping function compensates for the operation of the symbol distribution function and rearranges the alignment marker bits so that they appear on the PCS lanes intact and in the desired sequence. This preserves the properties of the alignment markers (e.g. DC balance, transition density) and provides a deterministic pattern for the purpose of synchronization.

For the 200GBASE-R PCS, an alignment marker group is composed of the alignment markers for all 8 PCS lanes plus an additional 65-bit pad and a 3-bit status field to yield the equivalent of four 257-bit blocks. For the 400GBASE-R PCS, an alignment marker group is composed of the alignment markers for all 16 PCS lanes plus an additional 133-bit pad and a 3-bit status field to yield the equivalent of eight 257-bit blocks. The alignment marker group is aligned to the beginning of two FEC messages, and interrupts any data transfer that is already in progress. The pad bits at the end of the alignment marker group shall be set to a free running PRBS9 pattern, defined by the polynomial $x^9 + x^5 + 1$. The initial value of the PRBS9 pattern generator may be any pattern other than all zeros.

Room for the alignment marker group is created by the transmit PCS (see 119.2.4.1). Special properties of the alignment marker group are that it is not scrambled, does not conform to the encoding rules as outlined in Figure 82-5 and is not transcoded. This is possible because the alignment marker group is added after encoding, transcoding, and scrambling, and removed before descrambling, transcoding, and 64B/66B decoding. The alignment marker group is not scrambled, which allows the receiver to directly search for and find the individual alignment markers, deskew the PCS lanes, and reassemble the aggregate stream before FEC decoding is performed. The alignment markers are formed from a known pattern that is defined to be balanced and with many transitions and therefore scrambling is not necessary.

The format of each PCS lane's alignment marker is shown in Figure 119–4. There is a portion that is common across all alignment markers (designated as CM_0 to CM_5), a unique portion per PCS lane (designated as UM_0 to UM_5), and finally a unique pad per PCS lane (designated as UP_0 to UP_2). Common synchronization logic independent of the received PCS lane number can be used with the common portion of the alignment marker. The unique pad (UP0 to UP2) within the alignment markers and the PRBS9 pad at the end of the alignment maker group are ignored on receive.

The content of the alignment markers shall be as shown in Table 119–1 for the 200GBASE-R PCS and as shown in Table 119–2 for the 400GBASE-R PCS. The contents depend on the PCS lane number and the octet number, with CM_0 through CM_5 being identical across all alignment markers to allow for common synchronization across lanes. The format shown in Table 119–1 defines how the alignment markers appear on a given PCS lane. In the FEC codewords, they appear in a permuted format due to the codeword interleaving that occurs before FEC codewords are distributed to PCS lanes.

The transmit alignment marker status field allows the local PCS to communicate the status of the FEC degraded feature to the remote PCS. If there is no extender sublayer between the PCS and the MAC, it is set as follows:

tx_am_sf<2:0> = {FEC_degraded_SER + rx_local_degraded,0,0}

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If there is a Clause 118 extender sublayer between the PCS and the MAC, it is set as follows:

tx_am_sf<2:0> = {PHY_XS:rx_rm_degraded, PHY_XS:FEC_degraded_SER, 0}

Where PHY_XS:rx_rm_degraded and PHY_XS:FEC_degraded_SER are the rx_rm_degraded and FEC_degraded_SER variables from the adjacent PHY_XS sublayer.

See 119.2.5.3 for more information on the optional FEC degrade feature.

119.2.4.4.1 AM creation for the 200GBASE-R PCS

For the 200GBASE-R PCS, the alignment marker mapping function creates a set of 8 alignment markers, and in combination with an additional 65-bit PRBS9 pad and a 3-bit status field, the PCS generates an alignment marker group. Let am_*x*<119:0> be the alignment marker for PCS lane *x*, *x*=0 to 7, where bit 0 is the first bit transmitted. The alignment markers shall be mapped to am_mapped<959:0> in a manner that yields the same result as the following process.

For *x*=0 to 7, am_*x*<119:0> is constructed as follows:

am_*x*<119:0> is set to CM_0, CM_1, CM_2, UP_0, CM_3, CM_4, CM_5, UP_1, UM_0, UM_1, UM_2, UP_2, UM_3, UM_4 and UM_5, as shown in Figure 119–4 (bits 119:0) using the values in Table 119–1 for PCS lane number *x*.

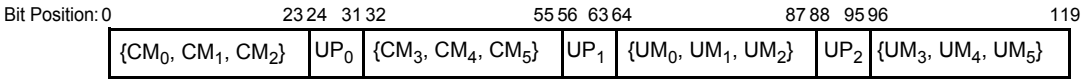


Figure 119–4—Alignment marker format

Table 119–1—200GBASE-R alignment marker encodings

PCS lane number	Encoding[a] {CM_0, CM_1, CM_2, UP_0, CM_3, CM_4, CM_5, UP_1, UM_0, UM_1, UM_2, UP_2, UM_3, UM_4, UM_5}
0	0x9A,0x4A,0x26,0x05,0x65,0xB5,0xD9,0xD6,0xB3,0xC0,0x8C,0x29,0x4C,0x3F,0x73
1	0x9A,0x4A,0x26,0x04,0x65,0xB5,0xD9,0x67,0x5A,0xDE,0x7E,0x98,0xA5,0x21,0x81
2	0x9A,0x4A,0x26,0x46,0x65,0xB5,0xD9,0xFE,0x3E,0xF3,0x56,0x01,0xC1,0x0C,0xA9
3	0x9A,0x4A,0x26,0x5A,0x65,0xB5,0xD9,0x84,0x86,0x80,0xD0,0x7B,0x79,0x7F,0x2F
4	0x9A,0x4A,0x26,0xE1,0x65,0xB5,0xD9,0x19,0x2A,0x51,0xF2,0xE6,0xD5,0xAE,0x0D
5	0x9A,0x4A,0x26,0xF2,0x65,0xB5,0xD9,0x4E,0x12,0x4F,0xD1,0xB1,0xED,0xB0,0x2E
6	0x9A,0x4A,0x26,0x3D,0x65,0xB5,0xD9,0xEE,0x42,0x9C,0xA1,0x11,0xBD,0x63,0x5E
7	0x9A,0x4A,0x26,0x22,0x65,0xB5,0xD9,0x32,0xD6,0x76,0x5B,0xCD,0x29,0x89,0xA4

[a] Each octet is transmitted LSB to MSB.

As an example, the variable am_0 is sent as (left most bit sent first, showing the first 32 bits transmitted of am_0):

01011001 01010010 01100100 10100000

The variable am_mapped is then derived from 10-bit interleaving the group of 8 alignment markers am_*x* per the following procedure:

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For all k=0 to 11
  For all j=0 to 3
    if even(k)
      am_mapped<80k+20j+9:80k+20j> = am_{2j}<10k+9:10k>
      am_mapped<80k+20j+19:80k+20j+10> = am_{2j+1}<10k+9:10k>
    else
      am_mapped<80k+20j+9:80k+20j> = am_{2j+1}<10k+9:10k>
      am_mapped<80k+20j+19:80k+20j+10> = am_{2j}<10k+9:10k>

The additional 65-bit pad is appended to variable am_mapped as follows:

am_mapped<1024:960> = PRBS9<64:0>

In this expression, PRBS9<0> is the first PRBS9 bit output of the 65-bit pad.

The 3-bit transmit alignment marker status field is then appended to the variable am_mapped as follows:

am_mapped<1027:1025> = tx_am_sf<2:0>

Alignment marker mapping is shown in Figure 119–5.

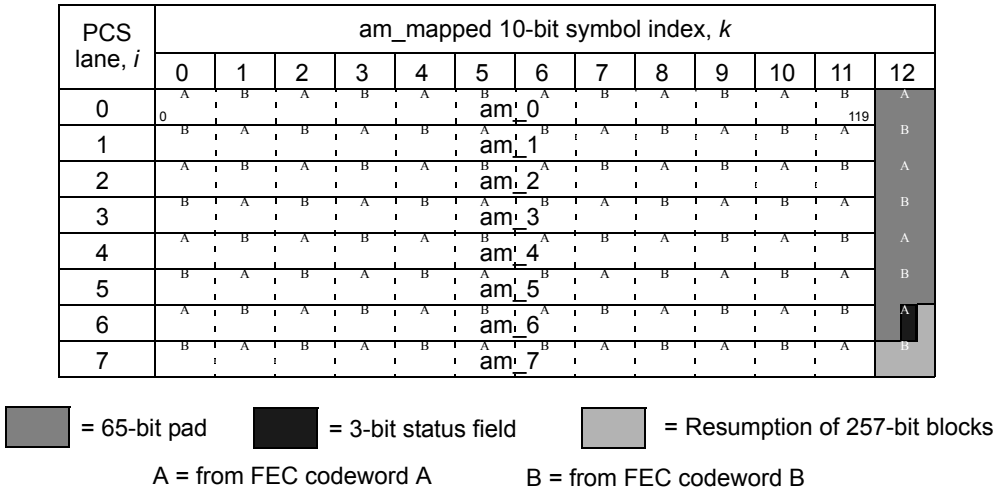


Figure 119–5—200GBASE-R alignment marker mapping to PCS lanes

The alignment marker group am_mapped<1027:0> shall be inserted so it appears in the output stream every 81 920 × 257-bit blocks. The variable tx_scrambled_am<10279:0> is constructed in one of two ways. Let the set of vectors tx_scrambled_i<256:0> represent consecutive values of tx_scrambled<256:0>.

For a 10280-bit block with an alignment marker group inserted:

tx_scrambled_am<1027:0> = am_mapped<1027:0>
For all i=0 to 35
  tx_scrambled_am<257i+1284:257i+1028> = tx_scrambled_i<256:0>

For a 10280-bit block without an alignment marker group:
For all i=0 to 39
  tx_scrambled_am<257i+256:257i> = tx_scrambled_i<256:0>

Alignment marker repetition rate is shown in Figure 119–6.

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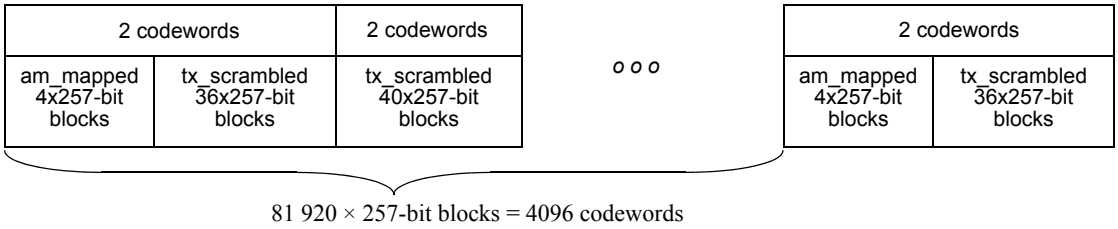


Figure 119–6—200GBASE-R alignment marker insertion period

119.2.4.4.2 AM creation for the 400GBASE-R PCS

For the 400GBASE-R PCS, the alignment marker mapping function creates a set of 16 alignment markers, and in combination with an additional 133-bit PRBS9 pad and a 3-bit status field, the PCS generates an alignment marker group.

Let am_*x*<119:0> be the alignment marker for PCS lane *x*, *x*=0 to 15, where bit 0 is the first bit transmitted. The alignment markers shall be mapped to am_mapped<1919:0> in a manner that yields the same result as the following process.

For *x*=0 to *15*, am_*x*<119:0> is constructed as follows:

am_*x*<119:0> is set to CM_0, CM_1, CM_2, UP_0, CM_3, CM_4, CM_5, UP_1, UM_0, UM_1, UM_2, UP_2, UM_3, UM_4 and UM_5, as shown in Figure 119–4 (bits 119:0) using the values in Table 119–2 for PCS lane number *x*.

Table 119–2—400GBASE-R alignment marker encodings

PCS lane number	Encoding[a] {CM_0,CM_1,CM_2,UP_0,CM_3,CM_4,CM_5,UP_1,UM_0,UM_1,UM_2,UP_2,UM_3,UM_4,UM_5}
0	0x9A,0x4A,0x26,0xB6,0x65,0xB5,0xD9,0xD9,0x01,0x71,0xF3,0x26,0xFE,0x8E,0x0C
1	0x9A,0x4A,0x26,0x04,0x65,0xB5,0xD9,0x67,0x5A,0xDE,0x7E,0x98,0xA5,0x21,0x81
2	0x9A,0x4A,0x26,0x46,0x65,0xB5,0xD9,0xFE,0x3E,0xF3,0x56,0x01,0xC1,0x0C,0xA9
3	0x9A,0x4A,0x26,0x5A,0x65,0xB5,0xD9,0x84,0x86,0x80,0xD0,0x7B,0x79,0x7F,0x2F
4	0x9A,0x4A,0x26,0xE1,0x65,0xB5,0xD9,0x19,0x2A,0x51,0xF2,0xE6,0xD5,0xAE,0x0D
5	0x9A,0x4A,0x26,0xF2,0x65,0xB5,0xD9,0x4E,0x12,0x4F,0xD1,0xB1,0xED,0xB0,0x2E
6	0x9A,0x4A,0x26,0x3D,0x65,0xB5,0xD9,0xEE,0x42,0x9C,0xA1,0x11,0xBD,0x63,0x5E
7	0x9A,0x4A,0x26,0x22,0x65,0xB5,0xD9,0x32,0xD6,0x76,0x5B,0xCD,0x29,0x89,0xA4
8	0x9A,0x4A,0x26,0x60,0x65,0xB5,0xD9,0x9F,0xE1,0x73,0x75,0x60,0x1E,0x8C,0x8A
9	0x9A,0x4A,0x26,0x6B,0x65,0xB5,0xD9,0xA2,0x71,0xC4,0x3C,0x5D,0x8E,0x3B,0xC3
10	0x9A,0x4A,0x26,0xFA,0x65,0xB5,0xD9,0x04,0x95,0xEB,0xD8,0xFB,0x6A,0x14,0x27
11	0x9A,0x4A,0x26,0x6C,0x65,0xB5,0xD9,0x71,0x22,0x66,0x38,0x8E,0xDD,0x99,0xC7
12	0x9A,0x4A,0x26,0x18,0x65,0xB5,0xD9,0x5B,0xA2,0xF6,0x95,0xA4,0x5D,0x09,0x6A
13	0x9A,0x4A,0x26,0x14,0x65,0xB5,0xD9,0xCC,0x31,0x97,0xC3,0x33,0xCE,0x68,0x3C
14	0x9A,0x4A,0x26,0xD0,0x65,0xB5,0xD9,0xB1,0xCA,0xFB,0xA6,0x4E,0x35,0x04,0x59
15	0x9A,0x4A,0x26,0xB4,0x65,0xB5,0xD9,0x56,0xA6,0xBA,0x79,0xA9,0x59,0x45,0x86

[a] Each octet is transmitted LSB to MSB.

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As an example, the variable am_0 is sent as (left most bit sent first, showing the first 32 bits transmitted of am_0):

01011001 01010010 01100100 01101101

The variable am_mapped is then derived from 10-bit interleaving the group of 16 alignment markers am_*x* per the following procedure:

For all *k*=0 to 11
  For all *j*=0 to 7
    if even(*k*)
      am_mapped<160*k*+20*j*+9:160*k*+20*j*> = am_{2*j*}<10*k*+9:10*k*>
      am_mapped<160*k*+20*j*+19:160*k*+20*j*+10> = am_{2*j*+1}<10*k*+9:10*k*>
    else
      am_mapped<160*k*+20*j*+9:160*k*+20*j*> = am_{2*j*+1}<10*k*+9:10*k*>
      am_mapped<160*k*+20*j*+19:160*k*+20*j*+10> = am_{2*j*}<10*k*+9:10*k*>

The additional 133-bit pad is appended to variable am_mapped as follows:

am_mapped<2052:1920> = PRBS9<132:0>

In this expression, PRBS9<0> is the first PRBS9 bit output of the 133-bit pad.

The 3-bit transmit alignment marker status field is then appended to the variable am_mapped as follows:

am_mapped<2055:2053> = tx_am_sf<2:0>

Alignment marker mapping is shown in Figure 119–7.

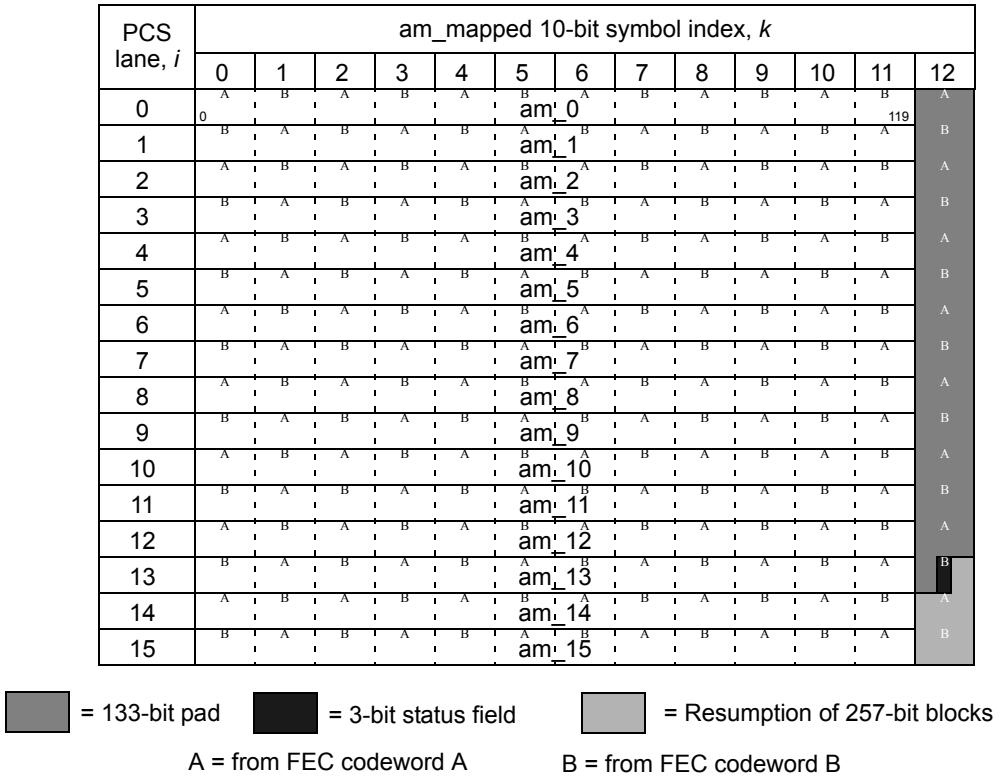


Figure 119–7—400GBASE-R alignment marker mapping to PCS lanes

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The alignment marker group am_mapped<2055:0> shall be inserted so it appears in the output stream every 163 840 × 257-bit blocks. The variable tx_scrambled_am<10279:0> is constructed in one of two ways. Let the set of vectors tx_scrambled_*i*<256:0> represent consecutive values of tx_scrambled<256:0>.

For a 10280-bit block with an alignment marker group inserted:
tx_scrambled_am<2055:0> = am_mapped<2055:0>
For all *i*=0 to 31
  tx_scrambled_am<257*i*+2312:257*i*+2056> = tx_scrambled_*i*<256:0>

For a 10280-bit block without an alignment marker group:
For all *i*=0 to 39
  tx_scrambled_am<257*i*+256:257*i*> = tx_scrambled_*i*<256:0>

Alignment marker repetition rate is shown in Figure 119–8.

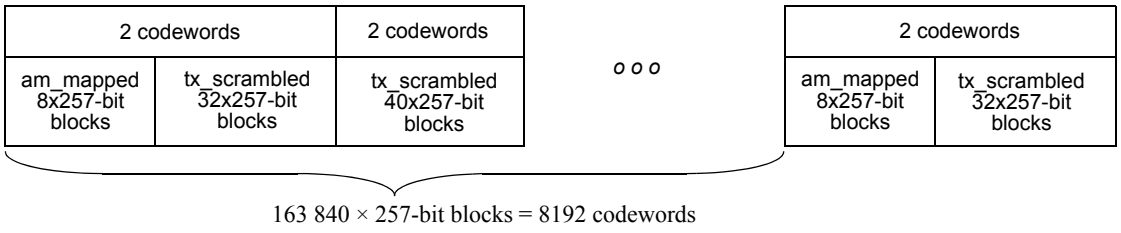


Figure 119–8—400GBASE-R alignment marker insertion period

119.2.4.5 Pre-FEC distribution

In order to improve error correction capability, each set of two consecutive Reed-Solomon FEC codewords is interleaved before being distributed to form the PCS lanes. To enable this interleaving, the Pre-FEC distribution function receives a 10280-bit block tx_scrambled_am, and shall perform a 10-bit symbol round robin distribution to form two 514-symbol FEC messages, m_A and m_B, which are subsequently each encoded by the RS FEC. The following describes the 10-bit round robin distribution process.

For all *i*=0 to 513

m_A<(513–*i*)> = tx_scrambled_am<(20*i*+9):(20*i*)>
m_B<(513–*i*)> = tx_scrambled_am<(20*i*+19):(20*i*+10)>

119.2.4.6 Reed-Solomon encoder

The PCS sublayer employs a Reed-Solomon code operating over the Galois Field GF(2^{10}) where the symbol size is 10 bits. The encoder processes *k* message symbols to generate 2*t* parity symbols, which are then appended to the message to produce a codeword of *n*=*k*+2*t* symbols. For the purposes of this clause, a particular Reed-Solomon code is denoted RS(*n*,*k*).

The PCS shall implement an RS(544,514) based FEC encoder. The PCS distributes a group of 40 × 257-bit blocks from tx_scrambled_am on a 10-bit round robin basis into two 5140-bit message blocks, $m_A$and m_B, as described in 119.2.4.5. These are then encoded using RS(544,514) encoder into codeword A and codeword B, respectively. The RS(544,514) code is based on the generating polynomial given by Equation (119–1).

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$$g(x) = \prod_{j=0}^{2t-1} (x - \alpha^j) = g_{2t}x^{2t} + g_{2t-1}x^{2t-1} + \ldots + g_1 x + g_0 \quad (119\text{–}1)$$

In Equation (119–1), α is a primitive element of the finite field defined by the polynomial $x^{10}+x^3+1$.

Equation (119–2) defines the message polynomial $m(x)$ whose coefficients are the message symbols m_{k-1} to m_0.

$$m(x) = m_{k-1}x^{n-1} + m_{k-2}x^{n-2} + \ldots + m_1 x^{2t+1} + m_0 x^{2t} \quad (119\text{–}2)$$

Each message symbol m_i is the bit vector $(m_{i,9}, m_{i,8}, \ldots, m_{i,1}, m_{i,0})$, which is identified with the element $m_{i,9}\alpha^9 + m_{i,8}\alpha^8 + \ldots + m_{i,1}\alpha + m_{i,0}$ of the finite field. The message symbols are composed of the variable m_A for codeword A and m_B for codeword B (including a group of alignment markers when appropriate). The first symbol input to the encoder is m_{k-1}.

Equation (119–3) defines the parity polynomial $p(x)$ whose coefficients are the parity symbols p_{2t-1} to p_0.

$$p(x) = p_{2t-1}x^{2t-1} + p_{2t-2}x^{2t-2} + \ldots + p_1 x + p_0 \quad (119\text{–}3)$$

The parity polynomial is the remainder from the division of $m(x)$ by $g(x)$. This may be computed using the shift register implementation illustrated in Figure 119–9. The outputs of the delay elements are initialized to zero prior to the computation of the parity for a given message. After the last message symbol, m_0, is processed by the encoder, the outputs of the delay elements are the parity symbols for that message.

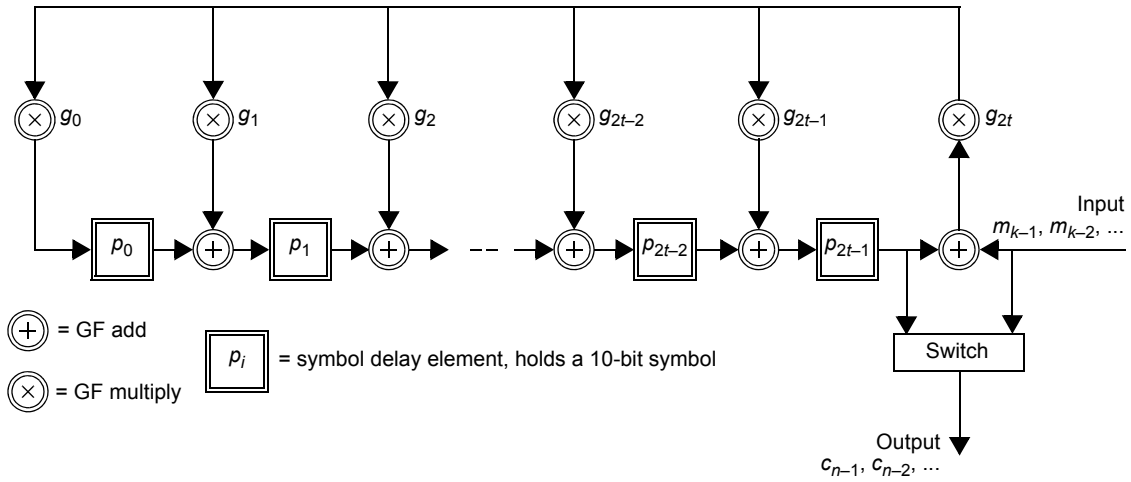


Figure 119–9—Reed-Solomon encoder functional model

The codeword polynomial $c(x)$ is then the sum of $m(x)$ and $p(x)$ where the coefficient of the highest power of x, $c_{n-1} = m_{k-1}$ is first and the coefficient of the lowest power of x, $c_0 = p_0$ is last.

The coefficients of the generator polynomial for the RS(544,514) code are presented in Table 119–3. Example codewords for the RS(544,514) code are provided in Annex 119A.

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Table 119–3—Coefficients of the generator polynomial g_i (decimal)

i	g_i	i	g_i	i	g_i
0	523	11	883	22	565
1	834	12	503	23	108
2	128	13	942	24	1
3	158	14	385	25	552
4	185	15	495	26	230
5	127	16	720	27	187
6	392	17	94	28	552
7	193	18	132	29	575
8	610	19	593	30	1
9	788	20	249		
10	361	21	282		

119.2.4.7 Symbol distribution

Once the data has been FEC encoded, two FEC codewords (c_A<543:0> and c_B<543:0>) are interleaved on a 10-bit basis before the data is distributed to each PCS lane.

The interleaving of two codewords for the 200GBASE-R PCS shall follow this procedure:

For all k=0 to 135
For all j=0 to 3
    if even(k)
        tx_out<$8k+2j$> = c_A<$543-4k-j$>
        tx_out<$8k+2j+1$> = c_B<$543-4k-j$>
    else
        tx_out<$8k+2j$> = c_B<$543-4k-j$>
        tx_out<$8k+2j+1$> = c_A<$543-4k-j$>

The interleaving of two codewords for the 400GBASE-R PCS shall follow this procedure:

For all k=0 to 67
For all j=0 to 7
    if even(k)
        tx_out<$16k+2j$> = c_A<$543-8k-j$>
        tx_out<$16k+2j+1$> = c_B<$543-8k-j$>
    else
        tx_out<$16k+2j$> = c_B<$543-8k-j$>
        tx_out<$16k+2j+1$> = c_A<$543-8k-j$>

Once the data has been Reed-Solomon encoded and interleaved, it shall be distributed to 8 PCS lanes for a 200GBASE-R PCS and to 16 PCS lanes for a 400GBASE-R PCS, one 10-bit symbol at a time, from the lowest to the highest PCS lane. The first bit transmitted from each 10-bit symbol is bit 0.

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119.2.4.8 Transmit bit ordering and distribution

The transmit bit ordering and distribution are illustrated in Figure 119–10 for a 200GBASE-R PCS.

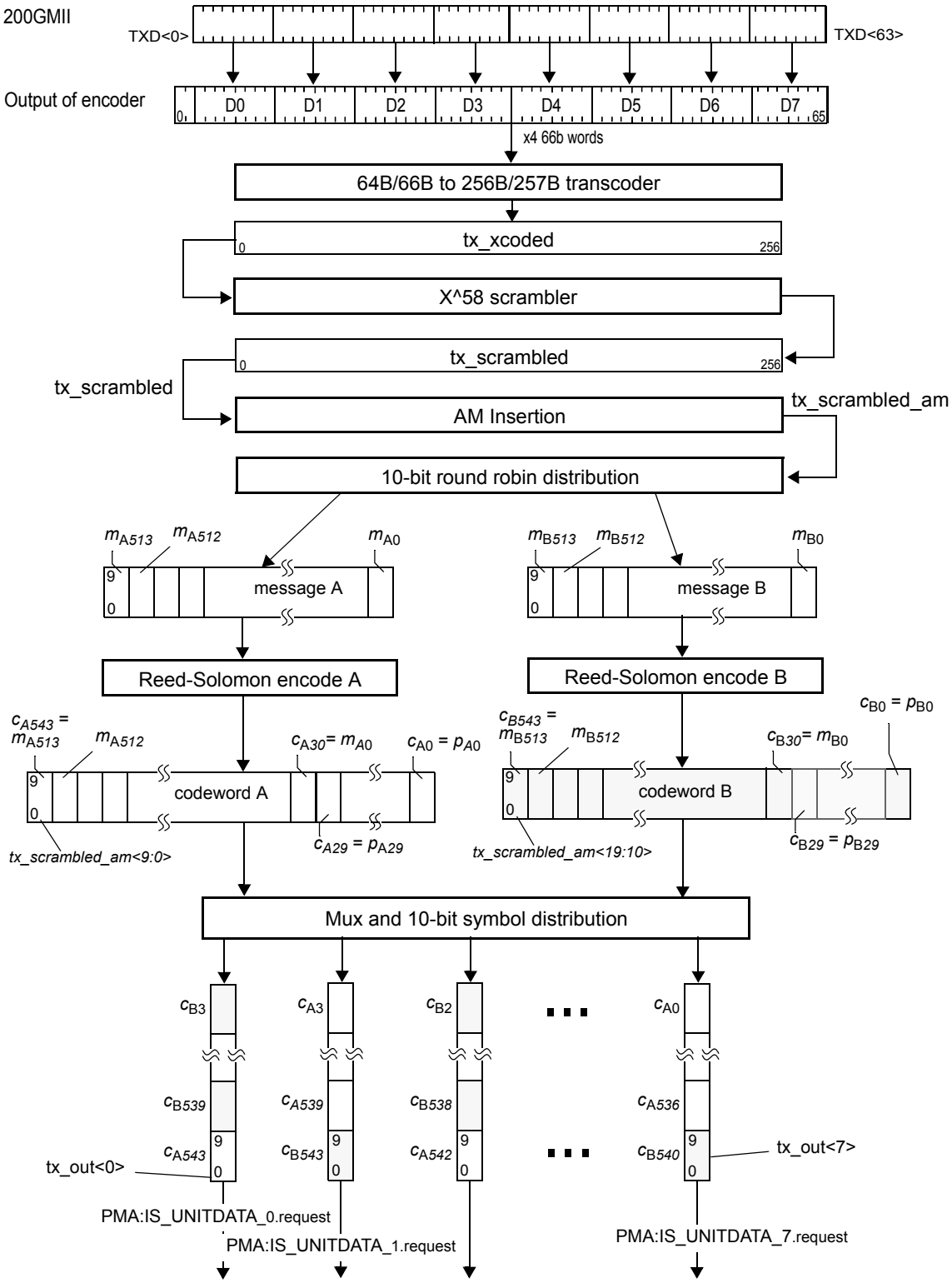


Figure 119–10—200GBASE-R Transmit bit ordering and distribution

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The transmit bit ordering and distribution is illustrated in Figure 119–11 for a 400GBASE-R PCS.

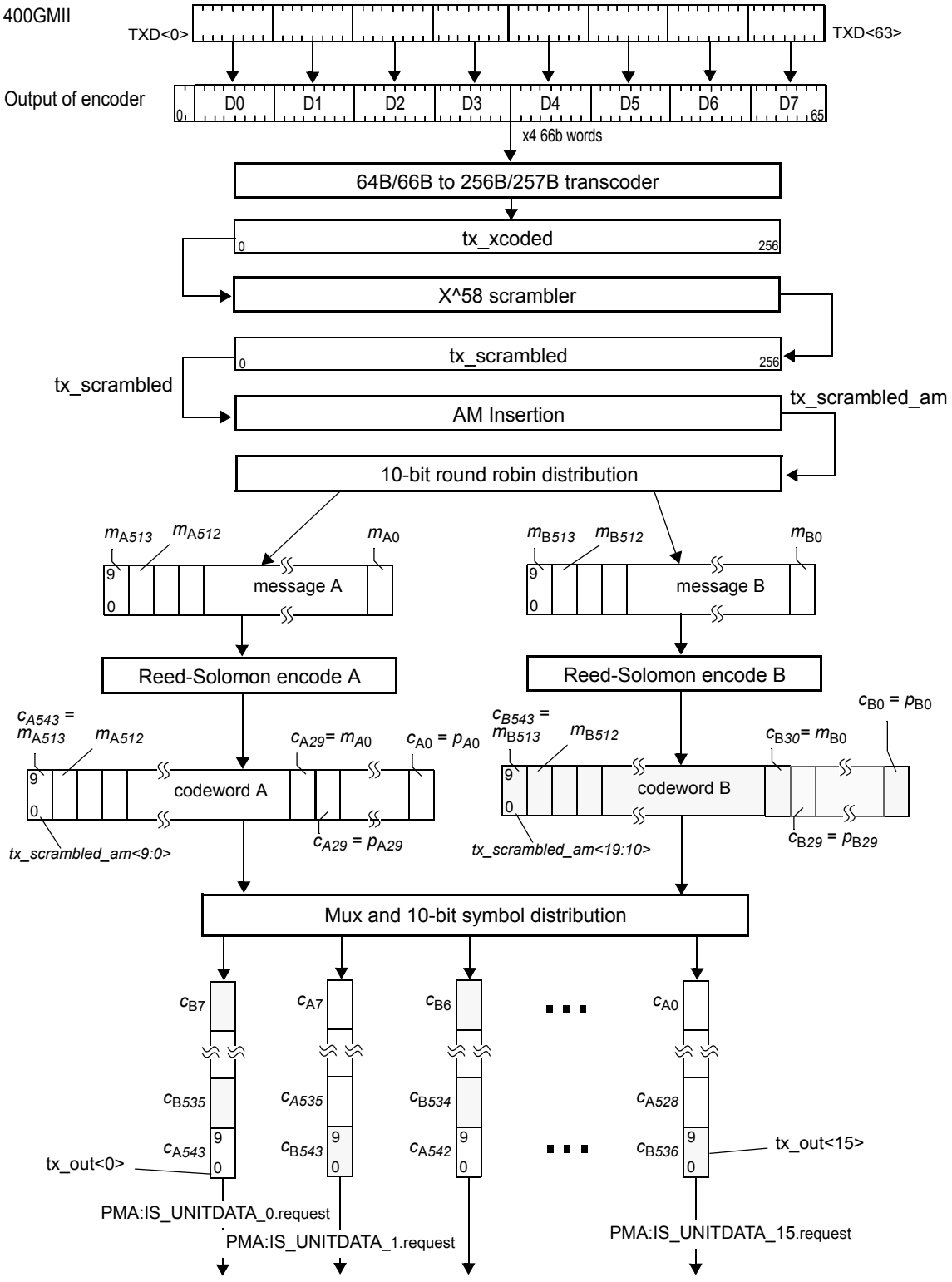


Figure 119–11—400GBASE-R Transmit bit ordering and distribution

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119.2.4.9 Test-pattern generators

The PCS shall have the ability to generate a scrambled idle test pattern which is suitable for receiver tests and for certain transmitter tests. When a scrambled idle pattern is enabled, the test pattern is generated by the PCS. The test pattern is an idle control block (block type=0x1E) with all idles as defined in Figure 82-5. The test pattern is sent continuously and is transcoded, scrambled, alignment markers are inserted and finally encapsulated by the FEC.

When the transmit channel is operating in test-pattern mode, the encoded bit stream is distributed to the PCS Lanes as in normal operation (see 119.2.4.7).

If a Clause 45 MDIO is implemented, then control of the test-pattern generation is from the BASE-R PCS test-pattern control register (bit 3.42.3).

119.2.5 Receive function

119.2.5.1 Alignment lock and deskew

The receive PCS forms *n* separate bit streams by concatenating the bits from each of the *n* PMA:IS_UNITDATA_*i*.indication primitives in the order they are received (where *n* = 8 for a 200GBASE-R PCS and *n* = 16 for a 400GBASE-R PCS). It obtains lock to the alignment markers as specified by the alignment marker lock state diagram shown in Figure 119–12. Note that alignment marker lock is achieved before FEC codewords are processed and therefore the alignment markers are processed in a high error probability environment.

After alignment marker lock is achieved on each of the *n* lanes (bit streams), all inter-lane Skew is removed as specified by the PCS synchronization state diagram shown in Figure 119–13. The PCS receive function shall support a maximum Skew of 180 ns, and maximum Skew Variation of 4 ns, between PCS lanes.

119.2.5.2 Lane reorder and de-interleave

PCS lanes can be received on different lanes of the service interface from which they were originally transmitted. The PCS receive function shall order the PCS lanes according to the PCS lane number. The PCS lane number is defined by the unique portion (UM_0 to UM_5) of the alignment marker that is mapped to each PCS lane (see 119.2.4.4).

After all PCS lanes are aligned, deskewed, and reordered, the two FEC codewords are de-interleaved to reconstruct the original stream of two FEC codewords.

119.2.5.3 Reed-Solomon decoder

The Reed-Solomon decoder extracts the message symbols from the codeword, corrects them as necessary, and discards the parity symbols.

The Reed-Solomon decoder shall be capable of correcting any combination of up to t=15 symbol errors in a codeword. The Reed-Solomon decoder shall also be capable of indicating when an errored codeword was not corrected. The probability that the decoder fails to indicate a codeword with t+1 errors as uncorrected is not expected to exceed 10^{-16}. This limit is also expected to apply for t+2 errors, t+3 errors, and so on.

If bypass error indication is not supported or not enabled, when the Reed-Solomon decoder determines that a codeword contains errors that were not corrected, it shall cause the PCS receive function to set every 66-bit block within the two associated codewords to an error block (set to EBLOCK_R). This may be achieved by setting the synchronization header to 11 for all 66-bit blocks created from these codewords by the 256B/257B to 64B/66B transcoder.

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The Reed-Solomon decoder may optionally provide the ability to bypass the error indication feature to reduce the delay contributed by the FEC function. The presence of this option is indicated by the assertion of the FEC_bypass_indication_ability variable (see 119.3). When the option is provided it is enabled by the assertion of the FEC_bypass_indication_enable variable (see 119.3).

When FEC_bypass_indication_enable is asserted, additional error monitoring is performed by the Reed-Solomon decoder to reduce the likelihood that errors in a packet are not detected. The Reed-Solomon decoder counts the number of symbol errors detected on all PCS lanes in consecutive non-overlapping blocks of 8192 codewords. When the number of symbol errors in a block of 8192 codewords exceeds 5560, the Reed-Solomon decoder shall assert hi_ser for a period of 60 ms to 75 ms.

The Reed-Solomon decoder may optionally provide the ability to signal a degradation of the received signal. The presence of this option is indicated by the assertion of the FEC_degraded_SER_ability variable (see 119.3). When the option is provided it is enabled by the assertion of the FEC_degraded_SER_enable variable (see 119.3).

When FEC_degraded_SER_enable is asserted, additional error monitoring is performed by the PCS. The Reed-Solomon decoder counts the number of symbol errors detected on all PCS lanes in consecutive non-overlapping blocks of FEC_degraded_SER_interval codewords (see 119.3.1), where the least significant bit of FEC_degraded_SER_interval is ignored (evaluated as 0) to make the number of codewords even. If the decoder determines that a codeword is uncorrectable, the number of symbol errors detected is increased by 16. When the number of symbol errors exceeds the threshold set in FEC_degraded_SER_activate_threshold (see 119.3.1), the FEC_degraded_SER bit (see 119.3.1) is set. At the end of each interval, if the number of symbol errors is less than FEC_degraded_SER_deactivate_threshold, the FEC_degraded_SER bit is cleared. If either FEC_degraded_SER_ability or FEC_degraded_SER_enable is de-asserted then the FEC_degraded_SER bit is cleared.

119.2.5.4 Post FEC interleave

After the Reed-Solomon decoder processes the data, data is interleaved on a 10-bit basis into rx_scrambled_am from two codewords corresponding to 40 transcoded blocks in order to recreate the transmitted data stream.

119.2.5.5 Alignment marker removal

For the 200GBASE-R PCS, every 81 920 x 257-bit blocks (corresponds to 4096 codewords) the first 1028 bits of rx_scrambled_am blocks is the vector am_rx<1027:0> where bit 0 is the first bit received. The specific codewords that include this vector are indicated by the alignment lock and deskew function. The 3-bit receive alignment marker status field is assigned from the variable am_rx as follows:

rx_am_sf<2:0> = am_rx<1027:1025>

For the 400GBASE-R PCS, every 163 840 x 257-bit blocks (corresponds to 8192 codewords) the first 2056 bits of rx_scrambled_am blocks is the vector am_rx<2055:0> where bit 0 is the first bit received. The specific codewords that include this vector are indicated by the alignment lock and deskew function. The 3-bit receive alignment marker status field is assigned from the variable am_rx as follows:

rx_am_sf<2:0> = am_rx<2055:2053>

The vector am_rx shall be removed from rx_scrambled_am to create rx_scrambled prior to descrambling.

119.2.5.6 Descrambler

The descrambler shall process rx_scrambled<256:0> to reverse the effect of the scrambler using the polynomial given in Equation (49-1).

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119.2.5.7 256B/257B to 64B/66B transcoder

The transcoder extracts a group of four 66-bit blocks, rx_coded_*j*<65:0> where *j*=0 to 3, from each 257-bit block rx_xcoded<256:0>. Bit 0 of the 257-bit block is the first bit received.

If rx_xcoded<0> is 1, rx_coded_*j*<65:0> for *j*=0 to 3 shall be derived as follows.

a) rx_coded_*j*<65:2> = rx_xcoded<(64*j*+64):(64*j*+1)> for *j*=0 to 3

b) rx_coded_*j*<0>=0 and rx_coded_*j*<1>=1 for all *j*=0 to 3

If rx_xcoded<0> is 0 and any rx_xcoded<*j*+1>=0 for *j*=0 to 3, rx_coded_*j*<65:0> for *j*=0 to 3 shall be derived as follows.

a1) Let *c* be the smallest value of *j* such that rx_xcoded<*j*+1>=0. In other words, rx_coded_*c* is the first 66-bit control block in the resulting group of four blocks.

b1) Let rx_payloads be a vector representing the payloads of the four 66-bit blocks. It is derived using the following expressions:
rx_payloads<(64*c*+3):0> = rx_xcoded<(64*c*+8):5>
rx_payloads<(64c+7):(64c+4)> is set to a value derived by cross-referencing rx_payloads<(64c+3):64c> using Figure 82-5. For example, if rx_payloads<(64c+3):64c> is 0xE then rx_payloads<(64c+7):(64c+4)> is 0x1. If no match to rx_payloads<(64c+3):64c> is found, rx_payloads<(64c+7):(64c+4)> is set to 0000
rx_payloads<255:(64*c*+8)> = rx_xcoded<256:(64*c*+9)>

c1) rx_coded_*j*<65:2> = rx_payloads<(64*j*+63):64*j*> for *j*=0 to 3

d1) If rx_xcoded<*j*+1>=0, rx_coded_*j*<0>=1 and rx_coded_*j*<1>=0 for *j*=0 to 3

e1) If rx_xcoded<*j*+1>=1, rx_coded_*j*<0>=0 and rx_coded_*j*<1>=1 for *j*=0 to 3

f1) If rx_payloads<(64c+7):(64c+4)> = 0000, rx_coded_c<1>=1 (invalidate synchronization header)

If rx_xcoded<0> is 0 and all rx_xcoded<*j*+1>=1 for *j*=0 to 3, rx_coded_*j*<65:0> for *j*=0 to 3 shall be derived as follows.

a2) Set *c* = 0

b2) Let rx_payloads be a vector representing the payloads of the four 66-bit blocks. It is derived using the following expressions.
rx_payloads<(64*c*+3):0> = rx_xcoded<(64*c*+8):5>
rx_payloads<(64*c*+7):(64*c*+4)> = 0000
rx_payloads<255:(64*c*+8)> = rx_xcoded<256:(64*c*+9)>

c2) rx_coded_*j*<65:2> = rx_payloads<(64*j*+63):(64*j*)> for *j*=0 to 3

d2) rx_coded_*j*<0>=0 and rx_coded_*j*<1>=0 for *j*=0 and 2

e2) rx_coded_*j*<0>=1 and rx_coded_*j*<1>=1 for *j*=1 and 3

The 66-bit blocks are transmitted in order from *j*=0 to 3. Bit 0 of each block is the first bit transmitted.

Note—The stream of 66-bit blocks generated by this process is used as the reference signal for de-mapping from OTN. See ITU-T G.709 [B50].

119.2.5.8 Decode and rate matching

The receive PCS decodes blocks to produce RXD<63:0> and RXC<7:0> for transmission to the 200GMII/400GMII. One 200GMII/400GMII data transfer is decoded from each block. The receive PCS may insert idle control characters to compensate for the removal of alignment markers. If the receive PCS spans multiple clock domains, it may also perform clock rate compensation via the deletion of idle control characters or sequence ordered sets or the insertion of idle control characters (see 82.2.3.6 and 82.2.3.9 for insertion and deletion rules).

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The PCS receive decodes blocks as specified in the receive state diagram shown in Figure 119–15.

119.2.6 Detailed functions and state diagrams

119.2.6.1 State diagram conventions

The body of this subclause is composed of state diagrams, including the associated definitions of variables, functions, and counters. Should there be a discrepancy between a state diagram and descriptive text, the state diagram prevails.

The notation used in the state diagrams follows the conventions of 21.5. The notation ++ after a counter or integer variable indicates that its value is to be incremented.

119.2.6.2 State variables

119.2.6.2.1 Constants

EBLOCK_R<71:0>
: 72-bit vector to be sent to the 200GMII/400GMII containing /E/ in all the eight character locations.

EBLOCK_T<65:0>
: 66-bit vector to be sent to the transcoder containing /E/ in all the eight character locations.

LBLOCK_R<71:0>
: 72-bit vector to be sent to the 200GMII/400GMII containing one Local Fault ordered set. The Local Fault ordered set is defined in 119.3.

LBLOCK_T<65:0>
: 66-bit vector to be sent to the transcoder containing one Local Fault ordered set.

119.2.6.2.2 Variables

align_status
: A variable set by the PCS synchronization process to reflect the status of PCS lane-to-lane alignment. Set to true when all lanes are synchronized and aligned and set to false when the PCS synchronization process is not complete.

all_locked
: A Boolean variable that is set to true when amps_lock<*x*> is true for all *x* and is set to false when amps_lock<*x*> is false for any *x*.

amp_counter_done
: Boolean variable that indicates that amp_counter has reached its terminal count.

amp_match
: Boolean variable that holds the output of the function AMP_COMPARE.

amp_valid
: Boolean variable that is set to true if the received 120-bit block is a valid alignment marker payload. The alignment marker payload, mapped to a PCS lane according to the process described in 119.2.4.4, consists of 96 known bits. The 48 bits of the common marker portion are compared on a nibble-wise basis (12 comparisons). If 9 or more nibbles in the candidate block match the corresponding known nibbles in the common portion of the alignment marker payload, the candidate block is considered a valid alignment marker payload.

amps_lock<*x*>
: Boolean variable that is set to true when the receiver has detected the location of the alignment marker payload sequence for a given lane on the PMA service interface, where $x = 0{:}7$ for 200GBASE-R and $x = 0{:}15$ for 400GBASE-R

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current_pcsl

A variable that holds the PCS lane number corresponding to the current alignment marker payload that is recognized on a given lane of the PMA service interface. It is compared to the variable first_pcsl to confirm that the location of the alignment marker payload sequence has been detected. The PCS lane number is determined by the alignment marker payloads based on the mapping defined in 119.2.4.4. The 48 bits that are in the positions of the unique marker bits in the received alignment marker payload are compared to the expected values for a given payload position and PCS lane on a nibblewise basis (12 comparisons). If 9 or more nibbles in the candidate block match the corresponding known nibbles for any payload position on a given PCS lane, then the PCS lane number is assigned accordingly. If a match is not found, then the AMP_COMPARE function sets amp_match to false.

cw_A_bad

A Boolean variable that is set to true if the Reed-Solomon decoder (see 119.2.5.3) fails to correct the current FEC codeword A and is set to false otherwise.

cw_B_bad

A Boolean variable that is set to true if the Reed-Solomon decoder (see 119.2.5.3) fails to correct the current FEC codeword B and is set to false otherwise.

deskew_done

A Boolean variable that is set to true when pcs_enable_deskew is set to true and the deskew process is completed. Otherwise, this variable is set to false.

first_pcsl

A variable that holds the PCS lane number that corresponds to the first alignment marker payload that is recognized on a given lane of the PMA service interface. It is compared to the PCS lane number corresponding to the second alignment marker payload that is tested.

hi_ser

When FEC_bypass_indication_enable is set to one, this bit is set to one for a period of 60 ms to 75 ms if the number of RS-FEC symbol errors in a window of 8192 codewords exceeds the threshold (see 119.2.5.3), and is set to zero otherwise. When FEC_bypass_indication_enable is set to zero, this bit is set to zero. This variable is mapped to the bit defined in 45.2.3.47k (3.801.2).

pcs_alignment_valid

Boolean variable that is set to true if all PCS lanes are aligned. PCS lanes are considered to be aligned when amps_lock<*x*> is true for all *x*, each PCS lane is locked to a unique alignment marker payload sequence (see 119.2.4.4), and the PCS lanes are deskewed. Otherwise, this variable is set to false.

pcs_enable_deskew

A Boolean variable that enables and disables the PCS synchronization process. Received bits may be discarded whenever deskew is enabled. It is set to true when deskew is enabled and set to false when deskew is disabled.

pcs_lane_mapping<*x*>

A variable that holds the value of the pcs_lane received on physical lane *x*.

PCS_status

A boolean variable that is true when align_status is true and is false otherwise.

reset

Boolean variable that controls the resetting of the PCS sublayer. It is true whenever a reset is necessary including when reset is initiated from the MDIO, during power on, and when the MDIO has put the PCS sublayer into low-power mode.

restart_lock

Boolean variable that is set by the PCS synchronization process to restart the alignment marker lock process on all PCS lanes. It is set to true after 3 consecutive uncorrected codewords are received (3_BAD state) or when 5 Alignment Markers in a row fail to match (5_BAD state) and set to false upon entry into the LOSS_OF_ALIGNMENT state.

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rx_coded<65:0>
: Vector containing the input to the 64B/66B decoder. The format for this vector is shown in Figure 82-5. The leftmost bit in the figure is rx_coded<0> and the rightmost bit is rx_coded<65>.

rx_local_degraded
: Boolean variable that is asserted true when the receiver detects rx_am_sf<1> asserted true for two consecutive alignment marker periods. It is deasserted when rx_am_sf<1> is deasserted for two consecutive alignment marker periods. If a Clause 45 MDIO is implemented, the status of this variable is reflected in bit 3.801.6.

rx_raw<71:0>
: Vector containing one 200GMII/400GMII transfer. RXC<0> through RXC<7> are from rx_raw<0> through rx_raw<7>, respectively. RXD<0> through RXD<63> are from rx_raw<8> through rx_raw<71>, respectively.

rx_rm_degraded
: Boolean variable that is asserted true when the receiver detects rx_am_sf<2> asserted true for two consecutive alignment marker periods. It is deasserted when rx_am_sf<2> is deasserted for two consecutive alignment marker periods. If a Clause 45 MDIO is implemented, the status of this variable is reflected in bit 3.801.5.

signal_ok
: Boolean variable that is set based on the most recently received value of PMA:IS_SIGNAL.indication(SIGNAL_OK). It is true if the value was OK and false if the value was FAIL.

slip_done
: Boolean variable that is set to true when the SLIP requested by the alignment marker lock state diagram has been completed indicating that the next candidate 120-bit block position can be tested.

test_amp
: Boolean variable that is set to true when a candidate block position is available for testing and false when the FIND_1ST state is entered.

test_cw
: Boolean variable that is set to true when a new FEC codeword is available for decoding and is set to false when the TEST_CW state is entered.

tx_coded<65:0>
: Vector containing the output from the 64B/66B encoder. The format for this vector is shown in Figure 82-5. The leftmost bit in the figure is tx_coded<0> and the rightmost bit is tx_coded<65>.

tx_raw<71:0>
: Vector containing one 200GMII/400GMII transfer. TXC<0> through TXC<7> are placed in tx_raw<0> through tx_raw<7>, respectively. TXD<0> through TXD<63> are placed in tx_raw<8> through tx_raw<71>, respectively.

119.2.6.2.3 Functions

AMP_COMPARE
: This function compares the values of first_pcsl and current_pcsl to determine if a valid alignment marker payload sequence has been detected and returns the result of the comparison using the variable amp_match. If current_pcsl and first_pcsl both found a match and indicate the same PCS lane number, amp_match is set to true.

DECODE(rx_coded<65:0>)
: Decodes the 66-bit vector returning rx_raw<71:0>, which is sent to the 200GMII/400GMII. The DECODE function shall decode the block as specified in 119.2.3.

ENCODE(tx_raw<71:0>)
: Encodes the 72-bit vector returning tx_coded<65:0>. The ENCODE function shall encode the block as specified in 119.2.3.

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R_TYPE(rx_coded<65:0>)

This function classifies the current rx_coded<65:0> vector as belonging to one of the following types, depending on its contents. The classification results are returned via the r_block_type variable.

Values: C; The vector contains a sync header of 10 and one of the following:

a) A block type field of 0x1E and eight valid control characters other than /E/ or /LI/;

b) A block type field of 0x4B.

LI; For EEE capability, the LI type is supported where the vector contains a sync header of 10, a block type field of 0x1E and eight control characters of 0x06 (/LI/).

S; The vector contains a sync header of 10 and the following:

a) A block type field of 0x78.

T; The vector contains a sync header of 10, a block type field of 0x87, 0x99, 0xAA, 0xB4, 0xCC, 0xD2, 0xE1 or 0xFF and all control characters are valid.

D; The vector contains a sync header of 01.

E; The vector does not meet the criteria for any other value.

Valid control characters are specified in Table 82-1.

NOTE—A PCS that does not support EEE classifies vectors containing one or more /LI/ control characters as type E.

R_TYPE_NEXT

This function classifies the 66-bit rx_coded vector that immediately follows the current rx_coded<65:0> vector as belonging to one of the five types defined in R_TYPE, depending on its contents. It is intended to perform a prescient end of packet check. The classification results are returned via the r_block_type_next variable.

SLIP

Causes the next candidate block position to be tested. The precise method for determining the next candidate block position is not specified and is implementation dependent. However, an implementation shall ensure that all possible block positions are evaluated.

T_TYPE = (tx_raw<71:0>)

This function classifies each 72-bit tx_raw vector as belonging to one of the following types depending on its contents. The classification results are returned via the t_block_type variable.

Values: C; The vector contains one of the following:

a) Eight valid control characters other than /O/, /S/, /T/, /LI/, and /E/;

b) One valid ordered set.

LI; For EEE capability, this vector contains eight /LI/ characters.

S; The vector contains an /S/ in its first character, and all characters following the /S/ are data characters.

T; The vector contains a /T/ in one of its characters, all characters before the /T/ are data characters, and all characters following the /T/ are valid control characters other than /O/, /S/ and /T/.

D; The vector contains eight data characters.

E; The vector does not meet the criteria for any other value.

A tx_raw character is a control character if its associated TXC bit is asserted. A valid control character is one containing an 200GMII/400GMII control code specified in Table 82-1. A valid ordered set consists of a valid /O/ character in the first character and data characters in the seven characters following the /O/. A valid /O/ is any character with a value for O code in Table 82-1.

NOTE—A PCS that does not support EEE classifies vectors containing one or more /LI/ control characters as type E.

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119.2.6.2.4 Counters

amp_bad_count
: Counts the number of consecutive alignment markers that don't match the expected values for a given PCS lane.

amp_counter
: This counter counts the interval of *i* FEC codewords containing normal alignment marker payload sequences (where *i* is 4096 for the 200GBASE-R PCS, and 8192 for the 400GBASE-R PCS).

cw_A_bad_count
: Counts the number of consecutive uncorrected FEC codewords for codeword A. This counter is set to zero when an FEC codeword A is received and cw_A_bad is false.

cw_B_bad_count
: Counts the number of consecutive uncorrected FEC codewords for codeword B. This counter is set to zero when an FEC codeword B is received and cw_B_bad is false.

119.2.6.3 State diagrams

The 200GBASE-R PCS shall implement eight alignment marker lock processes and the 400GBASE-R PCS shall implement sixteen alignment marker lock processes as depicted in Figure 119–12. An alignment marker lock process operates independently on each lane. The alignment marker lock state diagram shown in Figure 119–12 determines when the PCS has obtained alignment marker lock to the received bit stream for a given lane of the service interface. Each alignment marker lock process looks for two valid alignment markers 278 528 × 10-bit Reed-Solomon symbols apart, on a per PCS lane basis, to gain alignment marker lock. When the alignment marker lock process achieves lock for a lane, and if a Clause 45 MDIO is implemented, the PCS shall record the PCS lane number received on that lane of the service interface in the appropriate lane mapping register (3.400 to 3.415). Once in lock, a lane goes out of alignment marker lock only when restart_lock is signaled. This occurs when the PCS synchronization process determines that three uncorrectable codewords in a row are seen, or when the alignment marker lock process sees five alignment markers in a row that fail to match the expected pattern on a given lane.

The PCS shall run the synchronization process as depicted in Figure 119–13. The PCS synchronization process is responsible for determining if the PCS is capable of presenting coherent data to the 200GMII/400GMII. The synchronization process ensures that all PCS lanes have alignment marker lock, are locked to different alignment markers, and that the Skew is within the boundaries of what the PCS can deskew. The synchronization process, along with the alignment marker lock process, are restarted if three consecutive FEC codewords from the same codeword (A or B) are uncorrectable.

The Transmit state diagram shown in Figure 119–14 controls the encoding of 66-bit transmitted blocks. It makes exactly one transition for each 66-bit transmit block processed. Though the Transmit state diagram sends Local Fault ordered sets when reset is asserted, the scrambler is not guaranteed to be operational during reset. Thus, the Local Fault ordered sets may not appear on the PMA service interface.

The Receive state diagram shown in Figure 119–15 controls the decoding of received blocks. It makes exactly one transition for each receive 66-bit block processed.

The PCS shall perform the functions of alignment marker lock, PCS synchronization, Transmit, and Receive as specified in the respective state diagrams.

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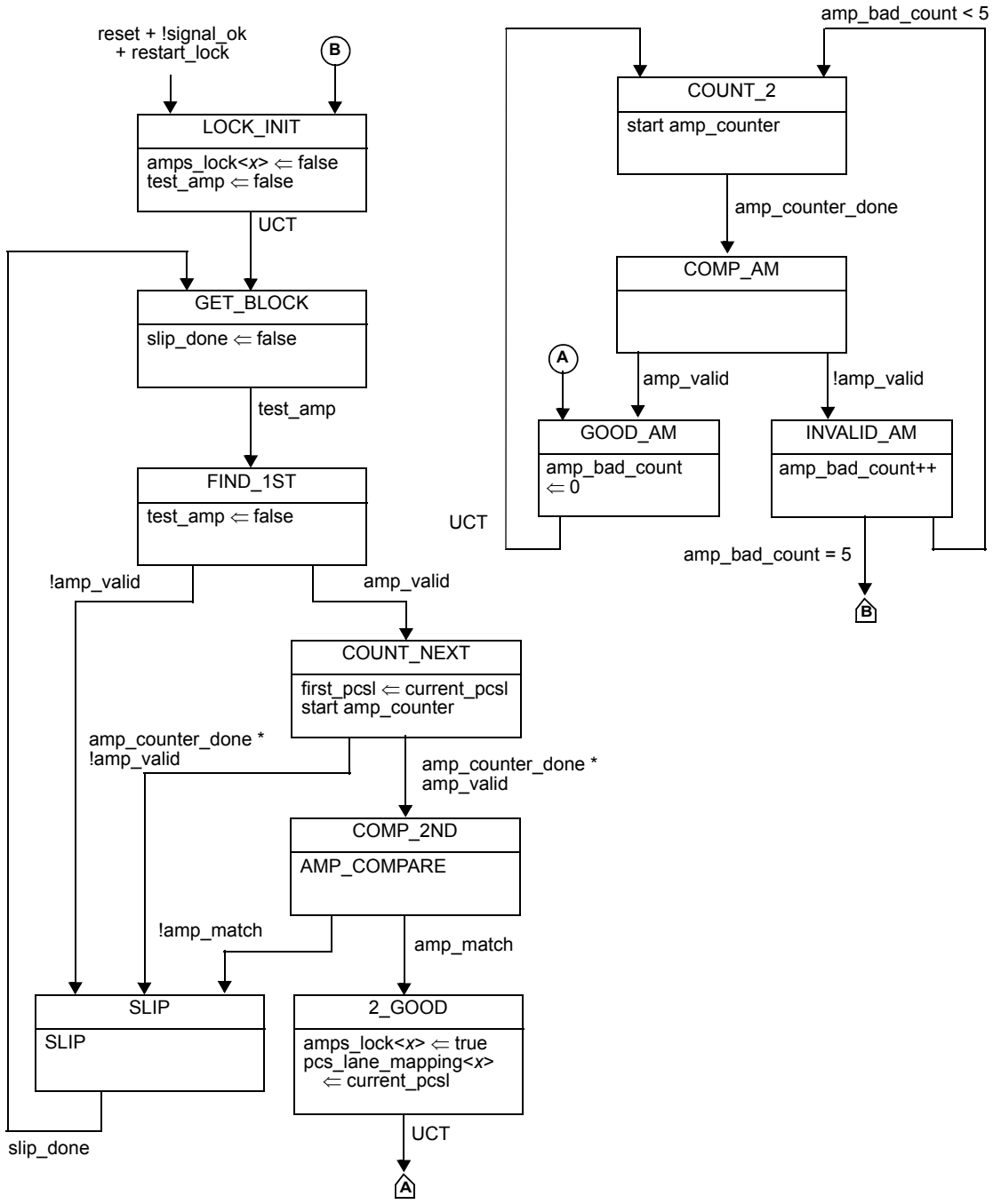


Figure 119–12—Alignment marker lock state diagram

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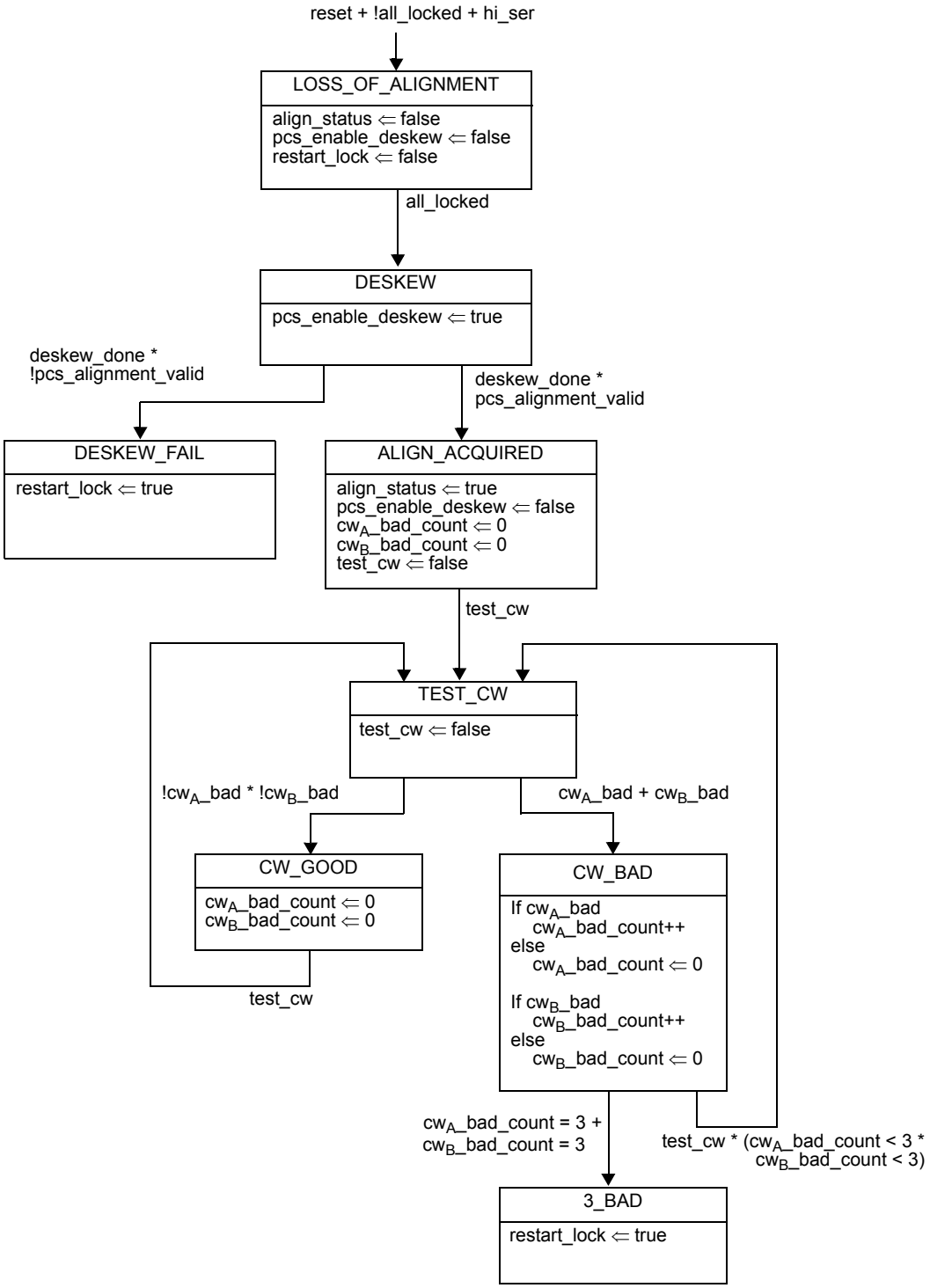


Figure 119–13—PCS synchronization state diagram

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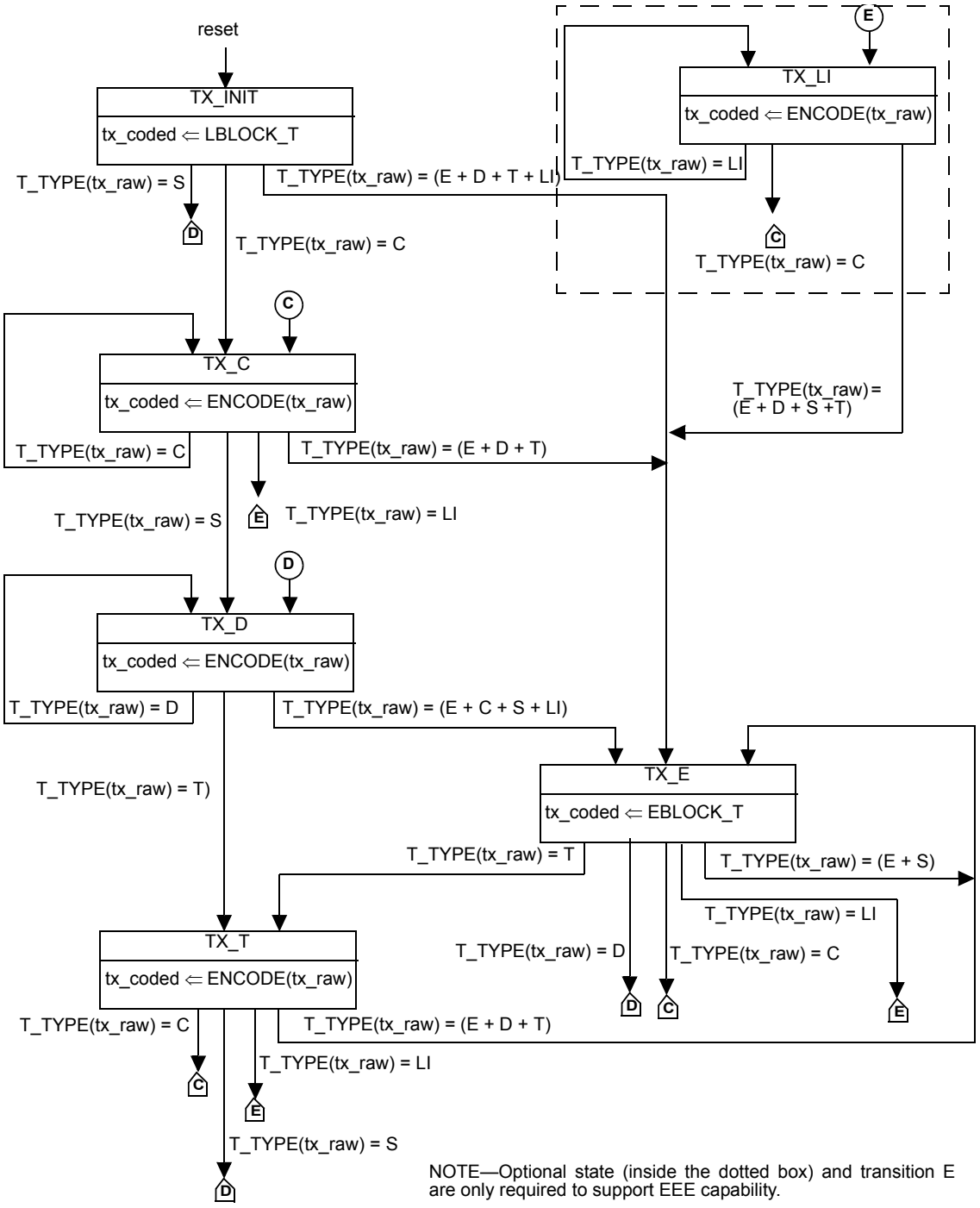


Figure 119–14—Transmit state diagram

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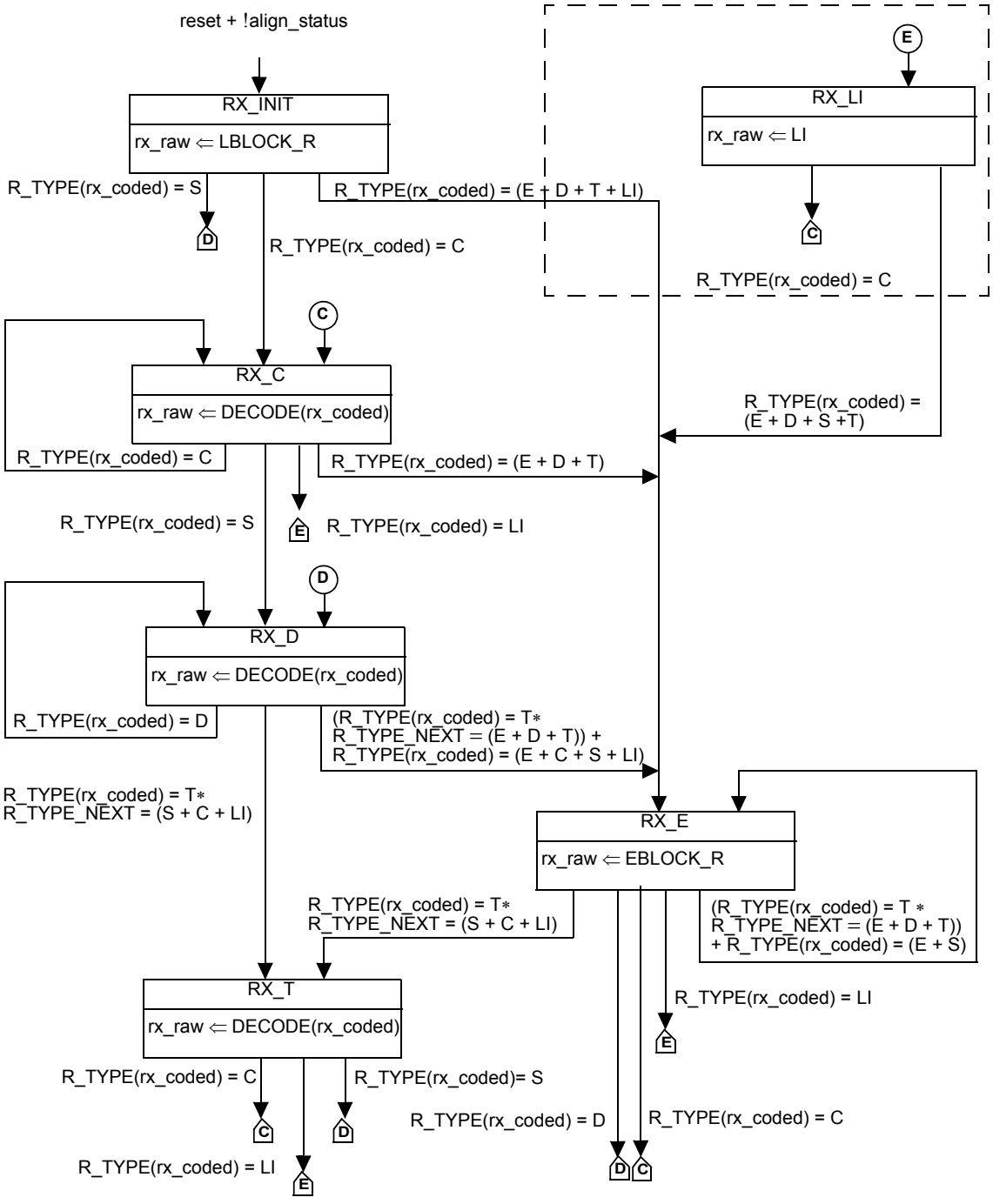


Figure 119–15—Receive state diagram

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119.3 PCS management

The following objects apply to PCS management. If an MDIO Interface is provided (see Clause 45), they are accessed via that interface. If not, it is recommended that an equivalent access is provided.

119.3.1 PCS MDIO function mapping

The optional MDIO capability described in Clause 45 defines several registers that provide control and status information for and about the PCS. If MDIO is implemented, it shall map MDIO control bits to PCS control variables, as shown in Table 119–4, and MDIO status bits to PCS status variables, as shown in Table 119–5.

Table 119–4—MDIO/PCS control variable mapping

MDIO control variable	PCS register name	Register/ bit number	PCS control variable
Reset	PCS control 1 register	3.0.15	reset
Loopback	PCS control 1 register	3.0.14	Loopback
Transmit test-pattern enable	BASE-R PCS test-pattern control register	3.42.3	tx_test_mode
LPI_FW	EEE control and capability	3.20.0	LPI_FW
PCS FEC bypass indication enable	PCS FEC control register	3.800.1	FEC_bypass_indication_enable
PCS FEC degraded SER enable	PCS FEC control register	3.800.2	FEC_degraded_SER_enable
PCS FEC degraded SER activate threshold	PCS FEC degraded SER activate threshold register	3.806, 3.807	FEC_degraded_SER_activate_threshold
PCS FEC degraded SER deactivate threshold	PCS FEC degraded SER deactivate threshold register	3.808, 3.809	FEC_degraded_SER_deactivate_threshold
PCS FEC degraded SER interval	PCS FEC degraded SER interval	3.810, 3.811	FEC_degraded_SER_interval

Table 119–5—MDIO/PCS status variable mapping

MDIO status variable	PCS register name	Register/ bit number	PCS status variable
BASE-R and MultiGBASE-T receive link status	BASE-R and MultiGBASE-T PCS status 1 register	3.32.12	PCS_status
Lane *x* aligned	Multi-lane BASE-R PCS alignment status 3 and 4 registers	3.52.7:0 3.53.7:0	am_lock<*x*>
PCS lane alignment status	Multi-lane BASE-R PCS alignment status 1 register	3.50.12	align_status
Lane *x* mapping	Lane *x* mapping register	3.400 through 3.415	pcs_lane_mapping<*x*>
PCS FEC bypass indication ability	PCS FEC status register	3.801.1	FEC_bypass_indication_ability

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Table 119–5—MDIO/PCS status variable mapping *(continued)*

MDIO status variable	PCS register name	Register/ bit number	PCS status variable
PCS FEC corrected codewords	PCS FEC corrected codewords counter register	3.802, 3.803	FEC_corrected_cw_counter
PCS FEC uncorrected codewords	PCS FEC uncorrected codewords counter register	3.804, 3.805	FEC_uncorrected_cw_counter
PCS FEC symbol errors, PCS lanes 0 to *x*	PCS FEC symbol error counter register, lanes 0 to *x*	3.600 to 3.631	FEC_symbol_error_counter_*i*
Tx LPI indication	PCS status 1	3.1.9	Tx LPI indication
Tx LPI received	PCS status 1	3.1.11	Tx LPI received
Rx LPI indication	PCS status 1	3.1.8	Rx LPI indication
Rx LPI received	PCS status 1	3.1.10	Rx LPI received
EEE wake error counter	EEE wake error counter	3.22	Wake_error_counter
PCS FEC degraded SER ability	PCS FEC status register	3.801.3	FEC_degraded_SER_ability
PCS FEC degraded SER	PCS FEC status register	3.801.4	FEC_degraded_SER
Local degraded SER received	PCS FEC status register	3.801.6	rx_local_degraded
Remote degraded SER received	PCS FEC status register	3.801.5	rx_rm_degraded
PCS FEC high SER	PCS FEC status register	3.801.2	hi_ser

119.4 Loopback

When the PCS is in loopback, the PCS shall accept data on the transmit path from the 200GMII/400GMII and return it on the receive path to the 200GMII/400GMII. In addition, the PCS shall transmit what it receives from the 200GMII/400GMII to the PMA sublayer, and shall ignore all data presented to it by the PMA sublayer. If a Clause 45 MDIO is implemented, then the PCS is placed in the loopback when the loopback bit from the PCS control 1 register (bit 3.0.14) is set to a one.

119.5 Delay constraints

The maximum delay contributed by the 200GBASE-R PCS (sum of transmit and receive delays at one end of the link) shall be no more than 160 256 BT (313 pause_quanta or 801.28 ns). The maximum delay contributed by the 400GBASE-R PCS (sum of transmit and receive delays at one end of the link) shall be no more than 320 000 BT (625 pause_quanta or 800 ns). A description of overall system delay constraints and the definitions for bit times and pause_quanta can be found in 116.4 and its references.

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119.6 Protocol implementation conformance statement (PICS) proforma for Clause 119, Physical Coding Sublayer (PCS) for 64B/66B, type 200GBASE-R and 400GBASE-R[4]

119.6.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Clause 119, Physical Coding Sublayer (PCS) for 64B/66B, type 200GBASE-R and 400GBASE-R, shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

119.6.2 Identification

119.6.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

119.6.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Clause 119, Physical Coding Sublayer (PCS) 64B/66B, type 200GBASE-R and 400GBASE-R
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[4]*Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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119.6.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
MII	200GMII or 400GMII logical interface	117, 119.1.4.1	Logical interface is supported	O	Yes [ ] No [ ]
*PCS200	PCS for 200GBASE-R	119.1.1		O.1	Yes [ ] No [ ]
*PCS400	PCS for 400GBASE-R	119.1.1		O.1	Yes [ ] No [ ]
*MD	MDIO	45, 119.3	Registers and interface supported	O	Yes [ ] No [ ]
*BI	Bypass indication	119.2.5.3	Capability is supported	O	Yes [ ] No [ ]
DC	Delay constraints	119.5	Conforms to delay constraints specified in 119.5	M	Yes [ ]
*EEE	EEE capability	119.2.3.3	Capability is supported	O	Yes [ ] No [ ]
JTM	Supports test-pattern mode	119.2.1		M	Yes [ ]
FDD	Support for optional FEC degraded detection	119.2.5.3	In the FEC decoder can optionally detect FEC SER degraded at a programmable threshold	O	Yes [ ] No [ ]

119.6.4 PICS proforma tables for Physical Coding Sublayer (PCS) 64B/66B, type 200GBASE-R and 400GBASE-R

119.6.4.1 Transmit function

Item	Feature	Subclause	Value/Comment	Status	Support
TF1	64B/66B to 256B/257B transcoder	119.2.4.2	tx_xcoded<256:0> constructed per 119.2.4.2	M	Yes [ ]
TF2	Transmission bit ordering	119.2.4.8	First bit transmitted is bit 0	M	Yes [ ]
TF3	Pad value	119.2.4.4	PRBS9	M	Yes [ ]
TF4	Alignment marker insertion	119.2.4.4		M	Yes [ ]
TF5	Pre-FEC distribution	119.2.4.5	Distribute the data to two FEC codewords	M	Yes [ ]
TF6	Reed-Solomon encoder	119.2.4.6	RS(544,514)	M	Yes [ ]
TF7	Symbol distribution	119.2.4.7	Distribution is based on 10b symbols	M	Yes [ ]

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119.6.4.2 Receive function

Item	Feature	Subclause	Value/Comment	Status	Support
RF1	Skew tolerance	119.2.5.1	Maximum Skew of 180 ns between PCS lanes and a maximum Skew Variation of 4 ns	M	Yes [ ]
RF2	Lane reorder and de-interleave	119.2.5.2	Order the PCS lanes according to the PCS lane number and de-interleave the FEC codewords	M	Yes [ ]
RF3	Reed-Solomon decoder	119.2.5.3	Corrects any combination of up to *t*=15 symbol errors in a codeword	M	Yes [ ]
RF4	Reed-Solomon decoder	119.2.5.3	Capable of indicating when a codeword was not corrected.	M	Yes [ ]
RF5	Error monitoring while error indication is bypassed	119.2.5.3	When the number of symbol errors in a block of 8192 codewords exceeds 5560 assert hi_ser for 60 ms to 75 ms	BI:M	Yes [ ] N/A [ ]
RF6	256B/257B to 64B/66B transcoder	119.2.5.7	rx_coded_*j*<65:0>, *j*=0 to 3 constructed per 119.2.5.7	M	Yes [ ]

119.6.4.3 64B/66B coding rules

Item	Feature	Subclause	Value/Comment	Status	Support
C1	Encoder (and ENCODE function) implements the code as specified	119.2.3, 119.2.6.2.3		M	Yes [ ]
C2	Decoder (and DECODE function) implements the code as specified	119.2.3, 119.2.6.2.3		M	Yes [ ]
C3	Only valid block types are transmitted	119.2.3.2		M	Yes [ ]
C4	Invalid block types are treated as an error	119.2.3.2		M	Yes [ ]
C5	Only valid control characters are transmitted	119.2.3.3		M	Yes [ ]
C6	Invalid control characters are treated as an error	119.2.3.3		M	Yes [ ]
C7	If EEE has not been negotiated, LPI is not transmitted	119.2.3.3		EEE:M	Yes [ ] N/A [ ]
C8	If EEE has not been negotiated, LPI is treated as an error if received	119.2.3.3		EEE:M	Yes [ ] N/A [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
C9	Idles do not interrupt data	119.2.3.5		M	Yes [ ]
C10	IDLE control code insertion and deletion	119.2.3.5	Insertion or Deletion in groups of 8 /I/s	M	Yes [ ]
C11	Sequence ordered set deletion	119.2.3.8	Only one whole ordered set of two consecutive sequence ordered sets may be deleted	M	Yes [ ]

119.6.4.4 Scrambler and descrambler

Item	Feature	Subclause	Value/Comment	Status	Support
S1	Scrambler	119.2.4.3	Performs as shown in Figure 49–8	M	Yes [ ]
S2	Descrambler	119.2.5.6	Performs as shown in Figure 49–10	M	Yes [ ]

119.6.4.5 Alignment markers

Item	Feature	Subclause	Value/Comment	Status	Support
AM1	Alignment marker insertion	119.2.4.4	Alignment markers are inserted periodically as in 119.2.4.4	M	Yes [ ]
AM2	Alignment marker form	119.2.4.4	Alignment markers are formed as described in 119.2.4.4	M	Yes [ ]
AM3	Lane mapping	119.2.6.3	PCS lane number is captured	MD:M	Yes [ ] N/A [ ]
AM4	Alignment marker removal	119.2.5.5	Alignment markers are removed prior to descrambling as described in 119.2.5.5	M	Yes [ ]

119.6.4.6 Test-pattern modes

Item	Feature	Subclause	Value/Comment	Status	Support
JT1	Scrambled idle transmit test-pattern generator is implemented	119.2.4.9		M	Yes [ ]

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119.6.4.7 Bit order

Item	Feature	Subclause	Value/Comment	Status	Support
B1	Transmit bit order	119.2.4.8	Placement of bits into the PCS lanes as shown in Figure 119–10 or Figure 119–11	M	Yes [ ]

119.6.4.8 Management

Item	Feature	Subclause	Value/Comment	Status	Support
M1	Alternate access to PCS management object is provided	119.3		O	Yes [ ] No [ ]
M2	Mapping of MDIO control bits and MDIO status bits	119.3.1	Table 119–4 and Table 119–5	MD:M	Yes [ ] N/A [ ]

119.6.4.9 State diagrams

Item	Feature	Subclause	Value/Comment	Status	Support
SM1	Alignment marker lock	119.2.6	Implements 8 alignment marker lock processes as depicted in Figure 119–12	PCS200:M	Yes [ ] N/A [ ]
SM2	Alignment marker lock	119.2.6	Implements 16 alignment marker lock processes as depicted in Figure 119–12	PCS400:M	Yes [ ] N/A [ ]
SM3	The SLIP function evaluates all possible block positions	119.2.6.2.3		M	Yes [ ]
SM4	PCS synchronization state diagram	119.2.6	Meets the requirements of Figure 119–13	M	Yes [ ]
SM5	Transmit process	119.2.6	Meets the requirements of Figure 119–14	M	Yes [ ]
SM6	Receive process	119.2.6	Meets the requirements of Figure 119–15	M	Yes [ ]

119.6.4.10 Loopback

Item	Feature	Subclause	Value/Comment	Status	Support
L1	Supports loopback	119.4		M	Yes [ ]
L2	When in loopback, transmits what it receives from the 200GMII/400GMII	119.4		M	Yes [ ]
L3	When in loopback, ignore all data presented by the PMA sublayer.	119.4		M	Yes [ ]

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119.6.4.11 Delay constraints

Item	Feature	Subclause	Value/Comment	Status	Support
TIM1	PCS delay constraint	119.5	No more than 160 256 BT for sum of transmit and receive path delays for 200GBASE-R.	PCS200: M	Yes [ ] N/A [ ]
TIM2	PCS delay constraint	119.5	No more than 320 000 BT for sum of transmit and receive path delays for 400GBASE-R.	PCS400: M	Yes [ ] N/A [ ]

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120. Physical Medium Attachment (PMA) sublayer, type 200GBASE-R and 400GBASE-R

120.1 Overview

120.1.1 Scope

This clause specifies the Physical Medium Attachment sublayer (PMA) for the families of 200 Gb/s and 400 Gb/s Physical Layer implementations, known as 200GBASE-R and 400GBASE-R. The PMA allows the PCS (specified in Clause 119) to connect in a media-independent way with a range of physical media, or for the DTE XS to connect to the PHY XS (specified in Clause 118). The 200GBASE-R PMA(s) can support any of the 200 Gb/s PMDs in Table 116–1, and the 400GBASE-R PMA(s) can support any of the 400 Gb/s PMDs in Table 116–2. The terms 200GBASE-R and 400GBASE-R are used when referring generally to Physical Layers using the PMA defined in this clause.

200GBASE-R and 400GBASE-R can be extended to support any full duplex medium requiring only that the PMD be compliant with the appropriate PMA interface.

The interfaces for the inputs of the 200GBASE-R and 400GBASE-R PMAs are defined in an abstract manner and do not imply any particular implementation. For 200GBASE-R PMAs, electrical interfaces connecting PMA sublayers, known as 200GAUI-n, are defined in Annex 120B, Annex 120C, Annex 120D, and Annex 120E. For 400GBASE-R PMAs, electrical interfaces connecting PMA sublayers, known as 400GAUI-n, are defined in Annex 120B, Annex 120C, Annex 120D, and Annex 120E.

120.1.2 Position of the PMA in the 200GBASE-R and 400GBASE-R sublayers

Figure 120–1 shows the relationship of the PMA sublayer (shown shaded) with other sublayers to the ISO Open System Interconnection (OSI) reference model.

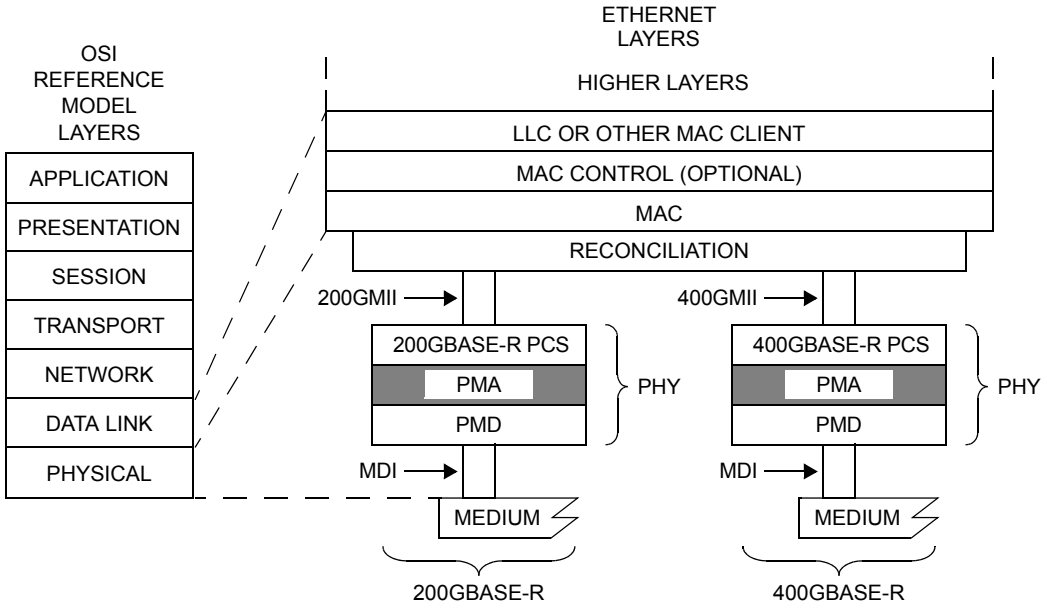
120.1.3 Summary of functions

The following is a summary of the principal functions implemented (when required) by the PMA in both the transmit and receive directions:

a) Adapt the PCSL formatted signal to the appropriate number of abstract or physical lanes.
b) Provide per input-lane clock and data recovery.
c) Provide bit-level multiplexing.
d) Provide clock generation.
e) Provide signal drivers.
f) Optionally provide local loopback to/from the PMA service interface.
g) Optionally provide remote loopback to/from the PMD service interface.
h) Optionally provide test-pattern generation and detection.
i) Tolerate Skew Variation.
j) Perform PAM4 encoding and decoding for 200GBASE-R PMAs where the number of physical lanes is 4, and for 400GBASE-R PMAs where the number of physical lanes is 4 or 8.

In addition, the PMA provides receive link status information in the receive direction.

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200GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
LLC = LOGICAL LINK CONTROL
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120–1—200GBASE-R and 400GBASE-R PMA relationship to the ISO/IEC Open Systems Interconnection (OSI) reference model and IEEE 802.3 Ethernet model

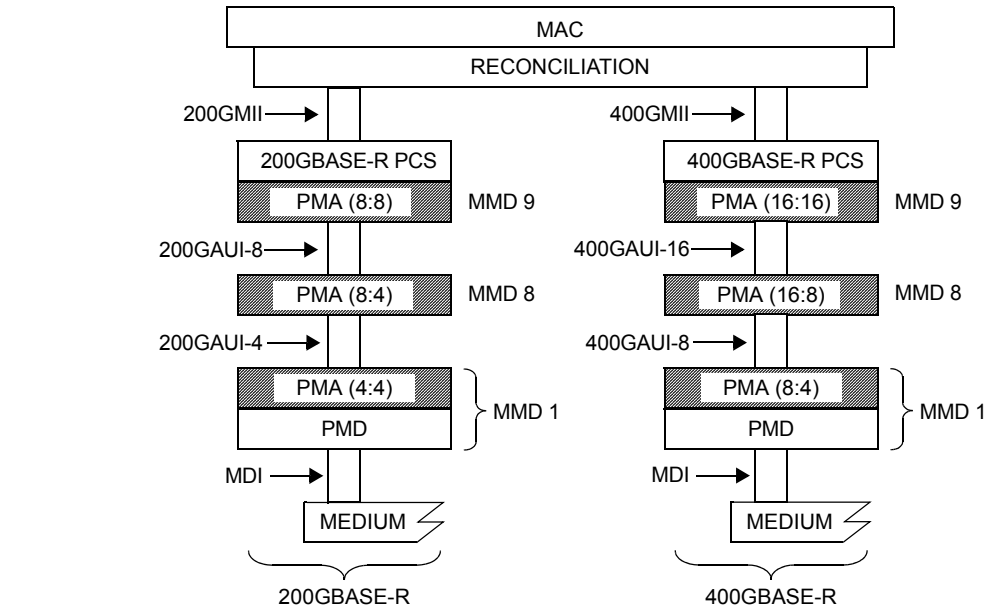
120.1.4 PMA sublayer positioning

An implementation may use one or more PMA sublayers to adapt the number and rate of the PCS lanes to the number and rate of the PMD lanes. The number of PMA sublayers required depends on the partitioning of functionality for a particular implementation. An example is illustrated in Figure 120–2. Additional examples are illustrated in Annex 120A. Each PMA maps the PCSLs from p PMA input lanes to q PMA output lanes in the Tx direction, and from q PMA input lanes to p PMA output lanes in the Rx direction.

Management Data Input/Output (MDIO) Manageable Device (MMD) addresses 1, 8, 9, and 10 are available for addressing multiple instances of PMA sublayers (see Table 45–1 for MMD device addresses). If the PMA sublayer that is closest to the PMD is packaged with the PMD, it shares MMD 1 with the PMD. More addressable instances of PMA sublayers, each one separated from lower addressable instances by chip-to-chip interfaces, may be implemented and addressed allocating MMD addresses to PMAs in increasing numerical order going from the PMD toward the MAC. The example shown in Figure 120–2 could be implemented with three addressable instances: MMD 1 addressing the lowest PMA sublayer (co-packaged with the PMD), MMD 8 addressing the PMA sublayer above the 200GAUI-4 below the 200GAUI-8 or above the 400GAUI-8 below the 400GAUI-16, and MMD 9 addressing the PMA sublayer below the PCS.

The number of input lanes and the number of output lanes for a PMA are always divisors of the number of PCSLs. For PMA sublayers supporting 200GBASE-R PMDs, the number of PCSLs is 8. For PMA sublayers supporting 400GBASE-R PMDs, the number of PCSLs is 16.

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200GAUI = 200 Gb/s ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120–2—Example 200GBASE-R and 400GBASE-R PMA layering

The following guidelines apply to the partitioning of PMAs:

a) The inter-sublayer service interface, defined in 116.3.1, is used for the PMA service interfaces supporting a flexible architecture with multiple PMA sublayers.
   1) An instance of this interface can only connect service interfaces with the same number of lanes, where the lanes operate at the same rate.

b) 200GAUI-n is a physical instantiation of the connection between two adjacent 200GBASE-R PMA sublayers with the exception of the inst:IS_SIGNAL.indication which is carried outside of this physically instantiated interface. 400GAUI-n is a physical instantiation of the connection between two adjacent 400GBASE-R PMA sublayers with the exception of the inst:IS_SIGNAL.indication which is carried outside of this physically instantiated interface.
   1) As physical instantiations, these define electrical and timing specification as well as requiring a receive re-timing function.
   2) 200GAUI-8 is a 26.5625 GBd by 8 lane NRZ physical instantiation of the 200 Gb/s connection. 400GAUI-16 is a 26.5625 GBd by 16 lane NRZ physical instantiation of the 400 Gb/s connection.
   3) 200GAUI-4 is a 26.5625 GBd by 4 lane PAM4 physical instantiation of the 200 Gb/s connection. 400GAUI-8 is a 26.5625 GBd by 8 lane PAM4 physical instantiation of the 400 Gb/s connection.

c) The abstract inter-sublayer service interface can be physically instantiated as a 200GAUI-n or 400GAUI-n, using associated PMAs to map to the appropriate number of lanes.

d) Opportunities for optional test-pattern generation, optional test-pattern detection, optional local loopback and optional remote loopback are dependent upon the location of the PMA sublayer in the implementation. See Figure 120–5.

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e) A minimum of one PMA sublayer is required in a PHY.

f) A maximum of four PMA sublayers are addressable as MDIO MMDs.

120.2 PMA interfaces

All PMA variants for 200GBASE-R and 400GBASE-R are based on a generic specification of a bit mux function that applies to all input/output lane counts and each direction of transmission. Each direction of transmission may employ one or more such bit muxes to adapt from the appropriate number of input lanes to the appropriate number of output lanes as illustrated in Figure 120–3.

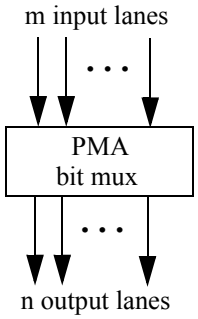


Figure 120–3—PMA bit mux used in both Tx and Rx directions

Conceptually, the PMA bit mux operates in one direction of transmission by demultiplexing PCSLs from m PMA input lanes and remultiplexing them into n PMA output lanes. The mapping of PCSLs from input to output lanes is not specified. See 120.5.2 and Figure 120–4 for details.

Figure 120–5 provides the functional block diagram of a PMA. The parameters of a PMA include the following:

— The numbers of input and output lanes in each direction.

— Whether the PMA is adjacent to a physically instantiated interface (200GAUI-n or 400GAUI-n above or below).

— Whether the PMA is adjacent to the PCS or DTE XS.

— Whether the PMA is adjacent to the PMD or PHY XS.

120.3 PMA service interface

The PMA service interface for 200GBASE-R and 400GBASE-R is an instance of the inter-sublayer service interface defined in 116.3.1. The PMA service interface primitives are summarized as follows:

PMA:IS_UNITDATA_*i*.request(tx_symbol)
PMA:IS_UNITDATA_*i*.indication(rx_symbol)
PMA:IS_SIGNAL.indication(SIGNAL_OK)

For a PMA service interface with p lanes, the primitives are defined for $i = 0$ to $p - 1$.

If the PMA client is the PCS or DTE XS, the PMA continuously sends eight (for 200GBASE-R) or sixteen (for 400GBASE-R) parallel bit streams to the PCS, each at the nominal signaling rate of the PCSL.

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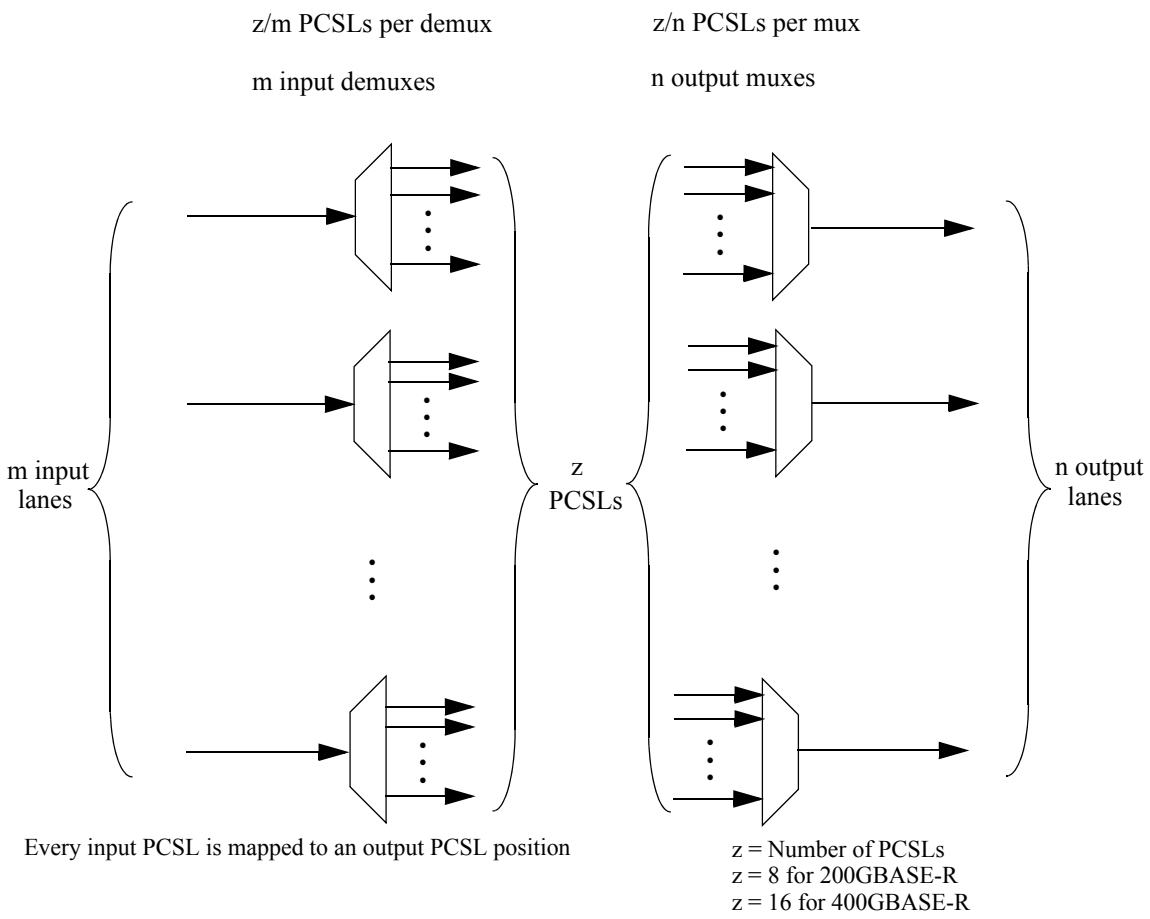
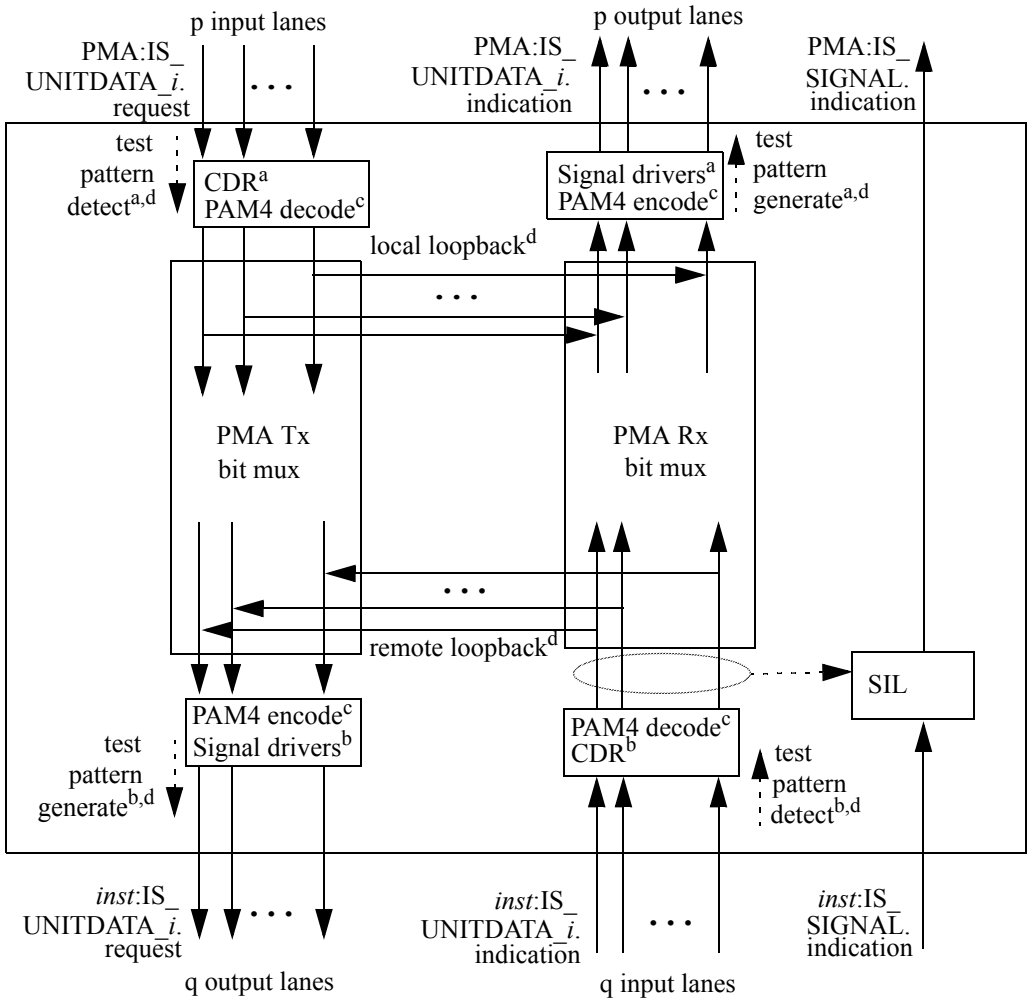


Figure 120–4—PMA bit mux operation used in both Tx and Rx directions

If a PMA client is another PMA, a 200GBASE-R PMA with p=8 or a 400GBASE-R PMA with p=16 physical input lanes receives bits from one PCSL on each of its input lanes. A 200GBASE-R PMA with p=4 physical input lanes receives PAM4 symbols on each of its input lanes at the PCSL rate, each symbol formed from two bits. A 400GBASE-R PMA with p=4 or p=8 physical input lanes receives PAM4 symbols on each of its input lanes at 8/p times the PCSL rate, each symbol formed from two bits. The bit stream represented by the input symbols carries z/p bit-multiplexed PCSLs on each physical input lane, where z=8 for 200GBASE-R and z=16 for 400GBASE-R. Skew may exist between different PCSLs received on the same physical input lane even though all PCSLs originate from the same synchronous source, so there is independence of arrival of bits from each PCSL on each physical input lane.

In the Tx direction, if the symbol from a PMA:IS_UNITDATA_*i*.request primitive is received over a physically instantiated interface (200GAUI-n or 400GAUI-n), clock and data are recovered on the lane receiving the symbol. If necessary, PAM4 symbols are converted to pairs of bits. Bits are routed through the PMA to an output lane through a process that may demultiplex PCSLs from the input, perform any necessary buffering to tolerate Skew Variation across input lanes, and multiplex PCSLs to output lanes. If necessary, pairs of bits on output lanes are converted to PAM4 symbols. The symbols are sent on an output lane to the sublayer below using the *inst*:IS_UNITDATA_*k*.request (*k* not necessarily equal to *i*) primitive (see 120.4).

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inst PMD, PMA, or PHY_XS, depending on which sublayer is below this PMA
SIL Signal Indication Logic

[a] If physically instantiated interface (200GAUI-n or 400GAUI-n) immediately above this PMA.
[b] If physically instantiated interface (200GAUI-n or 400GAUI-n) immediately below this PMA, or if this is the closest PMA to the PMD.
[c] If number of input or output lanes is 4 for a 200GBASE-R PMA, or 4 or 8 for a 400GBASE-R PMA.
[d] Optional.

Figure 120–5—PMA Functional Block Diagram

The PMA passes symbols from the input lanes to the output lanes in the Rx direction when data is being received from every input lane from the sublayer below the PMA that has a PCSL that is routed to a particular output lane at the PMA service interface, and (if necessary), buffers are filled to allow tolerating the Skew Variation that may appear between the input lanes. PCSLs are demultiplexed from the input lanes, remultiplexed to the output lanes, and symbols are transferred over each output lane to the PMA client via the PMA:IS_UNITDATA_*i*.indication primitive. If necessary, PAM4 symbols are converted to pairs of bits on the input lanes and/or pairs of bits are converted to PAM4 symbols on the output lanes.

The PMA:IS_SIGNAL.indication primitive is generated through a set of Signal Indication Logic (SIL) that reports signal health based on receipt of the *inst*:IS_SIGNAL.indication from the sublayer below, data being received on all of the input lanes from the sublayer below, buffers filled (if necessary) to accommodate Skew Variation, and symbols being sent to the PMA client on all of the output lanes. When these conditions

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are met, the SIGNAL_OK parameter sent to the PMA client via the PMA:IS_SIGNAL.indication primitive has the value OK. Otherwise, the SIGNAL_OK primitive has the value FAIL. In the case where the sublayer below the PMA is a PHY XS the PMA does not receive a PHY_XS:IS_SIGNAL.indication as an input to the SIL.

120.4 Service interface below PMA

Since the architecture supports multiple PMA sublayers for various PMD lane counts and device partitioning, there are several different sublayers that may appear below a PMA, including the PMD, an extender sublayer, or another PMA. The variable *inst* represents whichever sublayer appears below the PMA (e.g., another PMA or PMD).

The sublayer below the PMA utilizes the inter-sublayer service interface defined in 116.3. The service interface primitives provided to the PMA are summarized as follows:

inst:IS_UNITDATA_*i*.request(tx_symbol)
inst:IS_UNITDATA_*i*.indication(rx_symbol)
inst:IS_SIGNAL.indication(SIGNAL_OK)

The number of lanes q for the service interface matches the number of lanes expected by the PMA. The *inst*:IS_UNITDATA_*i* primitives are defined for each lane $i = 0$ to $q - 1$ of the service interface below the PMA. Note that electrical and timing specifications of the service interface are defined if the interface is physically instantiated (e.g., 200GAUI-n or 400GAUI-n), otherwise the service interface is specified only abstractly. The interface between the PMA and the sublayer below consists of q lanes for data transfer and a status indicating a good signal from the sublayer below the PMA (see Figure 120–5).

The PMA transfers symbols from the input lanes to the output lanes in the Tx direction when data is being received via the PMA:IS_UNITDATA_*i*.request primitive from every input lane from the PMA client at the PMA service interface (see 120.3) that has a PCSL that is routed to this output lane, and (if necessary), buffers are filled to provide the ability to tolerate the Skew Variation that may appear between the input lanes from the PMA client. PCSLs are demultiplexed from the input lanes, remultiplexed to the output lanes, and symbols are transferred over each output lane to the sublayer below the PMA.

In the Rx direction, if the symbol is received over a physically instantiated interface (200GAUI-n, 400GAUI-n, or physically instantiated PMD service interface), clock and data are recovered on the lane receiving the symbol. If necessary, PAM4 symbols received on the input lanes are converted to pairs of bits. The bits are routed through the PMA to an output lane toward the PMA client through a process that may demultiplex PCSLs from the input, perform any necessary buffering to tolerate Skew Variation across input lanes, and multiplex PCSLs to output lanes. If necessary, pairs of bits are converted to PAM4 symbols on the output lanes. Each symbol is sent on an output lane to the PMA client using the PMA:IS_UNITDATA_*k*.indication (*k* not necessarily equal to *i*) primitive at the PMA service interface.

In the case where the sublayer below the PMA is a PHY XS, there is an additional primitive:

PHY_XS:IS_SIGNAL.request(SIGNAL_OK)

The PHY_XS:IS_SIGNAL.request primitive is generated through a set of SIL that reports signal health based on data being received on all of the input lanes from the sublayer above, buffers filled (if necessary) to accommodate Skew Variation, and symbols being sent to the PHY XS on all of the output lanes. When these conditions are met, the SIGNAL_OK parameter sent to the PHY XS via the PHY_XS:IS_SIGNAL.request primitive has the value OK. Otherwise, the SIGNAL_OK primitive has the value FAIL.

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120.5 Functions within the PMA

The purpose of the PMA is to adapt the PCSL formatted signal to an appropriate number of abstract or physical lanes, to recover clock from the received signal (if appropriate), and optionally to provide test signals and loopback. Each input (Tx direction) or output (Rx direction) lane between the PMA and the PMA client carries one or more PCSLs that are bit-multiplexed. All input and output lanes between the PMA and the PMA client carry the same number of PCSLs and operate at the same nominal signaling rate. Likewise, each input (Rx direction) or output (Tx direction) lane between the PMA and the sublayer below the PMA carries one or more PCSLs that are bit-multiplexed. All input and output lanes between the PMA and the sublayer below the PMA carry the same number of PCSLs and operate at the same nominal signaling rate. As described in 120.1.4, the number of input lanes and the number of output lanes for a given PMA are divisors of 8 for 200GBASE-R, or 16 for 400GBASE-R, which are the number of PCSLs for the respective PHYs. The symbols received from input lanes or sent on output lanes are bits for 200GBASE-R interfaces where the number of lanes is 8, or 400GBASE-R interfaces where the number of lanes is 16. The symbols received from input lanes or sent on output lanes are PAM4 symbols for 200GBASE-R interfaces where the number of lanes is 4, or 400GBASE-R interfaces where the number of lanes is 8 or 4.

120.5.1 Per input-lane clock and data recovery

If the interface between the PMA client and the PMA is physically instantiated as 200GAUI-n or 400GAUI-n, the PMA shall meet the electrical and timing specifications in Annex 120B, Annex 120C, Annex 120D, or Annex 120E as appropriate. If the interface between the sublayer below the PMA and the PMA is physically instantiated as 200GAUI-n or 400GAUI-n, the PMA shall meet the electrical and timing specifications at the service interface below the PMA as specified in Annex 120B, Annex 120C, Annex 120D, or Annex 120E as appropriate.

Test patterns that are intended for transmitter testing, such as square wave, may not be correctly recovered by an adjacent PMA.

120.5.2 Bit-level multiplexing

The PMA provides bit-level multiplexing in both the Tx and Rx directions. In the Tx direction, the function is performed among the bits received from the PMA client via the PMA:IS_UNITDATA_*i*.request primitives (for PMA client lanes $i = 0$ to $p - 1$) with the result sent to the service interface below the PMA using the *inst*:IS_UNITDATA_*i*.request primitives (for service interface lanes $i = 0$ to $q - 1$), referencing the functional block diagram shown in Figure 120–5. The bit multiplexing behavior is illustrated in Figure 120–4.

The aggregate signal carried by the group of input lanes or the group of output lanes is arranged as a set of PCSLs. The number of PCSLs z is 8 for 200GBASE-R interfaces and 16 for 400GBASE-R interfaces. The nominal bit rate R of each PCSL is 26.5625 Gb/s.

For a PMA with m input lanes (Tx or Rx direction), each input lane carries, bit multiplexed, z/m PCSLs. Each input lane has a nominal bit rate of 26.5625 × z/m Gb/s. Note that the signaling (Baud) rate is equal to the bit rate when the number of physical lanes is 8 for 200GBASE-R or 16 for 400GBASE-R (bits are sent or received on the lanes). The Baud rate is equal to half of the bit rate when the number of physical lanes is 4 for 200GBASE-R or the number of physical lanes is 8 or 4 for 400GBASE-R (PAM4 symbols are sent or received on the lanes). If necessary, PAM4 symbols are converted to pairs of bits on the input lanes and/or pairs of bits are converted to PAM4 symbols on the output lanes. If bit x received on an input lane belongs to a particular PCSL, the next bit of that same PCSL is received on the same input lane at bit position $x+(z/m)$. The z/m PCSLs may arrive in any sequence on a given input lane.

For a PMA with n output lanes (Tx or Rx direction), each output lane carries, bit multiplexed, z/n PCSLs. Each output lane has a nominal signaling rate of 26.5625 × z/n Gb/s. Each PCSL is mapped from a position

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in the sequence on one of the m input lanes to a position in the sequence on one of the n output lanes. If bit *x* sent on an output lane belongs to a particular PCSL, the next bit of that same PCSL is sent on the same output lane at bit position *x* + (z/n). The PMA shall maintain the chosen sequence of PCSLs on all output lanes while it is receiving a valid stream of bits on all input lanes.

Each PCSL received in any temporal position on an input lane is transferred into a temporal position on an output lane. As the PCS (see Clause 119) has fully flexible receive logic, an implementation is free to perform the mapping of PCSLs from input lanes to output lanes without constraint. Figure 120–6 illustrates one possible bit ordering for a 400GBASE-R 8:4 PMA bit mux. Other bit orderings are also valid.

Note that since the number of input lanes and output lanes for a 200GBASE-R or 400GBASE-R PMA is always a power of two, many PMAs converting between different numbers of lanes normally simply multiplex two or four input lanes to one output lane, or demultiplex two or four output lanes from one input lane. However, any PMA implementation which produces an allowable order of bits from all PCSLs on the output lanes is valid.

NOTE—PMA output lanes composed of some specific combinations of four PCSLs with specific skew offsets (e.g., 400GBASE-R PCSLs 0, 2, 4, and 10 with delays 0, 1, 0, and 2 bits, respectively) may have reduced transition density.

120.5.3 Skew and Skew Variation

The Skew (relative delay) between the PCSLs must be kept within limits so that the information on the lanes can be reassembled by the PCS.

Any PMA that combines PCSLs from different input lanes onto the same output lane must tolerate Skew Variation between the input lanes without changing the PCSL positions on the output. Skew and Skew Variation are defined in 116.5. The limits for Skew and Skew Variation at physically instantiated interfaces are specified at Skew points SP1 and SP2 in the transmit direction and SP5 and SP6 in the receive direction as defined in 116.5 and illustrated in Figure 116–4 and Figure 116–5.

120.5.3.1 Skew generation toward SP1

In an implementation with one or more physically instantiated 200GAUI-n or 400GAUI-n interfaces, the PMA that sends data in the transmit direction toward the 200GAUI-n or 400GAUI-n that is closest to the PMD (SP1 in Figure 116–4 and Figure 116–5) shall produce no more than 29 ns of Skew between PCSLs toward the 200GAUI-n or 400GAUI-n, and no more than 200 ps of Skew Variation.

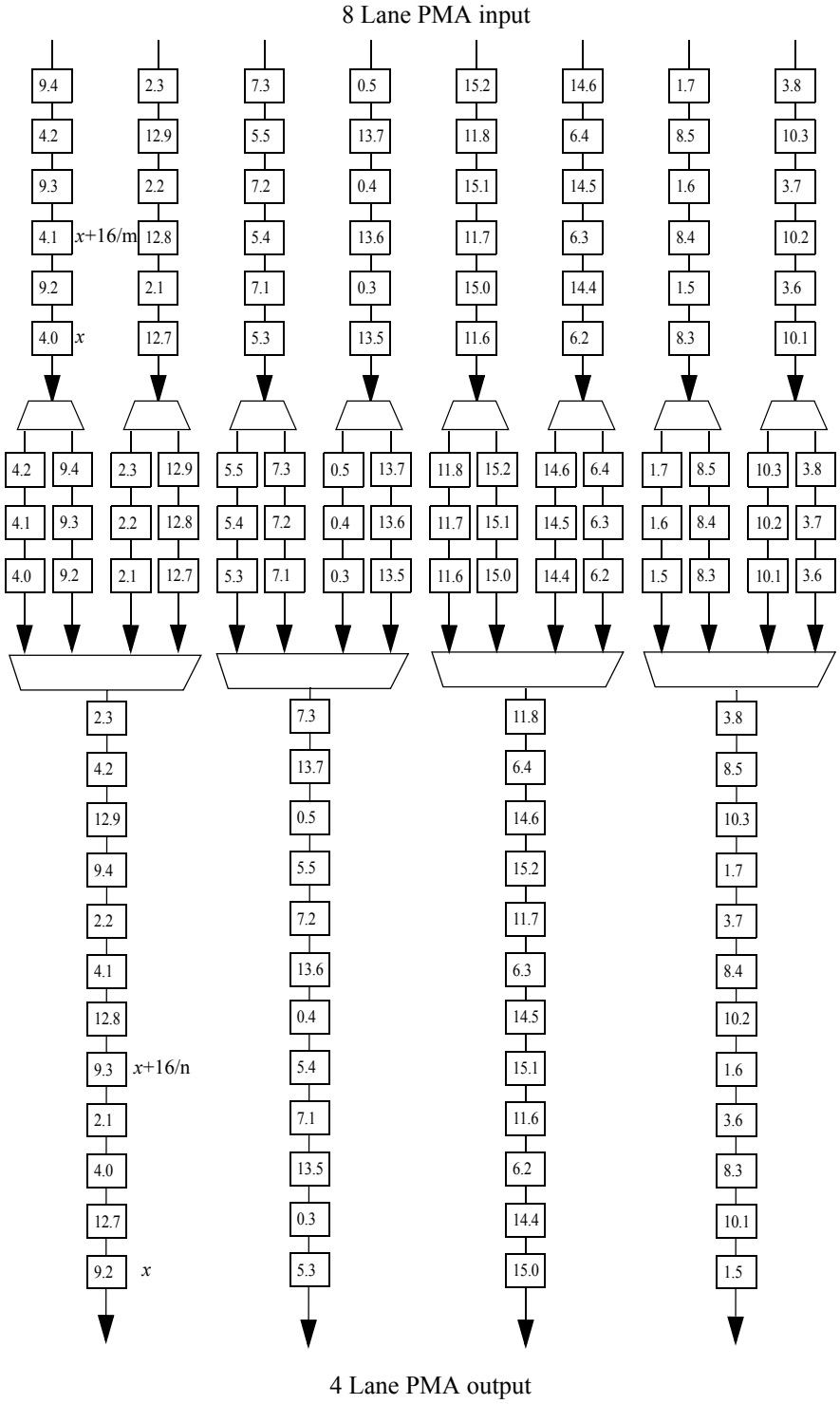
120.5.3.2 Skew tolerance at SP1

In an implementation with one or more physically instantiated 200GAUI-n or 400GAUI-n interfaces, the PMA service interface that receives data in the transmit direction from the 200GAUI-n or 400GAUI-n (SP1 in Figure 116–4 and Figure 116–5) shall tolerate the maximum amount of Skew Variation allowed at SP1 (200 ps) between input lanes while maintaining the bit ordering and position of each PCSL on each PMA lane in the transmit direction (toward the PMD).

120.5.3.3 Skew generation toward SP2

If the PMD service interface is physically instantiated so that the Skew at SP2 can be measured, the PMA adjacent to the PMD service interface shall generate no more than 43 ns of Skew, and no more than 400 ps of Skew Variation between output lanes toward the PMD service interface (SP2 in Figure 116–4 and Figure 116–5). If there is a physically instantiated 200GAUI-n or 400GAUI-n as well, then the Skew measured at SP1 is limited to no more than 29 ns of Skew and no more than 200 ps of Skew Variation.

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NOTE—i.k indicates bit k on PCSL i. Skew may exist between PCSLs.

Figure 120–6—Example 8:4 400GBASE-R PMA bit mux

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120.5.3.4 Skew tolerance at SP5

If the PMD service interface is physically instantiated so that the Skew at SP5 can be measured, the PMA adjacent to the PMD service interface (SP5) shall tolerate the maximum amount of Skew Variation allowed at SP5 (3.6 ns) between output lanes from the PMD service interface while maintaining the bit ordering and position of each PCSL on each PMA lane in the receive direction (toward the MAC).

120.5.3.5 Skew generation at SP6

In an implementation with one or more physically instantiated 200GAUI-n or 400GAUI-n interfaces, at SP6 (the receive direction of the 200GAUI-n or 400GAUI-n closest to the PCS), the PMA or group of PMAs between the PMD and the 200GAUI-n or 400GAUI-n closest to the PCS shall deliver no more than 160 ns of Skew, and no more than 3.8 ns of Skew Variation between output lanes toward the 200GAUI-n or 400GAUI-n in the Rx direction. If there is a physically instantiated PMD service interface that allows the Skew to be measured, the Skew measured at SP5 is limited to no more than 145 ns of Skew and no more than 3.6 ns of Skew Variation. If there is no physically instantiated PMD service interface, the Skew measured at SP4 is limited to no more than 134 ns of Skew, and no more than 3.4 ns of Skew Variation.

120.5.3.6 Skew tolerance at SP6

In an implementation with one or more physically instantiated 200GAUI-n or 400GAUI-n interfaces, the PMA between the 200GAUI-n or 400GAUI-n closest to the PCS and the PCS shall tolerate the maximum amount of Skew Variation allowed at SP6 (3.8 ns) between input lanes while maintaining the bit order and position of PCSLs on lanes sent in the receive direction towards the MAC.

120.5.4 Delay constraints

The maximum cumulative delay contributed by up to four PMA stages in a PHY (sum of transmit and receive delays at one end of the link) shall meet the values specified in Table 120–1. A description of overall system delay constraints and the definitions for bit-times and pause_quanta can be found in 116.4 and its references.

Table 120–1—Delay constraints

Sublayer	Maximum (bit time)	Maximum (pause_quanta)	Maximum (ns)
200GBASE-R PMA	18 432	36	92.16
400GBASE-R PMA	36 864	72	92.16

120.5.5 Clocking architecture

A PMA with m input lanes and n output lanes shall clock the output lanes such that the bit rate of the output lanes is m/n times the bit rate of the input lanes. For input or output lanes encoded as PAM4 (for 200GBASE-R, where the number of input or output lanes is 4, or for 400GBASE-R, where the number of input or output lanes is 4 or 8), the symbol rate is half of the bit rate. This applies in both the Tx and Rx directions of transmission. In the case where the interfaces between the PMA client and the PMA and/or the PMA and the sublayer below the PMA are physically instantiated, the PMA may derive its input clock(s) from the electrical interface on one or more of the input lanes, and generate the output clock(s) with an appropriate PLL multiplier/divider circuit.

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There is no requirement that the PMA clock all output lanes in unison. Examples of independent clocking of output lanes include the following:

— The case where the number of input and output lanes are equal (the PMA is provided for retiming and regeneration of the signal). This may be implemented without any rearrangement of PCSLs between input lanes and output lanes (although rearrangements are allowed), and such a PMA may be implemented by driving each output lane using the clock recovered from the corresponding input lane.

— Since the number of PCSLs is a power of 2, whenever the number of input and output lanes are unequal, each input lane could be mapped to multiple output lanes (in the case where the number of output lanes is greater), or multiple input lanes could be mapped to a single output lane.

120.5.6 Signal drivers

For cases where the interface between the PMA client and the PMA, or between the PMA and the sublayer below the PMA represent a physically instantiated interface, the PMA provides electrical signal drivers for that interface. The electrical and jitter/timing specifications for these interfaces appear in

— Annex 120B, which specifies the 200GAUI-8 and 400GAUI-16 interfaces for chip-to-chip applications.

— Annex 120C, which specifies the 200GAUI-8 and 400GAUI-16 interfaces for chip-to-module applications.

— Annex 120D, which specifies the 200GAUI-4 and 400GAUI-8 interfaces for chip-to-chip applications.

— Annex 120E, which specifies the 200GAUI-4 and 400GAUI-8 interfaces for chip-to-module applications.

For 200GAUI-8 or 400GAUI-16, the modulation format is NRZ. For 200GAUI-4 or 400GAUI-8, the modulation format is PAM4.

120.5.7 Gray mapping for PAM4 encoded lanes

For output lanes encoded as PAM4 (for 200GBASE-R, where the number of output lanes is 4, or for 400GBASE-R, where the number of output lanes is 4 or 8), the PMA transmit process shall map consecutive pairs of bits {A, B}, where A is the bit arriving first, to a Gray-coded symbol as follows:

{0, 0} maps to 0,
{0, 1} maps to 1,
{1, 1} maps to 2, and
{1, 0} maps to 3.

For input lanes encoded as PAM4 (for 200GBASE-R, where the number of input lanes 4, or for 400GBASE-R, where the number of input lanes is 4 or 8), the PMA receive process shall map Gray-coded PAM4 symbols to pairs of bits {A, B} where A is considered to be the first bit as follows:

0 maps to {0, 0},
1 maps to {0, 1},
2 maps to {1, 1}, and
3 maps to {1, 0}.

120.5.8 Link status

The PMA shall provide link status information to the PMA client using the PMA:IS_SIGNAL.indication primitive. The PMA continuously monitors the link status reported by the service interface below from the

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inst:IS_SIGNAL.indication primitive, and uses this as input to Signal Indication Logic (SIL) to determine the link status to report to the layer above. Other inputs to the SIL may include the status of clock and data recovery on the lanes from the service interface below the PMA and whether buffers/FIFOs have reached the required fill level to accommodate Skew Variation so that data is being sent on the output lanes.

120.5.9 PMA local loopback mode (optional)

PMA local loopback mode is optional. If it is implemented, it shall be as described in this subclause (120.5.9).

The PMA sublayer may provide a local loopback function. The function involves looping back each input lane to the corresponding output lane. Each bit received from the PMA:IS_UNITDATA_*i*.request(tx_symbol) primitive is looped back in the direction of the MAC using the PMA:IS_UNITDATA_*i*.indication(rx_symbol) primitive.

During local loopback, the PMA performs normal bit muxing of PCSLs per 120.5.2 onto the lanes in the Tx direction toward the service interface below the PMA.

Ability to perform this function is indicated by the Local_loopback_ability status variable. If a Clause 45 MDIO is implemented, this variable is accessible through bit 1.8.0 (45.2.1.7.15). A device is placed in local loopback mode when the Local_loopback_enable control variable is set to one, and removed from local loopback mode when this variable is set to zero. If a Clause 45 MDIO is implemented, this variable is accessible through PMA/PMD control 1 register (bit 1.0.0, see 45.2.1.1.5).

120.5.10 PMA remote loopback mode (optional)

PMA remote loopback mode is optional. If implemented, it shall be as described in this subclause (120.5.10).

Remote loopback, if provided, should be implemented in a PMA sublayer close enough to the PMD to maintain the bit sequence on each individual PMD lane. When remote loopback is enabled, each symbol received over a lane of the service interface below the PMA via *inst*:IS_UNITDATA_*i*.indication is looped back to the corresponding output lane toward the PMD via *inst*:IS_UNITDATA_*i*.request. Note that the service interface below the PMA can be provided by the PHY XS, PMD, or another PMA sublayer.

During remote loopback, the PMA performs normal bit muxing of PCSLs per 120.5.2 onto the lanes in the Rx direction towards the PMA client.

The ability to perform this function is indicated by the 200G_Remote_loopback_ability and 400G_Remote_loopback_ability status variables for the 200GBASE-R PMA and 400GBASE-R PMA, respectively. If a Clause 45 MDIO is implemented, the 200G_Remote_loopback_ability and 400G_Remote_loopback_ability variables are accessible through bit 1.23.15 (45.2.1.14e.1) and bit 1.24.15 (45.2.1.14f.1), respectively. A device is placed in remote loopback mode when the Remote_loopback_enable control variable is set to one, and removed from remote loopback mode when this variable is set to zero. If a Clause 45 MDIO is implemented, this variable is accessible through PMA/PMD Control register 1 (bit 1.0.1, see 45.2.1.1.4).

120.5.11 PMA test patterns (optional)

Where the output lanes of the PMA appear on a physically instantiated interface 200GAUI-n, 400GAUI-n or the PMD service interface (whether or not it is physically instantiated), the PMA may optionally generate and detect test patterns. These test patterns are used to test adjacent layer interfaces for an individual PMA sublayer or to perform testing between a physically instantiated interface of a PMA sublayer and external testing equipment.

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120.5.11.1 Test patterns for NRZ encoded signals

For a 200GBASE-R PMA with 8 output lanes or a 400GBASE-R PMA with 16 output lanes using 2-level NRZ encoding, the test patterns in this clause may be supported.

The ability to generate each of the respective test patterns in each direction of transmission are indicated by the PRBS9_Tx_generator_ability, PRBS9_Rx_generator_ability, PRBS31_Tx_generator_ability, PRBS31_Rx_generator_ability, and Square_wave_ability status variables, which if a Clause 45 MDIO is implemented are accessible through bits 1.1500.5, 1.1500.4, 1.1500.3, 1.1500.1, and 1.1500.12, respectively (see 45.2.1.123).

120.5.11.1.1 PRBS31 test pattern

The ability to check PRBS31 test patterns in each direction of transmission are indicated by the PRBS31_Tx_checker_ability and PRBS31_Rx_checker_ability status variables, which, if a Clause 45 MDIO is implemented, are accessible through bits 1.1500.2 and 1.1500.0, respectively (see 45.2.1.123).

If supported, when send Tx PRBS31 test pattern is enabled by the PRBS31_enable and PRBS_Tx_gen_enable control variables, the PMA shall generate a PRBS31 pattern (as defined in 49.2.8) on each of the lanes toward the service interface below the PMA via the *inst*:IS_UNITDATA_*i*.request primitive. To avoid correlated crosstalk, it is highly recommended that the PRBS31 patterns generated on each lane be generated from independent, random seeds or at a minimum offset of 20 000 UI between the PRBS31 sequence on any lane and any other lane. When send Tx PRBS31 test pattern is disabled, the PMA returns to normal operation performing bit multiplexing as described in 120.5.2. If a Clause 45 MDIO is implemented, the PRBS31_enable and PRBS_Tx_gen_enable control variables are accessible through bits 1.1501.7 and 1.1501.3 (see 45.2.1.124).

If supported, when send Rx PRBS31 test pattern is enabled by the PRBS31_enable and PRBS_Rx_gen_enable control variables, the PMA shall generate a PRBS31 pattern on each of the lanes toward the PMA client via the PMA:IS_UNITDATA_*i*.indication primitive. While this test pattern is enabled, the PMA also generates PMA:IS_SIGNAL.indication(SIGNAL_OK) toward the PMA client independent of the link status at the service interface below the PMA. When send Rx PRBS31 test pattern is disabled, the PMA returns to normal operation performing bit multiplexing as described in 120.5.2. If a Clause 45 MDIO is implemented, the PRBS31_enable and PRBS_Rx_gen_enable control variables are accessible through bits 1.1501.7 and 1.1501.1 (see 45.2.1.124).

If supported, when check Tx PRBS31 test pattern mode is enabled by the PRBS31_enable and PRBS_Tx_check_enable control variables, the PMA shall check for the PRBS31 pattern on each of the lanes received from the PMA client via the PMA:IS_UNITDATA_*i*.request primitive. The checker shall increment the test-pattern error counter by one for each incoming bit error in the PRBS31 pattern for isolated single bit errors. Implementations should be capable of counting at least one error whenever one or more errors occur in a sliding 1000-bit window. If a Clause 45 MDIO is implemented, the PRBS31_enable and PRBS_Tx_check_enable control variables are accessible through bits 1.1501.7 and 1.1501.2 (see 45.2.1.124). The Tx test-pattern error counters Ln0_PRBS_Tx_test_err_counter through Ln15_PRBS_Tx_test_err_counter count, per lane, errors in detecting the PRBS31 pattern on the lanes from the PMA client. If a Clause 45 MDIO is implemented, these counters are accessible through registers 1.1600 through 1.16015 (see 45.2.1.126). When check Tx PRBS31 test pattern is disabled, the PMA expects normal traffic and test-pattern error counting does not continue. While in check Tx PRBS31 test-pattern mode, bit multiplexing continues as described in 120.5.2. Note that bit multiplexing of per-lane PRBS31 may produce a signal which is not meaningful for downstream sublayers.

If supported, when check Rx PRBS31 test-pattern mode is enabled by the PRBS31_enable and PRBS_Rx_check_enable control variables, the PMA shall check for the PRBS31 pattern on each of the lanes received from the service interface below the PMA via the *inst*:IS_UNITDATA_*i*.indication primitive.

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If a Clause 45 MDIO is implemented, the PRBS31_enable and PRBS_Rx_check_enable control variables are accessible through bits 1.1501.7 and 1.1501.0 (see 45.2.1.124). The Rx test-pattern error counters Ln0_PRBS_Rx_test_err_counter through Ln15_PRBS_Rx_test_err_counter count, per lane, errors in detecting the PRBS31 pattern on the lanes from the service interface below the PMA. If a Clause 45 MDIO is implemented, these counters are accessible through registers 1.1700 through 1.1715 (see 45.2.1.127). While in check Rx PRBS31 test-pattern mode, the PMA:IS_SIGNAL.indication primitive does not indicate a valid signal. When check Rx PRBS31 test pattern is disabled, the PMA returns to normal operation performing bit multiplexing as described in 120.5.2.

120.5.11.1.2 PRBS9 test pattern

If supported, when send Tx PRBS9 test-pattern mode is enabled by the PRBS9_enable and Tx_gen_enable control variables, the PMA shall generate a PRBS9 pattern (as defined in Table 68–6) on each lane toward the service interface below the PMA via the *inst*:IS_UNITDATA_*i*.request primitive. If a Clause 45 MDIO is implemented, the PRBS9_enable and Tx_gen_enable control variables are accessible through bits 1.1501.6 and 1.1501.3 (see 45.2.1.124). When send Tx PRBS9 test-pattern mode is disabled, the PMA returns to normal operation performing bit multiplexing as described in 120.5.2.

If supported, when send Rx PRBS9 test-pattern mode is enabled by the PRBS9_enable and Rx_gen_enable control variables, the PMA shall generate a PRBS9 pattern on each lane toward the PMA client via the PMA:IS_UNITDATA_*i*.indication primitive. The PMA also generates PMA:IS_SIGNAL.indication(SIGNAL_OK) toward the PMA client independent of the link status at the service interface below the PMA. If a Clause 45 MDIO is implemented, the PRBS9_enable and Rx_gen_enable control variables are accessible through bits 1.1501.6 and 1.1501.1 (see 45.2.1.124). When send Rx PRBS9 test-pattern mode is disabled, the PMA returns to normal operation performing bit multiplexing as described in 120.5.2.

Note that PRBS9 is intended to be checked by external test gear, and no PRBS9 checking function is provided within the PMA.

120.5.11.1.3 Square wave test pattern

Transmit square wave test-pattern mode optionally applies to each lane of the Tx direction PMA towards a physically instantiated 200GAUI-8, 400GAUI-16 or towards an NRZ PMD service interface whether or not it is physically instantiated. The ability to perform this function is indicated by the Square_wave_ability status variable. If a Clause 45 MDIO is implemented, the Square_wave_ability status variable is accessible through the Square wave test ability bit 1.1500.12 (see 45.2.1.123). If implemented, the transmit square wave test-pattern mode is enabled by control variables Square_wave_enable_0 through Square_wave_enable_15. If a Clause 45 MDIO is implemented, these control variables are accessible through the square wave testing control and status register bits 1.1510.0 through 1.1510.15 (limited to the number of lanes of the service interface below the PMA, see 45.2.1.125). When enabled, the PMA shall generate a square wave test pattern (8 ones followed by 8 zeros) on the square wave enabled lanes toward the service interface below the PMA via the *inst*:IS_UNITDATA_*i*.request primitive. Lanes for which square wave is not enabled transmit normal data resulting from the bit multiplexing operations described in 120.5.2 or test patterns as determined by other registers. When transmit square wave test pattern is disabled for all lanes, the PMA performs normal operation performing bit multiplexing as described in 120.5.2 or transmit test patterns as determined by other registers.

120.5.11.2 Test patterns for PAM4 encoded signals

For a 200GBASE-R PMA with 4 output lanes or a 400GBASE-R PMA with 4 or 8 output lanes using PAM4 encoding, the test patterns described in this clause may optionally be supported.

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The patterns PRBS13Q and square wave (quaternary) can be enabled on a lane-by-lane basis. The patterns PRBS31Q and SSPRQ can be enabled on all lanes of an interface at once. If per-lane pattern(s) are enabled for a subset of the lanes and a per-interface pattern is also enabled, the per-lane patterns are generated only on the indicated lanes and the per-interface pattern is generated on the remaining lanes. The behavior if more than one per-lane pattern is enabled for the same lane or more than one per-interface pattern is enabled is not defined.

120.5.11.2.1 PRBS13Q test pattern

A PMA may optionally include a PRBS13Q pattern generator as specified in this subclause. The ability to perform this function is indicated by the PRBS13Q_gen_Tx_ability and PRBS13Q_gen_Rx_ability status variables, reflecting the ability to send this test pattern in the direction towards the PMD and towards the MAC, respectively. If a Clause 45 MDIO is implemented, the PRBS13Q_gen_Tx_ability and PRBS13Q_gen_Rx_ability status variables are accessible through the PRBS13Q Tx generator ability and PRBS13Q Rx generator ability bits 1.1500.11 and 1.1500.10 (see 45.2.1.123).

When the PRBS13Q test pattern enabled, it replaces the signal on the output lane(s) for which it is enabled. The PRBS13Q test pattern is a repeating 8191-symbol sequence formed by Gray coding pairs of bits from two repetitions of the PRBS13 pattern into PAM4 symbols as described in 120.5.7. The PRBS pattern generator produces the same result as the implementation shown in Figure 94-6, which implements the generator polynomial shown in Equation (94-3). Since the PRBS13 pattern is an odd number of bits in length, bits which are mapped as the first bit of a PAM4 symbol during one repetition of the PRBS13 sequence are mapped as the second bit of a PAM4 symbol during the next repetition of the PRBS13 sequence, and bits which are mapped as the second bit of a PAM4 symbol are mapped as the first bit of the following symbol in the next repetition of the PRBS13 sequence. For example, if the PRBS13 generator used to create the PRBS13Q sequence is initialized to a seed value of 0000010101011 (with the leftmost bit in S0 and the rightmost in S12), the PRBS13Q sequence begins with the following Gray coded PAM4 symbols, transmitted left to right: 10313202201111301031212312100121021210231311l2.

The PRBS13Q test pattern is enabled in the transmit direction by the Tx_gen_enable and PRBS13Q_enable_0 through PRBS13Q_enable_7 control variables (in the downstream direction towards the PMD), and in the receive direction by the Rx_gen_enable and PRBS13Q_enable_0 through PRBS13Q_enable_7 control variables (in the upstream direction towards the MAC). If the optional Clause 45 MDIO is implemented, the Tx_gen_enable and Rx_gen_enable control variables are accessible through the Tx generator enable and Rx generator enable bits 1.1501.3 and 1.1501.1 (see 45.2.1.124), and the PRBS13_enable_0 through PRBS13_enable_7 control variables are accessible through bits 1.1512.0 through 1.1512.7 (see 45.2.1.125a).

120.5.11.2.2 PRBS31Q test pattern

A PMA may optionally include a PRBS31Q pattern generator as specified in this subclause. The ability to generate PRBS31Q patterns in each direction of transmission are indicated by the PRBS31Q_gen_Tx_ability and PRBS31Q_gen_Rx_ability status variables, reflecting the ability to send this test pattern in the direction towards the PMD and towards the MAC, respectively. The ability to check PRBS31Q patterns in each direction of transmission are indicated by the PRBS31Q_Tx_checker_ability and PRBS31Q_Rx_checker_ability status variables. If a Clause 45 MDIO is implemented, the PRBS31Q_gen_Tx_ability, PRBS31Q_gen_Rx_ability, PRBS31Q_Tx_checker_ability and PRBS31Q_Rx_checker_ability status variables are accessible through the PRBS31Q Tx generator ability, PRBS31Q Rx generator ability, PRBS31Q Tx checker ability, and PRBS31Q Rx checker ability bits 1.1500.9, 1.1500.7, 1.1500.8, and 1.1500.6 (see 45.2.1.123).

The PRBS31Q test pattern is a repeating 2^{31}-1-symbol sequence formed by Gray coding pairs of bits from two repetitions of the PRBS31 pattern defined in 49.2.8 into PAM4 symbols as described in 120.5.7. Since the PRBS31 pattern is an odd number of bits in length, bits which are mapped as the first bit of a PAM4

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symbol during one repetition of the PRBS31 sequence are mapped as the second bit of a PAM4 symbol during the next repetition of the PRBS31 sequence, and bits which are mapped as the second bit of a PAM4 symbol during one repetition of the PRBS31 sequence are mapped as the first bit of the following symbol in the next repetition of the PRBS31 sequence. For example, if the PRBS31 generator used to create the PRBS31Q sequence is initialized to a seed value of all ones, the PRBS31Q sequence begins with the following Gray coded PAM4 symbols, transmitted left to right: 2222222222222012222222222220002222222222201201222. To avoid correlated crosstalk, it is highly recommended that the PRBS31 patterns used to generate the PRBS31Q pattern on each lane are generated from independent, random seeds, or at a minimum offset of 20 000 UI between the PRBS31 sequence used to generate the PRBS31Q pattern on any lane and any other lane. A PRBS31Q pattern checker operates by converting PAM4 symbols received on each input lane to pairs of bits as described in 120.5.7 and then using a PRBS31 pattern checker on the resulting bit stream. The checker shall increment the test-pattern error counter by one for each incoming bit error in the PRBS31 pattern for isolated single bit errors. Implementations should be capable of counting at least one error whenever one or more errors occur in a sliding 1000-bit window.

If supported, when send Tx PRBS31Q test pattern is enabled by the PRBS31Q_enable and PRBS_tx_gen enable control variables, the PMA shall generate a PRBS31Q pattern on each of the lanes toward the service interface below the PMA via the *inst*:IS_UNITDATA_*i*.request primitive. When send Tx PRBS31Q test pattern is disabled, the PMA returns to normal operation performing bit multiplexing as described in 120.5.2. If a Clause 45 MDIO is implemented, the PRBS31Q_enable and PRBS_Tx_gen_enable control variables are accessible through bits 1.1501.13 and 1.1501.3 (see 45.2.1.124).

If supported, when send Rx PRBS31Q test pattern is enabled by the PRBS31Q_enable and PRBS_rx_gen enable control variables, the PMA shall generate a PRBS31Q pattern on each of the lanes toward the PMA client via the PMA:IS_UNITDATA_*i*.indication primitive. The PMA also generates PMA:IS_SIGNAL.indication(SIGNAL_OK) toward the PMA client independent of the link status at the service interface below the PMA. If a Clause 45 MDIO is implemented, the PRBS31Q_enable and PRBS_Rx_gen_enable control variables are accessible through bits 1.1501.13 and 1.1501.1 (see 45.2.1.124). When send Rx PRBS31Q test pattern is disabled, the PMA returns to normal operation performing bit multiplexing as described in 120.5.2.

If supported, when check Tx PRBS31Q test pattern mode is enabled by the PRBS31Q_enable and PRBS_Tx_check_enable control variables, the PMA checks for the PRBS31Q pattern on each of the lanes received from the PMA client via the PMA:IS_UNITDATA_*i*.request primitive. If a Clause 45 MDIO is implemented, the PRBS31Q_enable and PRBS_Tx_check_enable control variables are accessible through bits 1.1501.13 and 1.1501.2 (see 45.2.1.124). The Tx test-pattern error counters Ln0_PRBS_Tx_test_err_counter through either Ln3_PRBS_Tx_test_error_counter or Ln7_PRBS_Tx_test_error_counter (depending on whether the number of lanes is 4 or 8) count, per lane, errors in detecting the PRBS31 pattern resulting from converting the PAM4 symbols received on each lane to pairs of bits. If a Clause 45 MDIO is implemented, these counters are accessible through registers 1.1600 through 1.1603 or 1.1607 (depending on whether the number of lanes is 4 or 8) (see 45.2.1.126). When check Tx PRBS31Q test pattern is disabled, the PMA expects normal traffic and test pattern error counting does not continue. While in check Tx PRBS31Q mode, bit multiplexing continues as described in 120.5.2. Note that bit multiplexing of per-lane PRBS31Q may produce a signal which is not meaningful for downstream sublayers.

If supported, when check Rx PRBS31Q test pattern mode is enabled by the PRBS31Q_enable and PRBS_Rx_check_enable control variables, the PMA checks for the PRBS31Q pattern on each of the lanes received from the service interface below the PMA via the *inst*:IS_UNITDATA_*i*.indication primitive. If a Clause 45 MDIO is implemented, the PRBS31Q_enable and PRBS_Rx_check_enable control variables are accessible through bits 1.1501.13 and 1.1501.0 (see 45.2.1.124). The Rx test-pattern error counters Ln0_PRBS_Rx_test_err_counter through either Ln3_PRBS_Rx_test_error_counter or Ln7_PRBS_Rx_test_error_counter (depending on whether the number of lanes is 4 or 8) count, per lane,

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errors in detecting the PRBS31 pattern resulting from converting the PAM4 symbols received on each lane to pairs of bits. If a Clause 45 MDIO is implemented, these counters are accessible through registers 1.1700 through 1.1703 or 1.1707 (depending on whether the number of lanes is 4 or 8) (see 45.2.1.127). While in check Rx PRBS31Q mode, the PMA:IS_SIGNAL.indication primitive does not indicate a valid signal. When check Rx PRBS31Q test pattern is disabled, the PMA returns to normal operation performing bit multiplexing as described in 120.5.2.

120.5.11.2.3 SSPRQ test pattern

A PMA may optionally include a short stress pattern random quaternary (SSPRQ) test pattern generator as specified in this subclause. The ability to perform this function is indicated by the SSPRQ_gen_Tx_ability status variable, reflecting the ability to send this test pattern in the direction towards the PMD. If a Clause 45 MDIO is implemented, the SSPRQ_gen_Tx_ability status variable is accessible through the SSPRQ Tx generator ability bit 1.1500.13 (see 45.2.1.123).

The SSPRQ pattern is a repeating 2^{16}–1 PAM4 symbol sequence constructed as follows:

— Bit sequence A is a 32768-bit sequence composed of three sections of the PRBS31 binary sequence (see 120.5.11.1.1) according to Table 120–2.

Table 120–2—SSPRQ bit sequence A

Pattern	Seed	Length
PRBS31	0x00000002	10924 bits
	0x34013FF7	10922 bits
	0x0CCCCCCC	10922 bits

— Each section of PRBS31 is generated as if produced by the shift register implementation shown in Figure 49-9 and the seed is a 31-bit hexadecimal value used to preset S30 through S0 (S30 is set to the MSB and S0 is set to the LSB) prior to the generation of the PRBS31 sequence for the indicated length of bits.

— Bit sequence B is a 65534-bit sequence formed by removing the first and last binary bit from a sequence consisting of two repetitions of bit sequence A.

— PAM4 sequence 1 is a 16384-symbol sequence formed by Gray coding bit sequence A according to 120.5.7.

— PAM4 sequence 2 is a 16384-symbol sequence formed by Gray coding bit sequence A according to 120.5.7, and then inverting each PAM4 symbol by replacing each PAM4 symbol *n* with the value 3–*n*.

— PAM4 sequence 3 is a 16383-symbol sequence formed by Gray coding the first 32766 bits of bit-sequence B according to 120.5.7.

— PAM4 sequence 4 is a 16384-symbol sequence formed by Gray coding the last 32768 bits of bit-sequence B according to 120.5.7, and then inverting each PAM4 symbol by replacing each PAM4 symbol *n* with the value 3–*n*.

— The repeating SSPRQ pattern formed by concatenating PAM4 sequences 1, 2, 3 and 4.

NOTE—A CSV file containing the entire PAM4 symbol sequence for the SSPRQ test pattern is available at http://standards.ieee.org/downloads/802.3/.

If supported, when send SSPRQ test pattern is enabled by the SSPRQ_enable control variable, the PMA shall generate an SSPRQ pattern on each of its lanes in the Tx direction towards the PMD with at least a 31 UI delay between the SSPRQ pattern on one lane and any other lane. If a Clause 45 MDIO is

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implemented, the SSPRQ_pattern_enable control variable is accessible through the SSPRQ pattern enable bit 1.1501.14 (see 45.2.1.124).

120.5.11.2.4 Square wave (quaternary) test pattern

Transmit square wave (quaternary) test-pattern mode optionally applies to each lane of the Tx direction PMA towards a physically instantiated 200GAUI-4, 400GAUI-8 or towards a PAM4 PMD service interface whether or not it is physically instantiated.

The ability to perform this function is indicated by the Square_wave_ability status variable. If a Clause 45 MDIO is implemented, the Square_wave_ability status variable is accessible through the Square wave test ability bit 1.1500.12 (see 45.2.1.123). If implemented, the transmit square wave test-pattern mode is enabled by control variables Square_wave_enable_0 through Square_wave_enable_8. If a Clause 45 MDIO is implemented, these control variables are accessible through the square wave testing control and status register bits 1.1510.0 through 1.1510.15 (limited to the number of lanes of the service interface below the PMA, see 45.2.1.125). When enabled, the PMA shall generate a square wave test pattern (8 threes followed by 8 zeros) on the square wave enabled lanes toward the service interface below the PMA via the *inst*:IS_UNITDATA_*i*.request primitive. Lanes for which square wave is not enabled transmit normal data resulting from the bit multiplexing operations described in 120.5.2 or test patterns as determined by other registers. When transmit square wave test pattern is disabled for all lanes, the PMA performs normal operation performing bit multiplexing as described in 120.5.2 or transmit test patterns as determined by other registers.

NOTE—A square wave transmitted over a 200GAUI-4 or 400GAUI-8 may not be correctly forwarded to the PMD transmitter output.

120.6 PMA MDIO function mapping

The optional MDIO capability described in Clause 45 describes several variables that provide control and status information for and about the PMA. Since a given implementation may employ more than one PMA sublayer, the PMA control and status information is organized into multiple addressable instances, one for each possible PMA sublayer. See 45.2.1 and 120.1.4 for the allocation of MMD addresses to PMA sublayers. Control and status registers for MMD 8, 9, and 10 use the Extended PMA control and status registers at identical locations to those for MMD 1.

Mapping of MDIO control variables to PMA control variables is shown in Table 120–3. Mapping of MDIO status variables to PMA status variables is shown in Table 120–4. Mapping of MDIO counter to PMA counters is shown in Table 120–5. These tables provide the register and bit numbers for the PMA addressed as MMD 1. For implementations with multiple PMA sublayers, additional PMA sublayers use the corresponding register and bit numbers in MMDs 8, 9, and 10 as necessary.

Table 120–3—MDIO/PMA control variable mapping

MDIO variable	PMA/PMD register name	Register/ bit number	PMA control variable
PMA remote loopback	PMA/PMD control 1	1.0.1	Remote_loopback_enable
PMA local loopback	PMA/PMD control 1	1.0.0	Local_loopback_enable
PRBS31 pattern enable	PRBS pattern testing control	1.1501.7	PRBS31_enable
PRBS9 pattern enable	PRBS pattern testing control	1.1501.6	PRBS9_enable
PRBS31Q pattern enable	PRBS pattern testing control	1.1501.13	PRBS31Q_pattern_enable

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Table 120–3—MDIO/PMA control variable mapping *(continued)*

MDIO variable	PMA/PMD register name	Register/ bit number	PMA control variable
SSPRQ pattern enable	PRBS pattern testing control	1.1501.14	SSPRQ_pattern_enable
Tx generator enable	PRBS pattern testing control	1.1501.3	PRBS_Tx_gen_enable
Tx checker enable	PRBS pattern testing control	1.1501.2	PRBS_Tx_check_enable
Rx generator enable	PRBS pattern testing control	1.1501.1	PRBS_Rx_gen_enable
Rx checker enable	PRBS pattern testing control	1.1501.0	PRBS_Rx_check_enable
Lane 0 SW enable	Square wave testing control	1.1510.0	Square_wave_enable_0
Lane 1 SW enable	Square wave testing control	1.1510.1	Square_wave_enable_1
Lane 2 SW enable	Square wave testing control	1.1510.2	Square_wave_enable_2
Lane 3 SW enable	Square wave testing control	1.1510.3	Square_wave_enable_3
Lane 4 SW enable	Square wave testing control	1.1510.4	Square_wave_enable_4
Lane 5 SW enable	Square wave testing control	1.1510.5	Square_wave_enable_5
Lane 6 SW enable	Square wave testing control	1.1510.6	Square_wave_enable_6
Lane 7 SW enable	Square wave testing control	1.1510.7	Square_wave_enable_7
Lane 8 SW enable	Square wave testing control	1.1510.8	Square_wave_enable_8
Lane 9 SW enable	Square wave testing control	1.1510.9	Square_wave_enable_9
Lane 10 SW enable	Square wave testing control	1.1510.10	Square_wave_enable_10
Lane 11 SW enable	Square wave testing control	1.1510.11	Square_wave_enable_11
Lane 12 SW enable	Square wave testing control	1.1510.12	Square_wave_enable_12
Lane 13 SW enable	Square wave testing control	1.1510.13	Square_wave_enable_13
Lane 14 SW enable	Square wave testing control	1.1510.14	Square_wave_enable_14
Lane 15 SW enable	Square wave testing control	1.1510.15	Square_wave_enable_15
Lane 0 PRBS13Q enable	PRBS13Q testing control	1.1512.0	PRBS13Q_enable_0
Lane 1 PRBS13Q enable	PRBS13Q testing control	1.1512.1	PRBS13Q_enable_1
Lane 2 PRBS13Q enable	PRBS13Q testing control	1.1512.2	PRBS13Q_enable_2
Lane 3 PRBS13Q enable	PRBS13Q testing control	1.1512.3	PRBS13Q_enable_3
Lane 4 PRBS13Q enable	PRBS13Q testing control	1.1512.4	PRBS13Q_enable_4
Lane 5 PRBS13Q enable	PRBS13Q testing control	1.1512.5	PRBS13Q_enable_5
Lane 6 PRBS13Q enable	PRBS13Q testing control	1.1512.6	PRBS13Q_enable_6
Lane 7 PRBS13Q enable	PRBS13Q testing control	1.1512.7	PRBS13Q_enable_7

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Table 120–4—MDIO/PMA status variable mapping

MDIO status variable	PMA/PMD register name	Register/bit number	PMA status variable
PMA remote loopback ability	200G PMA/PMD extended ability register	1.23.15	200G_Remote_loopback_ability
PMA remote loopback ability	400G PMA/PMD extended ability register	1.24.15	400G_Remote_loopback_ability
PMA local loopback ability	PMA/PMD status 2 register	1.8.0	Local_loopback_ability
PRBS9 Tx generator ability	Test-pattern ability register	1.1500.5	PRBS9_Tx_generator_ability
PRBS9 Rx generator ability	Test-pattern ability register	1.1500.4	PRBS9_Rx_generator_ability
PRBS31Tx generator ability	Test-pattern ability register	1.1500.3	PRBS31_Tx_generator_ability
PRBS31Tx checker ability	Test-pattern ability register	1.1500.2	PRBS31_Tx_checker_ability
PRBS31 Rx generator ability	Test-pattern ability register	1.1500.1	PRBS31_Rx_generator_ability
PRBS31 Rx checker ability	Test-pattern ability register	1.1500.0	PRBS31_Rx_checker_ability
Square wave test ability	Test-pattern ability register	1.1500.12	Square_wave_ability
PRBS13Q Tx generator ability	Test-pattern ability register	1.1500.11	PRBS13Q_gen_Tx_ability
PRBS13Q Rx generator ability	Test-pattern ability register	1.1500.10	PRBS13Q_gen_Rx_ability
PRBS31Q Tx generator ability	Test-pattern ability register	1.1500.9	PRBS31Q_gen_Tx_ability
PRBS31Q Rx generator ability	Test-pattern ability register	1.1500.7	PRBS31Q_gen_rx_ability
PRBS31Q Tx checker ability	Test-pattern ability register	1.1500.8	PRBS31Q_Tx_checker_ability
PRBS31Q Rx checker ability	Test-pattern ability register	1.1500.6	PRBS31Q_Rx_checker_ability
SSPRQ Tx generator ability	Test-pattern ability register	1.1500.13	SSPRQ_gen_Tx_ability

Table 120–5—MDIO/PMA counters mapping

MDIO variable	PMA/PMD register name	Register/bit number	PMA status variable
Error counter Tx, lane 0	PRBS Tx pattern testing error counter, lane 0	1.1600	Ln0_PRBS_Tx_test_err_counter
Error counter Tx, lane 1	PRBS Tx pattern testing error counter, lane 1	1.1601	Ln1_PRBS_Tx_test_err_counter
Error counter Tx, lane 2	PRBS Tx pattern testing error counter, lane 2	1.1602	Ln2_PRBS_Tx_test_err_counter
Error counter Tx, lane 3	PRBS Tx pattern testing error counter, lane 3	1.1603	Ln3_PRBS_Tx_test_err_counter
Error counter Tx, lane 4	PRBS Tx pattern testing error counter, lane 4	1.1604	Ln4_PRBS_Tx_test_err_counter

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Table 120–5—MDIO/PMA counters mapping *(continued)*

MDIO variable	PMA/PMD register name	Register/bit number	PMA status variable
Error counter Tx, lane 5	PRBS Tx pattern testing error counter, lane 5	1.1605	Ln5_PRBS_Tx_test_err_counter
Error counter Tx, lane 6	PRBS Tx pattern testing error counter, lane 6	1.1606	Ln6_PRBS_Tx_test_err_counter
Error counter Tx, lane 7	PRBS Tx pattern testing error counter, lane 7	1.1607	Ln7_PRBS_Tx_test_err_counter
Error counter Tx, lane 8	PRBS Tx pattern testing error counter, lane 8	1.1608	Ln8_PRBS_Tx_test_err_counter
Error counter Tx, lane 9	PRBS Tx pattern testing error counter, lane 9	1.1609	Ln9_PRBS_Tx_test_err_counter
Error counter Tx, lane 10	PRBS Tx pattern testing error counter, lane 10	1.1610	Ln10_PRBS_Tx_test_err_counter
Error counter Tx, lane 11	PRBS Tx pattern testing error counter, lane 11	1.1611	Ln11_PRBS_Tx_test_err_counter
Error counter Tx, lane 12	PRBS Tx pattern testing error counter, lane 12	1.1612	Ln12_PRBS_Tx_test_err_counter
Error counter Tx lane 13	PRBS Tx pattern testing error counter, lane 13	1.1613	Ln13_PRBS_Tx_test_err_counter
Error counter Tx lane 14	PRBS Tx pattern testing error counter, lane 14	1.1614	Ln14_PRBS_Tx_test_err_counter
Error counter Tx lane 15	PRBS Tx pattern testing error counter, lane 15	1.1615	Ln15_PRBS_Tx_test_err_counter
Error counter Rx, lane 0	PRBS Rx pattern testing error counter, lane 0	1.1700	Ln0_PRBS_Rx_test_err_counter
Error counter Rx, lane 1	PRBS Rx pattern testing error counter, lane 1	1.1701	Ln1_PRBS_Rx_test_err_counter
Error counter Rx, lane 2	PRBS Rx pattern testing error counter, lane 2	1.1702	Ln2_PRBS_Rx_test_err_counter
Error counter Rx, lane 3	PRBS Rx pattern testing error counter, lane 3	1.1703	Ln3_PRBS_Rx_test_err_counter
Error counter Rx, lane 4	PRBS Rx pattern testing error counter, lane 4	1.1704	Ln4_PRBS_Rx_test_err_counter
Error counter Rx, lane 5	PRBS Rx pattern testing error counter, lane 5	1.1705	Ln5_PRBS_Rx_test_err_counter
Error counter Rx, lane 6	PRBS Rx pattern testing error counter, lane 6	1.1706	Ln6_PRBS_Rx_test_err_counter
Error counter Rx, lane 7	PRBS Rx pattern testing error counter, lane 7	1.1707	Ln7_PRBS_Rx_test_err_counter
Error counter Rx, lane 8	PRBS Rx pattern testing error counter, lane 8	1.1708	Ln8_PRBS_Rx_test_err_counter
Error counter Rx, lane 9	PRBS Rx pattern testing error counter, lane 9	1.1709	Ln9_PRBS_Rx_test_err_counter

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Table 120–5—MDIO/PMA counters mapping *(continued)*

MDIO variable	PMA/PMD register name	Register/bit number	PMA status variable
Error counter Rx, lane 10	PRBS Rx pattern testing error counter, lane 10	1.1710	Ln10_PRBS_Rx_test_err_counter
Error counter Rx, lane 11	PRBS Rx pattern testing error counter, lane 11	1.1711	Ln11_PRBS_Rx_test_err_counter
Error counter Rx, lane 12	PRBS Rx pattern testing error counter, lane 12	1.1712	Ln12_PRBS_Rx_test_err_counter
Error counter Rx, lane 13	PRBS Rx pattern testing error counter, lane 13	1.1713	Ln13_PRBS_Rx_test_err_counter
Error counter Rx, lane 14	PRBS Rx pattern testing error counter, lane 14	1.1714	Ln14_PRBS_Rx_test_err_counter
Error counter Rx, lane 15	PRBS Rx pattern testing error counter, lane 15	1.1715	Ln15_PRBS_Rx_test_err_counter

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120.7 Protocol implementation conformance statement (PICS) proforma for Clause 120, Physical Medium Attachment (PMA) sublayer, type 200GBASE-R and 400GBASE-R[5]

120.7.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Clause 120, Physical Medium Attachment (PMA) sublayer, type 200GBASE-R and 400GBASE-R, shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

120.7.2 Identification

120.7.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

120.7.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Clause 120, Physical Medium Attachment (PMA) sublayer, type 200GBASE-R and 400GBASE-R
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[5]*Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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120.7.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
*PMA200	PMA for 200GBASE-R	120.1.1		O.1	Yes [ ] No [ ]
*PMA400	PMA for 400GBASE-R	120.1.1		O.1	Yes [ ] No [ ]
*LNS_UP-STRM	Number of lanes in direction of MAC	120.1.4	Divisor of number of PCS lanes	PMA200: M	8 [ ] 4 [ ]
				PMA400:M	16 [ ] 8 [ ]
*UP_NRZ	NRZ modulation used for PMA service interface	120.1.4		(PMA200* LNS_UPSTRM=8 or PMA400* LNS_UPSTRM=16) :M	Yes [ ] N/A [ ]
*UP_PAM4	PAM4 modulation used for PMA service interface	120.1.4		(PMA200* LNS_UPSTRM=4 or PMA400* LNS_UPSTRM=8) :M	Yes [ ] N/A [ ]
*LNS_DN-STRM	Number of lanes in the direction of PMD	120.1.4	Divisor of number of PCS lanes	PMA200: M	8 [ ] 4 [ ]
				PMA400:M	16 [ ] 8 [ ] 4 [ ]
*DN_NRZ	NRZ modulation used for service interface below the PMA	120.1.4		(PMA200* LNS_DNSTRM=8 or PMA400* LNS_DNSTRM=16) :M	Yes [ ] N/A [ ]
*DN_PAM4	PAM4 modulation used for service interface below the PMA	120.1.4		(PMA200* LNS_DNSTRM=4 or PMA400* LNS_DNSTRM<16) :M	Yes [ ] N/A [ ]
RX_CLOCK	Signaling rate of output lanes in Rx direction	120.5.5	26.5625 GBd	M	Yes [ ]
TX_CLOCK	Signaling rate of output lanes in Tx direction	120.5.5	26.5625 GBd	PMA200:M	Yes [ ] N/A [ ]
			LNS_DNSTRM= 8 or 16: 26.5625 GBd LNS_DNSTRM= 4: 53.125 GBd	PMA400:M	Yes [ ] N/A [ ]
LANE_ MAPPING	Maintain lane mapping while link is in operation	120.5.2	Maintain sequence of PCSLs on all output lanes	M	Yes [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
LNKS	PMA link status	120.5.8	Meets the requirements of 120.5.8	M	Yes [ ]
*LBL	PMA local loopback	120.5.9	Supports local loopback	O	Yes [ ] No [ ]
*LBR	PMA remote loopback	120.5.10	Supports remote loopback	O	Yes [ ] No [ ]
MD	MDIO	120.6	Registers and interface supported	O	Yes [ ] No [ ]
*USP1SP6	Physically instantiated 200GAUI-n or 400GAUI-n above (toward MAC)	120.5.3		O	Yes [ ] No [ ]
*DSP1SP6	Physically instantiated 200GAUI-n or 400GAUI-n below (toward PMD)	120.5.3		O	Yes [ ] No [ ]
*SP2SP5	Physically instantiated PMD service interface	120.5.3		O	Yes [ ] No [ ]
UNAUI	Electrical and timing requirements of Annex 120B, Annex 120C, Annex 120D, or Annex 120E as appropriate met by 200GAUI-n or 400GAUI-n of the PMA service interface	120.5.1, 120.5.5		USP1SP6:M	Yes [ ] N/A [ ]
DNAUI	Electrical and timing requirements of Annex 120B, Annex 120C, Annex 120D, or Annex 120E as appropriate met by 200GAUI-n or 400GAUI-n of the service interface below the PMA	120.5.1, 120.5.5		DSP1SP6:M	Yes [ ] N/A [ ]
DELAY200	Roundtrip delay limit for 200GBASE-R	120.5.4	No more than 18 432 BT or 36 pause_quanta	PMA200:M	Yes [ ] N/A [ ]
DELAY400	Roundtrip delay limit for 400GBASE-R	120.5.4	No more than 36 864 BT or 72 pause_quanta	PMA400:M	Yes [ ] N/A [ ]

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120.7.4 Skew generation and tolerance

Item	Feature	Subclause	Value/Comment	Status	Support
S1	Skew generation toward SP1 in Tx direction	120.5.3.1	≤ 29 ns	DSP1S P6:M	Yes [ ] N/A [ ]
S2	Skew variation generation toward SP1 in Tx direction	120.5.3.1	≤ 200 ps	DSP1S P6:M	Yes [ ] N/A [ ]
S3	Skew variation tolerance at SP1	120.5.3.2	Minimum 200 ps	USP1S P6:M	Yes [ ] N/A [ ]
S4	Skew generation toward SP2 in Tx direction	120.5.3.3	≤ 43 ns	SP2SP 5:M	Yes [ ] N/A [ ]
S5	Skew variation generation toward SP2 in Tx direction	120.5.3.3	≤ 400 ps	SP2SP 5:M	Yes [ ] N/A [ ]
S6	Skew variation tolerance at SP5	120.5.3.4	Minimum 3.6 ns	SP2SP 5:M	Yes [ ] N/A [ ]
S7	Skew generation toward SP6 in Rx direction	120.5.3.5	≤ 160 ns	USP1S P6:M	Yes [ ] N/A [ ]
S8	Skew variation generation toward SP6 in Rx direction	120.5.3.5	≤ 3.8 ns	USP1S P6:M	Yes [ ] N/A [ ]
S9	Skew variation tolerance at SP6	120.5.3.6	Minimum 3.8 ns	DSP1S P6:M	Yes [ ] N/A [ ]

120.7.5 Test patterns

Item	Feature	Subclause	Value/Comment	Status	Support
*JTP1	Physically Instantiated 200GAUI-n or 400GAUI-n between PMA and PMA client	120.5.11		O	Yes [ ] No [ ]
*JTP2	Physically Instantiated 200GAUI-n or 400GAUI-n between PMA and sublayer below the PMA, or adjacent to PMD	120.5.11		O	Yes [ ] No [ ]
J1	Send PRBS31 Tx	120.5.11.1.1		JTP2* DN_NRZ:O	Yes [ ] No [ ] N/A [ ]
J2	Send PRBS31 Rx	120.5.11.1.1		JTP1* UP_NRZ:O	Yes [ ] No [ ] N/A [ ]
J3	Check PRBS31 Tx	120.5.11.1.1		JTP1* UP_NRZ:O	Yes [ ] No [ ] N/A [ ]
J4	Check PRBS31 Rx	120.5.11.1.1		JTP2* DN_NRZ:O	Yes [ ] No [ ] N/A [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
J5	Send PRBS9 Tx	120.5.11.1.2		JTP2* DN_NRZ:O	Yes [ ] No [ ] N/A [ ]
J6	Send PRBS9 Rx	120.5.11.1.2		JTP1* UP_NRZ:O	Yes [ ] No [ ] N/A [ ]
J7	Send square wave Tx	120.5.11.1.3		JTP2* DN_NRZ:O	Yes [ ] No [ ] N/A [ ]
J8	Send PRBS13Q Tx	120.5.11.2.1		JTP2* DN_PAM4:O	Yes [ ] No [ ] N/A [ ]
J9	Send PRBS13Q Rx	120.5.11.2.1		JTP1* UP_PAM4:O	Yes [ ] No [ ] N/A [ ]
J10	Send PRBS31Q Tx	120.5.11.2.2		JTP2* DN_PAM4:O	Yes [ ] No [ ] N/A [ ]
J11	Send PRBS31Q Rx	120.5.11.2.2		JTP1* UP_PAM4:O	Yes [ ] No [ ] N/A [ ]
J12	Check PRBS31Q Tx	120.5.11.2.2		JTP1* UP_PAM4:O	Yes [ ] No [ ] N/A [ ]
J13	Check PRBS31Q Rx	120.5.11.2.2		JTP2* DN_PAM4:O	Yes [ ] No [ ] N/A [ ]
J14	Send SSPRQ Tx	120.5.11.2.3		DN_PAM4:O	Yes [ ] No [ ] N/A [ ]
J15	Send square wave (quaternary) Tx	120.5.11.2.4		DN_PAM4:O	Yes [ ] No [ ] N/A [ ]

120.7.6 Loopback modes

Item	Feature	Subclause	Value/Comment	Status	Support
LB1	PMA local loopback implemented	120.5.9	Meets the requirements of 120.5.9	LBL:M	Yes [ ] N/A [ ]
LB2	PMA remote loopback implemented	120.5.10	Meets the requirements of 120.5.10	LBR:M	Yes [ ] N/A [ ]

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121. Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-DR4

121.1 Overview

This clause specifies the 200GBASE-DR4 PMD together with the single-mode fiber medium. The optical signal generated by this PMD type is modulated using a 4-level pulse amplitude modulation (PAM4) format. When forming a complete Physical Layer, a PMD shall be connected to the appropriate PMA as shown in Table 121–1, to the medium through the MDI and optionally with the management functions that may be accessible through the management interface defined in Clause 45, or equivalent.

Table 121–1—Physical Layer clauses associated with the 200GBASE-DR4 PMD

Associated clause	200GBASE-DR4
117—RS	Required
117—200GMII[a]	Optional
118—200GMII Extender	Optional
119—PCS	Required
120—PMA	Required
120B—Chip-to-chip 200GAUI-8	Optional
120C—Chip-to-module 200GAUI-8	Optional
120D—Chip-to-chip 200GAUI-4	Optional
120E—Chip-to-module 200GAUI-4	Optional
78—Energy Efficient Ethernet	Optional

[a] The 200GMII is an optional interface. However, if the 200GMII is not implemented, a conforming implementation must behave functionally as though the RS and 200GMII were present.

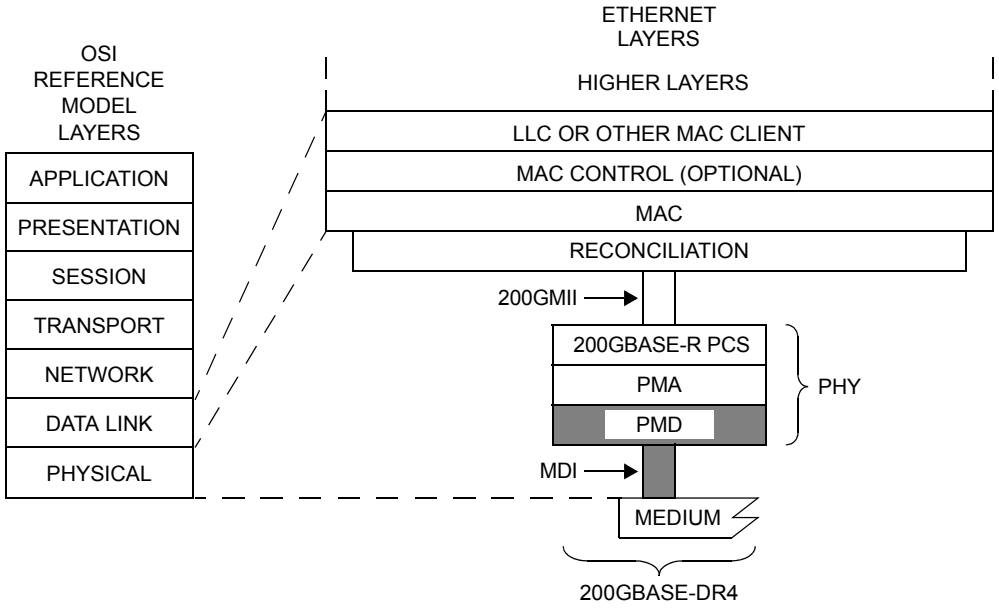
Figure 121–1 shows the relationship of the PMD and MDI (shown shaded) with other sublayers to the ISO/IEC Open System Interconnection (OSI) reference model. 200 Gb/s Ethernet is introduced in Clause 116 and the purpose of each PHY sublayer is summarized in 116.2.

200GBASE-DR4 PHYs with the optional Energy Efficient Ethernet (EEE) fast wake capability may enter the Low Power Idle (LPI) mode to conserve energy during periods of low link utilization (see Clause 78). The deep sleep mode of EEE is not supported.

121.1.1 Bit error ratio

The bit error ratio (BER) when processed according to Clause 120 shall be less than 2.4×10^{-4} provided that the error statistics are sufficiently random that this results in a frame loss ratio (see 1.4.223) of less than 1.7×10^{-12} for 64-octet frames with minimum interpacket gap when processed according to Clause 120 and then Clause 119. For a complete Physical Layer, the frame loss ratio may be degraded to 6.2×10^{-11} for 64-octet frames with minimum interpacket gap due to additional errors from the electrical interfaces.

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200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
LLC = LOGICAL LINK CONTROL
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT
DR = PMD FOR SINGLE-MODE FIBER — 500 m

Figure 121–1—200GBASE-DR4 PMD relationship to the ISO/IEC Open Systems Interconnection (OSI) reference model and IEEE 802.3 Ethernet model

If the error statistics are not sufficiently random to meet this requirement, then the BER shall be less than that required to give a frame loss ratio of less than 1.7×10^{-12} for 64-octet frames with minimum interpacket gap.

121.2 Physical Medium Dependent (PMD) service interface

This subclause specifies the services provided by the 200GBASE-DR4 PMD. The service interface for this PMD is described in an abstract manner and does not imply any particular implementation. The PMD service interface supports the exchange of encoded data between the PMA entity that resides just above the PMD, and the PMD entity. The PMD translates the encoded data to and from signals suitable for the specified medium.

The PMD service interface is an instance of the inter-sublayer service interface defined in 116.3. The PMD service interface primitives are summarized as follows:

PMD:IS_UNITDATA_*i*.request
PMD:IS_UNITDATA_*i*.indication
PMD:IS_SIGNAL.indication

The 200GBASE-DR4 PMD has four parallel symbol streams, hence $i = 0$ to 3.

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In the transmit direction, the PMA continuously sends four parallel symbol streams to the PMD, one per lane, each at a nominal signaling rate of 26.5625 GBd. The PMD then converts these streams of data units into the appropriate signals on the MDI.

In the receive direction, the PMD continuously sends four parallel symbol streams to the PMA corresponding to the signals received from the MDI, one per lane, each at a nominal signaling rate of 26.5625 GBd.

The SIGNAL_DETECT parameter defined in this clause maps to the SIGNAL_OK parameter in the PMD:IS_SIGNAL.indication(SIGNAL_OK) inter-sublayer service interface primitive defined in 116.3.

The SIGNAL_DETECT parameter can take on one of two values: OK or FAIL. When SIGNAL_DETECT = FAIL, the rx_symbol parameters are undefined.

NOTE—SIGNAL_DETECT = OK does not guarantee that the rx_symbol parameters are known to be good. It is possible for a poor quality link to provide sufficient light for a SIGNAL_DETECT = OK indication and still not meet the BER defined in 121.1.1.

121.3 Delay and Skew

121.3.1 Delay constraints

The sum of the transmit and receive delays at one end of the link contributed by the 200GBASE-DR4 PMD including 2 m of fiber in one direction shall be no more than 4096 bit times (8 pause_quanta or 20.48 ns). A description of overall system delay constraints and the definitions for bit times and pause_quanta can be found in 116.4 and its references.

121.3.2 Skew constraints

The Skew (relative delay between the lanes) and Skew Variation must be kept within limits so that the information on the lanes can be reassembled by the PCS. Skew and Skew Variation are defined in 116.5 and specified at the points SP1 to SP6 shown in Figure 116–4 and Figure 116–5.

If the PMD service interface is physically instantiated so that the Skew at SP2 can be measured, then the Skew at SP2 is limited to 43 ns and the Skew Variation at SP2 is limited to 400 ps.

The Skew at SP3 (the transmitter MDI) shall be less than 54 ns and the Skew Variation at SP3 shall be less than 600 ps.

The Skew at SP4 (the receiver MDI) shall be less than 134 ns and the Skew Variation at SP4 shall be less than 3.4 ns.

If the PMD service interface is physically instantiated so that the Skew at SP5 can be measured, then the Skew at SP5 shall be less than 145 ns and the Skew Variation at SP5 shall be less than 3.6 ns.

For more information on Skew and Skew Variation see 116.5. The measurements of Skew and Skew Variation are defined in 86.8.3.1.

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121.4 PMD MDIO function mapping

The optional MDIO capability described in Clause 45 defines several variables that may provide control and status information for and about the PMD. If the MDIO interface is implemented, the mapping of MDIO control variables to PMD control variables shall be as shown in Table 121–2 and the mapping of MDIO status variables to PMD status variables shall be as shown in Table 121–3.

Table 121–2—MDIO/PMD control variable mapping

MDIO control variable	PMA/PMD register name	Register/bit number	PMD control variable
Reset	PMA/PMD control 1 register	1.0.15	PMD_reset
Global PMD transmit disable	PMD transmit disable register	1.9.0	PMD_global_transmit_disable
PMD transmit disable 3 to PMD transmit disable 0	PMD transmit disable register	1.9.4 to 1.9.1	PMD_transmit_disable_3 to PMD_transmit_disable_0

Table 121–3—MDIO/PMD status variable mapping

MDIO status variable	PMA/PMD register name	Register/ bit number	PMD status variable
Fault	PMA/PMD status 1 register	1.1.7	PMD_fault
Transmit fault	PMA/PMD status 2 register	1.8.11	PMD_transmit_fault
Receive fault	PMA/PMD status 2 register	1.8.10	PMD_receive_fault
Global PMD receive signal detect	PMD receive signal detect register	1.10.0	PMD_global_signal_detect
PMD receive signal detect 3 to PMD receive signal detect 0	PMD receive signal detect register	1.10.4 to 1.10.1	PMD_signal_detect_3 to PMD_signal_detect_0

121.5 PMD functional specifications

The 200GBASE-DR4 PMD performs the Transmit and Receive functions, which convey data between the PMD service interface and the MDI.

121.5.1 PMD block diagram

The PMD block diagram is shown in Figure 121–2. For purposes of system conformance, the PMD sublayer is standardized at the points described in this subclause. The optical transmit signal is defined at the output end of a single-mode fiber patch cord (TP2), between 2 m and 5 m in length. Unless specified otherwise, all transmitter measurements and tests defined in 121.8 are made at TP2. The optical receive signal is defined at the output of the fiber optic cabling (TP3) at the MDI (see 121.11.3). Unless specified otherwise, all receiver measurements and tests defined in 121.8 are made at TP3.

TP1<0:3> and TP4<0:3> are informative reference points that may be useful to implementors for testing components (these test points will not typically be accessible in an implemented system).

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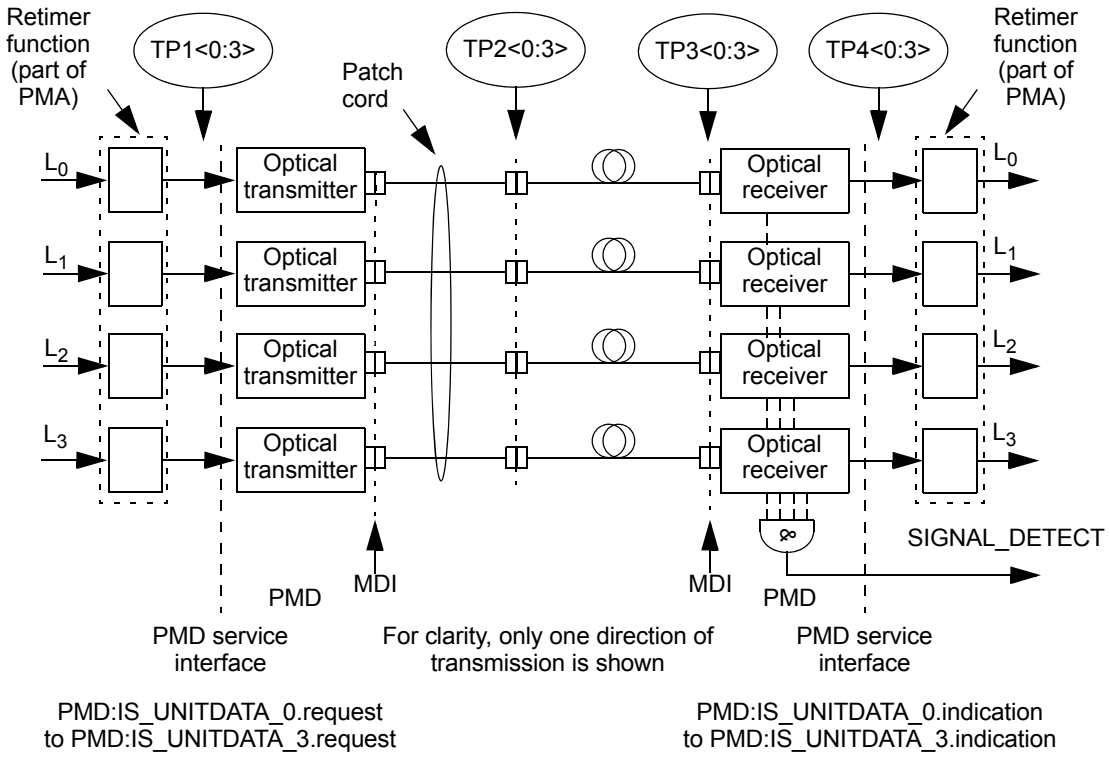


Figure 121–2—Block diagram for 200GBASE-DR4 transmit/receive paths

121.5.2 PMD transmit function

The PMD Transmit function shall convert the four symbol streams requested by the PMD service interface messages PMD:IS_UNITDATA_0.request to PMD:IS_UNITDATA_3.request into four separate optical signals. The four optical signals shall then be delivered to the MDI, which contains four parallel light paths for transmit, according to the transmit optical specifications in this clause. The highest optical power level in each signal stream shall correspond to tx_symbol = three and the lowest shall correspond to tx_symbol = zero.

121.5.3 PMD receive function

The PMD Receive function shall convert the four parallel optical signals received from the MDI into separate symbol streams for delivery to the PMD service interface using the messages PMD:IS_UNITDATA_0.indication to PMD:IS_UNITDATA_3.indication, all according to the receive optical specifications in this clause. The higher optical power level in each signal stream shall correspond to rx_symbol = three and the lowest shall correspond to rx_symbol = zero.

121.5.4 PMD global signal detect function

The PMD global signal detect function shall report the state of SIGNAL_DETECT via the PMD service interface. The SIGNAL_DETECT parameter is signaled continuously, while the PMD:IS_SIGNAL.indication message is generated when a change in the value of SIGNAL_DETECT occurs. The SIGNAL_DETECT parameter defined in this clause maps to the SIGNAL_OK parameter in the inter-sublayer service interface primitives defined in 116.3.

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SIGNAL_DETECT shall be a global indicator of the presence of optical signals on all four lanes. The value of the SIGNAL_DETECT parameter shall be generated according to the conditions defined in Table 121–4. The PMD receiver is not required to verify whether a compliant 200GBASE-DR4 signal is being received. This standard imposes no response time requirements on the generation of the SIGNAL_DETECT parameter.

Table 121–4—SIGNAL_DETECT value definition

Receive conditions	SIGNAL_DETECT value
For any lane; Average optical power at TP3 ≤ –16 dBm	FAIL
For all lanes; [(Optical power at TP3 ≥ average receive power, each lane (min) Table 121–7) AND (compliant 200GBASE–R signal input)]	OK
All other conditions	Unspecified

As an unavoidable consequence of the requirements for the setting of the SIGNAL_DETECT parameter, implementations must provide adequate margin between the input optical power level at which the SIGNAL_DETECT parameter is set to OK, and the inherent noise level of the PMD including the effects of crosstalk, power supply noise, etc.

Various implementations of the Signal Detect function are permitted by this standard, including implementations that generate the SIGNAL_DETECT parameter values in response to the amplitude of the modulation of the optical signal and implementations that respond to the average optical power of the modulated optical signal.

121.5.5 PMD lane-by-lane signal detect function

Various implementations of the Signal Detect function are permitted by this standard. When the MDIO is implemented, each PMD_signal_detect_*i*, where *i* represents the lane number in the range 0:3, shall be continuously set in response to the magnitude of the optical signal on its associated lane, according to the requirements of Table 121–4.

121.5.6 PMD reset function

If the MDIO interface is implemented, and if PMD_reset is asserted, the PMD shall be reset as defined in 45.2.1.1.1.

121.5.7 PMD global transmit disable function (optional)

The PMD_global_transmit_disable function is optional and allows all of the optical transmitters to be disabled.

a) When the PMD_global_transmit_disable variable is set to one, this function shall turn off all of the optical transmitters so that each transmitter meets the requirements of the average launch power of the OFF transmitter in Table 121–6.

b) If a PMD_fault is detected, then the PMD may set the PMD_global_transmit_disable to one, turning off the optical transmitter in each lane.

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121.5.8 PMD lane-by-lane transmit disable function (optional)

The PMD_transmit_disable_*i* (where *i* represents the lane number in the range 0:3) function is optional and allows the optical transmitters in each lane to be selectively disabled.

a) When a PMD_transmit_disable_*i* variable is set to one, this function shall turn off the optical transmitter associated with that variable so that the transmitter meets the requirements of the average launch power of the OFF transmitter in Table 121–6.

b) If a PMD_fault is detected, then the PMD may set each PMD_transmit_disable_*i* to one, turning off the optical transmitter in each lane.

If the optional PMD_transmit_disable_*i* function is not implemented in MDIO, an alternative method may be provided to independently disable each transmit lane for testing purposes.

121.5.9 PMD fault function (optional)

If the PMD has detected a local fault on any of the transmit or receive paths, the PMD shall set PMD_fault to one.

If the MDIO interface is implemented, PMD_fault shall be mapped to the fault bit as specified in 45.2.1.2.3.

121.5.10 PMD transmit fault function (optional)

If the PMD has detected a local fault on any transmit lane, the PMD shall set the PMD_transmit_fault variable to one.

If the MDIO interface is implemented, PMD_transmit_fault shall be mapped to the transmit fault bit as specified in 45.2.1.7.4.

121.5.11 PMD receive fault function (optional)

If the PMD has detected a local fault on any receive lane, the PMD shall set the PMD_receive_fault variable to one.

If the MDIO interface is implemented, PMD_receive_fault shall be mapped to the receive fault bit as specified in 45.2.1.7.5.

121.6 Lane assignments

There are no lane assignments (within a group of transmit or receive lanes) for 200GBASE-DR4. While it is expected that a PMD will map electrical lane *i* to optical lane *i* and vice versa, there is no need to define the physical ordering of the lanes, as the PCS sublayer is capable of receiving the lanes in any arrangement. The positioning of transmit and receive lanes at the MDI is specified in 121.11.3.1.

121.7 PMD to MDI optical specifications for 200GBASE-DR4

The operating range for the 200GBASE-DR4 PMD is defined in Table 121–5. A 200GBASE-DR4 compliant PMD operates on type B1.1, B1.3, or B6_a single-mode fibers according to the specifications defined in Table 121–14. A PMD that exceeds the operating range requirement while meeting all other optical specifications is considered compliant (e.g., a 200GBASE-DR4 PMD operating at 600 m meets the operating range requirement of 2 m to 500 m).

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Table 121–5—200GBASE-DR4 operating range

PMD type	Required operating range
200GBASE-DR4	2 m to 500 m

121.7.1 200GBASE-DR4 transmitter optical specifications

The 200GBASE-DR4 transmitter shall meet the specifications defined in Table 121–6 per the definitions in 121.8.

Table 121–6—200GBASE-DR4 transmit characteristics

Description	Value	Unit
Signaling rate, each lane (range)	26.5625 ± 100 ppm	GBd
Modulation format	PAM4	—
Lane wavelength (range)	1304.5 to 1317.5	nm
Side-mode suppression ratio (SMSR), (min)	30	dB
Average launch power, each lane (max)	3	dBm
Average launch power, each lane[a] (min)	–5.1	dBm
Outer Optical Modulation Amplitude (OMA_{outer}), each lane (max)	2.8	dBm
Outer Optical Modulation Amplitude (OMA_{outer}), each lane (min)[b]	–3	dBm
Launch power in OMA_{outer} minus TDECQ, each lane (min)	–4.4	dBm
Transmitter and dispersion eye closure for PAM4 (TDECQ), each lane (max)	3.4	dB
Average launch power of OFF transmitter, each lane (max)	–16	dBm
Extinction ratio, each lane (min)	3.5	dB
$RIN_{21.4}OMA$ (max)	–132	dB/Hz
Optical return loss tolerance (max)	21.4	dB
Transmitter reflectance[c] (max)	–26	dB

[a] Average launch power, each lane (min) is informative and not the principal indicator of signal strength. A transmitter with launch power below this value cannot be compliant; however, a value above this does not ensure compliance.
[b] Even if the TDECQ < 1.4 dB, the OMA_{outer} (min) must exceed this value.
[c] Transmitter reflectance is defined looking into the transmitter.

121.7.2 200GBASE-DR4 receive optical specifications

The 200GBASE-DR4 receiver shall meet the specifications defined in Table 121–7 per the definitions in 121.8.

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Table 121–7—200GBASE-DR4 receive characteristics

Description	Value	Unit
Signaling rate, each lane (range)	26.5625± 100 ppm	GBd
Modulation format	PAM4	—
Lane wavelengths (range)	1304.5 to 1317.5	nm
Damage threshold[a], each lane	4	dBm
Average receive power, each lane (max)	3	dBm
Average receive power, each lane[b] (min)	–8.1	dBm
Receive power (OMA_{outer}), each lane (max)	2.8	dBm
Receiver reflectance (max)	–26	dB
Receiver sensitivity (OMA_{outer}), each lane[c] (max)	–6.6	dBm
Stressed receiver sensitivity (OMA_{outer}), each lane[d] (max)	–4.1	dBm
Conditions of stressed receiver sensitivity test:[e]		
Stressed eye closure for PAM4 (SECQ), lane under test	3.4	dB
OMA_{outer} of each aggressor lane	2.8	dBm

[a] The receiver shall be able to tolerate, without damage, continuous exposure to an optical input signal having this average power level. The receiver does not have to operate correctly at this input power.
[b] Average receive power, each lane (min) is informative and not the principal indicator of signal strength. A received power below this value cannot be compliant; however, a value above this does not ensure compliance.
[c] Receiver sensitivity (OMA_{outer}), each lane (max) is informative and is defined for a transmitter with SECQ of 0.9 dB.
[d] Measured with conformance test signal at TP3 (see 121.8.9) for the BER specified in 121.1.1.
[e] These test conditions are for measuring stressed receiver sensitivity. They are not characteristics of the receiver.

121.7.3 200GBASE-DR4 illustrative link power budget

An illustrative power budget and penalties for 200GBASE-DR4 channels are shown in Table 121–8.

Table 121–8—200GBASE-DR4 illustrative link power budget

Parameter	Value	Unit
Power budget (for max TDECQ)	6.5	dB
Operating distance	500	m
Channel insertion loss[a]	3	dB
Maximum discrete reflectance	See 121.11.2.2	dB
Allocation for penalties[b] (for max TDECQ)	3.5	dB
Additional insertion loss allowed	0	dB

[a] The channel insertion loss is calculated using the maximum distance specified in Table 121–5 and cabled optical fiber attenuation of 0.5 dB/km at 1304.5 nm plus an allocation for connection and splice loss given in 121.11.2.1.
[b] Link penalties are used for link budget calculations. They are not requirements and are not meant to be tested.

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121.8 Definition of optical parameters and measurement methods

All transmitter optical measurements shall be made through a short patch cable, between 2 m and 5 m in length, unless otherwise specified.

121.8.1 Test patterns for optical parameters

While compliance is to be achieved in normal operation, specific test patterns are defined for measurement consistency and to enable measurement of some parameters. Table 121–10 gives the test patterns to be used in each measurement, unless otherwise specified, and also lists references to the subclauses in which each parameter is defined. Any of the test patterns given for a particular test in Table 121–10 may be used to perform that test. The test patterns used in this clause are shown in Table 121–9.

Table 121–9—Test patterns

Pattern	Pattern description	Defined in
Square wave	Square wave (8 threes, 8 zeros)	120.5.11.2.4
3	PRBS31Q	120.5.11.2.2
4	PRBS13Q	120.5.11.2.1
5	Scrambled idle	119.2.4.9
6	SSPRQ	120.5.11.2.3

Table 121–10—Test-pattern definitions and related subclauses

Parameter	Pattern	Related subclause
Wavelength	Square wave, 3, 4, 5, 6 or valid 200GBASE-R signal	121.8.2
Side mode suppression ratio	3, 5, 6 or valid 200GBASE-R signal	—
Average optical power	3, 5, 6 or valid 200GBASE-R signal	121.8.3
Outer Optical Modulation Amplitude (OMA_{outer})	4 or 6	121.8.4
Transmitter and dispersion eye closure for PAM4 (TDECQ)	6	121.8.5
Extinction ratio	4 or 6	121.8.6
$RIN_{21.4}OMA$	Square wave	121.8.7
Stressed receiver conformance test signal calibration	6	121.8.9.2
Stressed receiver sensitivity	3 or 5	121.8.9

121.8.2 Wavelength

The wavelength of each optical lane shall be within the range given in Table 121–6 if measured per TIA/EIA-455-127-A or IEC 61280-1-3. The lane under test is modulated using the test pattern defined in Table 121–10.

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121.8.3 Average optical power

The average optical power of each lane shall be within the limits given in Table 121–6 if measured using the methods given in IEC 61280-1-1. The average optical power is measured using the test pattern defined in Table 121–10, per the test setup in Figure 53-6.

121.8.4 Outer Optical Modulation Amplitude (OMA_{outer})

The OMA_{outer} of each lane shall be within the limits given in Table 121–6. The OMA_{outer} is measured using a test pattern specified for OMA_{outer} in Table 121–10 as the difference between the average optical launch power level P_3, measured over the central 2 UI of a run of 7 threes, and the average optical launch power level P_0, measured over the central 2 UI of a run of 6 zeros, as shown in Figure 121–3.

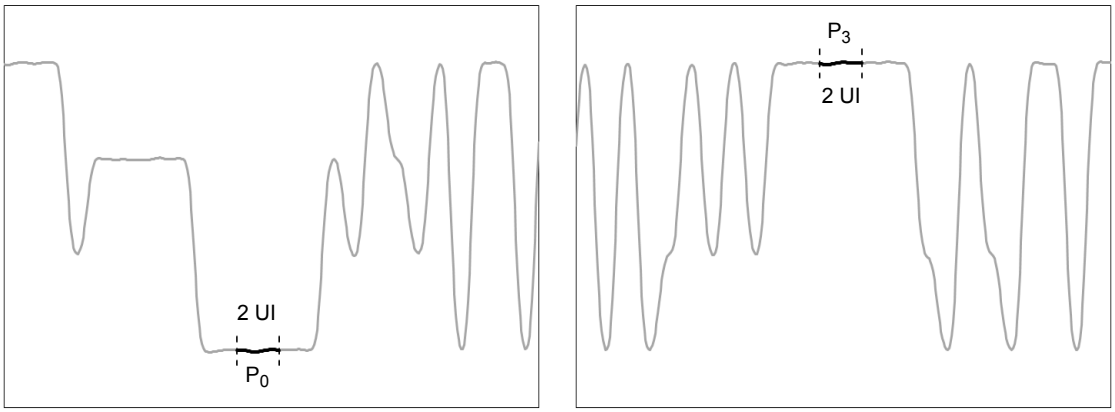


Figure 121–3—Example power levels P_0 and P_3 from PRBS13Q test pattern

121.8.5 Transmitter and dispersion eye closure for PAM4 (TDECQ)

The TDECQ of each lane shall be within the limits given in Table 121–6 if measured using the methods specified in 121.8.5.1, 121.8.5.2, and 121.8.5.3.

TDECQ is a measure of each optical transmitter's vertical eye closure when transmitted through a worst case optical channel (specified in 121.8.5.2), as measured through an optical to electrical converter (O/E) and oscilloscope with the combined frequency response given in 121.8.5.1, and equalized with the reference equalizer (as described in 121.8.5.4). The reference receiver and equalizer may be implemented in software or may be part of an oscilloscope.

Table 121–10 specifies the test patterns to be used for measurement of TDECQ.

121.8.5.1 TDECQ conformance test setup

A block diagram for the TDECQ conformance test is shown in Figure 121–4. Other equivalent measurement implementations may be used with suitable calibration.

Each optical lane is tested individually with all other lanes in operation and all lanes using the same test pattern. There shall be at least 31 UI delay between the test pattern on one lane and the pattern on any other lane, so that the symbols on each lane are not correlated within the PMD. The optical splitter and variable reflector are adjusted so that each transmitter is tested with the optical return loss specified in Table 121–11. The state of polarization of the back reflection is adjusted to create the greatest RIN. Each optical lane is tested with the optical channel described in 121.8.5.2. The combination of the O/E and the oscilloscope has

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a fourth-order Bessel-Thomson filter response with a bandwidth of approximately 13.28125 GHz. Compensation may be made for any deviation from an ideal fourth-order Bessel-Thomson response.

The test pattern (specified in Table 121–10) is transmitted repetitively by the optical lane under test and the oscilloscope is set up to capture the complete pattern for TDECQ analysis as described in 121.8.5.3. The clock recovery unit (CRU) has a corner frequency of 4 MHz and a slope of 20 dB/decade. The CRU can be implemented in hardware or software depending on oscilloscope technology.

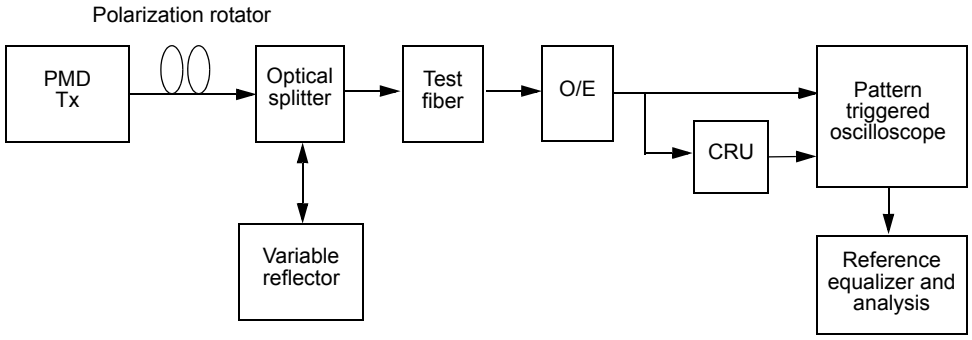


Figure 121–4—TDECQ conformance test block diagram

121.8.5.2 Channel requirements

The transmitter is tested using an optical channel that meets the requirements listed in Table 121–11.

Table 121–11—Transmitter compliance channel specifications

PMD type	Dispersion[a] (ps/nm)		Insertion loss[b]	Optical return loss[c]	Max mean DGD
	Minimum	Maximum			
200GBASE-DR4	$0.011625\times\lambda\times[1-(1324/\lambda)^4]$	$0.011625\times\lambda\times[1-(1300/\lambda)^4]$	Minimum	21.4 dB	0.5 ps

[a] The dispersion is measured for the wavelength of the device under test (λ in nm). The coefficient assumes 500 m for 200GBASE-DR4.
[b] There is no intent to stress the sensitivity of the O/E converter associated with the oscilloscope.
[c] The optical return loss is applied at TP2.

A 200GBASE-DR4 transmitter is to be compliant with a total dispersion at least as negative as the "minimum dispersion" and at least as positive as the "maximum dispersion" columns specified in Table 121–11 for the wavelength of the device under test. This may be achieved with channels consisting of fibers with lengths chosen to meet the dispersion requirements.

To verify that the fiber has the correct amount of dispersion, the measurement method defined in IEC 60793-1-42 may be used. The measurement is made in the linear power regime of the fiber.

The channel provides an optical return loss specified in Table 121–11. The state of polarization of the back reflection is adjusted to create the greatest RIN.

The mean DGD of the channel is to be less than the value specified in Table 121–11.

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121.8.5.3 TDECQ measurement method

The standard deviation of the noise of the O/E and oscilloscope combination, σ_S, is determined with no optical input signal and the same settings as used to capture the histograms described below.

OMA_{outer} is measured according to 121.8.4 on the equalized signal.

The test pattern specified for TDECQ (see Table 121–10) is transmitted repetitively by the optical lane under test and the oscilloscope is set up to capture samples from all symbols in the complete pattern without averaging.

If an equivalent-time sampling oscilloscope is used, the impact of the sampling process and the reference equalizer on transmitter noise must be compensated for, so that the correct magnitude of noise is present at the output of the equalizer.

The captured waveform is processed to find the largest noise that could be combined with the signal by an ideal reference receiver when optimally equalized by a reference equalizer. The optimal equalizer tap coefficients are dependent on the amount of noise that can be added to the signal, so finding the noise that can be added and the optimal equalizer setting is an iterative process. One way of doing this, using estimated PAM4 symbol error ratio as the figure of merit for the equalized signal, is described below.

The reference equalizer (specified in 121.8.5.4) is applied to the waveform. The sum of the equalizer tap coefficients is equal to 1. An eye diagram is formed from the equalized captured waveform.

The average optical power (P_{ave}) of the equalized eye diagram is determined, and the 0 UI and 1 UI crossing points are determined by the average of the eye diagram crossing times, as measured at P_{ave}, as illustrated in Figure 121–5.

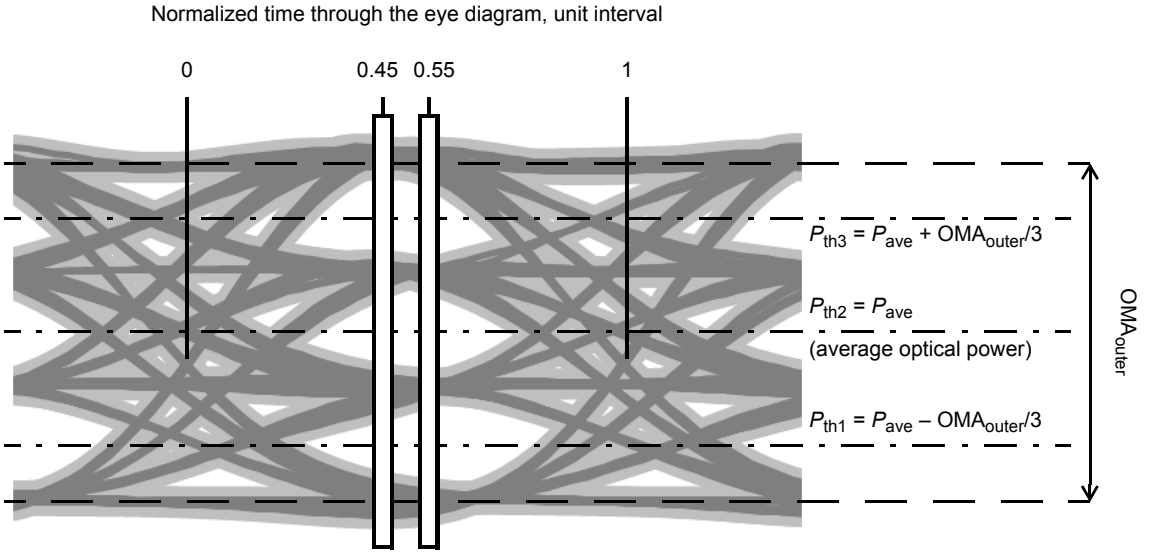


Figure 121–5—Illustration of the TDECQ measurement

Two vertical histograms are measured through the eye diagram, nominally centered at 0.45 UI and 0.55 UI. Each of the histogram windows spans all of the modulation levels of the eye diagram, as illustrated in Figure 121–5. The precise time position of the pair of histograms is adjusted to minimize TDECQ while keeping the histograms spaced 0.1 UI apart.

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Each histogram window has a width of 0.04 UI. Each histogram window has outer height boundaries which are set beyond the extremes of the eye diagram (so that no further samples would be captured by increasing the vertical separation of the height boundaries).

The sub-eye threshold levels P_{th1}, P_{th2}, and P_{th3}, are determined from the OMA_{outer} and the average optical power of the eye diagram (P_{ave}) as defined in Equation (121–1), Equation (121–2), and Equation (121–3) and illustrated in Figure 121–5.

$$P_{th1} = P_{ave} - \frac{OMA_{outer}}{3} \tag{121–1}$$

$$P_{th2} = P_{ave} \tag{121–2}$$

$$P_{th3} = P_{ave} + \frac{OMA_{outer}}{3} \tag{121–3}$$

Each captured histogram is processed to, in effect, combine the PAM4 waveform with noise, in order to produce an estimate of the partial PAM4 symbol error ratio (SER) for each sub-eye. One way of doing this is described below.

The left and right histograms are each normalized, and can be represented as a series of equally spaced optical power values (y_i) with separation Δy, each with an associated fraction $F(y_i)$, equal to the number of samples captured in that power interval divided by the total number of samples in that histogram. The sum of all $F(y_i)$ for each histogram is equal to 1.

From the left normalized histogram $F(y_i)$, three cumulative probability functions are created, $CF_{L1}(y_i)$, $CF_{L2}(y_i)$, and $CF_{L3}(y_i)$, one around each sub-eye threshold. The right histogram is treated similarly to create three cumulative probability functions $CF_{R1}(y_i)$, $CF_{R2}(y_i)$, and $CF_{R3}(y_i)$. $CF_{L1}(y_i)$ is defined in Equation (121–4).

$$CF_{L1}(y_i) = \begin{cases} \sum_{y=P_{th1}}^{y_i} F(y) & \text{for } y_i \geq P_{th1} \\ \sum_{y=y_i}^{P_{th1}} F(y) & \text{for } y_i < P_{th1} \end{cases} \tag{121–4}$$

Each element of the cumulative probability function, $CF_{L1}(y_i)$, is multiplied by a value $G_{th1}(y_i)$, and then summed to calculate an approximation for SER_{L1}, the partial SER for threshold 1. Each element of the cumulative probability function, $CF_{L2}(y_i)$, is multiplied by a value $G_{th2}(y_i)$, and then summed to calculate an approximation for SER_{L2}. Each element of the cumulative probability function, $CF_{L3}(y_i)$, is multiplied by a value $G_{th3}(y_i)$, and then summed to calculate an approximation for SER_{L3}. The sum of the three partial SERs is the SER associated with the left histogram, SER_L.

Each element of the cumulative probability function, $CF_{R1}(y_i)$, is multiplied by a value $G_{th1}(y_i)$, and then summed to calculate an approximation for SER_{R1}, the partial SER for threshold 1. $CF_{R2}(y_i)$ and $CF_{R3}(y_i)$ are treated similarly to calculate SER_{R2}, and SER_{R3}, the partial SERs for threshold 2 and threshold 3. The sum of the three partial SERs is the SER associated with the right histogram, SER_R.

$G_{th1}(y_i)$ is equivalent to a Gaussian probability density function with an RMS value of σ_G, centered around the sub-eye threshold P_{th1}. $G_{th1}(y_i)$ is given by Equation (121–5) and can be estimated by Equation (121–6).

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$$G_{\text{th1}}(y_i) = \int_{y_i-\frac{\Delta y}{2}}^{y_i+\frac{\Delta y}{2}} \frac{1}{C_{\text{eq}}\sigma_G\sqrt{2\pi}} \times \text{e}^{-\left(\frac{y-P_{\text{th1}}}{C_{\text{eq}}\sigma_G\sqrt{2}}\right)^2} dy \quad (121\text{–}5)$$

$$G_{\text{th1}}(y_i) = \frac{1}{C_{\text{eq}}\sigma_G\sqrt{2\pi}} \times \text{e}^{-\left(\frac{y_i-P_{\text{th1}}}{C_{\text{eq}}\sigma_G\sqrt{2}}\right)^2} \times \Delta y \quad (121\text{–}6)$$

$G_{\text{th2}}(y_i)$ and $G_{\text{th3}}(y_i)$ are similar Gaussian probability density functions with the same RMS value of σ_G, centered around the sub-eye thresholds P_{th2} and P_{th3} respectively. $G_{\text{th2}}(y_i)$ and $G_{\text{th3}}(y_i)$ are given by Equation (121–7) and Equation (121–8) respectively.

$$G_{\text{th2}}(y_i) = \int_{y_i-\frac{\Delta y}{2}}^{y_i+\frac{\Delta y}{2}} \frac{1}{C_{\text{eq}}\sigma_G\sqrt{2\pi}} \times \text{e}^{-\left(\frac{y-P_{\text{th2}}}{C_{\text{eq}}\sigma_G\sqrt{2}}\right)^2} dy \quad (121\text{–}7)$$

$$G_{\text{th3}}(y_i) = \int_{y_i-\frac{\Delta y}{2}}^{y_i+\frac{\Delta y}{2}} \frac{1}{C_{\text{eq}}\sigma_G\sqrt{2\pi}} \times \text{e}^{-\left(\frac{y-P_{\text{th3}}}{C_{\text{eq}}\sigma_G\sqrt{2}}\right)^2} dy \quad (121\text{–}8)$$

where

C_{eq} is a coefficient which accounts for the reference equalizer noise enhancement

The value of C_{eq} can be calculated from the product of the normalized noise power density spectrum $N(f)$ at the input of the reference equalizer and the normalized frequency response $H_{\text{eq}}(f)$ of the reference equalizer, as shown in Equation (121–9).

$$C_{\text{eq}} = \sqrt{\int_f N(f) \times \left|H_{\text{eq}}(f)\right|^2 df} \quad (121\text{–}9)$$

where

$N(f)$ is the normalized noise power density spectrum equivalent to white noise filtered by a fourth-order Bessel-Thomson response filter with a bandwidth of 13.28125 GHz.

and

$$\int_f N(f)df = H_{\text{eq}}(f = 0) = 1 \quad (121\text{–}10)$$

The equalizer tap coefficients are iteratively adjusted and SER_L and SER_R calculated until the largest of SER_L and SER_R is minimized. Then, if the larger of SER_L and SER_R is greater than the target SER of 4.8×10^{-4}, the value of σ_G is decreased and the process of equalizer optimization is repeated; If the larger of SER_L and SER_R is lower than the target SER of 4.8×10^{-4}, then the value of σ_G is increased and the process of equalizer optimization is repeated.

When the larger of SER_L and SER_R is equal to the target SER of 4.8×10^{-4}, and the value of σ_G cannot be increased by further optimization of the equalizer tap coefficients, then TDECQ is calculated.

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The RMS noise, R, that could be added by a receiver is given by Equation (121–11).

$$R = \sqrt{\sigma_G^{\,2} + \sigma_S^{\,2}} \tag{121–11}$$

TDECQ is given by Equation (121–12).

$$TDECQ = 10\log_{10}\left(\frac{OMA_{\text{outer}}}{6} \times \frac{1}{Q_t R}\right) \tag{121–12}$$

where

OMA_{outer} is the Outer Optical Modulation Amplitude as defined in 121.8.4

Q_t is 3.414 consistent with the BER and target symbol error ratio for Gray coded PAM4

121.8.5.4 TDECQ reference equalizer

The reference equalizer for 200GBASE-DR4 is a 5 tap, T spaced, feed-forward equalizer (FFE), where T is the symbol period. The sum of the equalizer tap coefficients is equal to 1.

NOTE—This reference equalizer is part of the TDECQ test and does not imply any particular receiver equalizer implementation.

121.8.6 Extinction ratio

The extinction ratio of each lane shall be within the limits given in Table 121–6 if measured using a test pattern specified for extinction ratio in Table 121–10. The extinction ratio of a PAM4 optical signal is defined as the ratio of the average optical launch power level P_3, measured over the central 2 UI of a run of 7 threes, and the average optical launch power level P_0, measured over the central 2 UI of a run of 6 zeros, as shown in Figure 121–3.

121.8.7 Relative intensity noise ($RIN_{21.4}OMA$)

RIN shall be as defined by the measurement methodology of 52.9.6 with the following exceptions:

a) The optical return loss is 21.4 dB.

b) Each lane may be tested individually with the sum of the optical power from all of the lanes not under test being below –30 dBm.

c) The upper –3 dB limit of the measurement apparatus is to be approximately equal to the signaling rate (i.e., 26.6 GHz).

d) The test pattern is according to Table 121–10.

121.8.8 Receiver sensitivity

Receiver sensitivity, which is defined for an input signal with SECQ of 0.9 dB (e.g., an ideal input signal without overshoot), is informative and compliance is not required. If measured, the test signal should have negligible impairments such as intersymbol interference (ISI), rise/fall times, jitter and RIN. Instead, the normative requirement for receivers is stressed receiver sensitivity.

121.8.9 Stressed receiver sensitivity

Stressed receiver sensitivity shall be within the limits given in Table 121–7 if measured using the method defined in 121.8.9.1 and 121.8.9.3, with the conformance test signal at TP3 as described in 121.8.9.2, using the test pattern specified for stressed receiver sensitivity in Table 121–10. The BER is required to be met for the lane under test on its own.

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Stressed receiver sensitivity is defined with all transmit and receive lanes in operation. Any of the patterns specified for stressed receiver sensitivity in Table 121–10 is sent from the transmit section of the PMD under test. The signal being transmitted is asynchronous to the received signal.

121.8.9.1 Stressed receiver conformance test block diagram

A block diagram for the receiver conformance test is shown in Figure 121–6. The patterns used for the received conformance signal are specified in Table 121–10. The optical test signal is conditioned (stressed) using the stressed receiver methodology defined in 121.8.9.2 and has sinusoidal jitter applied as specified in 121.8.9.4. A suitable test set is needed to characterize the signal used to test the receiver. Stressed receiver conformance test signal verification is described in 121.8.9.3.

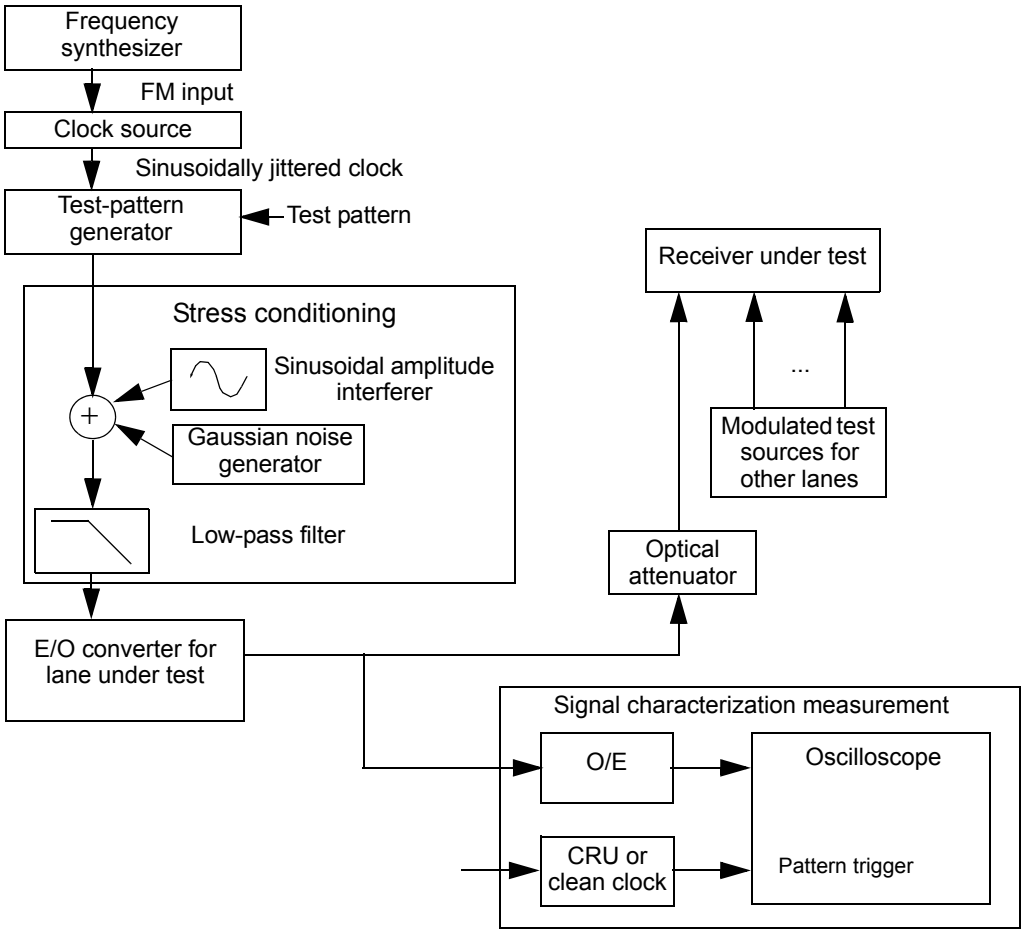


Figure 121–6—Stressed receiver conformance test block diagram

The low-pass filter is used to create ISI. The combination of the low-pass filter and the E/O converter should have a frequency response that results in at least half of the dB value of the stressed eye closure (SECQ) specified in Table 121–7 before the sinusoidal and Gaussian noise terms are added, according to the methods specified in 121.8.9.2. The sinusoidal amplitude interferer causes additional eye closure, but in conjunction with the finite edge rates, also causes some jitter.

The sinusoidally jittered clock represents other forms of jitter and also verifies that the receiver under test can track low-frequency jitter. The sinusoidal amplitude interferer may be set at any frequency between

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100 MHz and 2 GHz, although care should be taken to avoid harmonic relationships between the sinusoidal interferer, the sinusoidal jitter, the signaling rate, and the pattern repetition rate. The Gaussian noise generator, the amplitude of the sinusoidal interferer, and the low-pass filter are adjusted so that the SECQ specified in Table 121–7 is met, according to the methods specified in 121.8.9.2.

For improved visibility for calibration, all elements in the signal path (cables, DC blocks, E/O converter, etc.) should have wide and smooth frequency response, and linear phase response, throughout the spectrum of interest. Baseline wander and overshoot and undershoot should be negligible.

121.8.9.2 Stressed receiver conformance test signal characteristics and calibration

The conformance test signal is used to validate that each lane of the PMD receiver meets BER requirements with near worst-case waveforms at TP3.

The primary parameters of the stressed receiver conformance test signal are its stressed eye closure (SECQ), and the sinusoidal jitter applied, as specified in 121.8.9.4. The SECQ of the stressed receiver conformance test signal is measured according to 121.8.5, except that the test fiber is not used. An example stressed receiver conformance test setup is shown in Figure 121–6; however, alternative test setups that generate equivalent stress conditions may be used.

The following steps describe a possible method for setting up and calibrating a stressed receiver conformance test signal when using a stressed receiver conformance test setup as shown in Figure 121–6:

1) Set the signaling rate of the test pattern generator to meet the requirements in Table 121–6.
2) With the sinusoidal jitter, sinusoidal interferer, and the Gaussian noise generator turned off, set the extinction ratio of the E/O converter to approximately the minimum specified in Table 121–6.
3) The required value of SECQ is given in Table 121–7. With the sinusoidal jitter, sinusoidal interferer, and Gaussian noise generator turned off, at least half of the dB value of SECQ should be created by the selection of the appropriate bandwidth for the combination of the low-pass filter and the E/O converter. Any remaining SECQ must be created with a combination of sinusoidal jitter, sinusoidal interference, and Gaussian noise. Sinusoidal jitter is added as specified in Table 121–12.

When calibrating the conformance signal, the sinusoidal jitter frequency should be between 50 MHz and 10 times LB as defined in Table 121–12. Sinusoidal jitter amplitude may be calibrated by measuring the jitter on the oscilloscope, while transmitting the square wave pattern, as defined in Table 121–9, and using a clean clock in place of the CRU to trigger the oscilloscope.

Iterate the adjustments of the sinusoidal interferer, the Gaussian noise generator, and extinction ratio, until the required value of SECQ is met, while also meeting the following conditions: the extinction ratio is approximately the minimum specified in Table 121–6; and sinusoidal jitter is as specified in Table 121–12.

Each receiver lane is conformance tested in turn. The source for the lane under test is adjusted to supply a signal at the input to the receiver under test at the "Stressed receiver sensitivity (OMA_{outer}), each lane (max)" specified in Table 121–7, and the test sources for the other lanes are set to the "OMA_{outer} of each aggressor lane" specified in Table 121–7.

121.8.9.3 Stressed receiver conformance test signal verification

The SECQ of the stressed receiver conformance test signal is measured according to 121.8.5, except that the test fiber is not used. The clock output from the clock source in Figure 121–6 is modulated with the sinusoidal jitter. To use an oscilloscope to calibrate the final stressed eye jitter that includes the sinusoidal jitter component, a separate clock source (clean clock of Figure 121–6) is required that is synchronized to the source clock, but not modulated with the jitter source.

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Care should be taken when characterizing the test signal because excessive noise/jitter in the measurement system would result in an input signal that does not fully stress the receiver under test. Running the receiver tolerance test with a signal that is under-stressed may result in the deployment of non-compliant receivers. The noise/jitter introduced by the O/E, filters, and oscilloscope should be negligible or the results should be corrected for its effects. While the details of test equipment are beyond the scope of this standard, it is recommended that the implementer fully characterize the test equipment and apply appropriate guard bands so that the stressed receiver conformance input signal meets the stress and sinusoidal jitter specified in 121.8.9.2 and 121.8.9.4.

121.8.9.4 Sinusoidal jitter for receiver conformance test

The sinusoidal jitter is used to test receiver jitter tolerance. The amplitude of the applied sinusoidal jitter is dependent on frequency as specified in Table 121–12 and is illustrated in Figure 121–7.

Table 121–12—Applied sinusoidal jitter

Frequency range	Sinusoidal jitter, peak-to-peak (UI)
$f < 40$ kHz	Not specified
40 kHz $< f \leq 4$ MHz	$2 \times 10^5 / f$
4 MHz $< f <$ 10 LB[a]	0.05

[a] LB = loop bandwidth; upper frequency bound for added sine jitter should be at least 10 times the loop bandwidth of the receiver being tested.

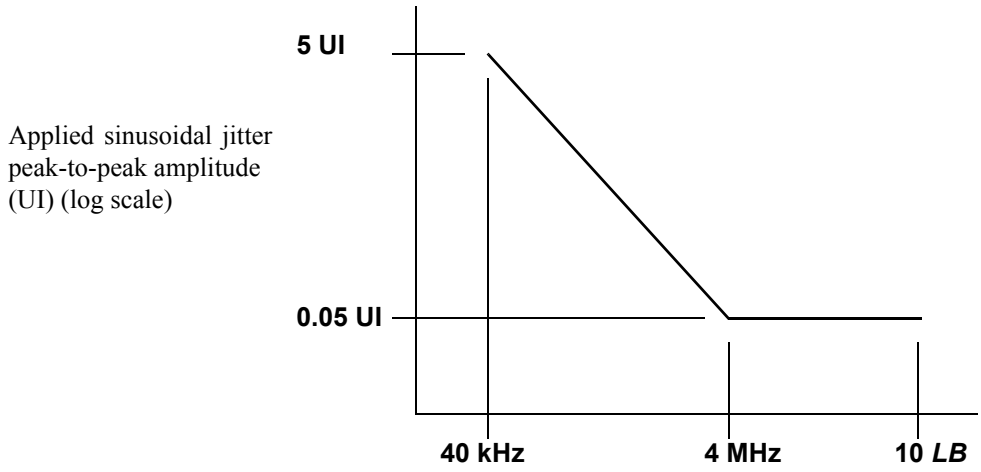


Figure 121–7—Illustration of the mask of the sinusoidal component of jitter tolerance

121.9 Safety, installation, environment, and labeling

121.9.1 General safety

All equipment subject to this clause shall conform to IEC 60950-1.

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121.9.2 Laser safety

200GBASE-DR4 optical transceivers shall conform to Hazard Level 1 laser requirements as defined in IEC 60825-1 and IEC 60825-2, under any condition of operation. This includes single fault conditions whether coupled into a fiber or out of an open bore.

Conformance to additional laser safety standards may be required for operation within specific geographic regions.

Laser safety standards and regulations require that the manufacturer of a laser product provide information about the product's laser, safety features, labeling, use, maintenance, and service. This documentation explicitly defines requirements and usage restrictions on the host system necessary to meet these safety certifications.[6]

121.9.3 Installation

It is recommended that proper installation practices, as defined by applicable local codes and regulation, be followed in every instance in which such practices are applicable.

121.9.4 Environment

Normative specifications in this clause shall be met by a system integrating a 200GBASE-DR4 PMD over the life of the product while the product operates within the manufacturer's range of environmental, power, and other specifications.

It is recommended that manufacturers indicate in the literature associated with the PHY the operating environmental conditions to facilitate selection, installation, and maintenance.

It is recommended that manufacturers indicate, in the literature associated with the components of the optical link, the distance and operating environmental conditions over which the specifications of this clause will be met.

121.9.5 Electromagnetic emission

A system integrating a 200GBASE-DR4 PMD shall comply with applicable local and national codes for the limitation of electromagnetic interference.

121.9.6 Temperature, humidity, and handling

The optical link is expected to operate over a reasonable range of environmental conditions related to temperature, humidity, and physical handling (such as shock and vibration). Specific requirements and values for these parameters are considered to be beyond the scope of this standard.

121.9.7 PMD labeling requirements

It is recommended that each PHY (and supporting documentation) be labeled in a manner visible to the user, with at least the applicable safety warnings and the applicable port type designation (e.g., 200GBASE-DR4).

Labeling requirements for Hazard Level 1 lasers are given in the laser safety standards referenced in 121.9.2.

[6]A host system that fails to meet the manufacturer's requirements and/or usage restrictions may emit laser radiation in excess of the safety limits of one or more safety standards. In such a case, the host manufacturer is required to obtain its own laser safety certification.

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121.10 Fiber optic cabling model

The fiber optic cabling model is shown in Figure 121–8.

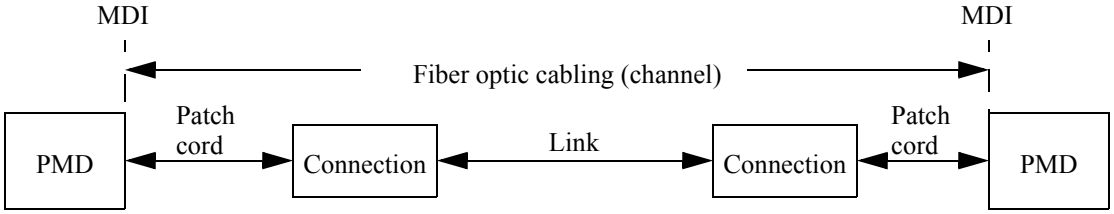


Figure 121–8—Fiber optic cabling model

The channel insertion loss is given in Table 121–13. A channel may contain additional connectors as long as the optical characteristics of the channel, such as attenuation, dispersion, reflections, and polarization mode dispersion meet the specifications. Insertion loss measurements of installed fiber cables are made in accordance with ANSI/TIA/EIA-526-7/method A-1. The fiber optic cabling model (channel) defined here is the same as a simplex fiber optic link segment. The term channel is used here for consistency with generic cabling standards.

Table 121–13—Fiber optic cabling (channel) characteristics

Description	200GBASE-DR4	Unit
Operating distance (max)	500	m
Channel insertion loss[a, b](max)	3	dB
Channel insertion loss (min)	0	dB
Positive dispersion[b] (max)	0.8	ps/nm
Negative dispersion[b] (min)	–0.93	ps/nm
DGD_max[c]	2.24	ps
Optical return loss (min)	37	dB

[a] These channel insertion loss values include cable, connectors, and splices.
[b] Over the wavelength range 1304.5 nm to 1317.5 nm.
[c] Differential Group Delay (DGD) is the time difference at reception between the fractions of a pulse that were transmitted in the two principal states of polarization of an optical signal. DGD_max is the maximum differential group delay that the system must tolerate.

121.11 Characteristics of the fiber optic cabling (channel)

The 200GBASE-DR4 fiber optic cabling shall meet the specifications defined in Table 121–14. The fiber optic cabling consists of one or more sections of fiber optic cable and any intermediate connections required to connect sections together.

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121.11.1 Optical fiber cable

The fiber optic cable requirements are satisfied by cables containing IEC 60793-2-50 type B1.1 (dispersion un-shifted single-mode), type B1.3 (low water peak single-mode), or type B6_a (bend insensitive) fibers or the requirements in Table 121–14 where they differ.

Table 121–14—Optical fiber and cable characteristics

Description	Value	Unit
Nominal fiber specification wavelength	1310	nm
Cabled optical fiber attenuation (max)	0.5[a]	dB/km
Zero dispersion wavelength (λ_0)	$1300 \leq \lambda_0 \leq 1324$	nm
Dispersion slope (max) (S_0)	0.093	ps/nm^2 km

[a] The 0.5 dB/km attenuation is provided for Outside Plant cable as defined in ANSI/TIA 568-C.3.

121.11.2 Optical fiber connection

An optical fiber connection, as shown in Figure 121–8, consists of a mated pair of optical connectors.

121.11.2.1 Connection insertion loss

The maximum link distance is based on an allocation of 2.75 dB total connection and splice loss. For example, this allocation supports five connections with an average insertion loss per connection of 0.5 dB. Connections with different loss characteristics may be used provided the requirements of Table 121–13 are met.

121.11.2.2 Maximum discrete reflectance

The maximum value for each discrete reflectance shall be less than or equal to the value shown in Table 121–15 corresponding to the number of discrete reflectances above –55 dB within the channel. For numbers of discrete reflectances in between two numbers shown in the table, the lower of the two corresponding maximum discrete reflectance values applies.

Table 121–15—Maximum value of each discrete reflectance

Number of discrete reflectances above –55 dB	Maximum value for each discrete reflectance
1	–37 dB
2	–42 dB
4	–45 dB
6	–47 dB
8	–48 dB
10	–49 dB

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121.11.3 Medium Dependent Interface (MDI)

The 200GBASE-DR4 PMD is coupled to the fiber optic cabling at the MDI. The MDI is the interface between the PMD and the "fiber optic cabling" (as shown in Figure 121–8). The 200GBASE-DR4 PMD is coupled to the fiber optic cabling through one connector plug into the MDI optical receptacle as shown in Figure 121–10. Example constructions of the MDI include the following:

a) PMD with a connectorized fiber pigtail plugged into an adapter

b) PMD receptacle

121.11.3.1 Optical lane assignments

The four transmit and four receive optical lanes of 200GBASE-DR4 shall occupy the positions depicted in Figure 121–9 when looking into the MDI receptacle with the connector keyway feature on top. The interface contains eight active lanes within twelve total positions. The transmit optical lanes occupy the left-most four positions. The receive optical lanes occupy the right-most four positions. The four center positions are unused.

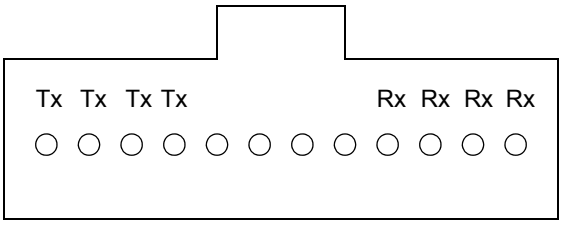


Figure 121–9—200GBASE-DR4 optical lane assignments

121.11.3.2 Medium Dependent Interface (MDI) requirements

The MDI shall meet the dimensional specifications of IEC 61754-7-1 interface 7-1-9: *MPO device receptacle, angled interface*. The plug terminating the optical fiber cabling shall meet the dimensional specifications of IEC 61754-7-1 interface 7-1-1: *MPO female plug connector, down-angled interface for 2 to 12 fibres*. The MDI shall optically mate with the plug on the optical fiber cabling. Figure 121–10 shows an MPO female plug connector with down-angled interface, and an MDI as an active device receptacle with angled interface.

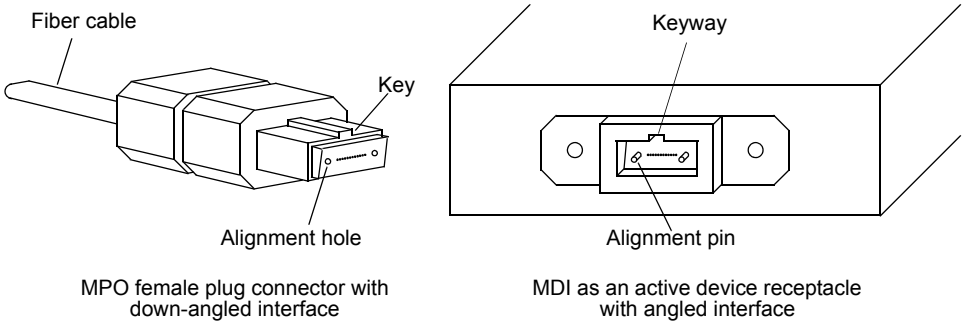


Figure 121–10—MPO female plug with down-angled interface and MDI active device receptacle with angled interface

The MDI shall meet the interface performance specifications of IEC 61753-021-2 for performance level D/2.

NOTE—Transmitter compliance testing is performed at TP2 as defined in 121.5.1, not at the MDI.

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121.12 Protocol implementation conformance statement (PICS) proforma for Clause 121, Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-DR4[7]

121.12.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Clause 121, Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-DR4, shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

121.12.2 Identification

121.12.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

121.12.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Clause 121, Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-DR4
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[7] *Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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121.12.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
*INS	Installation / cable	121.10	Items marked with INS include installation practices and cable specifications not applicable to a PHY manufacturer	O	Yes [ ] No [ ]
TP1	Reference point TP1 exposed and available for testing	121.5.1	This point may be made available for use by implementers to certify component conformance	O	Yes [ ] No [ ]
TP4	Reference point TP4 exposed and available for testing	121.5.1	This point may be made available for use by implementers to certify component conformance	O	Yes [ ] No [ ]
DC	Delay constraints	121.3.1	Device conforms to delay constraints	M	Yes [ ]
SC	Skew constraints	121.3.2	Device conforms to Skew and Skew Variation constraints	M	Yes [ ]
*MD	MDIO capability	121.4	Registers and interface supported	O	Yes [ ] No [ ]

121.12.4 PICS proforma tables for Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-DR4

121.12.4.1 PMD functional specifications

Item	Feature	Subclause	Value/Comment	Status	Support
F1	Compatible with 200GBASE-R PCS and PMA	121.1		M	Yes [ ]
F2	Integration with management functions	121.1		O	Yes [ ] No [ ]
F3	Bit error ratio	121.1.1	Meets the BER specified in 121.1.1	M	Yes [ ]
F4	Transmit function	121.5.2	Conveys symbols from PMD service interface to MDI	M	Yes [ ]
F5	Mapping between optical signal and logical signal for transmitter	121.5.2	Highest optical power corresponds to tx_symbol = three	M	Yes [ ]
F6	Receive function	121.5.3	Conveys symbols from MDI to PMD service interface	M	Yes [ ]
F7	Conversion of four optical signals to four electrical signals	121.5.3	For delivery to the PMD service interface	M	Yes [ ]
F8	Mapping between optical signal and logical signal for receiver	121.5.3	Highest optical power corresponds to rx_symbol = three	M	Yes [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
F9	Global Signal Detect function	121.5.4	Report to the PMD service interface the message PMD:IS_SIGNAL.indication (SIGNAL_DETECT)	M	Yes [ ]
F10	Global Signal Detect behavior	121.5.4	SIGNAL_DETECT is a global indicator of the presence of optical signals on all four lanes	M	Yes [ ]
F11	Lane-by-lane Signal Detect function	121.5.5	Sets PMD_signal_detect_*i* values on a lane-by-lane basis per requirements of Table 121–4	MD:O	Yes [ ] No [ ] N/A [ ]
F12	PMD reset function	121.5.6	Resets the PMD sublayer	MD:O	Yes [ ] No [ ] N/A [ ]

121.12.4.2 Management functions

Item	Feature	Subclause	Value/Comment	Status	Support
M1	Management register set	121.4		MD:M	Yes [ ] N/A [ ]
M2	Global transmit disable function	121.5.7	Disables all of the optical transmitters with the PMD_global_transmit_disable variable	MD:O	Yes [ ] No [ ] N/A [ ]
M3	PMD_lane_by_lane_transmit_disable function	121.5.8	Disables the optical transmitter on the lane associated with the PMD_transmit_disable_*i* variable	MD:O	Yes [ ] No [ ] N/A [ ]
M4	PMD_fault function	121.5.9	Sets PMD_fault to one if any local fault is detected	MD:O	Yes [ ] No [ ] N/A [ ]
M5	PMD_transmit_fault function	121.5.10	Sets PMD_transmit_fault to one if a local fault is detected on any transmit lane	MD:O	Yes [ ] No [ ] N/A [ ]
M6	PMD_receive_fault function	121.5.11	Sets PMD_receive_fault to one if a local fault is detected on any receive lane	MD:O	Yes [ ] No [ ] N/A [ ]

121.12.4.3 PMD to MDI optical specifications for 200GBASE-DR4

Item	Feature	Subclause	Value/Comment	Status	Support
DR1	Transmitter meets specifications in Table 121–6	121.7.1	Per definitions in 121.8	M	Yes [ ]
DR2	Receiver meets specifications in Table 121–7	121.7.2	Per definitions in 121.8	M	Yes [ ]

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121.12.4.4 Optical measurement methods

Item	Feature	Subclause	Value/Comment	Status	Support
OM1	Measurement cable	121.8	2 m to 5 m in length	M	Yes [ ]
OM2	Center wavelength	121.8.2	Per TIA/EIA-455-127-A or IEC 61280-1-3 under modulated conditions	M	Yes [ ]
OM3	Average optical power	121.8.3	Per IEC 61280-1-1	M	Yes [ ]
OM4	OMA measurements	121.8.4	Each lane	M	Yes [ ]
OM5	Transmitter and dispersion eye closure for PAM4 (TDECQ)	121.8.5	Each lane	M	Yes [ ]
OM6	Extinction ratio	121.8.6	Each lane	M	Yes [ ]
OM7	Stressed receiver sensitivity	121.8.9	Each lane	M	Yes [ ]

121.12.4.5 Environmental specifications

Item	Feature	Subclause	Value/Comment	Status	Support
ES1	General safety	121.9.1	Conforms to IEC 60950-1	M	Yes [ ]
ES2	Laser safety—IEC Hazard Level 1	121.9.2	Conforms to Hazard Level 1 laser requirements defined in IEC 60825-1 and IEC 60825-2	M	Yes [ ]
ES3	Electromagnetic interference	121.9.5	Complies with applicable local and national codes for the limitation of electromagnetic interference	M	Yes [ ]

121.12.4.6 Characteristics of the fiber optic cabling and MDI

Item	Feature	Subclause	Value/Comment	Status	Support
OC1	Fiber optic cabling	121.11	Meets requirements specified in Table 121–13	INS:M	Yes [ ] N/A [ ]
OC2	Maximum discrete reflectance	121.11.2.2	Meets requirements specified in Table 121–15	INS:M	Yes [ ] N/A [ ]
OC3	MDI layout	121.11.3.1	Optical lane assignments per Figure 121–9	M	Yes [ ]
OC4	MDI dimensions	121.11.3.2	Per IEC 61754-7-1 interface 7-1-9	M	Yes [ ]
OC5	Cabling connector dimensions	121.11.3.2	Per IEC 61754-7-1 interface 7-1-1	INS:M	Yes [ ] N/A [ ]
OC6	MDI mating	121.11.3.2	MDI optically mates with plug on the cabling	M	Yes [ ]

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122. Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8

122.1 Overview

This clause specifies the 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and the 400GBASE-LR8 PMDs together with the single-mode fiber medium. The optical signals generated by these four PMD types are modulated using a 4-level pulse amplitude modulation (PAM4) format. When forming a complete Physical Layer, a PMD shall be connected to the appropriate PMA as shown in Table 122–1, to the medium through the MDI and optionally with the management functions that may be accessible through the management interface defined in Clause 45, or equivalent.

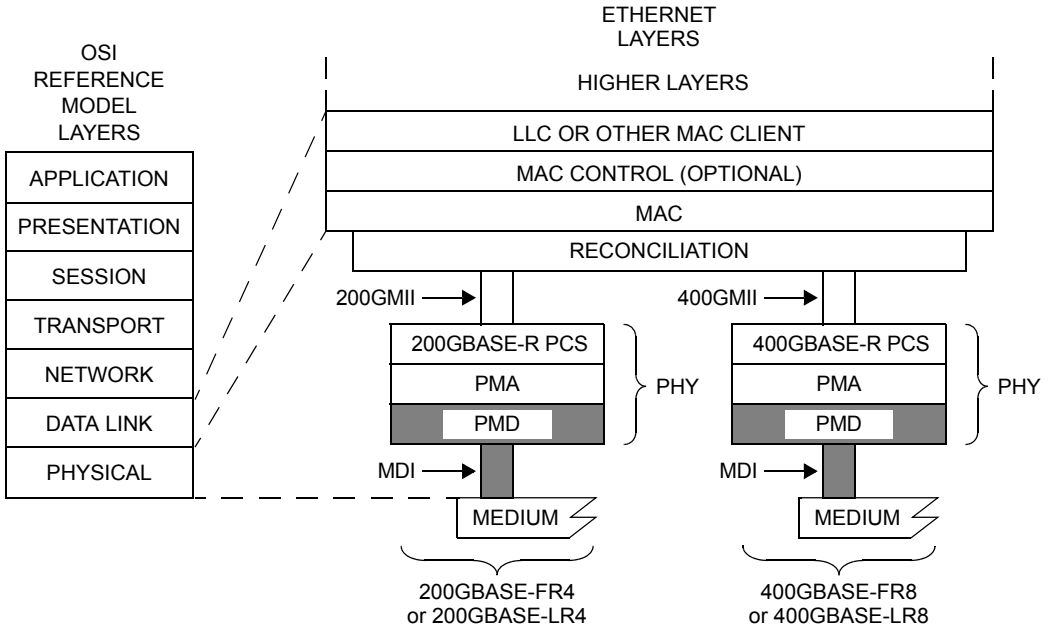
Table 122–1—Physical Layer clauses associated with the 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 PMDs

Associated clause	200GBASE-FR4, 200GBASE-LR4	400GBASE-FR8, 400GBASE-LR8
117—RS	Required	Required
117—200GMII[a]	Optional	Not applicable
117—400GMII[a]	Not applicable	Optional
118—200GMII Extender	Optional	Not applicable
118—400GMII Extender	Not applicable	Optional
119—PCS for 200GBASE-R	Required	Not applicable
119—PCS for 400GBASE-R	Not applicable	Required
120—PMA for 200GBASE-R	Required	Not applicable
120—PMA for 400GBASE-R	Not applicable	Required
120B—Chip-to-chip 200GAUI-8	Optional	Not applicable
120B—Chip-to-chip 400GAUI-16	Not applicable	Optional
120C—Chip-to-module 200GAUI-8	Optional	Not applicable
120C—Chip-to-module 400GAUI-16	Not applicable	Optional
120D—Chip-to-chip 200GAUI-4	Optional	Not applicable
120D—Chip-to-chip 400GAUI-8	Not applicable	Optional
120E—Chip-to-module 200GAUI-4	Optional	Not applicable
120E—Chip-to-module 400GAUI-8	Not applicable	Optional
78—Energy Efficient Ethernet	Optional	Optional

[a] 200GMII and 400GMII are optional interfaces. However, if the appropriate interface is not implemented, a conforming implementation must behave functionally as though the RS and 200GMII or 400GMII were present.

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Figure 122–1 shows the relationship of the PMD and MDI (shown shaded) with other sublayers to the ISO/IEC Open System Interconnection (OSI) reference model. 200 Gb/s and 400 Gb/s Ethernet are introduced in Clause 116 and the purpose of each PHY sublayer is summarized in 116.2.



400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
LLC = LOGICAL LINK CONTROL
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

FR = PMD FOR SINGLE-MODE FIBER — 2 km
LR = PMD FOR SINGLE-MODE FIBER — 10 km

Figure 122–1—200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 PMDs relationship to the ISO/IEC Open Systems Interconnection (OSI) reference model and IEEE 802.3 Ethernet model

200GBASE-FR4 and 200GBASE-LR4 use four lanes, while 400GBASE-FR8 and 400GBASE-LR8 use eight lanes. In this clause, where there are four or eight items (depending on PMD type) such as lanes, the items are numbered from 0 to $n - 1$, and an example item is numbered i. Thus n is 4 or 8.

200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 PHYs with the optional Energy Efficient Ethernet (EEE) fast wake capability may enter the Low Power Idle (LPI) mode to conserve energy during periods of low link utilization (see Clause 78). The deep sleep mode of EEE is not supported.

122.1.1 Bit error ratio

The bit error ratio (BER) when processed according to Clause 120 shall be less than 2.4×10^{-4} provided that the error statistics are sufficiently random that this results in a frame loss ratio (see 1.4.223) of less than 1.7×10^{-12} for 64-octet frames with minimum interpacket gap when processed according to Clause 120 and then Clause 119. For a complete Physical Layer, the frame loss ratio may be degraded to 6.2×10^{-11} for 64-octet frames with minimum interpacket gap due to additional errors from the electrical interfaces.

If the error statistics are not sufficiently random to meet this requirement, then the BER shall be less than that required to give a frame loss ratio of less than 1.7×10^{-12} for 64-octet frames with minimum interpacket gap.

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122.2 Physical Medium Dependent (PMD) service interface

This subclause specifies the services provided by the 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 PMDs. The service interfaces for these PMDs are described in an abstract manner and do not imply any particular implementation. The PMD service interface supports the exchange of encoded data between the PMA entity that resides just above the PMD, and the PMD entity. The PMD translates the encoded data to and from signals suitable for the specified medium.

The PMD service interface is an instance of the inter-sublayer service interface defined in 116.3. The PMD service interface primitives are summarized as follows:

PMD:IS_UNITDATA_*i*.request
PMD:IS_UNITDATA_*i*.indication
PMD:IS_SIGNAL.indication

The 200GBASE-FR4 and 200GBASE-LR4 PMDs have four parallel symbol streams, in which case $i = 0$ to 3, and the 400GBASE-FR8 and 400GBASE-LR8 PMDs have eight parallel symbol streams, in which case $i = 0$ to 7.

In the transmit direction, the PMA continuously sends n parallel symbol streams to the PMD, one per lane, each at a nominal signaling rate of 26.5625 GBd. The PMD then converts these streams of data units into the appropriate signals on the MDI.

In the receive direction, the PMD continuously sends n parallel symbol streams to the PMA corresponding to the signals received from the MDI, one per lane, each at a nominal signaling rate of 26.5625 GBd.

The SIGNAL_DETECT parameter defined in this clause maps to the SIGNAL_OK parameter in the PMD:IS_SIGNAL.indication(SIGNAL_OK) inter-sublayer service interface primitive defined in 116.3.

The SIGNAL_DETECT parameter can take on one of two values: OK or FAIL. When SIGNAL_DETECT = FAIL, the rx_symbol parameters are undefined.

NOTE—SIGNAL_DETECT = OK does not guarantee that the rx_symbol parameters are known to be good. It is possible for a poor quality link to provide sufficient light for a SIGNAL_DETECT = OK indication and still not meet the BER defined in 122.1.1.

122.3 Delay and Skew

122.3.1 Delay constraints

The sum of the transmit and receive delays at one end of the link contributed by the 200GBASE-FR4 or 200GBASE-LR4 PMD including 2 m of fiber in one direction shall be no more than 4096 bit times (8 pause_quanta or 20.48 ns). The sum of the transmit and receive delays at one end of the link contributed by the 400GBASE-FR8 or 400GBASE-LR8 PMD including 2 m of fiber in one direction shall be no more than 8192 bit times (16 pause_quanta or 20.48 ns). A description of overall system delay constraints and the definitions for bit times and pause_quanta can be found in 116.4 and its references.

122.3.2 Skew constraints

The Skew (relative delay between the lanes) and Skew Variation must be kept within limits so that the information on the lanes can be reassembled by the PCS. Skew and Skew Variation are defined in 116.5 and specified at the points SP1 to SP6 shown in Figure 116–4 and Figure 116–5.

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If the PMD service interface is physically instantiated so that the Skew at SP2 can be measured, then the Skew at SP2 is limited to 43 ns and the Skew Variation at SP2 is limited to 400 ps.

The Skew at SP3 (the transmitter MDI) shall be less than 54 ns and the Skew Variation at SP3 shall be less than 600 ps.

The Skew at SP4 (the receiver MDI) shall be less than 134 ns and the Skew Variation at SP4 shall be less than 3.4 ns.

If the PMD service interface is physically instantiated so that the Skew at SP5 can be measured, then the Skew at SP5 shall be less than 145 ns and the Skew Variation at SP5 shall be less than 3.6 ns.

For more information on Skew and Skew Variation see 116.5. The measurements of Skew and Skew Variation are defined in 87.8.2.

122.4 PMD MDIO function mapping

The optional MDIO capability described in Clause 45 defines several variables that may provide control and status information for and about the PMD. If the MDIO interface is implemented, the mapping of MDIO control variables to PMD control variables shall be as shown in Table 122–2 and the mapping of MDIO status variables to PMD status variables shall be as shown in Table 122–3.

Table 122–2—MDIO/PMD control variable mapping

MDIO control variable	PMA/PMD register name	Register/bit number	PMD control variable
Reset	PMA/PMD control 1 register	1.0.15	PMD_reset
Global PMD transmit disable	PMD transmit disable register	1.9.0	PMD_global_transmit_disable
PMD transmit disable *n*–1 to PMD transmit disable 0	PMD transmit disable register	1.9.8 to 1.9.1	PMD_transmit_disable_*n*–1 to PMD_transmit_disable_0

Table 122–3—MDIO/PMD status variable mapping

MDIO status variable	PMA/PMD register name	Register/ bit number	PMD status variable
Fault	PMA/PMD status 1 register	1.1.7	PMD_fault
Transmit fault	PMA/PMD status 2 register	1.8.11	PMD_transmit_fault
Receive fault	PMA/PMD status 2 register	1.8.10	PMD_receive_fault
Global PMD receive signal detect	PMD receive signal detect register	1.10.0	PMD_global_signal_detect
PMD receive signal detect *n*–1 to PMD receive signal detect 0	PMD receive signal detect register	1.10.8 to 1.10.1	PMD_signal_detect_*n*–1 to PMD_signal_detect_0

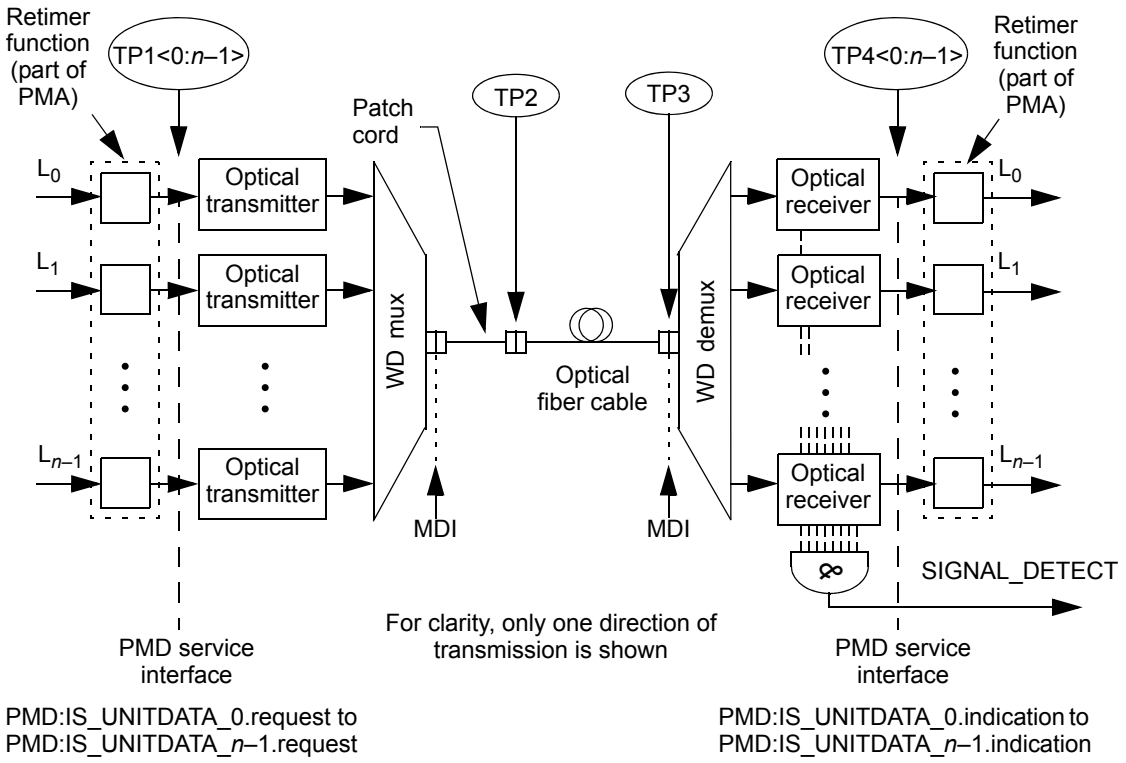
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122.5 PMD functional specifications

The 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 PMDs perform the Transmit and Receive functions, which convey data between the PMD service interface and the MDI.

122.5.1 PMD block diagram

The PMD block diagram is shown in Figure 122–2. For purposes of system conformance, the PMD sublayer is standardized at the points described in this subclause. The optical transmit signal is defined at the output end of a single-mode fiber patch cord (TP2), between 2 m and 5 m in length. Unless specified otherwise, all transmitter measurements and tests defined in 122.8 are made at TP2. The optical receive signal is defined at the output of the fiber optic cabling (TP3) at the MDI (see 122.11.3). Unless specified otherwise, all receiver measurements and tests defined in 122.8 are made at TP3.



NOTE—Specification of the retimer function and the electrical implementation of the PMD service interface is beyond the scope of this standard.

Figure 122–2—Block diagram for 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 transmit/receive paths

TP1<0:*n*–1> and TP4<0:*n*–1> are informative reference points that may be useful to implementers for testing components (these test points will not typically be accessible in an implemented system).

122.5.2 PMD transmit function

The PMD Transmit function shall convert the *n* symbol streams requested by the PMD service interface messages PMD:IS_UNITDATA_0.request to PMD:IS_UNITDATA_*n*–1.request into *n* separate optical

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signals. The *n* optical signals shall then be wavelength division multiplexed and delivered to the MDI, all according to the transmit optical specifications in this clause. The highest optical power level in each signal stream shall correspond to tx_symbol = three and the lowest shall correspond to tx_symbol = zero.

122.5.3 PMD receive function

The PMD Receive function shall demultiplex the composite optical signal received from the MDI into *n* separate optical signals. The *n* optical signals shall then be converted into *n* symbol streams for delivery to the PMD service interface using the messages PMD:IS_UNITDATA_0.indication to PMD:IS_UNITDATA_*n*–1.indication, all according to the receive optical specifications in this clause. The higher optical power level in each signal shall correspond to rx_symbol = three and the lowest shall correspond to rx_symbol = zero.

122.5.4 PMD global signal detect function

The PMD global signal detect function shall report the state of SIGNAL_DETECT via the PMD service interface. The SIGNAL_DETECT parameter is signaled continuously, while the PMD:IS_SIGNAL.indication message is generated when a change in the value of SIGNAL_DETECT occurs. The SIGNAL_DETECT parameter defined in this clause maps to the SIGNAL_OK parameter in the inter-sublayer service interface primitives defined in 116.3.

SIGNAL_DETECT shall be a global indicator of the presence of optical signals on all *n* lanes. The value of the SIGNAL_DETECT parameter shall be generated according to the conditions defined in Table 122–4. The PMD receiver is not required to verify whether a compliant 200GBASE-R or 400GBASE-R signal is being received. This standard imposes no response time requirements on the generation of the SIGNAL_DETECT parameter.

Table 122–4—SIGNAL_DETECT value definition

Receive conditions	SIGNAL_DETECT value
For any lane; Average optical power at TP3 ≤ –30 dBm	FAIL
For all lanes; [(Optical power at TP3 ≥ average receive power, each lane (min) in Table 122–11 for 200GBASE-FR4 and 200GBASE-LR4 or Table 122–12 for 400GBASE-FR8 and 400GBASE-LR8) AND (compliant 200GBASE-R or 400GBASE–R signal input)]	OK
All other conditions	Unspecified

As an unavoidable consequence of the requirements for the setting of the SIGNAL_DETECT parameter, implementations must provide adequate margin between the input optical power level at which the SIGNAL_DETECT parameter is set to OK, and the inherent noise level of the PMD including the effects of crosstalk, power supply noise, etc.

Various implementations of the Signal Detect function are permitted by this standard, including implementations that generate the SIGNAL_DETECT parameter values in response to the amplitude of the modulation of the optical signal and implementations that respond to the average optical power of the modulated optical signal.

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122.5.5 PMD lane-by-lane signal detect function

Various implementations of the Signal Detect function are permitted by this standard. When the MDIO is implemented, each PMD_signal_detect_*i*, where *i* represents the lane number in the range 0:*n*–1, shall be continuously set in response to the magnitude of the optical signal on its associated lane, according to the requirements of Table 122–4.

122.5.6 PMD reset function

If the MDIO interface is implemented, and if PMD_reset is asserted, the PMD shall be reset as defined in 45.2.1.1.1.

122.5.7 PMD global transmit disable function (optional)

The PMD_global_transmit_disable function is optional and allows all of the optical transmitters to be disabled.

a) When the PMD_global_transmit_disable variable is set to one, this function shall turn off all of the optical transmitters so that each transmitter meets the requirements of the average launch power of the OFF transmitter in Table 122–9 for 200GBASE-FR4 and 200GBASE-LR4 and Table 122–10 for 400GBASE-FR8 and 400GBASE-LR8.

b) If a PMD_fault is detected, then the PMD may set the PMD_global_transmit_disable to one, turning off the optical transmitter in each lane.

122.5.8 PMD lane-by-lane transmit disable function

The PMD_transmit_disable_*i* (where *i* represents the lane number in the range 0:3 for 200GBASE-FR4 and 200GBASE-LR4 and 0:7 for 400GBASE-FR8 and 400GBASE-LR8) function allows the optical transmitters in each lane to be selectively disabled.

a) When a PMD_transmit_disable_*i* variable is set to one, this function shall turn off the optical transmitter associated with that variable so that the transmitter meets the requirements of the average launch power of the OFF transmitter in Table 122–9 for 200GBASE-FR4 and 200GBASE-LR4 and Table 122–10 for 400GBASE-FR8 and 400GBASE-LR8.

b) If a PMD_fault is detected, then the PMD may set each PMD_transmit_disable_*i* to one, turning off the optical transmitter in each lane.

If the PMD_transmit_disable_*i* function is not implemented in MDIO, an alternative method shall be provided to independently disable each transmit lane for testing purposes.

122.5.9 PMD fault function (optional)

If the PMD has detected a local fault on any of the transmit or receive paths, the PMD shall set PMD_fault to one.

If the MDIO interface is implemented, PMD_fault shall be mapped to the fault bit as specified in 45.2.1.2.3.

122.5.10 PMD transmit fault function (optional)

If the PMD has detected a local fault on any transmit lane, the PMD shall set the PMD_transmit_fault variable to one.

If the MDIO interface is implemented, PMD_transmit_fault shall be mapped to the transmit fault bit as specified in 45.2.1.7.4.

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122.5.11 PMD receive fault function (optional)

If the PMD has detected a local fault on any receive lane, the PMD shall set the PMD_receive_fault variable to one.

If the MDIO interface is implemented, PMD_receive_fault shall be mapped to the receive fault bit as specified in 45.2.1.7.5.

122.6 Wavelength-division-multiplexed lane assignments

The wavelength range for each lane of the 200GBASE-FR4 PMD is defined in Table 122–5. The 200GBASE-FR4 center wavelengths are members of the CWDM wavelength grid defined in ITU-T G.694.2 and are spaced at 20 nm.

The wavelength range for each lane of the 200GBASE-LR4 PMD is defined in Table 122–6. The wavelength range for each lane of the 400GBASE-FR8 and 400GBASE-LR8 PMDs is defined in Table 122–7. The 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 center frequencies are members of the frequency grid for 100 GHz spacing and above defined in ITU-T G.694.1 and are spaced at 800 GHz.

Table 122–5—200GBASE-FR4 wavelength-division-multiplexed lane assignments

Lane	Center wavelength	Wavelength range
L_0	1271 nm	1264.5 to 1277.5 nm
L_1	1291 nm	1284.5 to 1297.5 nm
L_2	1311 nm	1304.5 to 1317.5 nm
L_3	1331 nm	1324.5 to 1337.5 nm

Table 122–6—200GBASE-LR4 wavelength-division-multiplexed lane assignments

Lane	Center frequency	Center wavelength	Wavelength range
L_0	231.4 THz	1295.56 nm	1294.53 to 1296.59 nm
L_1	230.6 THz	1300.05 nm	1299.02 to 1301.09 nm
L_2	229.8 THz	1304.58 nm	1303.54 to 1305.63 nm
L_3	229 THz	1309.14 nm	1308.09 to 1310.19 nm

NOTE—There is no requirement to associate a particular electrical lane with a particular optical lane, as the PCS is capable of receiving lanes in any arrangement.

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Table 122–7—400GBASE-FR8 and 400GBASE-LR8 wavelength-division-multiplexed lane assignments

Lane	Center frequency	Center wavelength	Wavelength range
L_0	235.4 THz	1273.54 nm	1272.55 to 1274.54 nm
L_1	234.6 THz	1277.89 nm	1276.89 to 1278.89 nm
L_2	233.8 THz	1282.26 nm	1281.25 to 1283.27 nm
L_3	233 THz	1286.66 nm	1285.65 to 1287.68 nm
L_4	231.4 THz	1295.56 nm	1294.53 to 1296.59 nm
L_5	230.6 THz	1300.05 nm	1299.02 to 1301.09 nm
L_6	229.8 THz	1304.58 nm	1303.54 to 1305.63 nm
L_7	229 THz	1309.14 nm	1308.09 to 1310.19 nm

122.7 PMD to MDI optical specifications for 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8

The operating ranges for the 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 PMDs are defined in Table 122–8. A 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, or 400GBASE-LR8 compliant PMD operates on type B1.1, B1.3, or B6_a single-mode fibers according to the specifications defined in Table 122–18. A PMD that exceeds the operating range requirement while meeting all other optical specifications is considered compliant (e.g., a 400GBASE-FR8 PMD operating at 2.5 km meets the operating range requirement of 2 m to 2 km). The 400GBASE-LR8 PMD interoperates with the 400GBASE-FR8 PMD provided that the channel requirements for 400GBASE-FR8 are met.

Table 122–8—200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 operating ranges

PMD type	Required operating range
200GBASE-FR4 and 400GBASE-FR8	2 m to 2 km
200GBASE-LR4 and 400GBASE-LR8	2 m to 10 km

122.7.1 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 transmitter optical specifications

The 200GBASE-FR4 transmitter shall meet the specifications defined in Table 122–9 per the definitions in 122.8. The 200GBASE-LR4 transmitter shall meet the specifications defined in Table 122–9 per the definitions in 122.8. The 400GBASE-FR8 transmitter shall meet the specifications defined in Table 122–10 per the definitions in 122.8. The 400GBASE-LR8 transmitter shall meet the specifications defined in Table 122–10 per the definitions in 122.8.

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Table 122–9—200GBASE-FR4 and 200GBASE-LR4 transmit characteristics

Description	200GBASE-FR4	200GBASE-LR4	Unit
Signaling rate, each lane (range)	26.5625 ± 100 ppm		GBd
Modulation format	PAM4		—
Lane wavelengths (range)	1264.5 to 1277.5 1284.5 to 1297.5 1304.5 to 1317.5 1324.5 to 1337.5	1294.53 to 1296.59 1299.02 to 1301.09 1303.54 to 1305.63 1308.09 to 1310.19	nm
Side-mode suppression ratio (SMSR), (min)	30		dB
Total average launch power (max)	10.7	11.3	dBm
Average launch power, each lane (max)	4.7	5.3	dBm
Average launch power, each lane[a] (min)	–4.2	–3.4	dBm
Outer Optical Modulation Amplitude (OMA_{outer}), each lane (max)	4.5	5.1	dBm
Outer Optical Modulation Amplitude (OMA_{outer}), each lane (min)[b]	–1.2	–0.4	dBm
Difference in launch power between any two lanes (OMA_{outer}) (max)	4		dB
Launch power in OMA_{outer} minus TDECQ, each lane (min): for extinction ratio ≥ 4.5 dB for extinction ratio < 4.5 dB	 –2.6 –2.5	 –1.8 –1.7	 dBm dBm
Transmitter and dispersion eye closure for PAM4 (TDECQ), each lane (max)	3.3	3.4	dB
Average launch power of OFF transmitter, each lane (max)	–30		dBm
Extinction ratio, each lane (min)	3.5		dB
$RIN_{16.5}OMA$ (max)	–132	—	dB/Hz
$RIN_{15.1}OMA$ (max)	—	–132	dB/Hz
Optical return loss tolerance (max)	16.5	15.1	dB
Transmitter reflectance[c] (max)	–26		dB

[a] Average launch power, each lane (min) is informative and not the principal indicator of signal strength. A transmitter with launch power below this value cannot be compliant; however, a value above this does not ensure compliance.
[b] Even if the TDECQ < 1.4 dB for an extinction ratio of ≥ 4.5 dB or TDECQ < 1.3 dB for an extinction ratio of < 4.5 dB, the OMA_{outer} (min) must exceed this value.
[c] Transmitter reflectance is defined looking into the transmitter.

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Table 122–10—400GBASE-FR8 and 400GBASE-LR8 transmit characteristics

Description	400GBASE-FR8	400GBASE-LR8	Unit
Signaling rate, each lane (range)	26.5625 ± 100 ppm		GBd
Modulation format	PAM4		—
Lane wavelengths (range)	1272.55 to 1274.54 1276.89 to 1278.89 1281.25 to 1283.27 1285.65 to 1287.68 1294.53 to 1296.59 1299.02 to 1301.09 1303.54 to 1305.63 1308.09 to 1310.19		nm
Side-mode suppression ratio (SMSR), (min)	30		dB
Total average launch power (max)	13.2		dBm
Average launch power, each lane[a] (max)	5.3		dBm
Average launch power, each lane[b] (min)	–3.5	–2.8	dBm
Outer Optical Modulation Amplitude (OMA_{outer}), each lane (max)	5.5	5.7	dBm
Outer Optical Modulation Amplitude (OMA_{outer}), each lane (min)[c]	–0.5	0.2	dBm
Difference in launch power between any two lanes (OMA_{outer}) (max)	4		dB
Launch power in OMA_{outer} minus TDECQ, each lane (min): for extinction ratio ≥ 4.5 dB for extinction ratio < 4.5 dB	 –1.9 –1.8	 –1.2 –1.1	 dBm dBm
Transmitter and dispersion eye closure for PAM4 (TDECQ), each lane (max)	3.1	3.3	dB
Average launch power of OFF transmitter, each lane (max)	–30		dBm
Extinction ratio, each lane (min)	3.5		dB
$RIN_{16.5}OMA$ (max)	–132	—	dB/Hz
$RIN_{15.1}OMA$ (max)	—	–132	dB/Hz
Optical return loss tolerance (max)	16.5	15.1	dB
Transmitter reflectance[d] (max)	–26		dB

[a] As the total average launch power limit has to be met, not all of the lanes can operate at the maximum average launch power, each lane.
[b] Average launch power, each lane (min) is informative and not the principal indicator of signal strength. A transmitter with launch power below this value cannot be compliant; however, a value above this does not ensure compliance.
[c] Even if the TDECQ < 1.4 dB for an extinction ratio of ≥ 4.5 dB or TDECQ < 1.3 dB for an extinction ratio of < 4.5 dB, the OMA_{outer} (min) must exceed this value.
[d] Transmitter reflectance is defined looking into the transmitter.

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122.7.2 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 receive optical specifications

The 200GBASE-FR4 receiver shall meet the specifications defined in Table 122–11 per the definitions in 122.8. The 200GBASE-LR4 receiver shall meet the specifications defined in Table 122–11 per the definitions in 122.8. The 400GBASE-FR8 receiver shall meet the specifications defined in Table 122–12 per the definitions in 122.8. The 400GBASE-LR8 receiver shall meet the specifications defined in Table 122–12 per the definitions in 122.8.

Table 122–11—200GBASE-FR4 and 200GBASE-LR4 receive characteristics

Description	200GBASE-FR4	200GBASE-LR4	Unit
Signaling rate, each lane (range)	26.5625 ± 100 ppm		GBd
Modulation format	PAM4		—
Lane wavelengths (range)	1264.5 to 1277.5 1284.5 to 1297.5 1304.5 to 1317.5 1324.5 to 1337.5	1294.53 to 1296.59 1299.02 to 1301.09 1303.54 to 1305.63 1308.09 to 1310.19	nm
Damage threshold[a], each lane	5.7	6.3	dBm
Average receive power, each lane (max)	4.7	5.3	dBm
Average receive power, each lane[b] (min)	–8.2	–9.7	dBm
Receive power (OMA_{outer}), each lane (max)	4.5	5.1	dBm
Difference in receive power between any two lanes (OMA_{outer}) (max)	4.1	4.2	dB
Receiver reflectance (max)	–26		dB
Receiver sensitivity (OMA_{outer}), each lane[c] (max)	–6	–7.7	dBm
Stressed receiver sensitivity (OMA_{outer}), each lane[d] (max)	–3.6	–5.2	dBm
Conditions of stressed receiver sensitivity test:[e]			
Stressed eye closure for PAM4 (SECQ), lane under test	3.3	3.4	dB
OMA_{outer} of each aggressor lane	0.5	–1	dBm

[a] The receiver shall be able to tolerate, without damage, continuous exposure to an optical input signal having this average power level.
[b] Average receive power, each lane (min) is informative and not the principal indicator of signal strength. A received power below this value cannot be compliant; however, a value above this does not ensure compliance.
[c] Receiver sensitivity (OMA_{outer}), each lane (max) is informative and is defined for a transmitter with SECQ of 0.9 dB.
[d] Measured with conformance test signal at TP3 (see 122.8.9) for the BER specified in 122.1.1.
[e] These test conditions are for measuring stressed receiver sensitivity. They are not characteristics of the receiver.

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Table 122–12—400GBASE-FR8 and 400GBASE-LR8 receive characteristics

Description	400GBASE-FR8	400GBASE-LR8	Unit
Signaling rate, each lane (range)	26.5625 ± 100 ppm		GBd
Modulation format	PAM4		—
Lane wavelengths (range)	1272.55 to 1274.54 1276.89 to 1278.89 1281.25 to 1283.27 1285.65 to 1287.68 1294.53 to 1296.59 1299.02 to 1301.09 1303.54 to 1305.63 1308.09 to 1310.19		nm
Damage threshold[a], each lane	6.3		dBm
Average receive power, each lane (max)	5.3		dBm
Average receive power, each lane[b] (min)	–7.5	–9.1	dBm
Receive power (OMA_{outer}), each lane (max)	5.7		dBm
Difference in receive power between any two lanes (OMA_{outer}) (max)	4.1	4.5	dB
Receiver reflectance (max)	–26		dB
Receiver sensitivity (OMA_{outer}), each lane[c] (max)	–5.3	–7.1	dBm
Stressed receiver sensitivity (OMA_{outer}), each lane[d] (max)	–3.1	–4.7	dBm
Conditions of stressed receiver sensitivity test:[e]			
Stressed eye closure for PAM4 (SECQ), lane under test	3.1	3.3	dB
OMA_{outer} of each aggressor lane	1	–0.2	dBm

[a] The receiver shall be able to tolerate, without damage, continuous exposure to an optical input signal having this average power level.
[b] Average receive power, each lane (min) is informative and not the principal indicator of signal strength. A received power below this value cannot be compliant; however, a value above this does not ensure compliance.
[c] Receiver sensitivity (OMA_{outer}), each lane (max) is informative and is defined for a transmitter with SECQ of 0.9 dB.
[d] Measured with conformance test signal at TP3 (see 122.8.9) for the BER specified in 122.1.1.
[e] These test conditions are for measuring stressed receiver sensitivity. They are not characteristics of the receiver.

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122.7.3 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 illustrative link power budgets

Illustrative power budgets and penalties for 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 channels are shown in Table 122–13.

Table 122–13—200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 illustrative link power budgets

Parameter	200GBASE-FR4	400GBASE-FR8	200GBASE-LR4	400GBASE-LR8	Unit
Power budget (for maximum TDECQ): for extinction ratio ≥ 4.5 dB for extinction ratio < 4.5 dB	 7.6 7.7	 7.4 7.5	 10.2 10.3	 10.1 10.2	 dB dB
Operating distance	2		10		km
Channel insertion loss	4[a]		6.3		dB
Maximum discrete reflectance	See 122.11.2.2		See 122.11.2.2		dB
Allocation for penalties[b] (for maximum TDECQ): for extinction ratio ≥ 4.5 dB for extinction ratio < 4.5 dB	 3.6 3.7	 3.4 3.5	 3.9 4	 3.8 3.9	 dB dB
Additional insertion loss allowed	0		0		dB

[a] The channel insertion loss is calculated using the maximum distance specified in Table 122–8 for 200GBASE-FR4 and 400GBASE-FR8 and fiber attenuation of 0.5 dB/km plus an allocation for connection and splice loss given in 122.11.2.1.

[b] Link penalties are used for link budget calculations. They are not requirements and are not meant to be tested.

122.8 Definition of optical parameters and measurement methods

All transmitter optical measurements shall be made through a short patch cable, between 2 m and 5 m in length, unless otherwise specified.

122.8.1 Test patterns for optical parameters

While compliance is to be achieved in normal operation, specific test patterns are defined for measurement consistency and to enable measurement of some parameters. Table 122–15 gives the test patterns to be used in each measurement, unless otherwise specified, and also lists references to the subclauses in which each parameter is defined. Any of the test patterns given for a particular test in Table 122–15 may be used to perform that test. The test patterns used in this clause are shown in Table 122–14.

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Table 122–14—Test patterns

Pattern	Pattern description	Defined in
Square wave	Square wave (8 threes, 8 zeros)	120.5.11.2.4
3	PRBS31Q	120.5.11.2.2
4	PRBS13Q	120.5.11.2.1
5	Scrambled idle	119.2.4.9
6	SSPRQ	120.5.11.2.3

Table 122–15—Test-pattern definitions and related subclauses

Parameter	Pattern	Related subclause
Wavelength	Square wave, 3, 4, 5, 6 or valid 200GBASE-R or 400GBASE-R signal	122.8.2
Side mode suppression ratio	3, 5, 6 or valid 200GBASE-R or 400GBASE-R signal	—
Average optical power	3, 5, 6 or valid 200GBASE-R or 400GBASE-R signal	122.8.3
Outer Optical Modulation Amplitude (OMA_{outer})	4 or 6	122.8.4
Transmitter and dispersion eye closure for PAM4 (TDECQ)	6	122.8.5
Extinction ratio	4 or 6	122.8.6
$RIN_{15.1}OMA$ and $RIN_{16.5}OMA$	Square wave	122.8.7
Stressed receiver conformance test signal calibration	6	122.8.9.2
Stressed receiver sensitivity	3 or 5	122.8.9

122.8.2 Wavelength

The wavelength of each optical lane shall be within the ranges given in Table 122–5 for 200GBASE-FR4, in Table 122–6 for 200GBASE-LR4 and Table 122–7 for 400GBASE-FR8 and 400GBASE-LR8, if measured per TIA/EIA-455-127-A or IEC 61280-1-3. The lane under test is modulated using the test pattern defined in Table 122–15.

122.8.3 Average optical power

The average optical power of each lane shall be within the limits given in Table 122–9 for 200GBASE-FR4 and 200GBASE-LR4 and in Table 122–10 for 400GBASE-FR8 and 400GBASE-LR8 if measured using the methods given in IEC 61280-1-1, with the sum of the optical power from all of the lanes not under test below –30 dBm, per the test setup in Figure 53–6.

122.8.4 Outer Optical Modulation Amplitude (OMA_{outer})

The OMA_{outer} of each lane shall be within the limits given in Table 122–9 for 200GBASE-FR4 and 200GBASE-LR4 and in Table 122–10 for 400GBASE-FR8 and 400GBASE-LR8. The OMA_{outer} is measured using a test pattern specified for OMA_{outer} in Table 122–15. It is the difference between the

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average optical launch power level P_3, measured over the central 2 UI of a run of 7 threes, and the average optical launch power level P_0, measured over the central 2 UI of a run of 6 zeros, as shown in Figure 122–3. For this measurement the sum of the optical power from all of the lanes not under test is below –30 dBm, or if other lanes are operating, a suitable optical filter may be used to separate the lane under test.

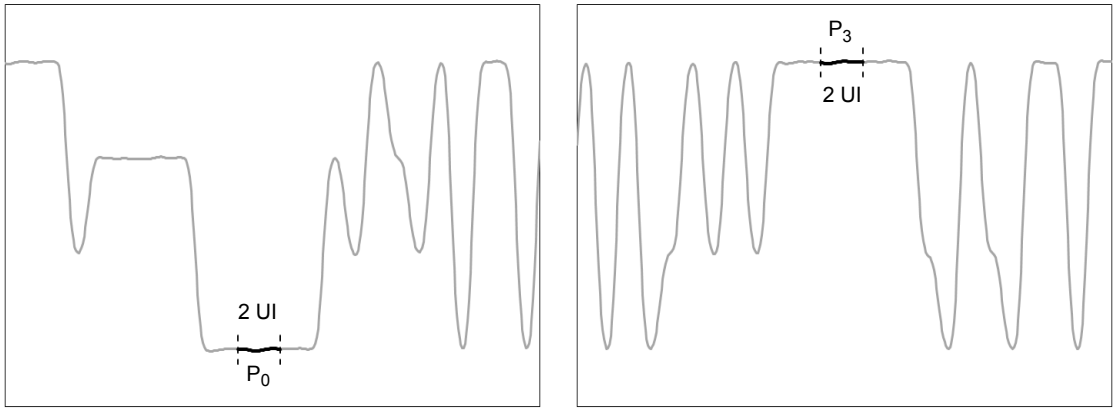


Figure 122–3—Power levels P_0 and P_3 from PRBS13Q test pattern

122.8.5 Transmitter and dispersion eye closure for PAM4 (TDECQ)

The TDECQ of each lane shall be within the limits given in Table 122–9 for 200GBASE-FR4 and 200GBASE-LR4 and in Table 122–10 for 400GBASE-FR8 and 400GBASE-LR8 if measured using the methods specified in 122.8.5.1, 122.8.5.2, and 122.8.5.3.

TDECQ is a measure of each optical transmitter's vertical eye closure when transmitted through a worst case optical channel (specified in 122.8.5.2), as measured through an optical to electrical converter (O/E) with a bandwidth equivalent to a reference receiver, and equalized with the reference equalizer (as described in 122.8.5.4). The reference receiver and equalizer may be implemented in software or may be part of the oscilloscope.

Table 122–15 specifies the test patterns to be used for measurement of TDECQ.

122.8.5.1 TDECQ conformance test setup

A block diagram for the TDECQ conformance test is shown in Figure 122–4. Other equivalent measurement implementations may be used with suitable calibration.

Each optical lane is tested individually with all other lanes in operation and all lanes using the same test pattern. There shall be at least 31 UI delay between the test pattern on one lane and the pattern on any other lane, so that the symbols on each lane are not correlated within the PMD. The optical splitter and variable reflector are adjusted so that each transmitter is tested with the optical return loss specified in Table 122–16. The state of polarization of the back reflection is adjusted to create the greatest RIN. The optical filter is used to separate the lane under test from the others. Each optical lane is tested with the optical channel described in 122.8.5.2. The combination of the O/E and the oscilloscope has a fourth-order Bessel-Thomson filter response with a bandwidth of approximately 13.28125 GHz. Compensation may be made for any deviation from an ideal fourth-order Bessel-Thomson response.

The optical filter passband ripple shall be limited to 0.5 dB peak-to-peak and the isolation is chosen such that the ratio of the power in the lane being measured to the sum of the powers of all of the other lanes is greater than 20 dB (see ITU-T G.959.1 Annex B).

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The test pattern (specified in Table 122–15) is transmitted repetitively by the optical lane under test and the oscilloscope is set up to capture the complete pattern for TDECQ analysis as described in 122.8.5.3. The clock recovery unit (CRU) has a corner frequency of 4 MHz and a slope of 20 dB/decade. The CRU can be implemented in hardware or software depending on oscilloscope technology.

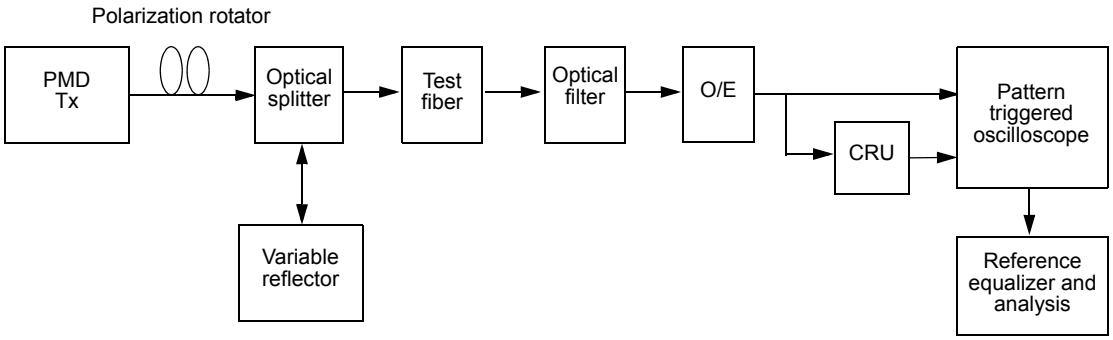


Figure 122–4—TDECQ conformance test block diagram

122.8.5.2 Channel requirements

The transmitter is tested using an optical channel that meets the requirements listed in Table 122–16.

Table 122–16—Transmitter compliance channel specifications

PMD type	Dispersion[a] (ps/nm)		Insertion loss[b]	Optical return loss[c]	Max mean DGD
	Minimum	Maximum			
200GBASE-FR4 or 400GBASE-FR8	$0.0465 \cdot \lambda \cdot [1 - (1324 / \lambda)^4]$	$0.0465 \cdot \lambda \cdot [1 - (1300 / \lambda)^4]$	Minimum	17.8 dB	0.8 ps
200GBASE-LR4 or 400GBASE-LR8	$0.2325 \cdot \lambda \cdot [1 - (1324 / \lambda)^4]$	$0.2325 \cdot \lambda \cdot [1 - (1300 / \lambda)^4]$	Minimum	15.7 dB	0.8 ps

[a] The dispersion is measured for the wavelength of the device under test (λ in nm). The coefficient assumes 2 km for 200GBASE-FR4 and 400GBASE-FR8 and 10 km for 200GBASE-LR4 and 400GBASE-LR8.
[b] There is no intent to stress the sensitivity of the O/E converter associated with the oscilloscope.
[c] The optical return loss is applied at TP2.

A 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8 or 400GBASE-LR8 transmitter is to be compliant with a total dispersion at least as negative as the "minimum dispersion" and at least as positive as the "maximum dispersion" columns specified in Table 122–16 for the wavelength of the device under test. This may be achieved with channels consisting of fibers with lengths chosen to meet the dispersion requirements.

To verify that the fiber has the correct amount of dispersion, the measurement method defined in IEC 60793-1-42 may be used. The measurement is made in the linear power regime of the fiber.

The channel provides an optical return loss specified in Table 122–16. The state of polarization of the back reflection is adjusted to create the greatest RIN.

The mean DGD of the channel is to be less than the value specified in Table 122–16.

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122.8.5.3 TDECQ measurement method

TDECQ for 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 is measured as described in 121.8.5.3 with the exception that the reference equalizer is as specified in 122.8.5.4.

122.8.5.4 TDECQ reference equalizer

The reference equalizer for 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 is a 5 tap, T spaced, feed-forward equalizer (FFE), where T is the symbol period. The sum of the equalizer tap coefficients is equal to 1.

NOTE—This reference equalizer is part of the TDECQ test and does not imply any particular receiver equalizer implementation.

122.8.6 Extinction ratio

The extinction ratio of each lane shall be within the limits given in Table 122–9 for 200GBASE-FR4 and 200GBASE-LR4 and in Table 122–10 for 400GBASE-FR8 and 400GBASE-LR8 if measured using a test pattern specified for extinction ratio in Table 122–15 with the sum of the optical power from all of the lanes not under test being below –30 dBm, or if other lanes are operating, a suitable optical filter may be used to separate the lane under test. The extinction ratio of a PAM4 optical signal is defined as the ratio of the average optical launch power level P_3, measured over the central 2 UI of a run of 7 threes, and the average optical launch power level P_0, measured over the central 2 UI of a run of 6 zeros, as shown in Figure 122–3.

122.8.7 Relative intensity noise ($RIN_{16.5}OMA$ and $RIN_{15.1}OMA$)

RIN shall be as defined by the measurement methodology of 52.9.6 with the following exceptions:

a) The optical return loss is 16.5 dB for 200GBASE-FR4 and 400GBASE-FR8 and 15.1 dB for 200GBASE-LR4 and 400GBASE-LR8.

b) Each lane may be tested individually with the sum of the optical power from all of the lanes not under test being below –30 dBm, or if other lanes are operating, a suitable optical filter may be used to separate the lane under test.

c) The upper –3 dB limit of the measurement apparatus is to be approximately equal to the signaling rate (i.e., 26.6 GHz).

d) The test pattern is according to Table 122–15.

122.8.8 Receiver sensitivity

Receiver sensitivity, which is defined for an input signal with SECQ of 0.9 dB (e.g., an ideal input signal without overshoot), is informative and compliance is not required. If measured, the test signal should have negligible impairments such as intersymbol interference (ISI), rise/fall times, jitter and RIN. Instead, the normative requirement for receivers is stressed receiver sensitivity.

122.8.9 Stressed receiver sensitivity

Stressed receiver sensitivity shall be within the limits given in Table 122–11 for 200GBASE-FR4 and 200GBASE-LR4 and in Table 122–12 for 400GBASE-FR8 and 400GBASE-LR8 if measured using the method defined in 122.8.9.1 and 122.8.9.3, with the conformance test signal at TP3 as described in 122.8.9.2, using the test pattern specified for SRS in Table 122–15. The BER is required to be met for the lane under test on its own.

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Stressed receiver sensitivity is defined with all transmit and receive lanes in operation. Any of the patterns specified for SRS in Table 122–15 is sent from the transmit section of the PMD under test. The signal being transmitted is asynchronous to the received signal.

122.8.9.1 Stressed receiver conformance test block diagram

A block diagram for the receiver conformance test is shown in Figure 122–5. The patterns used for the received conformance signal are specified in Table 122–15. The optical test signal is conditioned (stressed) using the stressed receiver methodology defined in 122.8.9.2 and has sinusoidal jitter applied as specified in 121.8.9.4. A suitable test set is needed to characterize the signal used to test the receiver. Stressed receiver conformance test signal verification is described in 122.8.9.3.

The low-pass filter is used to create ISI. The combination of the low-pass filter and the E/O converter should have a frequency response that results in at least half of the dB value of the stressed eye closure (SECQ) specified in Table 122–11 for 200GBASE-FR4 and 200GBASE-LR4 and in Table 122–12 for 400GBASE-FR8 and 400GBASE-LR8 before the sinusoidal and Gaussian noise terms are added, according to the methods specified in 122.8.9.2. The sinusoidal amplitude interferer causes additional eye closure, but in conjunction with the finite edge rates, also causes some jitter.

The sinusoidally jittered clock represents other forms of jitter and also verifies that the receiver under test can track low-frequency jitter. The sinusoidal amplitude interferer may be set at any frequency between 100 MHz and 2 GHz, although care should be taken to avoid harmonic relationships between the sinusoidal interferer, the sinusoidal jitter, the signaling rate, and the pattern repetition rate. The Gaussian noise generator, the amplitude of the sinusoidal interferer, and the low-pass filter are adjusted so that the SECQ specified in Table 122–11 or Table 122–12 is met, according to the methods specified in 122.8.9.2.

For improved visibility for calibration, all elements in the signal path (cables, DC blocks, E/O converter, etc.) should have wide and smooth frequency response, and linear phase response, throughout the spectrum of interest. Baseline wander and overshoot and undershoot should be negligible.

122.8.9.2 Stressed receiver conformance test signal characteristics and calibration

The stressed receiver conformance test signal characteristics and calibration methods are as described in 121.8.9.2 with the following exceptions:

- The SECQ of the stressed receiver conformance test signal is measured according to 122.8.5, except that the test fiber is not used.
- An example stressed receiver conformance test setup is shown in Figure 122–5; however, alternative test setups that generate equivalent stress conditions may be used.
- The signaling rate of the test pattern generator and the extinction ratio of the E/O converter are as given in Table 122–9 for 200GBASE-FR4 and 200GBASE-LR4 and in Table 122–10 for 400GBASE-FR8 and 400GBASE-LR8.
- The required values of the “Stressed receiver sensitivity (OMA_{outer}), each lane (max)”, “Stressed eye closure for PAM4 (SECQ), lane under test”, and “OMAouter of each aggressor lane” are as given in Table 122–11 for 200GBASE-FR4 and 200GBASE-LR4 and in Table 122–12 for 400GBASE-FR8 and 400GBASE-LR8.

122.8.9.3 Stressed receiver conformance test signal verification

The SECQ of the stressed receiver conformance test signal is measured according to 122.8.5, except that the test fiber is not used. The clock output from the clock source in Figure 122–5 is modulated with the sinusoidal jitter. To use an oscilloscope to calibrate the final stressed eye jitter that includes the sinusoidal jitter component, a separate clock source (clean clock of Figure 122–5) is required that is synchronized to the source clock, but not modulated with the jitter source.

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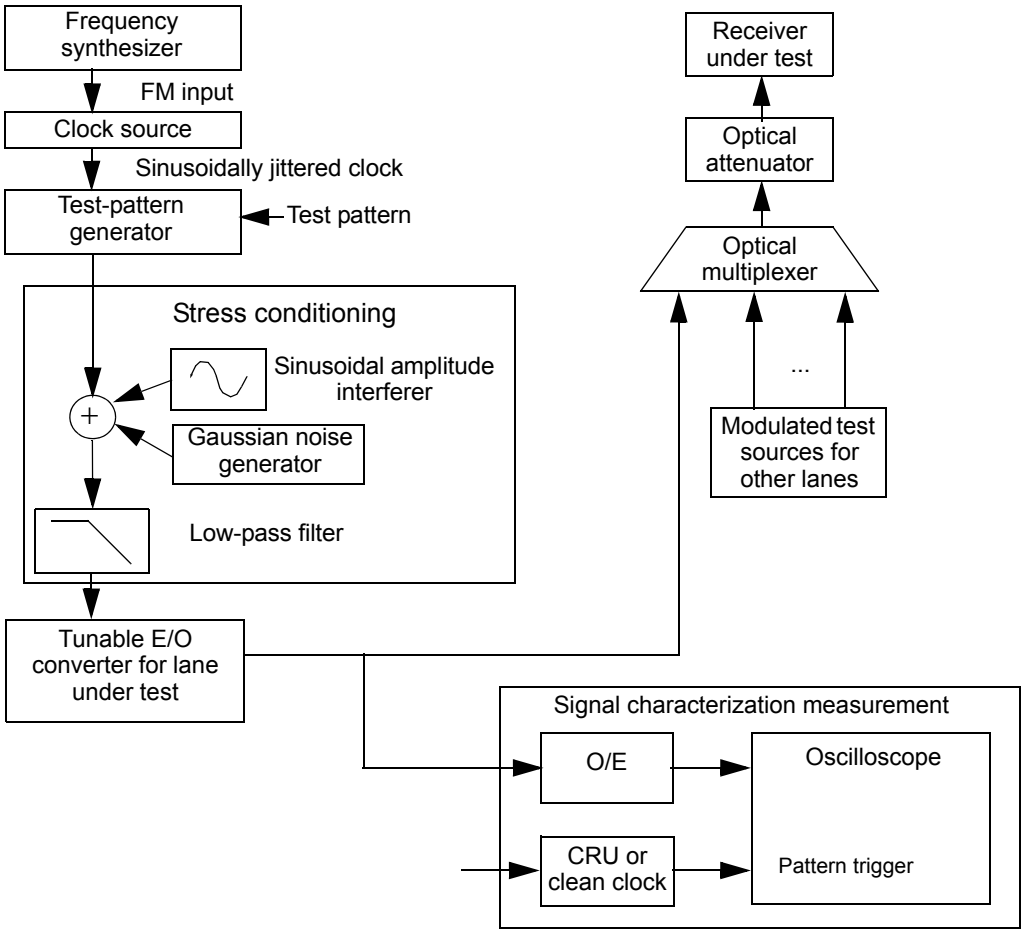


Figure 122–5—Stressed receiver conformance test block diagram

Care should be taken when characterizing the test signal because excessive noise/jitter in the measurement system would result in an input signal that does not fully stress the receiver under test. Running the receiver tolerance test with a signal that is under-stressed may result in the deployment of non-compliant receivers. The noise/jitter introduced by the O/E, filters, and oscilloscope should be negligible or the results should be corrected for its effects. While the details of test equipment are beyond the scope of this standard, it is recommended that the implementer fully characterize the test equipment and apply appropriate guard bands so that the stressed receiver conformance input signal meets the stress and sinusoidal jitter specified in 122.8.9.2 and 121.8.9.4.

122.9 Safety, installation, environment, and labeling

122.9.1 General safety

All equipment subject to this clause shall conform to IEC 60950-1.

122.9.2 Laser safety

200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 optical transceivers shall conform to Hazard Level 1 laser requirements as defined in IEC 60825-1 and IEC 60825-2, under any condition of operation. This includes single fault conditions whether coupled into a fiber or out of an open bore.

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Conformance to additional laser safety standards may be required for operation within specific geographic regions.

Laser safety standards and regulations require that the manufacturer of a laser product provide information about the product's laser, safety features, labeling, use, maintenance, and service. This documentation explicitly defines requirements and usage restrictions on the host system necessary to meet these safety certifications.[8]

122.9.3 Installation

It is recommended that proper installation practices, as defined by applicable local codes and regulation, be followed in every instance in which such practices are applicable.

122.9.4 Environment

Normative specifications in this clause shall be met by a system integrating a 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, or 400GBASE-LR8 PMD over the life of the product while the product operates within the manufacturer's range of environmental, power, and other specifications.

It is recommended that manufacturers indicate in the literature associated with the PHY the operating environmental conditions to facilitate selection, installation, and maintenance.

It is recommended that manufacturers indicate, in the literature associated with the components of the optical link, the distance and operating environmental conditions over which the specifications of this clause will be met.

122.9.5 Electromagnetic emission

A system integrating a 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, or 400GBASE-LR8 PMD shall comply with applicable local and national codes for the limitation of electromagnetic interference.

122.9.6 Temperature, humidity, and handling

The optical link is expected to operate over a reasonable range of environmental conditions related to temperature, humidity, and physical handling (such as shock and vibration). Specific requirements and values for these parameters are considered to be beyond the scope of this standard.

122.9.7 PMD labeling requirements

It is recommended that each PHY (and supporting documentation) be labeled in a manner visible to the user, with at least the applicable safety warnings and the applicable port type designation (e.g., 400GBASE-FR8).

Labeling requirements for Hazard Level 1 lasers are given in the laser safety standards referenced in 122.9.2.

[8]A host system that fails to meet the manufacturer's requirements and/or usage restrictions may emit laser radiation in excess of the safety limits of one or more safety standards. In such a case, the host manufacturer is required to obtain its own laser safety certification.

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122.10 Fiber optic cabling model

The fiber optic cabling model is shown in Figure 122–6.

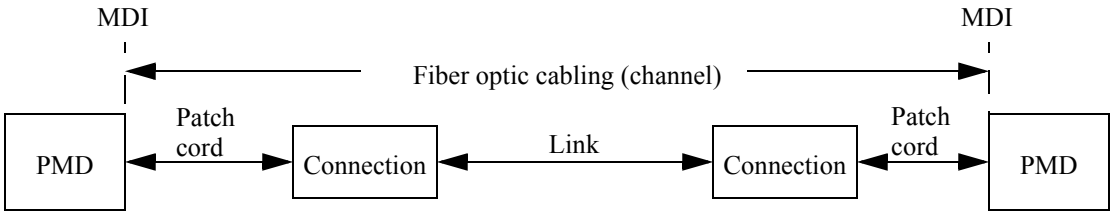


Figure 122–6—Fiber optic cabling model

The channel insertion loss is given in Table 122–17. A channel may contain additional connectors as long as the optical characteristics of the channel, such as attenuation, dispersion, reflections, and polarization mode dispersion meet the specifications. Insertion loss measurements of installed fiber cables are made in accordance with ANSI/TIA/EIA-526-7/method A-1. The fiber optic cabling model (channel) defined here is the same as a simplex fiber optic link segment. The term channel is used here for consistency with generic cabling standards.

Table 122–17—Fiber optic cabling (channel) characteristics

Description	200GBASE-FR4	400GBASE-FR8	200GBASE-LR4	400GBASE-LR8	Unit
Operating distance (max)	2		10		km
Channel insertion loss[a,b] (max)	4		6.3[c]		dB
Channel insertion loss (min)	0		0		dB
Positive dispersion[b] (max)	6.7	1.9	9.5		ps/nm
Negative dispersion[b] (min)	–11.9	–10.2	–28.4	–50.8	ps/nm
DGD_max[d]	3		8		ps
Optical return loss (min)	25		22		dB

[a] These channel insertion loss values include cable, connectors, and splices.
[b] Over the wavelength range 1264.5 nm to 1337.5 nm for 200GBASE-FR4, 1294.53 nm to 1310.19 nm for 200GBASE-LR4, and 1272.55 nm to 1310.19 nm for 400GBASE-FR8 and 400GBASE-LR8.
[c] Using 0.46 dB/km at 1272.55 nm attenuation for optical fiber cables derived from Appendix I of ITU-T G.695 may not support operation at 10 km for 400GBASE-LR8 under worst case conditions.
[d] Differential Group Delay (DGD) is the time difference at reception between the fractions of a pulse that were transmitted in the two principal states of polarization of an optical signal. DGD_max is the maximum differential group delay that the system must tolerate.

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122.11 Characteristics of the fiber optic cabling (channel)

The 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8 fiber optic cabling shall meet the specifications defined in Table 122–17. The fiber optic cabling consists of one or more sections of fiber optic cable and any intermediate connections required to connect sections together.

122.11.1 Optical fiber cable

The fiber optic cable requirements are satisfied by cables containing IEC 60793-2-50 type B1.1 (dispersion un-shifted single-mode), type B1.3 (low water peak single-mode), or type B6_a (bend insensitive) fibers or the requirements in Table 122–18 where they differ.

Table 122–18—Optical fiber and cable characteristics

Description	Value	Unit
Nominal fiber specification wavelength	1310	nm
Cabled optical fiber attenuation (max)	0.47[a] or 0.5[b]	dB/km
Zero dispersion wavelength (λ_0)	$1300 \leq \lambda_0 \leq 1324$	nm
Dispersion slope (max) (S_0)	0.093	ps/nm^2 km

[a] The 0.47 dB/km at 1264.5 nm attenuation for optical fiber cables is derived from Appendix I of ITU-T G.695.
[b] The 0.5 dB/km attenuation is provided for Outside Plant cable as defined in ANSI/TIA 568-C.3. Using 0.5 dB/km may not support operation 10 km for 200GBASE-LR4 or 400GBASE-LR8.

122.11.2 Optical fiber connection

An optical fiber connection, as shown in Figure 122–6, consists of a mated pair of optical connectors.

122.11.2.1 Connection insertion loss

The maximum link distance for 200GBASE-LR4 and 400GBASE-LR8 is based on an allocation of 2 dB total connection and splice loss. For example, this allocation supports four connections with an average insertion loss per connection of 0.5 dB. The maximum link distance for 200GBASE-FR4 and 400GBASE-FR8 is based on an allocation of 3 dB total connection and splice loss. Connections with different loss characteristics may be used provided the requirements of Table 122–17 are met.

122.11.2.2 Maximum discrete reflectance

The maximum value for each discrete reflectance shall be less than or equal to the value shown in Table 122–19 corresponding to the number of discrete reflectances above –55 dB within the channel. For numbers of discrete reflectances in between two numbers shown in the table, the lower of the two corresponding maximum discrete reflectance values applies.

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Table 122–19—Maximum value of each discrete reflectance

Number of discrete reflectances above –55 dB	Maximum value for each discrete reflectance	
	200GBASE-FR4 or 400GBASE-FR8	200GBASE-LR4 or 400GBASE-LR8
1	–25 dB	–22 dB
2	–31 dB	–29 dB
4	–35 dB	–33 dB
6	–38 dB	–35 dB
8	–40 dB	–37 dB
10	–41 dB	–39 dB

122.11.3 Medium Dependent Interface (MDI) requirements

The 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, or 400GBASE-LR8 PMD is coupled to the fiber optic cabling at the MDI. The MDI is the interface between the PMD and the "fiber optic cabling" (as shown in Figure 122–6). Examples of an MDI include the following:

a) Connectorized fiber pigtail
b) PMD receptacle

When the MDI is a connector plug and receptacle connection, it shall meet the interface performance specifications of IEC 61753-1-1 and IEC 61753-021-2.

NOTE—Transmitter compliance testing is performed at TP2 as defined in 122.5.1, not at the MDI.

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122.12 Protocol implementation conformance statement (PICS) proforma for Clause 122, Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8[9]

122.12.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Clause 122, Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8, shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

122.12.2 Identification

122.12.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

122.12.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Clause 122, Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[9] *Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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122.12.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
*FR4	200GBASE-FR4 PMD	122.7	Device supports requirements for 200GBASE-FR4 PHY	O.1	Yes [ ] No [ ]
*LR4	200GBASE-LR4 PMD	122.7	Device supports requirements for 200GBASE-LR4 PHY	O.1	Yes [ ] No [ ]
*FR8	400GBASE-FR8 PMD	122.7	Device supports requirements for 400GBASE-FR8 PHY	O.1	Yes [ ] No [ ]
*LR8	400GBASE-LR8 PMD	122.7	Device supports requirements for 400GBASE-LR8 PHY	O.1	Yes [ ] No [ ]
*INS	Installation / cable	122.10	Items marked with INS include installation practices and cable specifications not applicable to a PHY manufacturer	O	Yes [ ] No [ ]
TP1	Reference point TP1 exposed and available for testing	122.5.1	This point may be made available for use by implementers to certify component conformance	O	Yes [ ] No [ ]
TP4	Reference point TP4 exposed and available for testing	122.5.1	This point may be made available for use by implementers to certify component conformance	O	Yes [ ] No [ ]
DC	Delay constraints	122.5.1	Device conforms to delay constraints	M	Yes [ ]
SC	Skew constraints	122.3.2	Device conforms to Skew and Skew Variation constraints	M	Yes [ ]
*MD	MDIO capability	122.4	Registers and interface supported	O	Yes [ ] No [ ]

122.12.4 PICS proforma tables for Physical Medium Dependent (PMD) sublayer and medium, type 200GBASE-FR4, 200GBASE-LR4, 400GBASE-FR8, and 400GBASE-LR8

122.12.4.1 PMD functional specifications

Item	Feature	Subclause	Value/Comment	Status	Support
F1	Compatible with 200GBASE-R or 400GBASE-R PCS and PMA	122.1		M	Yes [ ]
F2	Integration with management functions	122.1		O	Yes [ ] No [ ]
F3	Bit error ratio	122.1.1	Meets the BER specified in 122.1.1	M	Yes [ ]
F4	Transmit function	122.5.2	Conveys symbols from PMD service interface to MDI	M	Yes [ ]
F5	Optical multiplexing and delivery to the MDI	122.5.2	Optically multiplexes the four or eight optical signals for delivery to the MDI	M	Yes [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
F6	Mapping between optical signal and logical signal for transmitter	122.5.2	Highest optical power corresponds to tx_symbol = three	M	Yes [ ]
F7	Receive function	122.5.3	Conveys symbols from MDI to PMD service interface	M	Yes [ ]
F8	Conversion of four or eight optical signals to eight electrical signals	122.5.3	For delivery to the PMD service interface	M	Yes [ ]
F9	Mapping between optical signal and logical signal for receiver	122.5.3	Highest optical power corresponds to rx_symbol = three	M	Yes [ ]
F10	Global Signal Detect function	122.5.4	Report to the PMD service interface the message PMD:IS_SIGNAL.indication (SIGNAL_DETECT)	M	Yes [ ]
F11	Global Signal Detect behavior	122.5.4	SIGNAL_DETECT is a global indicator of the presence of optical signals on all four or eight lanes	M	Yes [ ]
F12	Lane-by-lane Signal Detect function	122.5.5	Sets PMD_signal_detect_*i* values on a lane-by-lane basis per requirements of Table 122–4	MD:O	Yes [ ] No [ ] N/A [ ]
F13	PMD reset function	122.5.6	Resets the PMD sublayer	MD:O	Yes [ ] No [ ] N/A [ ]

122.12.4.2 Management functions

Item	Feature	Subclause	Value/Comment	Status	Support
M1	Management register set	122.4		MD:M	Yes [ ] N/A [ ]
M2	Global transmit disable function	122.5.7	Disables all of the optical transmitters with the PMD_global_transmit_disable variable	MD:O	Yes [ ] No [ ] N/A [ ]
M3	PMD_lane_by_lane_transmit_ disable function	122.5.8	Disables the optical transmitter via MDIO on the lane associated with the PMD_transmit_disable_*i* variable	MD:O.2	Yes [ ] No [ ] N/A [ ]
M4	PMD lane-by-lane transmit disable	122.5.8	Disables each optical transmitter independently if M3 = No	O.2	Yes [ ] N/A [ ]
M5	PMD_fault function	122.5.9	Sets PMD_fault to one if any local fault is detected	MD:O	Yes [ ] No [ ] N/A [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
M6	PMD_transmit_fault function	122.5.10	Sets PMD_transmit_fault to one if a local fault is detected on any transmit lane	MD:O	Yes [ ] No [ ] N/A [ ]
M7	PMD_receive_fault function	122.5.11	Sets PMD_receive_fault to one if a local fault is detected on any receive lane	MD:O	Yes [ ] No [ ] N/A [ ]

122.12.4.3 PMD to MDI optical specifications for 200GBASE-FR4

Item	Feature	Subclause	Value/Comment	Status	Support
FRF1	Transmitter meets specifications in Table 122–9	122.7.1	Per definitions in 122.8	FR4:M	Yes [ ] N/A [ ]
FRF2	Receiver meets specifications in Table 122–11	122.7.2	Per definitions in 122.8	FR4:M	Yes [ ] N/A [ ]

122.12.4.4 PMD to MDI optical specifications for 200GBASE-LR4

Item	Feature	Subclause	Value/Comment	Status	Support
LRF1	Transmitter meets specifications in Table 122–9	122.7.1	Per definitions in 122.8	LR4:M	Yes [ ] N/A [ ]
LRF2	Receiver meets specifications in Table 122–11	122.7.2	Per definitions in 122.8	LR4:M	Yes [ ] N/A [ ]

122.12.4.5 PMD to MDI optical specifications for 400GBASE-FR8

Item	Feature	Subclause	Value/Comment	Status	Support
FRE1	Transmitter meets specifications in Table 122–10	122.7.1	Per definitions in 122.8	FR8:M	Yes [ ] N/A [ ]
FRE2	Receiver meets specifications in Table 122–12	122.7.2	Per definitions in 122.8	FR8:M	Yes [ ] N/A [ ]

122.12.4.6 PMD to MDI optical specifications for 400GBASE-LR8

Item	Feature	Subclause	Value/Comment	Status	Support
LRE1	Transmitter meets specifications in Table 122–10	122.7.1	Per definitions in 122.8	LR8:M	Yes [ ] N/A [ ]
LRE2	Receiver meets specifications in Table 122–12	122.7.2	Per definitions in 122.8	LR8:M	Yes [ ] N/A [ ]

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122.12.4.7 Optical measurement methods

Item	Feature	Subclause	Value/Comment	Status	Support
OM1	Measurement cable	122.8	2 m to 5 m in length	M	Yes [ ]
OM2	Center wavelength	122.8.2	Per TIA/EIA-455-127-A or IEC 61280-1-3 under modulated conditions	M	Yes [ ]
OM3	Average optical power	122.8.3	Per IEC 61280-1-1	M	Yes [ ]
OM4	OMA measurements	122.8.4	Each lane	M	Yes [ ]
OM5	Transmitter and dispersion eye closure for PAM4 (TDECQ)	122.8.5	Each lane	M	Yes [ ]
OM6	Extinction ratio	122.8.6	Each lane	M	Yes [ ]
OM7	RIN_{xx}OMA measurement procedure	122.8.7	Each lane	M	Yes [ ]
OM8	Stressed receiver sensitivity	122.8.9	Each lane	M	Yes [ ]

122.12.4.8 Environmental specifications

Item	Feature	Subclause	Value/Comment	Status	Support
ES1	General safety	122.9.1	Conforms to IEC 60950-1	M	Yes [ ]
ES2	Laser safety—IEC Hazard Level 1	122.9.2	Conforms to Hazard Level 1 laser requirements defined in IEC 60825-1 and IEC 60825-2	M	Yes [ ]
ES3	Electromagnetic interference	122.9.5	Complies with applicable local and national codes for the limitation of electromagnetic interference	M	Yes [ ]

122.12.4.9 Characteristics of the fiber optic cabling and MDI

Item	Feature	Subclause	Value/Comment	Status	Support
OC1	Fiber optic cabling	122.11	Meets requirements specified in Table 122–17	INS:M	Yes [ ] N/A [ ]
OC2	Maximum discrete reflectance	122.11.2.2	Meets requirements specified in Table 122–19	INS:M	Yes [ ] N/A [ ]
OC3	MDI requirements	122.11.3	Meets IEC 61753-1-1 and IEC 61753-021-2	INS:M	Yes [ ] N/A [ ]

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123. Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-SR16

123.1 Overview

This clause specifies the 400GBASE-SR16 PMD together with the multimode fiber medium. The PMD sublayer provides a point-to-point 400 Gb/s Ethernet link over 16 pairs of multimode fiber, up to at least 100 m. When forming a complete Physical Layer, a PMD shall be connected to the appropriate PMA as shown in Table 123–1, to the medium through the MDI and optionally with the management functions that may be accessible through the management interface defined in Clause 45, or equivalent.

Table 123–1—Physical Layer clauses associated with the 400GBASE-SR16 PMD

Associated clause	400GBASE-SR16
117—RS	Required
117—400GMII[a]	Optional
118—400GMII Extender	Optional
119—PCS	Required
120—PMA	Required
120B—Chip-to-chip 400GAUI-16	Optional
120C—Chip-to-module 400GAUI-16	Optional
120D—Chip-to-chip 400GAUI-8	Optional
120E—Chip-to-module 400GAUI-8	Optional
78—Energy Efficient Ethernet	Optional

[a] The 400GMII is an optional interface. However, if the 400GMII is not implemented, a conforming implementation must behave functionally as though the RS and 400GMII were present.

Figure 123–1 shows the relationship of the PMD and MDI (shown shaded) with other sublayers to the ISO/IEC Open System Interconnection (OSI) reference model. 400 Gb/s Ethernet is introduced in Clause 116 and the purpose of each PHY sublayer is summarized in 116.2.

400GBASE-SR16 PHYs with the optional Energy Efficient Ethernet (EEE) fast wake capability may enter the Low Power Idle (LPI) mode to conserve energy during periods of low link utilization (see Clause 78). The deep sleep mode of EEE is not supported.

Further relevant information may be found in Clause 1 (terminology and conventions, references, definitions and abbreviations) and Annex A (bibliography, referenced as [B1], [B2], etc.).

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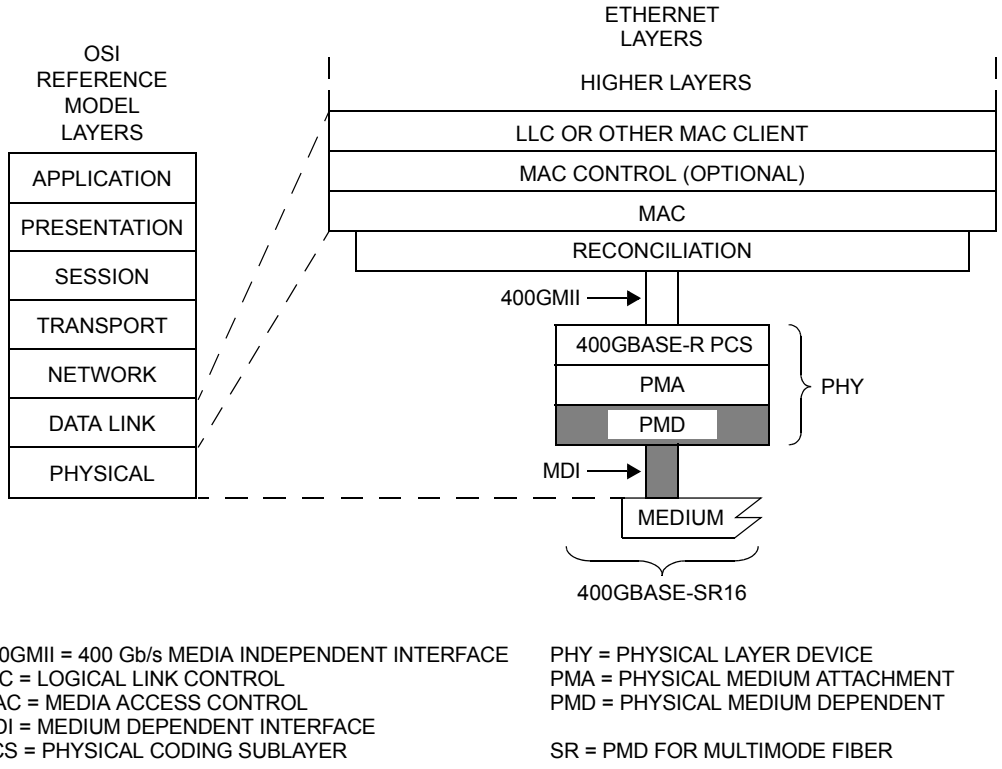


Figure 123–1—400GBASE-SR16 PMD relationship to the ISO/IEC Open Systems Interconnection (OSI) reference model and IEEE 802.3 Ethernet model

123.1.1 Bit error ratio

The bit error ratio (BER) shall be less than 2.4×10^{-4} provided that the error statistics are sufficiently random that this results in a frame loss ratio (see 1.4.223) of less than 1.7×10^{-12} for 64-octet frames with minimum interpacket gap when processed according to Clause 119. For a complete Physical Layer, the frame loss ratio may be degraded to 6.2×10^{-11} for 64-octet frames with minimum interpacket gap due to additional errors from the electrical interfaces.

If the error statistics are not sufficiently random to meet this requirement, then the BER shall be less than that required to give a frame loss ratio of less than 1.7×10^{-12} for 64-octet frames with minimum interpacket gap.

123.2 Physical Medium Dependent (PMD) service interface

This subclause specifies the services provided by the 400GBASE-SR16 PMD. The service interface for this PMD is described in an abstract manner and does not imply any particular implementation. The PMD service interface supports the exchange of encoded data between the PMA entity that resides just above the PMD, and the PMD entity. The PMD translates the encoded data to and from signals suitable for the specified medium.

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The PMD service interface is an instance of the inter-sublayer service interface defined in 116.3. The PMD service interface primitives are summarized as follows:

PMD:IS_UNITDATA_*i*.request
PMD:IS_UNITDATA_*i*.indication
PMD:IS_SIGNAL.indication

The 400GBASE-SR16 PMD has 16 parallel bit streams, hence $i = 0$ to 15.

In the transmit direction, the PMA continuously sends 16 parallel bit streams to the PMD, one per lane, each at a nominal signaling rate of 26.5625 GBd. The PMD converts these streams of bits into appropriate signals on the MDI.

In the receive direction, the PMD continuously sends 16 parallel bit streams to the PMA corresponding to the signals received from the MDI, one per lane, each at a nominal signaling rate of 26.5625 GBd.

The SIGNAL_DETECT parameter defined in this clause maps to the SIGNAL_OK parameter in the PMD:IS_SIGNAL.indication(SIGNAL_OK) inter-sublayer service primitive defined in 116.3.

The SIGNAL_DETECT parameter can take on one of two values: OK or FAIL. When SIGNAL_DETECT = FAIL, the rx_symbol parameters are undefined.

NOTE—SIGNAL_DETECT = OK does not guarantee that the rx_symbol parameters are known to be good. It is possible for a poor quality link to provide sufficient light for a SIGNAL_DETECT = OK indication and still not meet the BER defined in 123.1.1.

123.3 Delay and Skew

123.3.1 Delay constraints

An upper bound to the delay through the PMA and PMD is required for predictable operation of the MAC Control PAUSE operation. The sum of the transmit and receive delays at one end of the link contributed by the 400GBASE-SR16 PMD including 2 m of fiber in one direction shall be no more than 8192 bit times (16 pause_quanta or 20.48 ns). A description of overall system delay constraints and the definitions for bit times and pause_quanta can be found in 116.4 and its references.

123.3.2 Skew constraints

The Skew (relative delay) between the lanes must be kept within limits so that the information on the lanes can be reassembled by the PCS sublayer. Skew and Skew Variation are defined in 116.5 and specified at the points SP1 to SP6 shown in Figure 116–4 and Figure 116–5.

If the PMD service interface is physically instantiated so that the Skew at SP2 can be measured, then the Skew at SP2 is limited to 43 ns and the Skew Variation at SP2 is limited to 400 ps.

The Skew at SP3 (the transmitter MDI) shall be less than 54 ns and the Skew Variation at SP3 shall be less than 600 ps.

The Skew at SP4 (the receiver MDI) shall be less than 134 ns and the Skew Variation at SP4 shall be less than 3.4 ns.

If the PMD service interface is physically instantiated so that the Skew at SP5 can be measured, then the Skew at SP5 shall be less than 145 ns and the Skew Variation at SP5 shall be less than 3.6 ns.

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For more information on Skew and Skew Variation see 116.5. The measurements of Skew and Skew Variation are defined in 86.8.3.1.

123.4 PMD MDIO function mapping

The optional MDIO capability described in Clause 45 defines several variables that may provide control and status information for and about the PMD. If the MDIO interface is implemented, the mapping of MDIO control variables to PMD control variables shall be as shown in Table 123–2 and the mapping of MDIO status variables to PMD status variables shall be as shown in Table 123–3.

Table 123–2—MDIO/PMD control variable mapping

MDIO control variable	PMA/PMD register name	Register/bit number	PMD control variable
Reset	PMA/PMD control 1 register	1.0.15	PMD_reset
Global PMD transmit disable	PMD transmit disable register	1.9.0	PMD_global_transmit_disable
PMD transmit disable 15	PMD transmit disable extension register	1.27.0	PMD_transmit_disable_15
PMD transmit disable 14 to PMD transmit disable 0	PMD transmit disable register	1.9.15 to 1.9.1	PMD_transmit_disable_14 to PMD_transmit_disable_0

Table 123–3—MDIO/PMD status variable mapping

MDIO status variable	PMA/PMD register name	Register/ bit number	PMD status variable
Fault	PMA/PMD status 1 register	1.1.7	PMD_fault
Transmit fault	PMA/PMD status 2 register	1.8.11	PMD_transmit_fault
Receive fault	PMA/PMD status 2 register	1.8.10	PMD_receive_fault
Global PMD receive signal detect	PMD receive signal detect register	1.10.0	PMD_global_signal_detect
PMD receive signal detect 15	PMD receive signal detect extension register	1.28.0	PMD_signal_detect_15
PMD receive signal detect 14 to PMD receive signal detect 0	PMD receive signal detect register	1.10.15 to 1.10.1	PMD_signal_detect_14 to PMD_signal_detect_0

123.5 PMD functional specifications

The 400GBASE-SR16 PMD performs the Transmit and Receive functions, which convey data between the PMD service interface and the MDI.

123.5.1 PMD block diagram

The PMD block diagram is shown in Figure 123–2. For purposes of system conformance, the PMD sublayer is standardized at the points described in this subclause. The optical transmit signal is defined at the output end of a multimode fiber patch cord (TP2), between 2 m and 5 m in length. Unless specified otherwise, all transmitter measurements and tests defined in 123.8 are made at TP2. The optical receive signal is defined at

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the output of the fiber optic cabling (TP3) at the MDI (see 123.11.3). Unless specified otherwise, all receiver measurements and tests defined in 123.8 are made at TP3.

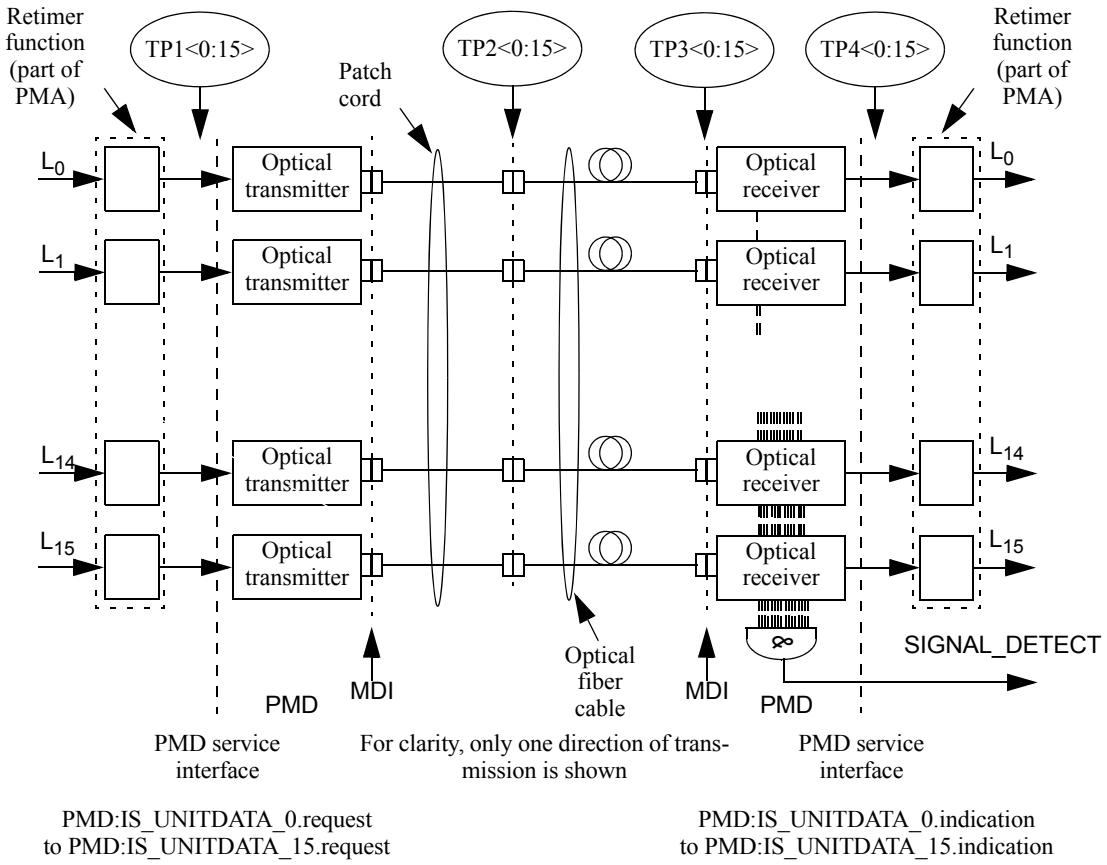


Figure 123–2—Block diagram for 400GBASE-SR16 transmit/receive paths

TP1 and TP4 are informative reference points that may be useful to implementers for testing components (these test points will not typically be accessible in an implemented system).

123.5.2 PMD transmit function

The PMD Transmit function shall convert the 16 bit streams requested by the PMD service interface messages PMD:IS_UNITDATA_0.request to PMD:IS_UNITDATA_15.request into 16 optical signals. The optical signals shall then be delivered to the MDI which contains 16 parallel light paths for transmit, according to the transmit optical specifications in this clause. The higher optical power level in each signal stream shall correspond to tx_symbol = one.

123.5.3 PMD receive function

The PMD Receive function shall convert the 16 optical signals received from the MDI into separate bit streams for delivery to the PMD service interface using the messages PMD:IS_UNITDATA_0.indication to PMD:IS_UNITDATA_15.indication, all according to the receive optical specifications in this clause. The higher optical power level in the signal stream shall correspond to rx_symbol = one.

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123.5.4 PMD global signal detect function

The PMD global signal detect function shall report the state of SIGNAL_DETECT via the PMD service interface. The SIGNAL_DETECT parameter is signaled continuously, while the PMD:IS_SIGNAL.indication message is generated when a change in the value of SIGNAL_DETECT occurs. The SIGNAL_DETECT parameter defined in this clause maps to the SIGNAL_OK parameter in the inter-sublayer service interface primitives defined in 116.3.

SIGNAL_DETECT shall be an indicator of the presence of the optical signals on all 16 lanes. The value of the SIGNAL_DETECT parameter shall be generated according to the conditions defined in Table 123–4. The PMD receiver is not required to verify whether a compliant 400GBASE-SR16 signal is being received. This standard imposes no response time requirements on the generation of the SIGNAL_DETECT parameter.

Table 123–4—SIGNAL_DETECT value definition

Receive conditions	SIGNAL_DETECT value
For any lane; Average optical power at TP3 ≤ –30 dBm	FAIL
For all lanes; [(Optical power at TP3 ≥ average receive power, each lane (min) in Table 95–7) AND (compliant 400GBASE-SR16 signal input)]	OK
All other conditions	Unspecified

As an unavoidable consequence of the requirements for the setting of the SIGNAL_DETECT parameter, implementations must provide adequate margin between the input optical power level at which the SIGNAL_DETECT parameter is set to OK, and the inherent noise level of the PMD including the effects of crosstalk, power supply noise, etc.

Various implementations of the PMD global signal detect function are permitted by this standard, including implementations that generate the SIGNAL_DETECT parameter values in response to the amplitude of the modulation of the optical signals and implementations that respond to the average optical power of the modulated optical signals. When the MDIO is implemented, the SIGNAL_DETECT value shall be continuously set in response to the magnitude of the optical signals, according to the requirements of Table 123–4.

123.5.5 PMD lane-by-lane signal detect function

Various implementations of the Signal Detect function are permitted by this standard. When the MDIO is implemented, each PMD_signal_detect_*i*, where *i* represents the lane number in the range 0:15, shall be continuously set in response to the magnitude of the optical signal on its associated lane, according to the requirements of Table 123–4.

123.5.6 PMD reset function

If the MDIO interface is implemented, and if PMD_reset is asserted, the PMD shall be reset as defined in 45.2.1.1.1.

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123.5.7 PMD global transmit disable function (optional)

The PMD_global_transmit_disable function is optional and allows all of the optical transmitters to be disabled.

a) When the PMD_global_transmit_disable variable is set to one, this function shall turn off all of the optical transmitters so that each transmitter meets the requirements of the average launch power of the OFF transmitter in Table 95–6.

b) If a PMD_fault is detected, then the PMD may set the PMD_global_transmit_disable to one, turning off the optical transmitter in each lane.

123.5.8 PMD lane-by-lane transmit disable function (optional)

The PMD_transmit_disable_*i* (where *i* represents the lane number in the range 0:15) function is optional and allows the optical transmitter in each lane to be selectively disabled.

a) When a PMD_transmit_disable_*i* variable is set to one, this function shall turn off the optical transmitter associated with that variable so that the transmitter meets the requirements of the average launch power of the OFF transmitter in Table 95–6.

b) If a PMD_fault is detected, then the PMD may set each PMD_transmit_disable_*i* to one, turning off the optical transmitter in each lane.

If the optional PMD_transmit_disable_*i* function is not implemented in MDIO, an alternative method may be provided to independently disable each transmit lane.

123.5.9 PMD fault function (optional)

If the PMD has detected a local fault on any of the transmit or receive paths, the PMD shall set PMD_fault to one.

If the MDIO interface is implemented, PMD_fault shall be mapped to the fault bit as specified in 45.2.1.2.3.

123.5.10 PMD transmit fault function (optional)

If the PMD has detected a local fault on the transmitter, the PMD shall set PMD_transmit_fault to one.

If the MDIO interface is implemented, PMD_transmit_fault shall be mapped to the transmit fault bit as specified in 45.2.1.7.4.

123.5.11 PMD receive fault function (optional)

If the PMD has detected a local fault on the receiver, the PMD shall set the PMD_receive_fault variable to one.

If the MDIO interface is implemented, PMD_receive_fault shall be mapped to the receive fault bit as specified in 45.2.1.7.5.

123.6 Lane assignments

There are no lane assignments (within a group of transmit or receive lanes) for 400GBASE-SR16. While it is expected that a PMD will map electrical lane *i* to optical lane *i* and vice versa, there is no need to define the physical ordering of the lanes, as the PCS sublayer is capable of receiving the lanes in any arrangement. The positioning of transmit and receive lanes at the MDI is specified in 123.11.3.

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123.7 PMD to MDI optical specifications for 400GBASE-SR16

The operating range for the 400GBASE-SR16 PMD is defined in Table 123–5. A 400GBASE-SR16 compliant PMD operates on 50/125 μm multimode fibers, type A1a.2 (OM3), type A1a.3 (OM4), or type A1a.4 (OM5), according to the specifications defined in Table 123–6. A PMD that exceeds the operating range requirement while meeting all other optical specifications is considered compliant (e.g., a 400GBASE-SR16 PMD operating at 120 m meets the operating range requirement of 0.5 m to 100 m). The operating range shown in Table 123–5 is the same as 100GBASE-SR4 (See Clause 95).

Table 123–5—400GBASE-SR16 operating range

PMD type	Required operating range[a]
400GBASE-SR16	0.5 m to 70 m for OM3
	0.5 m to 100 m for OM4
	0.5 m to 100 m for OM5

[a] The PCS FEC correction function may not be bypassed for any operating distance.

123.7.1 400GBASE-SR16 transmitter optical specifications

The optical characteristics of each lane of a 400GBASE-SR16 transmitter shall be the same as those of a single lane of 100GBASE-SR4, as specified in Table 95–6 and 95.7.1, with the exception that the "signaling rate, each lane" is 26.5625 GBd ± 100 ppm, TDEC is as specified in 123.8.5, and the BER requirement is as specified in 123.1.1.

123.7.2 400GBASE-SR16 receive optical specifications

The optical characteristics of each lane of a 400GBASE-SR16 receiver shall be the same as those of a single lane of 100GBASE-SR4, as specified in Table 95–7 and 95.7.2, with the exception that the "signaling rate, each lane" is 26.5625 GBd ± 100 ppm, stressed receiver sensitivity is as specified in 123.8.8, and the BER requirement is as specified in 123.1.1.

123.7.3 400GBASE-SR16 illustrative link power budget

The illustrative power budget and penalties for each lane of a 400GBASE-SR16 link are the same as those of a single lane of 100GBASE-SR4, as specified in Table 95–8 and 95.7.3.

123.8 Definition of optical parameters and measurement methods

All transmitter optical measurements shall be made through a short patch cable, between 2 m and 5 m in length, unless otherwise specified. Optical parameters and measurement methods for 400GBASE-SR16 are the same as for 100GBASE-SR4.

123.8.1 Test patterns for optical parameters

The test patterns used in this clause shall be the same as those used for 100GBASE-SR4, as described in 95.8.1 and shown in Table 95–9, with the exception that pattern 5 defined in 82.2.11 is encoded by Clause 119 PCS for 400GBASE-SR16. Table 95–10 as modified by this subclause (123.8.1) shows the test patterns to be used in each measurement, unless otherwise specified, and also lists references to the subclauses in which each parameter is defined.

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123.8.2 Center wavelength and spectral width

The center wavelength and RMS spectral width shall be within the range given in Table 95-6 if measured per TIA/EIA-455-127-A or IEC 61280-1-3. The transmitter is modulated using one of the test patterns specified in Table 95–10 as modified by 123.8.1.

123.8.3 Average optical power

The average optical power shall be within the limits given in Table 95–6 if measured using the methods given in IEC 61280-1-1. The average optical power is measured using one of the test patterns specified in Table 95–10 as modified by 123.8.1.

123.8.4 Optical Modulation Amplitude (OMA)

OMA shall be within the limits given in Table 95–6 if measured as defined in 52.9.5 for measurement with a square wave (8 ones, 8 zeros) test pattern or as defined in 68.6.2 (from the variable MeasuredOMA in 68.6.6.2) for measurement with a PRBS9 test pattern. See 123.8.1 for test pattern information.

123.8.5 Transmitter and dispersion eye closure (TDEC)

TDEC is a measure of the optical transmitter's vertical eye closure, as described in 95.8.5. TDEC shall be within the limits given in Table 95–6 if measured using the methods specified in 95.8.5.1 and 95.8.5.2, with the exceptions that the clock recovery unit (CRU) has a corner frequency of 4 MHz and in Equation 95–6, 3.8906R is replaced by 3.4917R, for consistency with the BER of 2.4×10^{-4} given in 123.1.1. Table 95–10 as modified by 123.8.1 specifies the test patterns to be used for measurement of TDEC.

123.8.6 Extinction ratio

The extinction ratio shall be within the limits given in Table 95–6 if measured using the methods specified in IEC 61280-2-2. The extinction ratio is measured using one of the test patterns specified for extinction ratio in Table 95–10 as modified by 123.8.1.

NOTE—Extinction ratio and OMA are defined with different test patterns (see Table 95–10).

123.8.7 Transmitter optical waveform (transmit eye)

The transmitter optical waveform of a port transmitting the test pattern specified in Table 95–10 as modified by 123.8.1 shall meet the transmitter eye mask coordinates and hit ratio specified in Table 95–6 if measured according to the methods specified in 95.8.7 with the exception that the clock recovery unit's high frequency corner bandwidth is 4 MHz.

123.8.8 Stressed receiver sensitivity

Stressed receiver sensitivity shall be within the limits given in Table 95–7 if measured using the methods defined in 95.8.8 with the following exceptions:

- The signaling rate of the test pattern generator is set to the rate defined in 123.7.2.
- The clock recovery unit (CRU) has a corner bandwidth of 4 MHz.
- Sinusoidal jitter is added as specified in Table 87–13 instead of Table 95–11.
- When using 95.8.5 to measure the SEC of the stressed receiver conformance test signal, the BER requirement is as specified in 123.1.1 and 3.4917R replaces 3.8906R in Equation 95–6.
- The hit ratio for the stressed receiver eye mask definition is changed from 5×10^{-5} to 2.4×10^{-4}.

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123.9 Safety, installation, environment, and labeling

123.9.1 General safety

All equipment subject to this clause shall conform to IEC 60950-1.

123.9.2 Laser safety

400GBASE-SR16 optical transceivers shall conform to Hazard Level 1M laser requirements as defined in IEC 60825-1 and IEC 60825-2, under any condition of operation. This includes single fault conditions whether coupled into a fiber or out of an open bore.

Conformance to additional laser safety standards may be required for operation within specific geographic regions.

Laser safety standards and regulations require that the manufacturer of a laser product provide information about the product's laser, safety features, labeling, use, maintenance, and service. This documentation explicitly defines requirements and usage restrictions on the host system necessary to meet these safety certifications.[10]

123.9.3 Installation

It is recommended that proper installation practices, as defined by applicable local codes and regulation, be followed in every instance in which such practices are applicable.

123.9.4 Environment

Normative specifications in this clause shall be met by a system integrating a 400GBASE-SR16 PMD over the life of the product while the product operates within the manufacturer's range of environmental, power, and other specifications.

It is recommended that manufacturers indicate, in the literature associated with the PHY, the operating environmental conditions to facilitate selection, installation, and maintenance.

It is recommended that manufacturers indicate, in the literature associated with the components of the optical link, the distance and operating environmental conditions over which the specifications of this clause will be met.

123.9.5 Electromagnetic emission

A system integrating a 400GBASE-SR16 PMD shall comply with applicable local and national codes for the limitation of electromagnetic interference.

123.9.6 Temperature, humidity, and handling

The optical link is expected to operate over a reasonable range of environmental conditions related to temperature, humidity, and physical handling (such as shock and vibration). Specific requirements and values for these parameters are considered to be beyond the scope of this standard.

[10]A host system that fails to meet the manufacturer's requirements and/or usage restrictions may emit laser radiation in excess of the safety limits of one or more safety standards. In such a case, the host manufacturer is required to obtain its own laser safety certification.

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123.9.7 PMD labeling requirements

It is recommended that each PHY (and supporting documentation) be labeled in a manner visible to the user, with at least the applicable safety warnings and the applicable port type designation (e.g. 400GBASE-SR16).

Labeling requirements for Hazard Level 1M lasers are given in the laser safety standards referenced in 123.9.2.

123.10 Fiber optic cabling model

The fiber optic cabling model is shown in Figure 123–3. The fiber type and length are the same as 100BASE-SR4 (See Clause 95).

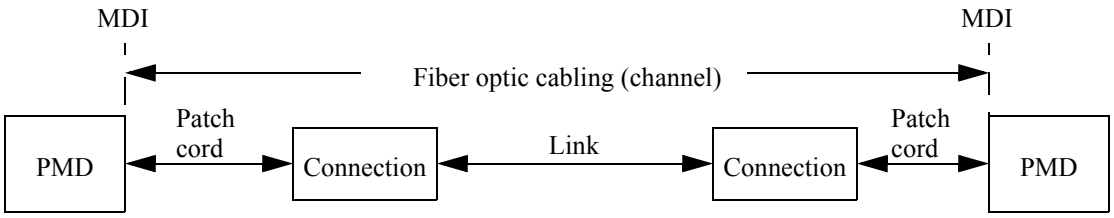


Figure 123–3—Fiber optic cabling model

The channel insertion loss is given in Table 123–6. A channel may contain additional connectors as long as the optical characteristics of the channel, such as attenuation, modal dispersion, reflections and losses of all connectors and splices meet the specifications. Insertion loss measurements of installed fiber cables are made in accordance with IEC 61280-4-1:2009. As OM4 and OM5 optical fiber meet the requirements for OM3, a channel compliant to the "OM3" column may use OM4 or OM5 optical fiber, or a combination of OM3, OM4 and OM5. The fiber optic cabling model (channel) defined here is the same as a simplex fiber optic link segment. The term channel is used here for consistency with generic cabling standards.

Table 123–6—Fiber optic cabling (channel) characteristics for 400GBASE-SR16

Description	OM3	OM4	OM5	Unit
Operating distance (max)	70	100	100	m
Channel insertion loss[a] (max)	1.8	1.9	1.9	dB
Channel insertion loss (min)	0			dB

[a] These channel insertion loss values include cable loss plus 1.5 dB allocated for connection and splice loss, over the wavelength range 840 nm to 860 nm.

123.11 Characteristics of the fiber optic cabling (channel)

The 400GBASE-SR16 fiber optic cabling shall meet the specifications defined in Table 123–6. The fiber optic cabling consists of one or more sections of fiber optic cable and any intermediate connections required to connect sections together. The requirements for the optical fiber, connection insertion loss, and maximum discrete reflectance are the same as 100GBASE-SR4 (see Clause 95).

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123.11.1 Optical fiber cable

The fiber contained within the 400GBASE-SR16 fiber optic cabling shall comply with the specifications and parameters of Table 123–7. A variety of multimode cable types may satisfy these requirements, provided the resulting channel also meets the specifications of Table 123–6.

Table 123–7—Optical fiber and cable characteristics

Description	OM3[a]	OM4[b]	OM5[c]	Unit
Nominal core diameter	50			μm
Nominal fiber specification wavelength	850			nm
Effective modal bandwidth (min)[d]	2000	4700		MHz.km
Cabled optical fiber attenuation (max)	3.5			dB/km
Zero dispersion wavelength (λ_0)	$1295 \le \lambda_0 \le 1340$		$1297 \le \lambda_0 \le 1328$	nm
Chromatic dispersion slope (max) (S_0)	0.105 for $1295 \le \lambda_0 \le 1310$ and $0.000375 \times (1590 - \lambda_0)$ for $1310 \le \lambda_0 \le 1340$		$-412/(840(1-(\lambda_0/840)^4))$ for $1297 \le \lambda_0 \le 1328$	ps/nm^2 km

[a] IEC 60793-2-10 type A1a.2
[b] IEC 60793-2-10 type A1a.3
[c] IEC 60793-2-10 type A1a.4
[d] When measured with the launch conditions specified in Table 95–6

123.11.2 Optical fiber connection

An optical fiber connection, as shown in Figure 123–3, consists of a mated pair of optical connectors.

123.11.2.1 Connection insertion loss

The maximum link distance is based on an allocation of 1.5 dB total connection and splice loss. For example, this allocation supports 3 connections with an average insertion loss per connection of 0.5 dB. Connections with lower loss characteristics may be used provided the requirements of Table 123–6 are met. However, the loss of a single connection shall not exceed 0.75 dB

123.11.2.2 Maximum discrete reflectance

The maximum discrete reflectance shall be less than –20 dB.

123.11.3 Medium Dependent Interface (MDI)

The 400GBASE-SR16 PMD is coupled to the fiber optic cabling at the MDI. The MDI is the interface between the PMD and the "fiber optic cabling" (as shown in Figure 123–3). Examples of an MDI include the following:

a) Connectorized fiber pigtail
b) PMD receptacle

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123.11.3.1 Optical lane assignments

The 16 transmit and 16 receive optical lanes of 400GBASE-SR16 shall occupy the positions depicted in Figure 123–4 when looking into the MDI receptacle with the connector keyway feature on top. The interface contains 32 active lanes. The transmit optical lanes occupy the upper 16 positions. The receive optical lanes occupy the lower 16 positions.

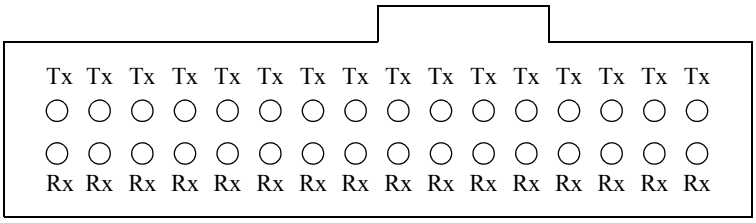


Figure 123–4—400GBASE-SR16 optical lane assignments

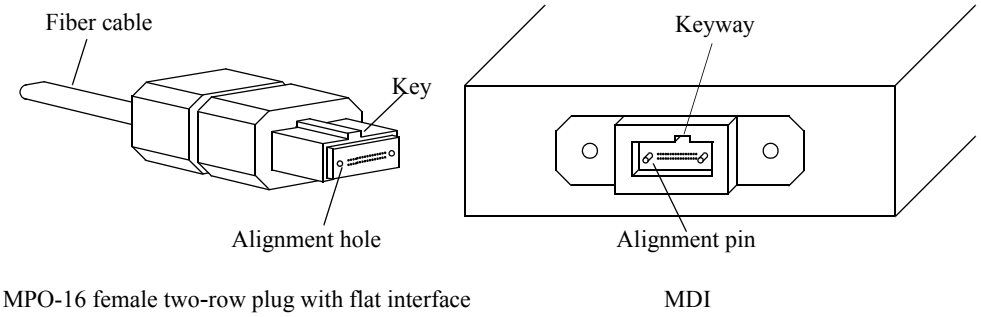


Figure 123–5—MPO-16 female two-row plug with flat interface, and an MDI

123.11.3.2 Medium Dependent Interface (MDI) requirements

The MDI adapter or receptacle shall meet the dimensional specifications for designation FOCIS 18 A-k-0 as defined in ANSI/TIA-604-18. The plug terminating the optical fiber cabling shall meet the dimensional specifications of designation FOCIS 18 P-2x16-1-0-2-2 as defined in ANSI/TIA-604-18. The MDI shall optically mate with the plug on the optical fiber cabling. Figure 123–5 shows an MPO-16 female two-row plug with flat interface, and an MDI. The MDI shall meet the interface performance specifications of IEC 61753-1 and IEC 61753-022-2 for performance Class Cm/2m.

NOTE—Transmitter compliance testing is performed at TP2 as defined in 123.5.1, not at the MDI.

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123.12 Protocol implementation conformance statement (PICS) proforma for Clause 123, Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-SR16[11]

123.12.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Clause 123, Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-SR16, shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

123.12.2 Identification

123.12.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

123.12.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Clause 123, Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-SR16
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[11] *Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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123.12.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
*INS	Installation / cable	123.10	Items marked with INS include installation practices and cable specifications not applicable to a PHY manufacturer	O	Yes [ ] No [ ]
TP1	Reference point TP1 exposed and available for testing	123.5.1	This point may be made available for use by implementers to certify component conformance	O	Yes [ ] No [ ]
TP4	Reference point TP4 exposed and available for testing	123.5.1	This point may be made available for use by implementers to certify component conformance	O	Yes [ ] No [ ]
DC	Delay constraints	123.3.1	Device conforms to delay constraints	M	Yes [ ]
SC	Skew constraints	123.3.2	Device conforms to Skew and Skew Variation constraints	M	Yes [ ]
*MD	MDIO capability	123.4	Registers and interface supported	O	Yes [ ] No [ ]

123.12.4 PICS proforma tables for Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-SR16

123.12.4.1 PMD functional specifications

Item	Feature	Subclause	Value/Comment	Status	Support
F1	Compatible with 400GBASE-R PCS and PMA	123.1		M	Yes [ ]
F2	Integration with management functions	123.1		O	Yes [ ] No [ ]
F3	Bit error ratio	123.1.1	Meets the BER specified in 123.1.1	M	Yes [ ]
F4	Transmit function	123.5.2	Conveys bits from PMD service interface to MDI	M	Yes [ ]
F5	Mapping between optical signal and logical signal for transmitter	123.5.2	Higher optical power is a one	M	Yes [ ]
F6	Receive function	123.5.3	Conveys bits from MDI to PMD service interface	M	Yes [ ]
F7	Conversion of 16 optical signals to 16 electrical signals	123.5.3	For delivery to the PMD service interface	M	Yes [ ]
F8	Mapping between optical signal and logical signal for receiver	123.5.3	Higher optical power is a one	M	Yes [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
F9	Global Signal Detect function	123.5.4	Report to the PMD service interface the message PMD:IS_SIGNAL.indication (SIGNAL_DETECT)	M	Yes [ ]
F10	Global Signal Detect behavior	123.5.4	SIGNAL_DETECT is a global indicator of the presence of optical signals on all 16 lanes	M	Yes [ ]
F11	Lane-by-lane Signal Detect function	123.5.5	Sets PMD_signal_detect_*i* values on a lane-by-lane basis per requirements of Table 123–4	MD:O	Yes [ ] No [ ] N/A [ ]
F12	PMD reset function	123.5.6	Resets the PMD sublayer	MD:O	Yes [ ] No [ ] N/A [ ]

123.12.4.2 Management functions

Item	Feature	Subclause	Value/Comment	Status	Support
M1	Management register set	123.4		MD:M	Yes [ ] N/A [ ]
M2	Global transmit disable function	123.5.7	Disables all of the optical transmitters with the PMD_global_transmit_disable variable	MD:O	Yes [ ] No [ ] N/A [ ]
M3	PMD_lane_by_lane_transmit_disable function	123.5.8	Disables the optical transmitter on the lane associated with the PMD_transmit_disable_*i* variable	MD:O	Yes [ ] No [ ] N/A [ ]
M4	PMD_fault function	123.5.9	Sets PMD_fault to one if any local fault is detected	MD:O	Yes [ ] No [ ] N/A [ ]
M5	PMD_transmit_fault function	123.5.10	Sets PMD_transmit_fault to one if a local fault is detected on any transmit lane	MD:O	Yes [ ] No [ ] N/A [ ]
M6	PMD_receive_fault function	123.5.11	Sets PMD_receive_fault to one if a local fault is detected on any receive lane	MD:O	Yes [ ] No [ ] N/A [ ]

123.12.4.3 PMD to MDI optical specifications for 400GBASE-SR16

Item	Feature	Subclause	Value/Comment	Status	Support
SR1	Transmitter meets specifications in 95.7.1 with the exceptions in 123.7.1	123.7.1	Per definitions in 123.8	M	Yes [ ]
SR2	Receiver meets specifications in 95.7.2 with the exceptions in 123.7.2	123.7.2	Per definitions in 123.8	M	Yes [ ]

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123.12.4.4 Optical measurement methods

Item	Feature	Subclause	Value/Comment	Status	Support
OM1	Measurement cable	123.8	2 m to 5 m in length	M	Yes [ ]
OM2	Center wavelength and spectral width	123.8.2	Per TIA/EIA-455-127-A or IEC 61280-1-3 under modulated conditions	M	Yes [ ]
OM3	Average optical power	123.8.3	Per IEC 61280-1-1	M	Yes [ ]
OM4	OMA measurements	123.8.4	Each lane	M	Yes [ ]
OM5	Transmitter and dispersion eye closure (TDEC)	123.8.5	Each lane	M	Yes [ ]
OM6	Extinction ratio	123.8.6	Per IEC 61280-2-2	M	Yes [ ]
OM7	Transmit eye	123.8.7	Each lane	M	Yes [ ]
OM8	Stressed receiver sensitivity	123.8.8		M	Yes [ ]

123.12.4.5 Environmental specifications

Item	Feature	Subclause	Value/Comment	Status	Support
ES1	General safety	123.9.1	Conforms to IEC 60950-1	M	Yes [ ]
ES2	Laser safety—IEC Hazard Level 1M	123.9.2	Conforms to Hazard Level 1M laser requirements defined in IEC 60825-1 and IEC 60825-2	M	Yes [ ]
ES3	Electromagnetic interference	123.9.5	Complies with applicable local and national codes for the limitation of electromagnetic interference	M	Yes [ ]

123.12.4.6 Characteristics of the fiber optic cabling and MDI

Item	Feature	Subclause	Value/Comment	Status	Support
OC1	Fiber optic cabling	123.11	Meets requirements specified in Table 123–6	INS:M	Yes [ ] N/A [ ]
OC2	Optical fiber characteristics	123.11.1	Per Table 123–7	INS:M	Yes [ ] N/A [ ]
OC3	Maximum discrete reflectance	123.11.2.2	Less than –20 dB	INS:M	Yes [ ] N/A [ ]
OC4	MDI layout	123.11.3.1	Optical lane assignments per Figure 123–4	M	Yes [ ]
OC5	MDI dimensions	123.11.3.2	Meets FOCIS 18 A-k-0 as defined in ANSI/TIA-604-18	M	Yes [ ]
OC6	Cabling connector dimensions	123.11.3.2	Meets FOCIS 18 P-2x16-1-0-2-2 as defined in ANSI/TIA-604-18	INS:M	Yes [ ] N/A [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
OC7	MDI mating	123.11.3.2	MDI optically mates with plug on the cabling	M	Yes [ ]
OC8	MDI requirements	123.11.3.2	Meets IEC 61753-1 and IEC 61753-022-2	INS:M	Yes [ ] N/A [ ]

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124. Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-DR4

124.1 Overview

This clause specifies the 400GBASE-DR4 PMD together with the single-mode fiber medium. The optical signal generated by this PMD type is modulated using a 4-level pulse amplitude modulation (PAM4) format. When forming a complete Physical Layer, a PMD shall be connected to the appropriate PMA as shown in Table 124–1, to the medium through the MDI and optionally with the management functions that may be accessible through the management interface defined in Clause 45, or equivalent.

Table 124–1—Physical Layer clauses associated with the 400GBASE-DR4 PMD

Associated clause	400GBASE-DR4
117—RS	Required
117—400GMII[a]	Optional
118—400GMII Extender	Optional
119—PCS	Required
120—PMA	Required
120B—Chip-to-chip 400GAUI-16	Optional
120C—Chip-to-module 400GAUI-16	Optional
120D—Chip-to-chip 400GAUI-8	Optional
120E—Chip-to-module 400GAUI-8	Optional
78—Energy Efficient Ethernet	Optional

[a] The 400GMII is an optional interface. However, if the 400GMII is not implemented, a conforming implementation must behave functionally as though the RS and 400GMII were present.

Figure 124–1 shows the relationship of the PMD and MDI (shown shaded) with other sublayers to the ISO/IEC Open System Interconnection (OSI) reference model. 400 Gb/s Ethernet is introduced in Clause 116 and the purpose of each PHY sublayer is summarized in 116.2.

400GBASE-DR4 PHYs with the optional Energy Efficient Ethernet (EEE) fast wake capability may enter the Low Power Idle (LPI) mode to conserve energy during periods of low link utilization (see Clause 78). The deep sleep mode of EEE is not supported.

124.1.1 Bit error ratio

The bit error ratio (BER) when processed according to Clause 120 shall be less than 2.4×10^{-4} provided that the error statistics are sufficiently random that this results in a frame loss ratio (see 1.4.223) of less than 1.7×10^{-12} for 64-octet frames with minimum interpacket gap when processed according to Clause 120 and then Clause 119. For a complete Physical Layer, the frame loss ratio may be degraded to 6.2×10^{-11} for 64-octet frames with minimum interpacket gap due to additional errors from the electrical interfaces.

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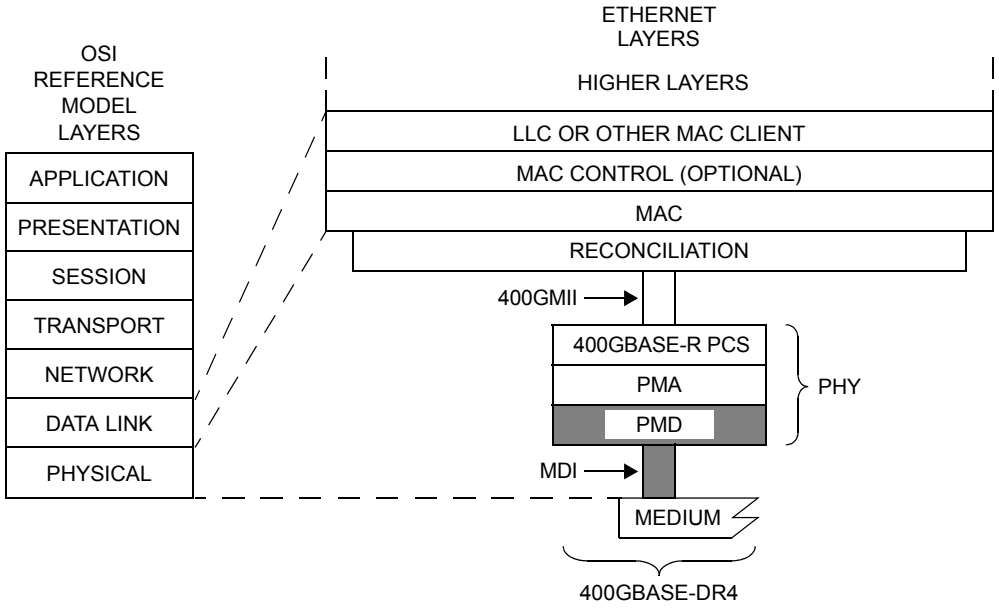


Figure 124–1—400GBASE-DR4 PMD relationship to the ISO/IEC Open Systems Interconnection (OSI) reference model and IEEE 802.3 Ethernet model

If the error statistics are not sufficiently random to meet this requirement, then the BER shall be less than that required to give a frame loss ratio of less than 1.7×10^{-12} for 64-octet frames with minimum interpacket gap.

124.2 Physical Medium Dependent (PMD) service interface

This subclause specifies the services provided by the 400GBASE-DR4 PMD. The service interface for this PMD is described in an abstract manner and does not imply any particular implementation. The PMD service interface supports the exchange of encoded data between the PMA entity that resides just above the PMD, and the PMD entity. The PMD translates the encoded data to and from signals suitable for the specified medium.

The PMD service interface is an instance of the inter-sublayer service interface defined in 116.3. The PMD service interface primitives are summarized as follows:

PMD:IS_UNITDATA_*i*.request
PMD:IS_UNITDATA_*i*.indication
PMD:IS_SIGNAL.indication

The 400GBASE-DR4 PMD has four parallel symbol streams, hence $i = 0$ to 3.

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In the transmit direction, the PMA continuously sends four parallel symbol streams to the PMD, one per lane, each at a nominal signaling rate of 53.125 GBd. The PMD then converts these streams of data units into the appropriate signals on the MDI.

In the receive direction, the PMD continuously sends four parallel symbol streams to the PMA corresponding to the signals received from the MDI, one per lane, each at a nominal signaling rate of 53.125 GBd. See NOTE at the end of 120.5.2 concerning the transition density of lanes operating at this nominal signaling rate.

The SIGNAL_DETECT parameter defined in this clause maps to the SIGNAL_OK parameter in the PMD:IS_SIGNAL.indication(SIGNAL_OK) inter-sublayer service interface primitive defined in 116.3.

The SIGNAL_DETECT parameter can take on one of two values: OK or FAIL. When SIGNAL_DETECT = FAIL, the rx_symbol parameters are undefined.

NOTE—SIGNAL_DETECT = OK does not guarantee that the rx_symbol parameters are known to be good. It is possible for a poor quality link to provide sufficient light for a SIGNAL_DETECT = OK indication and still not meet the BER defined in 124.1.1.

124.3 Delay and Skew

124.3.1 Delay constraints

The sum of the transmit and receive delays at one end of the link contributed by the 400GBASE-DR4 PMD including 2 m of fiber in one direction shall be no more than 8192 bit times (16 pause_quanta or 20.48 ns). A description of overall system delay constraints and the definitions for bit times and pause_quanta can be found in 116.4 and its references.

124.3.2 Skew constraints

The Skew (relative delay between the lanes) and Skew Variation must be kept within limits so that the information on the lanes can be reassembled by the PCS. Skew and Skew Variation are defined in 116.5 and specified at the points SP1 to SP6 shown in Figure 116–4 and Figure 116–5.

If the PMD service interface is physically instantiated so that the Skew at SP2 can be measured, then the Skew at SP2 is limited to 43 ns and the Skew Variation at SP2 is limited to 400 ps.

The Skew at SP3 (the transmitter MDI) shall be less than 54 ns and the Skew Variation at SP3 shall be less than 600 ps.

The Skew at SP4 (the receiver MDI) shall be less than 134 ns and the Skew Variation at SP4 shall be less than 3.4 ns.

If the PMD service interface is physically instantiated so that the Skew at SP5 can be measured, then the Skew at SP5 shall be less than 145 ns and the Skew Variation at SP5 shall be less than 3.6 ns.

For more information on Skew and Skew Variation see 116.5. The measurements of Skew and Skew Variation are defined in 86.8.3.1.

124.4 PMD MDIO function mapping

The optional MDIO capability described in Clause 45 defines several variables that may provide control and status information for and about the PMD. If the MDIO interface is implemented, the mapping of MDIO

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control variables to PMD control variables shall be as shown in Table 124–2 and the mapping of MDIO status variables to PMD status variables shall be as shown in Table 124–3.

Table 124–2—MDIO/PMD control variable mapping

MDIO control variable	PMA/PMD register name	Register/bit number	PMD control variable
Reset	PMA/PMD control 1 register	1.0.15	PMD_reset
Global PMD transmit disable	PMD transmit disable register	1.9.0	PMD_global_transmit_disable
PMD transmit disable 3 to PMD transmit disable 0	PMD transmit disable register	1.9.4 to 1.9.1	PMD_transmit_disable_3 to PMD_transmit_disable_0

Table 124–3—MDIO/PMD status variable mapping

MDIO status variable	PMA/PMD register name	Register/ bit number	PMD status variable
Fault	PMA/PMD status 1 register	1.1.7	PMD_fault
Transmit fault	PMA/PMD status 2 register	1.8.11	PMD_transmit_fault
Receive fault	PMA/PMD status 2 register	1.8.10	PMD_receive_fault
Global PMD receive signal detect	PMD receive signal detect register	1.10.0	PMD_global_signal_detect
PMD receive signal detect 3 to PMD receive signal detect 0	PMD receive signal detect register	1.10.4 to 1.10.1	PMD_signal_detect_3 to PMD_signal_detect_0

124.5 PMD functional specifications

The 400GBASE-DR4 PMD performs the Transmit and Receive functions, which convey data between the PMD service interface and the MDI.

124.5.1 PMD block diagram

The PMD block diagram is shown in Figure 124–2. For purposes of system conformance, the PMD sublayer is standardized at the points described in this subclause. The optical transmit signal is defined at the output end of a single-mode fiber patch cord (TP2), between 2 m and 5 m in length. Unless specified otherwise, all transmitter measurements and tests defined in 124.8 are made at TP2. The optical receive signal is defined at the output of the fiber optic cabling (TP3) at the MDI (see 124.11.3). Unless specified otherwise, all receiver measurements and tests defined in 124.8 are made at TP3.

TP1<0:3> and TP4<0:3> are informative reference points that may be useful to implementors for testing components (these test points will not typically be accessible in an implemented system).

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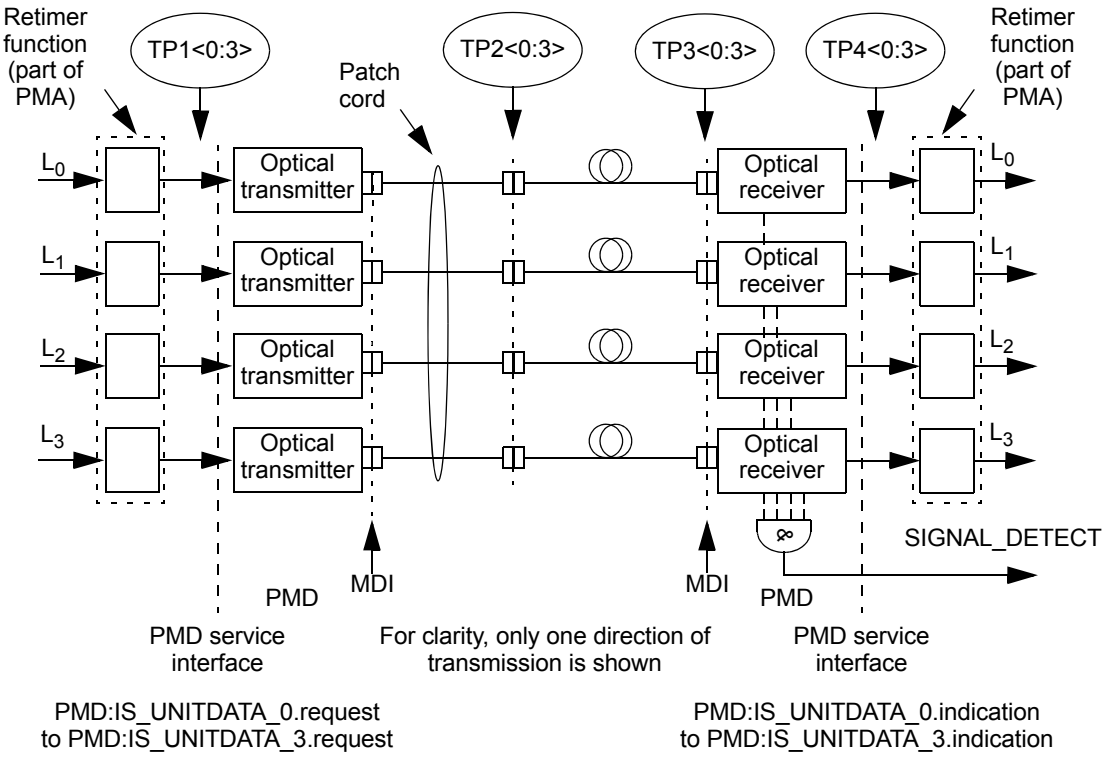


Figure 124–2—Block diagram for 400GBASE-DR4 transmit/receive paths

124.5.2 PMD transmit function

The PMD Transmit function shall convert the four symbol streams requested by the PMD service interface messages PMD:IS_UNITDATA_0.request to PMD:IS_UNITDATA_3.request into four separate optical signals. The four optical signals shall then be delivered to the MDI, which contains four parallel light paths for transmit, according to the transmit optical specifications in this clause. The highest optical power level in each signal stream shall correspond to tx_symbol = three and the lowest shall correspond to tx_symbol = zero.

124.5.3 PMD receive function

The PMD Receive function shall convert the four parallel optical signals received from the MDI into separate symbol streams for delivery to the PMD service interface using the messages PMD:IS_UNITDATA_0.indication to PMD:IS_UNITDATA_3.indication, all according to the receive optical specifications in this clause. The higher optical power level in each signal stream shall correspond to rx_symbol = three and the lowest shall correspond to rx_symbol = zero.

124.5.4 PMD global signal detect function

The PMD global signal detect function shall report the state of SIGNAL_DETECT via the PMD service interface. The SIGNAL_DETECT parameter is signaled continuously, while the PMD:IS_SIGNAL.indication message is generated when a change in the value of SIGNAL_DETECT occurs. The SIGNAL_DETECT parameter defined in this clause maps to the SIGNAL_OK parameter in the inter-sublayer service interface primitives defined in 116.3.

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SIGNAL_DETECT shall be a global indicator of the presence of optical signals on all four lanes. The value of the SIGNAL_DETECT parameter shall be generated according to the conditions defined in Table 124–4. The PMD receiver is not required to verify whether a compliant 400GBASE-DR4 signal is being received. This standard imposes no response time requirements on the generation of the SIGNAL_DETECT parameter.

Table 124–4—SIGNAL_DETECT value definition

Receive conditions	SIGNAL_DETECT value
For any lane; Average optical power at TP3 ≤ –15 dBm	FAIL
For all lanes; [(Optical power at TP3 ≥ average receive power, each lane (min) Table 124–7) AND (compliant 400GBASE–R signal input)]	OK
All other conditions	Unspecified

As an unavoidable consequence of the requirements for the setting of the SIGNAL_DETECT parameter, implementations must provide adequate margin between the input optical power level at which the SIGNAL_DETECT parameter is set to OK, and the inherent noise level of the PMD including the effects of crosstalk, power supply noise, etc.

Various implementations of the Signal Detect function are permitted by this standard, including implementations that generate the SIGNAL_DETECT parameter values in response to the amplitude of the modulation of the optical signal and implementations that respond to the average optical power of the modulated optical signal.

124.5.5 PMD lane-by-lane signal detect function

Various implementations of the Signal Detect function are permitted by this standard. When the MDIO is implemented, each PMD_signal_detect_*i*, where *i* represents the lane number in the range 0:3, shall be continuously set in response to the magnitude of the optical signal on its associated lane, according to the requirements of Table 124–4.

124.5.6 PMD reset function

If the MDIO interface is implemented, and if PMD_reset is asserted, the PMD shall be reset as defined in 45.2.1.1.1.

124.5.7 PMD global transmit disable function (optional)

The PMD_global_transmit_disable function is optional and allows all of the optical transmitters to be disabled.

a) When the PMD_global_transmit_disable variable is set to one, this function shall turn off all of the optical transmitters so that each transmitter meets the requirements of the average launch power of the OFF transmitter in Table 124–6.

b) If a PMD_fault is detected, then the PMD may set the PMD_global_transmit_disable to one, turning off the optical transmitter in each lane.

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124.5.8 PMD lane-by-lane transmit disable function (optional)

The PMD_transmit_disable_*i* (where *i* represents the lane number in the range 0:3) function is optional and allows the optical transmitters in each lane to be selectively disabled.

a) When a PMD_transmit_disable_*i* variable is set to one, this function shall turn off the optical transmitter associated with that variable so that the transmitter meets the requirements of the average launch power of the OFF transmitter in Table 124–6.

b) If a PMD_fault is detected, then the PMD may set each PMD_transmit_disable_*i* to one, turning off the optical transmitter in each lane.

If the optional PMD_transmit_disable_*i* function is not implemented in MDIO, an alternative method may be provided to independently disable each transmit lane for testing purposes.

124.5.9 PMD fault function (optional)

If the PMD has detected a local fault on any of the transmit or receive paths, the PMD shall set PMD_fault to one.

If the MDIO interface is implemented, PMD_fault shall be mapped to the fault bit as specified in 45.2.1.2.3.

124.5.10 PMD transmit fault function (optional)

If the PMD has detected a local fault on any transmit lane, the PMD shall set the PMD_transmit_fault variable to one.

If the MDIO interface is implemented, PMD_transmit_fault shall be mapped to the transmit fault bit as specified in 45.2.1.7.4.

124.5.11 PMD receive fault function (optional)

If the PMD has detected a local fault on any receive lane, the PMD shall set the PMD_receive_fault variable to one.

If the MDIO interface is implemented, PMD_receive_fault shall be mapped to the receive fault bit as specified in 45.2.1.7.5.

124.6 Lane assignments

There are no lane assignments (within a group of transmit or receive lanes) for 400GBASE-DR4. While it is expected that a PMD will map electrical lane *i* to optical lane *i* and vice versa, there is no need to define the physical ordering of the lanes, as the PCS sublayer is capable of receiving the lanes in any arrangement. The positioning of transmit and receive lanes at the MDI is specified in 124.11.3.1.

124.7 PMD to MDI optical specifications for 400GBASE-DR4

The operating range for the 400GBASE-DR4 PMD is defined in Table 124–5. A 400GBASE-DR4 compliant PMD operates on type B1.1, B1.3, or B6_a single-mode fibers according to the specifications defined in Table 124–12. A PMD that exceeds the operating range requirement while meeting all other optical specifications is considered compliant (e.g., a 400GBASE-DR4 PMD operating at 600 m meets the operating range requirement of 2 m to 500 m).

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Table 124–5—400GBASE-DR4 operating range

PMD type	Required operating range
400GBASE-DR4	2 m to 500 m

124.7.1 400GBASE-DR4 transmitter optical specifications

The 400GBASE-DR4 transmitter shall meet the specifications defined in Table 124–6 per the definitions in 124.8.

Table 124–6—400GBASE-DR4 transmit characteristics

Description	Value	Unit
Signaling rate, each lane (range)	53.125 ± 100 ppm	GBd
Modulation format	PAM4	—
Lane wavelength (range)	1304.5 to 1317.5	nm
Side-mode suppression ratio (SMSR), (min)	30	dB
Average launch power, each lane (max)	4	dBm
Average launch power, each lane[a] (min)	–2.9	dBm
Outer Optical Modulation Amplitude (OMA_{outer}), each lane (max)	4.2	dBm
Outer Optical Modulation Amplitude (OMA_{outer}), each lane (min)[b]	–0.8	dBm
Launch power in OMA_{outer} minus TDECQ, each lane (min)	–2.2	dBm
Transmitter and dispersion eye closure for PAM4 (TDECQ), each lane (max)	3.4	dB
Average launch power of OFF transmitter, each lane (max)	–15	dBm
Extinction ratio, each lane (min)	3.5	dB
$RIN_{21.4}OMA$ (max)	–136	dB/Hz
Optical return loss tolerance (max)	21.4	dB
Transmitter reflectance[c] (max)	–26	dB

[a] Average launch power, each lane (min) is informative and not the principal indicator of signal strength. A transmitter with launch power below this value cannot be compliant; however, a value above this does not ensure compliance.
[b] Even if the TDECQ < 1.4 dB, the OMA_{outer} (min) must exceed these values.
[c] Transmitter reflectance is defined looking into the transmitter.

124.7.2 400GBASE-DR4 receive optical specifications

The 400GBASE-DR4 receiver shall meet the specifications defined in Table 124–7 per the definitions in 124.8. See NOTE at the end of 120.5.2 concerning the transition density of lanes operating at this nominal signaling rate.

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Table 124–7—400GBASE-DR4 receive characteristics

Description	Value	Unit
Signaling rate, each lane (range)	53.125 ± 100 ppm	GBd
Modulation format	PAM4	—
Lane wavelengths (range)	1304.5 to 1317.5	nm
Damage threshold[a], each lane	5	dBm
Average receive power, each lane (max)	4	dBm
Average receive power, each lane[b] (min)	–5.9	dBm
Receive power (OMA_{outer}), each lane (max)	4.2	dBm
Receiver reflectance (max)	–26	dB
Receiver sensitivity (OMA_{outer}), each lane[c] (max)	–4.4	dBm
Stressed receiver sensitivity (OMA_{outer}), each lane[d] (max)	–1.9	dBm
Conditions of stressed receiver sensitivity test:[e]		
Stressed eye closure for PAM4 (SECQ), lane under test	3.4	dB
OMA_{outer} of each aggressor lane	4.2	dBm

[a] The receiver shall be able to tolerate, without damage, continuous exposure to an optical input signal having this average power level. The receiver does not have to operate correctly at this input power.
[b] Average receive power, each lane (min) is informative and not the principal indicator of signal strength. A received power below this value cannot be compliant; however, a value above this does not ensure compliance.
[c] Receiver sensitivity (OMA_{outer}), each lane (max) is informative and is defined for a transmitter with SECQ of 0.9 dB.
[d] Measured with conformance test signal at TP3 (see 124.8.9) for the BER specified in 124.1.1.
[e] These test conditions are for measuring stressed receiver sensitivity. They are not characteristics of the receiver.

124.7.3 400GBASE-DR4 illustrative link power budget

An illustrative power budget and penalties for 400GBASE-DR4 channels are shown in Table 124–8.

Table 124–8—400GBASE-DR4 illustrative link power budget

Parameter	Value	Unit
Power budget (for max TDECQ)	6.5	dB
Operating distance	500	m
Channel insertion loss[a]	3	dB
Maximum discrete reflectance	See 124.11.2.2	dB
Allocation for penalties[b] (for max TDECQ)	3.5	dB
Additional insertion loss allowed	0	dB

[a] The channel insertion loss is calculated using the maximum distance specified in Table 124–5 and cabled optical fiber attenuation of 0.5 dB/km at 1304.5 nm plus an allocation for connection and splice loss given in 124.11.2.1.
[b] Link penalties are used for link budget calculations. They are not requirements and are not meant to be tested.

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124.8 Definition of optical parameters and measurement methods

All transmitter optical measurements shall be made through a short patch cable, between 2 m and 5 m in length, unless otherwise specified.

124.8.1 Test patterns for optical parameters

While compliance is to be achieved in normal operation, specific test patterns are defined for measurement consistency and to enable measurement of some parameters. Table 124–10 gives the test patterns to be used in each measurement, unless otherwise specified, and also lists references to the subclauses in which each parameter is defined. Any of the test patterns given for a particular test in Table 124–10 may be used to perform that test. The test patterns used in this clause are shown in Table 124–9.

Table 124–9—Test patterns

Pattern	Pattern description	Defined in
Square wave	Square wave (8 threes, 8 zeros)	120.5.11.2.4
3	PRBS31Q	120.5.11.2.2
4	PRBS13Q	120.5.11.2.1
5	Scrambled idle	119.2.4.9
6	SSPRQ	120.5.11.2.3

Table 124–10—Test-pattern definitions and related subclauses

Parameter	Pattern	Related subclause
Wavelength	Square wave, 3, 4, 5, 6 or valid 400GBASE-R signal	124.8.2
Side mode suppression ratio	3, 5, 6 or valid 400GBASE-R signal	—
Average optical power	3, 5, 6 or valid 400GBASE-R signal	124.8.3
Outer Optical Modulation Amplitude (OMA_{outer})	4 or 6	124.8.4
Transmitter and dispersion eye closure for PAM4 (TDECQ)	6	124.8.5
Extinction ratio	4 or 6	124.8.6
$RIN_{21.4}OMA$	Square wave	124.8.7
Stressed receiver conformance test signal calibration	6	124.8.9
Stressed receiver sensitivity	3 or 5	124.8.9

124.8.2 Wavelength

The wavelength of each optical lane shall be within the range given in Table 124–6 if measured per TIA/EIA-455-127-A or IEC 61280-1-3. The lane under test is modulated using the test pattern defined in Table 124–10.

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124.8.3 Average optical power

The average optical power of each lane shall be within the limits given in Table 124–6 if measured using the methods given in IEC 61280-1-1. The average optical power is measured using the test pattern defined in Table 124–10, per the test setup in Figure 53-6.

124.8.4 Outer Optical Modulation Amplitude (OMA_{outer})

The OMA_{outer} of each lane shall be within the limits given in Table 124–6. The OMA_{outer} is measured using a test pattern specified for OMA_{outer} in Table 124–10 as the difference between the average optical launch power level P_3, measured over the central 2 UI of a run of 7 threes, and the average optical launch power level P_0, measured over the central 2 UI of a run of 6 zeros, as shown in Figure 124–3.

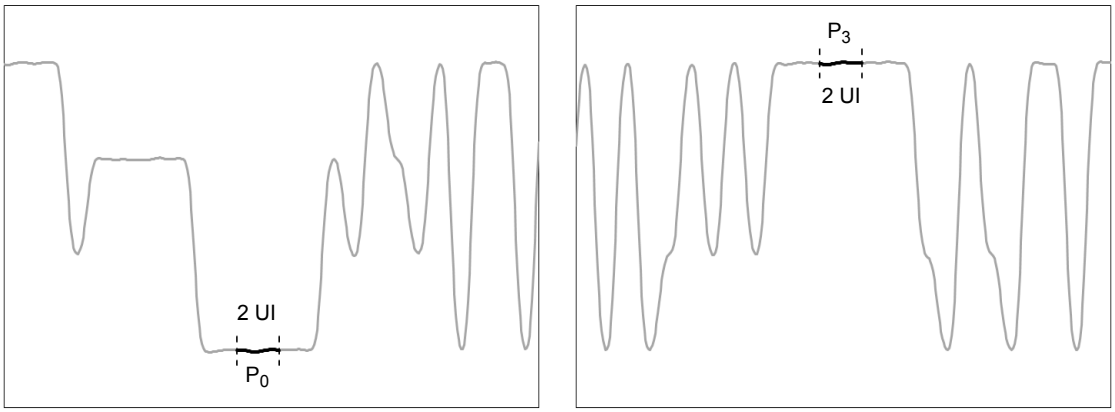


Figure 124–3—Example power levels P_0 and P_3 from PRBS13Q test pattern

124.8.5 Transmitter and dispersion eye closure for PAM4 (TDECQ)

The TDECQ of each lane shall be within the limits given in Table 124–6 if measured using the methods specified in 121.8.5.1, 121.8.5.2, and 121.8.5.3 using a reference equalizer as described in 121.8.5.4, with the following exceptions:

— The signaling rate of the test pattern generator is as given in Table 124–6 and uses the test pattern specified for TDECQ in Table 124–10.

— The combination of the O/E converter and the oscilloscope has a fourth-order Bessel-Thomson filter response with a bandwidth of approximately 26.5625 GHz.

— The normalized noise power density spectrum $N(f)$ is equivalent to white noise filtered by a fourth-order Bessel-Thomson response filter with a bandwidth of 26.5625 GHz.

124.8.6 Extinction ratio

The extinction ratio of each lane shall be within the limits given in Table 124–6 if measured using a test pattern specified for extinction ratio in Table 124–10. The extinction ratio of a PAM4 optical signal is defined as the ratio of the average optical launch power level P_3, measured over the central 2 UI of a run of 7 threes, and the average optical launch power level P_0, measured over the central 2 UI of a run of 6 zeros, as shown in Figure 124–3.

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124.8.7 Relative intensity noise ($RIN_{21.4}OMA$)

RIN shall be as defined by the measurement methodology of 52.9.6 with the following exceptions:

a) The optical return loss is 21.4 dB.

b) Each lane may be tested individually with the sum of the optical power from all of the lanes not under test being below –30 dBm.

c) The upper –3 dB limit of the measurement apparatus is to be approximately equal to the signaling rate (i.e., 53.2 GHz).

d) The test pattern is according to Table 124–10.

124.8.8 Receiver sensitivity

Receiver sensitivity, which is defined for an input signal with SECQ of 0.9 dB (e.g., an ideal input signal without overshoot), is informative and compliance is not required. If measured, the test signal should have negligible impairments such as intersymbol interference (ISI), rise/fall times, jitter and RIN. Instead, the normative requirement for receivers is stressed receiver sensitivity.

124.8.9 Stressed receiver sensitivity

Stressed receiver sensitivity shall be within the limits given in Table 124–7 if measured using the method defined in 121.8.9 with the following exceptions:

— The SECQ of the stressed receiver conformance test signal is measured according to 124.8.5, except that the test fiber is not used.

— The signaling rate of the test pattern generator and the extinction ratio of the E/O converter are as given in Table 124–6 using test patterns specified in Table 124–10.

— The required values of the “Stressed receiver sensitivity (OMA_{outer}), each lane (max)”, “Stressed eye closure for PAM4 (SECQ), lane under test”, and “OMAouter of each aggressor lane” are as given in Table 124–7.

124.9 Safety, installation, environment, and labeling

124.9.1 General safety

All equipment subject to this clause shall conform to IEC 60950-1.

124.9.2 Laser safety

400GBASE-DR4 optical transceivers shall conform to Hazard Level 1 laser requirements as defined in IEC 60825-1 and IEC 60825-2, under any condition of operation. This includes single fault conditions whether coupled into a fiber or out of an open bore.

Conformance to additional laser safety standards may be required for operation within specific geographic regions.

Laser safety standards and regulations require that the manufacturer of a laser product provide information about the product’s laser, safety features, labeling, use, maintenance, and service. This documentation explicitly defines requirements and usage restrictions on the host system necessary to meet these safety certifications.[12]

[12]A host system that fails to meet the manufacturer’s requirements and/or usage restrictions may emit laser radiation in excess of the safety limits of one or more safety standards. In such a case, the host manufacturer is required to obtain its own laser safety certification.

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124.9.3 Installation

It is recommended that proper installation practices, as defined by applicable local codes and regulation, be followed in every instance in which such practices are applicable.

124.9.4 Environment

Normative specifications in this clause shall be met by a system integrating a 400GBASE-DR4 PMD over the life of the product while the product operates within the manufacturer's range of environmental, power, and other specifications.

It is recommended that manufacturers indicate in the literature associated with the PHY the operating environmental conditions to facilitate selection, installation, and maintenance.

It is recommended that manufacturers indicate, in the literature associated with the components of the optical link, the distance and operating environmental conditions over which the specifications of this clause will be met.

124.9.5 Electromagnetic emission

A system integrating a 400GBASE-DR4 PMD shall comply with applicable local and national codes for the limitation of electromagnetic interference.

124.9.6 Temperature, humidity, and handling

The optical link is expected to operate over a reasonable range of environmental conditions related to temperature, humidity, and physical handling (such as shock and vibration). Specific requirements and values for these parameters are considered to be beyond the scope of this standard.

124.9.7 PMD labeling requirements

It is recommended that each PHY (and supporting documentation) be labeled in a manner visible to the user, with at least the applicable safety warnings and the applicable port type designation (e.g., 400GBASE-DR4).

Labeling requirements for Hazard Level 1 lasers are given in the laser safety standards referenced in 124.9.2.

124.10 Fiber optic cabling model

The fiber optic cabling model is shown in Figure 124–4.

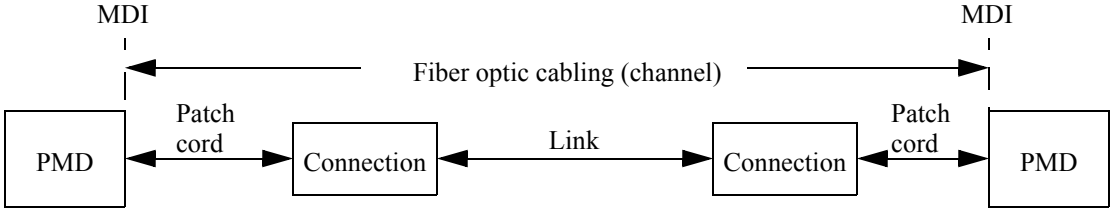


Figure 124–4—Fiber optic cabling model

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The channel insertion loss is given in Table 124–11. A channel may contain additional connectors as long as the optical characteristics of the channel, such as attenuation, dispersion, reflections, and polarization mode dispersion meet the specifications. Insertion loss measurements of installed fiber cables are made in accordance with ANSI/TIA/EIA-526-7/method A-1. The fiber optic cabling model (channel) defined here is the same as a simplex fiber optic link segment. The term channel is used here for consistency with generic cabling standards.

Table 124–11—Fiber optic cabling (channel) characteristics

Description	400GBASE-DR4	Unit
Operating distance (max)	500	m
Channel insertion loss[a,b](max)	3	dB
Channel insertion loss (min)	0	dB
Positive dispersion[b] (max)	0.8	ps/nm
Negative dispersion[b] (min)	–0.93	ps/nm
DGD_max[c]	2.24	ps
Optical return loss (min)	37	dB

[a] These channel insertion loss values include cable, connectors, and splices.
[b] Over the wavelength range 1304.5 nm to 1317.5 nm
[c] Differential Group Delay (DGD) is the time difference at reception between the fractions of a pulse that were transmitted in the two principal states of polarization of an optical signal. DGD_max is the maximum differential group delay that the system must tolerate.

124.11 Characteristics of the fiber optic cabling (channel)

The 400GBASE-DR4 fiber optic cabling shall meet the specifications defined in Table 124–12. The fiber optic cabling consists of one or more sections of fiber optic cable and any intermediate connections required to connect sections together.

124.11.1 Optical fiber cable

The fiber optic cable requirements are satisfied by cables containing IEC 60793-2-50 type B1.1 (dispersion un-shifted single-mode), type B1.3 (low water peak single-mode), or type B6_a (bend insensitive) fibers or the requirements in Table 124–12 where they differ.

Table 124–12—Optical fiber and cable characteristics

Description	Value	Unit
Nominal fiber specification wavelength	1310	nm
Cabled optical fiber attenuation (max)	0.5[a]	dB/km
Zero dispersion wavelength (λ_0)	$1300 \leq \lambda_0 \leq 1324$	nm
Dispersion slope (max) (S_0)	0.093	ps/nm^2 km

[a] The 0.5 dB/km attenuation is provided for Outside Plant cable as defined in ANSI/TIA 568-C.3.

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124.11.2 Optical fiber connection

An optical fiber connection, as shown in Figure 124–4, consists of a mated pair of optical connectors.

124.11.2.1 Connection insertion loss

The maximum link distance is based on an allocation of 2.75 dB total connection and splice loss. For example, this allocation supports five connections with an average insertion loss per connection of 0.5 dB. Connections with different loss characteristics may be used provided the requirements of Table 124–11 are met.

124.11.2.2 Maximum discrete reflectance

The maximum value for each discrete reflectance shall be less than or equal to the value shown in Table 124–13 corresponding to the number of discrete reflectances above –55 dB within the channel. For numbers of discrete reflectances in between two numbers shown in the table, the lower of the two corresponding maximum discrete reflectance values applies.

Table 124–13—Maximum value of each discrete reflectance

Number of discrete reflectances above –55 dB	Maximum value for each discrete reflectance
1	–37 dB
2	–42 dB
4	–45 dB
6	–47 dB
8	–48 dB
10	–49 dB

124.11.3 Medium Dependent Interface (MDI)

The 400GBASE-DR4 PMD is coupled to the fiber optic cabling at the MDI. The MDI is the interface between the PMD and the "fiber optic cabling" (as shown in Figure 124–4). The 400GBASE-DR4 PMD is coupled to the fiber optic cabling through one connector plug into the MDI optical receptacle as shown in Figure 124–6. Example constructions of the MDI include the following:

a) PMD with a connectorized fiber pigtail plugged into an adapter
b) PMD receptacle

124.11.3.1 Optical lane assignments

The four transmit and four receive optical lanes of 400GBASE-DR4 shall occupy the positions depicted in Figure 124–5 when looking into the MDI receptacle with the connector keyway feature on top. The interface contains eight active lanes within twelve total positions. The transmit optical lanes occupy the left-most four positions. The receive optical lanes occupy the right-most four positions. The four center positions are unused.

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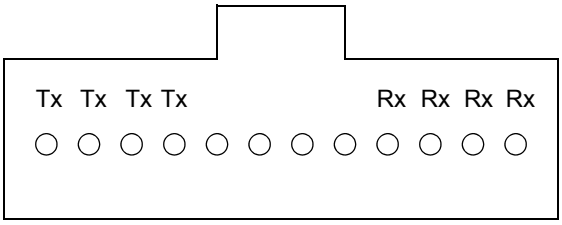


Figure 124–5—400GBASE-DR4 optical lane assignments

124.11.3.2 Medium Dependent Interface (MDI) requirements

The MDI shall meet the dimensional specifications of IEC 61754-7-1 interface 7-1-9: *MPO device receptacle, angled interface*. The plug terminating the optical fiber cabling shall meet the dimensional specifications of IEC 61754-7-1 interface 7-1-1: *MPO female plug connector, down-angled interface for 2 to 12 fibres*. The MDI shall optically mate with the plug on the optical fiber cabling. Figure 124–6 shows an MPO female plug connector with down-angled interface, and an MDI as an active device receptacle with angled interface.

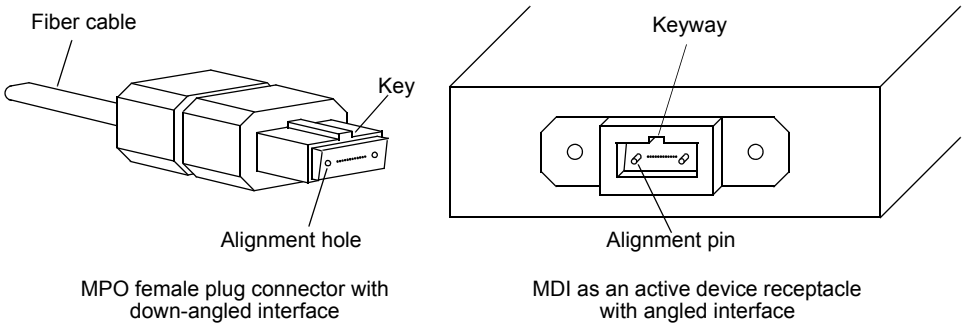


Figure 124–6—MPO female plug with down-angled interface and MDI active device receptacle with angled interface

The MDI shall meet the interface performance specifications of IEC 61753-021-2 for performance level D/2.

NOTE—Transmitter compliance testing is performed at TP2 as defined in 124.5.1, not at the MDI.

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124.12 Protocol implementation conformance statement (PICS) proforma for Clause 124, Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-DR4[13]

124.12.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Clause 124, Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-DR4, shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

124.12.2 Identification

124.12.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

124.12.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Clause 124, Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-DR4
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[13] *Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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124.12.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
*INS	Installation / cable	124.10	Items marked with INS include installation practices and cable specifications not applicable to a PHY manufacturer	O	Yes [ ] No [ ]
TP1	Reference point TP1 exposed and available for testing	124.5.1	This point may be made available for use by implementers to certify component conformance	O	Yes [ ] No [ ]
TP4	Reference point TP4 exposed and available for testing	124.5.1	This point may be made available for use by implementers to certify component conformance	O	Yes [ ] No [ ]
DC	Delay constraints	124.3.1	Device conforms to delay constraints	M	Yes [ ]
SC	Skew constraints	124.3.2	Device conforms to Skew and Skew Variation constraints	M	Yes [ ]
*MD	MDIO capability	124.4	Registers and interface supported	O	Yes [ ] No [ ]

124.12.4 PICS proforma tables for Physical Medium Dependent (PMD) sublayer and medium, type 400GBASE-DR4

124.12.4.1 PMD functional specifications

Item	Feature	Subclause	Value/Comment	Status	Support
F1	Compatible with 400GBASE-R PCS and PMA	124.1		M	Yes [ ]
F2	Integration with management functions	124.1		O	Yes [ ] No [ ]
F3	Bit error ratio	124.1.1	Meets the BER specified in 124.1.1	M	Yes [ ]
F4	Transmit function	124.5.2	Conveys symbols from PMD service interface to MDI	M	Yes [ ]
F5	Mapping between optical signal and logical signal for transmitter	124.5.2	Highest optical power corresponds to tx_symbol = three	M	Yes [ ]
F6	Receive function	124.5.3	Conveys symbols from MDI to PMD service interface	M	Yes [ ]
F7	Conversion of four optical signals to four electrical signals	124.5.3	For delivery to the PMD service interface	M	Yes [ ]
F8	Mapping between optical signal and logical signal for receiver	124.5.3	Highest optical power corresponds to rx_symbol = three	M	Yes [ ]

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Item	Feature	Subclause	Value/Comment	Status	Support
F9	Global Signal Detect function	124.5.4	Report to the PMD service interface the message PMD:IS_SIGNAL.indication (SIGNAL_DETECT)	M	Yes [ ]
F10	Global Signal Detect behavior	124.5.4	SIGNAL_DETECT is a global indicator of the presence of optical signals on all four lanes	M	Yes [ ]
F11	Lane-by-lane Signal Detect function	124.5.5	Sets PMD_signal_detect_*i* values on a lane-by-lane basis per requirements of Table 124–4	MD:O	Yes [ ] No [ ] N/A [ ]
F12	PMD reset function	124.5.6	Resets the PMD sublayer	MD:O	Yes [ ] No [ ] N/A [ ]

124.12.4.2 Management functions

Item	Feature	Subclause	Value/Comment	Status	Support
M1	Management register set	124.4		MD:M	Yes [ ] N/A [ ]
M2	Global transmit disable function	124.5.7	Disables all of the optical transmitters with the PMD_global_transmit_disable variable	MD:O	Yes [ ] No [ ] N/A [ ]
M3	PMD_lane_by_lane_transmit_disable function	124.5.8	Disables the optical transmitter on the lane associated with the PMD_transmit_disable_*i* variable	MD:O	Yes [ ] No [ ] N/A [ ]
M4	PMD_fault function	124.5.9	Sets PMD_fault to one if any local fault is detected	MD:O	Yes [ ] No [ ] N/A [ ]
M5	PMD_transmit_fault function	124.5.10	Sets PMD_transmit_fault to one if a local fault is detected on any transmit lane	MD:O	Yes [ ] No [ ] N/A [ ]
M6	PMD_receive_fault function	124.5.11	Sets PMD_receive_fault to one if a local fault is detected on any receive lane	MD:O	Yes [ ] No [ ] N/A [ ]

124.12.4.3 PMD to MDI optical specifications for 400GBASE-DR4

Item	Feature	Subclause	Value/Comment	Status	Support
DR1	Transmitter meets specifications in Table 124–6	124.7.1	Per definitions in 124.8	M	Yes [ ]
DR2	Receiver meets specifications in Table 124–7	124.7.2	Per definitions in 124.8	M	Yes [ ]

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124.12.4.4 Optical measurement methods

Item	Feature	Subclause	Value/Comment	Status	Support
OM1	Measurement cable	124.8	2 m to 5 m in length	M	Yes [ ]
OM2	Center wavelength	124.8.2	Per TIA/EIA-455-127-A or IEC 61280-1-3 under modulated conditions	M	Yes [ ]
OM3	Average optical power	124.8.3	Per IEC 61280-1-1	M	Yes [ ]
OM4	OMA measurements	124.8.4	Each lane	M	Yes [ ]
OM5	Transmitter and dispersion eye closure for PAM4 (TDECQ)	124.8.5	Each lane	M	Yes [ ]
OM6	Extinction ratio	124.8.6	Each lane	M	Yes [ ]
OM7	Stressed receiver sensitivity	124.8.9	Each lane	M	Yes [ ]

124.12.4.5 Environmental specifications

Item	Feature	Subclause	Value/Comment	Status	Support
ES1	General safety	124.9.1	Conforms to IEC 60950-1	M	Yes [ ]
ES2	Laser safety—IEC Hazard Level 1	124.9.2	Conforms to Hazard Level 1 laser requirements defined in IEC 60825-1 and IEC 60825-2	M	Yes [ ]
ES3	Electromagnetic interference	124.9.5	Complies with applicable local and national codes for the limitation of electromagnetic interference	M	Yes [ ]

124.12.4.6 Characteristics of the fiber optic cabling and MDI

Item	Feature	Subclause	Value/Comment	Status	Support
OC1	Fiber optic cabling	124.11	Meets requirements specified in Table 124–11	INS:M	Yes [ ] N/A [ ]
OC2	Maximum discrete reflectance	124.11.2.2	Meets requirements specified in Table 124–13	INS:M	Yes [ ] N/A [ ]
OC3	MDI layout	124.11.3.1	Optical lane assignments per Figure 124–5	M	Yes [ ]
OC4	MDI dimensions	124.11.3.2	Per IEC 61754-7-1 interface 7-1-9	M	Yes [ ]
OC5	Cabling connector dimensions	124.11.3.2	Per IEC 61754-7-1 interface 7-1-1	INS:M	Yes [ ] N/A [ ]
OC6	MDI mating	124.11.3.2	MDI optically mates with plug on the cabling	M	Yes [ ]

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Annex A

(informative)

Bibliography

Insert the following reference after [B55]:

[B55a] OIF-CEI-04.0, Common Electrical I/O (CEI)—Electrical and Jitter Interoperability agreements for 6G+ bps, 11G+ bps and 25G+ bps I/O.

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NOTE—This annex is numbered in correspondence to its associated clause; i.e., Annex 4A corresponds to Clause 4.

Annex 4A

(normative)

Simplified full duplex media access control

4A.4 Specific implementations

4A.4.2 MAC parameters

Change Note 4 below Table 4A-2 as follows:

NOTE 4—For 40 Gb/s, ~~and~~ 100 Gb/s, 200 Gb/s, and 400 Gb/s operation, the received interpacket gap (the spacing between two packets, from the last bit of the FCS field of the first packet to the first bit of the Preamble of the second packet) can have a minimum value of 8 BT (bit times), as measured at the XLGMII, ~~or~~ CGMII, 200GMII, or 400GMII receive signals at the DTE due to clock tolerance and lane alignment requirements.

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Annex 31B

(normative)

MAC Control PAUSE operation

31B.3 Detailed specification of PAUSE operation

31B.3.7 Timing considerations for PAUSE operation

Insert two new paragraphs in 31B.3.7 immediately after the paragraph starting "At operating speeds of 100 Gb/s," as follows:

At operating speeds of 200 Gb/s, a station shall not begin to transmit a (new) frame more than 453 pause_quanta after the reception of a valid PAUSE frame that contains a non-zero value of pause_time, as measured at the MDI.

At operating speeds of 400 Gb/s, a station shall not begin to transmit a (new) frame more than 905 pause_quanta after the reception of a valid PAUSE frame that contains a non-zero value of pause_time, as measured at the MDI.

Insert two new paragraphs at the end of 31B.3.7 as follows:

200 Gb/s – max_overrun = 28992 + frame_length.

400 Gb/s – max_overrun = 57920 + frame_length.

31B.4 Protocol implementation conformance statement (PICS) proforma for MAC Control PAUSE operation[14]

31B.4.3 Major capabilities/options

Insert rows for items *MIIi and *MIIj at the bottom of the table in 31B.4.3 (as modified by IEEE Std 802.3by-2016) as follows (unchanged rows not shown):

Item	Feature	Subclause	Value/Comment	Status	Support
…					
*MIIi	At operating speeds of 200 Gb/s.	31B.3.7	N/A	O	Yes [ ] No [ ]
*MIIj	At operating speeds of 400 Gb/s.	31B.3.7	N/A	O	Yes [ ] No [ ]

[14] *Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this annex so that it can be used for its intended purpose and may further publish the completed PICS.

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31B.4.6 PAUSE command MAC timing considerations

Insert items TIM10 and TIM11 at the bottom of the table in 31B.4.6 (as modified by IEEE Std 802.3by-2016) as follows (unchanged rows not shown):

Item	Feature	Subclause	Value/Comment	Status	Support
…					
TIM10	Measurement point for station at 200 Gb/s		Delay at MDI ≤ 453 pause_quanta	MIIi: M	N/A [ ] M: Yes [ ]
TIM11	Measurement point for station at 400 Gb/s		Delay at MDI ≤ 905 pause_quanta	MIIj: M	N/A [ ] M: Yes [ ]

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Annex 93A

(normative)

Specification methods for electrical channels

93A.1 Channel Operating Margin

Change the row for "Device package model" in Table 93A–1, insert a new row "Continuous time filter, DC gain 2" after the row for "Continuous time filter, DC gain", and insert a new row "Continuous time filter, low-frequency pole/zero" after the row for "Continuous time filter, pole frequencies", as follows (unchanged rows not shown):

Table 93A–1—COM parameters

Parameter	Reference	Symbol	Units
…			
Device package model Single-ended device capacitance Transmission line length Single-ended package capacitance at package-to-board interface Transmission line characteristic impedance[a]	93A.1.2	 C_d z_p C_p Z_c	 nF mm nF Ω
…			
Continuous time filter, DC gain 2[b] Minimum value Maximum value Step size	93A.1.4.3	g_{DC2}	 dB dB dB
…			
Continuous time filter, low frequency pole/zero[c]	93A.1.4.3	f_{LF}	GHz
…			

[a]Some clauses that invoke this method do not provide a value for Z_c. See 93A.1.2.
[b]Some clauses that invoke this method do not provide a value for g_{DC2}. See 93A.1.4.3.
[c]Some clauses that invoke this method do not provide a value for f_{LF}. See 93A.1.4.3.

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Insert the following rows at the bottom of Table 93A–2 (as modified by IEEE Std 802.3by-2016) as follows (unchanged rows not shown):

Table 93A–2—Physical Layer specifications that employ COM

Physical Layer	Parameter values
…	
200GAUI-8 (Annex 120B)	Table 83D–6[a]
200GAUI-4 (Annex 120D)	Table 120D–8
400GAUI-16 (Annex 120B)	Table 83D–6[a]
400GAUI-8 (Annex 120D)	Table 120D–8

[a]With the exceptions given in 120B.4.

93A.1.2 Transmitter and receiver device package models

93A.1.2.3 Two-port network for the package transmission line

Change the last two sentences of the first paragraph of 93A.1.2.3 as follows:

The values of the parameters that appear in these equations are defined in Table 93A–3. Where a value for Z_c is not provided by the clause that invokes this method, it takes the value 78.2 Ω. The units of f are GHz.

Change Table 93A-3 to remove the row for Z_C as follows:

Table 93A–3—Transmission line model parameters and values

Parameter	Value	Units
γ_0	0	1/mm
a_1	1.734×10^{-3}	$ns^{1/2}$/mm
a_2	1.455×10^{-4}	ns/mm
τ	6.141×10^{-3}	ns/mm
~~Z_e~~	~~78.2~~	~~Ω~~

93A.1.4 Filters

93A.1.4.3 Receiver equalizer

Change the text of 93A.1.4.3 as follows:

$H_{ctf}(f)$ is defined by Equation (93A–22). Where values for g_{DC2} and f_{LF} are not provided by the clause that invokes this method, g_{DC2} takes the value 0 and f_{LF} takes the value 1.

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Replace Equation (93A-22) with the following:

$$H_{ctf}(f) = \frac{\left(10^{\frac{g_{DC}}{20}} + j\frac{f}{f_z}\right)\left(10^{\frac{g_{DC2}}{20}} + j\frac{f}{f_{LF}}\right)}{\left(1 + j\frac{f}{f_{p1}}\right)\left(1 + j\frac{f}{f_{p2}}\right)\left(1 + j\frac{f}{f_{LF}}\right)} \quad \text{(93A–22)}$$

93A.1.6 Determination of variable equalizer parameters

Change the first paragraph of 93A.1.6 and item a) as follows:

COM is a function of the variables $c(-1)$, $c(1)$, ~~g_{DC}~~, and g_{DC2}. The following procedure is used to determine the values of these variables that are used to calculate COM.

a) Compute the pulse response $h^{(k)}(t)$ of each signal path k for a given $c(-1)$, $c(1)$, ~~g_{DC}~~, and g_{DC2} using the procedure defined in 93A.1.5.

Change the last paragraph of 93A.1.6 as follows:

The FOM is calculated for each permitted combination of $c(-1)$, $c(1)$, ~~g_{DC}~~, and g_{DC2} values per Table 93A–1. The combination of values that maximizes the FOM, including the corresponding value of t_s, is used for the calculation of the interference and noise amplitude in 93A.1.7 and the calculation of COM in 93A.1.

93A.1.7 Interference and noise amplitude

Change the first paragraph of 93A.1.7 as follows:

Given the values of $c(-1)$, $c(1)$, ~~g_{DC}~~, g_{DC2}, and t_s derived in 93A.1.6, compute the combined interference and noise distribution $p(y)$ per 93A.1.7.3. The corresponding cumulative distribution function is $P(y)$ as defined by Equation (93A–37).

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Insert Annex 119A and Annex 120A to Annex 120E as follows:

Annex 119A

(informative)

200GBASE-R and 400GBASE-R PCS FEC codeword examples

This annex provides example PCS FEC codewords produced by the 64B/66B to 256B/257B transcoding, scrambling, pre-FEC distribution, and Reed-Solomon encoding defined in Clause 119. This annex presents data in a tabular form. The contents of the tables are hexadecimal representations of the data and are transmitted to the symbol distribution function from left to right within each row starting from the top row and ending at the bottom row. The most significant bit of each hex symbol is transmitted first.

A constant transmission of Idle control characters is sent to the 256B/257B transcoder resulting in a continuous stream of 257-bit blocks, tx_xcoded<0:256>, shown in hex:

tx_xcoded<0:256> = 00700000000000000780000000000000078000000000000007800000000000000

In this example, an alignment marker is due for insertion and the scrambler seed (see Equation (49-1)) just before the 257-bit block was scrambled was:
S<0:57> = 24e6959d0fa5dbd

After scrambling and inserting the alignment marker group am_mapped, the variable tx_scrambled_am is produced as shown in Table 119A–1 for 200GBASE-R and Table 119A–2 for 400GBASE-R. In these tables, italicized characters are alignment markers and bold characters are padding for the alignment markers. Just prior to insertion, the seed of the PRBS9 generator used for padding was P<0:8> = 0x100.

After distributing into two message blocks, and PCS FEC encoding, codewords c_A and c_B are produced. c_A and c_B may be illustrated as two 5440-bit representations cx_A and cx_B as follows:

For all i=0 to 543:
cx_A <(10i+9):(10i)> = c_A<i>
cx_B <(10i+9):(10i)> = c_B<i>

The expanded codewords are shown in Table 119A–3 and Table 119A–4 for 200GBASE-R and Table 119A–5 and Table 119A–6 for 400GBASE-R (bold characters are parity).

The example codewords have tx_am_sf<2:0> = {0,0,0}.

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Table 119A–1—Example tx_scrambled with alignment marker group for 200GBASE-R

Index <*i*:*j*>	tx_scrambled_am<*i*:*j*>
<0:256>	*0b2acab2acab2acab2aca9324c9324c9324c9324c942108c22c90a2697a225314*
<257:513>	*1aa6a9ba6e98a629ab6adab6adab6adab6ad9be6d9b26ddb76e5b46def365bdf1*
<514:770>	*0c42a493a18d4f681a02679e4456dd1c98cfc35417a7b1742db5334a1bd2041b7*
<771:1027>	*07ace48b2fd61333bfb75c3dc524204cebd0955d1b04967a***00846139561bd3720**
<1028:1284>	08ccd177010357e8dab65b6f9160968075341ac958cfc828d8372dc66242646d8
<1285:1541>	026feaa09736e915dd54c5f7e600ecfcbd7d7c207941c5d7182d8b95fe412ed12
<1542:1798>	014cce446415f10cdc7bbafb1cc4b442ef5737afb4f3bed48d240c9095f5fd3a6
<1799:2055>	10219fbe135ede0b73c15238c150007886562e30a50cb27df341954db8ca1c564
<2056:2312>	16598ead3c229838cb9bee7f793139ee1f89fdad9071687cbedf89b4e527a03cd
<2313:2569>	1c2044f260a85ee6eb506c25e7c1196298548b4b3576f1b084c83f2e3cb5fc5f6
<2570:2826>	1c9d40ef27471536fe9eb5e3bc3870d46755080ea6d29f09cf8fa697a738dd3ec
<2827:3083>	193fa8a7675ce03706a49019b93544e19dd66ce78e6a94bebd07eccae7df2076d
<3084:3340>	081919f234c00959e5affc9f83c909805378b5f3f2e4917f3e48b5de4758cea42
<3341:3597>	1d378c193cdd3d0f781abc3529bf76a66b7c73a7230b517d96daff6a39f9ea9b6
<3598:3854>	05e852b6917e9d1410b6509e778629a516fb2bceaba79f12241dbc94cdaffff02
<3855:4111>	1d98f5991744c9fc5393efd6299558cb4f28df3bf3c8636dba505e3a296c7bd2e
<4112:4368>	00faacfbfdbc1c29ced046c73c9aaafe3de8849adad16e864814846bb8cb889ad
<4369:4625>	04d11bfe0e8def5ecfdbe45d1ee4607b4b44df874f8e976dd828fcfd38158a8ff
<4626:4882>	00ee28b635049ba9506024ba13862a1d6bae22d0db5ddb1f4dbe0f80e9edbbcc6
<4883:5139>	1cd0769bc76446da04136e7c6d19905dfb81ac3bfd931426901028d729e529f5a
<5140:5396>	14c3ff33243d7d2b5137b63359fd25b82cde769d0f10d2e60a0386e60820be8f4
<5397:5653>	1993f1fe2d6c60459f26a74bd0265c2959899e4c5b24196afd51db241cfcf9091
<5654:5910>	18ee61ea5a0d584de8ac602618874bba60a907115513bc960f6be32c68f1f2458
<5911:6167>	042251fd6ef16662f5c99d9779b20b865b1401c96034c99550401bcb2792e5638
<6168:6424>	0ce3ae06476b769f33e5577ce4ba6d62071cabe3ea9561d6077fd23909568a25a
<6425:6681>	1bbc46ac3df26a5bb0ea4fdbcbfb42738a84007619c594e24ab396afb5f05fbff
<6682:6938>	076c96a407e106d494aa6b0c5168e1f335b812d7bc530ce8c6fc13f6c54ec2165
<6939:7195>	056cc06afe2bd2a9894c4d1ad9e632f861208ad3894eff0de8b0291c486ea3281
<7196:7452>	094534a897031e6307e32c1906c85e99cd4696fc818d2b5eaf26a576de149ae12
<7453:7709>	085ae9048eddc5401827faa937411ff35c10287236ffdc87e7e7e334b038ee3f9
<7710:7966>	014179e0e2359d3b961a1302fa4b7ac384f0564b15d7504dadbe3c65ee8f6f13a
<7967:8223>	08d40f4b092621cca4919e81d0c15014204622d41814894dbe389c149e5e2257a
<8224:8480>	0e4772c6718da5361abfaad60f5c32d304740de671045cdad6617125f4bbf457d
<8481:8737>	071a7d2b02b41f13971a22fe8a29f0a03c1cec43ca0411d7872f18d12e0a2bd06
<8738:8994>	1344403844b360b4a61968d045941ff821511c2e95455c300eed811d314ad7018

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Table 119A–1—Example tx_scrambled with alignment marker group for 200GBASE-R *(continued)*

Index <*i*:*j*>	tx_scrambled_am<*i*:*j*>
<8995:9251>	0275a5c4a36f6b8b13afafffcdfdbd9b13f00488494886cdeb13b31fc9f7941cf
<9252:9508>	168e51b4882712decd042307b45ebff549e07c92fd3c3dab5a655cc419bba06f1
<9509:9765>	02b88c8d61954691b36009d54637647f750e1b99733303ea76e0805b182d3b9d0
<9766:10022>	1c50996262a56fb5d7631247c2f52b5150c17ba61feb56a37be13f2abc226aa0a
<10023:10279>	1a1f1cf493816e6f69603ff83e895ba5aff2ecb8e9090cb28ae93c232756fedaf

Table 119A–2—Example tx_scrambled with alignment marker group for 400GBASE-R

Index <*i*:*j*>	tx_scrambled_am<*i*:*j*>
<0:256>	*0b2acab2acab2acab2acab2acab2acab2acab2aca9324c9324c9324c9324c9324*
<257:513>	*192649926499264991b421184592144d2f445014d517435064290242ca669aa6a*
<514:770>	*1374dd314c5354d5354dd314c5334dd356d5b56d5b56d5b56d5b56d5b56d5b56d*
<771:1027>	*0b6adab66f9b96c9b76ddb96d1b76d1bf6e1b26cdbd6d9b8b8196f7c588549274*
<1028:1284>	*063530ac615723489c7aa5333d91c8099e79115b747311f9c333ae74adeaefbf9*
<1285:1541>	*1d7e1aa0bd3d8ba16daab83c86f1c6a4c3d959e99a91de9020db9eb3922fa719f*
<1542:1798>	*036fa98f2dab3957b8b0999dfdbae1ee2924183bb157354862e1154102985e84a*
<1799:2055>	*15d1b04967a30d5178ce48f0569849a***00846139561bd37228569fb24b7e4d4cc0**
<2056:2312>	08ccd177010357e8dab65b6f9160968075341ac958cfc828d8372dc66242646d8
<2313:2569>	026feaa09736e915dd54c5f7e600ecfcbd7d7c207941c5d7182d8b95fe412ed12
<2570:2826>	014cce446415f10cdc7bbafb1cc4b442ef5737afb4f3bed48d240c9095f5fd3a6
<2827:3083>	10219fbe135ede0b73c15238c150007886562e30a50cb27df341954db8ca1c564
<3084:3340>	16598ead3c229838cb9bee7f793139ee1f89fdad9071687cbedf89b4e527a03cd
<3341:3597>	1c2044f260a85ee6eb506c25e7c1196298548b4b3576f1b084c83f2e3cb5fc5f6
<3598:3854>	1c9d40ef27471536fe9eb5e3bc3870d46755080ea6d29f09cf8fa697a738dd3ec
<3855:4111>	193fa8a7675ce03706a49019b93544e19dd66ce78e6a94bebd07eccae7df2076d
<4112:4368>	081919f234c00959e5affc9f83c909805378b5f3f2e4917f3e48b5de4758cea42
<4369:4625>	1d378c193cdd3d0f781abc3529bf76a66b7c73a7230b517d96daff6a39f9ea9b6
<4626:4882>	05e852b6917e9d1410b6509e778629a516fb2bceaba79f12241dbc94cdaffff02
<4883:5139>	1d98f5991744c9fc5393efd6299558cb4f28df3bf3c8636dba505e3a296c7bd2e
<5140:5396>	00faacfbfdbc1c29ced046c73c9aaafe3de8849adad16e864814846bb8cb889ad
<5397:5653>	04d11bfe0e8def5ecfdbe45d1ee4607b4b44df874f8e976dd828fcfd38158a8ff
<5654:5910>	00ee28b635049ba9506024ba13862a1d6bae22d0db5ddb1f4dbe0f80e9edbbcc6
<5911:6167>	1cd0769bc76446da04136e7c6d19905dfb81ac3bfd931426901028d729e529f5a
<6168:6424>	14c3ff33243d7d2b5137b63359fd25b82cde769d0f10d2e60a0386e60820be8f4
<6425:6681>	1993f1fe2d6c60459f26a74bd0265c2959899e4c5b24196afd51db241cfcf9091
<6682:6938>	18ee61ea5a0d584de8ac602618874bba60a907115513bc960f6be32c68f1f2458

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Table 119A–2—Example tx_scrambled with alignment marker group for 400GBASE-R *(continued)*

Index <*i*:*j*>	tx_scrambled_am<*i*:*j*>
<6939:7195>	042251fd6ef16662f5c99d9779b20b865b1401c96034c99550401bcb2792e5638
<7196:7452>	0ce3ae06476b769f33e5577ce4ba6d62071cabe3ea9561d6077fd23909568a25a
<7453:7709>	1bbc46ac3df26a5bb0ea4fdbcbfb42738a84007619c594e24ab396afb5f05fbff
<7710:7966>	076c96a407e106d494aa6b0c5168e1f335b812d7bc530ce8c6fc13f6c54ec2165
<7967:8223>	056cc06afe2bd2a9894c4d1ad9e632f861208ad3894eff0de8b0291c486ea3281
<8224:8480>	094534a897031e6307e32c1906c85e99cd4696fc818d2b5eaf26a576de149ae12
<8481:8737>	085ae9048eddc5401827faa937411ff35c10287236ffdc87e7e7e334b038ee3f9
<8738:8994>	014179e0e2359d3b961a1302fa4b7ac384f0564b15d7504dadbe3c65ee8f6f13a
<8995:9251>	08d40f4b092621cca4919e81d0c15014204622d41814894dbe389c149e5e2257a
<9252:9508>	0e4772c6718da5361abfaad60f5c32d304740de671045cdad6617125f4bbf457d
<9509:9765>	071a7d2b02b41f13971a22fe8a29f0a03c1cec43ca0411d7872f18d12e0a2bd06
<9766:10022>	1344403844b360b4a61968d045941ff821511c2e95455c300eed811d314ad7018
<10023:10279>	0275a5c4a36f6b8b13afafffcdfdbd9b13f00488494886cdeb13b31fc9f7941cf

Table 119A–3—RS(544,514) codeword A for 200GBASE-R

Index <*i*:*j*>	cx_A<*i*:*j*>
<5439:5120>	a6a9aa6a9a64992649921486284bd26519565d946d5b56d5b57d936ed8b6cf7efa842ede6024f1db
<5119:4800>	63156bc9a159a5edb273fd0ccbb63b810bd2ea92401c8f6204458108bf4d44c2d1675209d8331cc8
<4799:4480>	40d556ceeaa30193f3af9e2e8b43ea7494144fd4767ae467a1d9f9656c125225f1007dedfce7100a
<4479:4160>	84474c2bbea631da8f34e5832ce9fce47f0add1ce9f2325278204197782b7a662a1b4bf6c844feb4
<4159:3840>	6f8155cb65af1ee61d57ae3ca3e74b6399b457ae3b092274729af1e4a60bd4c4f0adf3018f39fe78
<3839:3520>	00da3a7fd18904fb9b2e9832e1ef0ffd999ce031bebd69c6552f4b672e84f2230d15eacba247b6d9
<3519:3200>	1ff19a27932706b6a8f273e1376c7ada74baaf7fa1c16e72aabdeb2b4424b13a65af62171eb7d9e2
<3199:2880>	e048b2e3a541afc1a83fa290552920391c28ead81b7ede02edced9689a059d98beeb586fe420214e
<2879:2560>	e5065667d74558d5fe83b711e338380822fc991f0613ed2e999a89e24fd5dbbce90b3bb490d6a032
<2559:2240>	a5e8311ee4e0e631e0d129376199937b9d0a36934ca0a4f749870ebb71f25524e35a71be30df7138
<2239:1920>	8b76b5613eddbc9fd3390226172b35df7e8dcd2b10d491acb499e40478b907d1b0876ad58bda458b
<1919:1600>	19c61d4729ec1408494c515223c4fc4c34259fed58eaf5290f439d6710a8e4325f8a0e4e2ffe7f31
<1599:1280>	1c27f3d223b942cbe9af1e46915eda4c7784baf092533e630ba006260a9171d2811ce51a763c36ab
<1279:960>	d79968151cd9d0c7d05f8e75282fa7f4651c04dc053ae8f5d2bd516438360c85805020871150180d
<959:640>	518076b85275b5cfffdbf21124497b336f9ce14e08b69105bcfe8f2a4b872d467eed1e3126189401
<639:320>	62b8970ba6c0ddbda32d70b4c4a9d691ba115b41f0eb50faaaac85abce815bffc45da5dd09734c97
<319:0>	9337f**1a9130b2cf27ea48eb65e77eadceab99c502fbc45e7f255a8ada28cca6d50f528d5740a36f4**

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Table 119A–4—RS(544,514) codeword B for 200GBASE-R

Index <*i*:*j*>	cx_B<*i*:*j*>
<5439:5120>	a6a9aa6a9a6499264992109a2c8a226519565d946d5b56d5b56dbb69dbb66986849163c0bcc51271
<5119:4800>	1fbd07455bc28405e74486bf7c3892732a40d979621aaec9981df56ab7b620d706b398629a748db1
<4799:4480>	fda9025eabfbf70af443e098e7442722e6613211ef31b1695efafb958926d7ecbf9ab2d06a083380
<4479:4160>	6a5434d0b3b2a14d26b88858eefa4fef3c81370bf6f960bc3d3ca155d8360f94644aac36013f1e3f
<4159:3840>	c9dee472fe3d438314215b3212f9cbe5ff26e41c562602aee673559d7cdfbe5b813162523ad3f084
<3839:3520>	d967d24fe7758d784b1ecf30b5614a5bbed394adda56fe7c6da53d11454df70b26faae7cf0494bfe
<3519:3200>	ba26b22e8ff7e65d32c5cfdb10a517de381be41ee71b1b263f421adc2c943b8882c87fdeff3e8a27
<3199:2880>	4b5fbc7fb6fc41cc54ee637b2304baa33ad5a37597fe0bc867e0b8fb632063c261df708c825b169e
<2879:2560>	5affe84aa5bd8d6b4bd9172580a33ba0997e3f6b68a5602d43645a3a60ab091f38e2f0582f603023
<2559:2240>	2ec1255527eb6c627d08bf9a3af66e4da9880301a9a00e4da98e530ddbe67df2ec0bea5283597d48
<2239:1920>	48a3df0e565c3db0b551b81a3491693ebfd49bc05b0aa51b0eed7d6866c67e6e55340c47e552326d
<1919:1600>	f4c121cbfdd14e2574a0a5a0718da66c0cb2c4095ab276dec984a09eed8057c2eb3d40db213f9cd2
<1599:1280>	1dc5083f3534c196903835175d938fbbd1e2b03484649816a0042d12836c3929eea7711ce52fa835
<1279:960>	98710cf510d6a47ba5f5f954023c5845d0fe0fc2208e12c450580046925ab4f44ffd154b875bb971
<959:640>	5a9033a7db8d3d7fb36600e48363233f8a72dd6c1c99b18bd155df097ead54cc1c09d4ac62ab36b9
<639:320>	bb2ef760cc95c100639d850c8fb4ddf1295854bb6bfd87e48782a3ec919d06b07dabfd1d0245150f
<319:0>	6afd6**8ce7558fb5c8911150c62f97b29d912bca53cccf26f707b9af277c6fd62055979796d2b8f85**

Table 119A–5—RS(544,514) codeword A for 400GBASE-R

Index <*i*:*j*>	cx_A<*i*:*j*>
<5439:5120>	a6a9aa6a9aa6a9aa6a9a6499264992649926499264992d886284bd280ba2609026519565d946559565196
<5119:4800>	6d5b56d5b56d5b56d5b57d936ed8b62d876cd9b6077efa842e86550894abde6024f1dbc4598972ea
<4799:4480>	bcd56bc9a11d7d8256a659a5edb273e5f63e326a3bcccbb63b3054625443810bd2ea92c331e3c619
<4479:4160>	401c8f61447e7edcca66077d56afed4815c9afe6d8b691236c7f7a4894aa7ecdcebc10380639d209
<4159:3840>	48b9904d84bbdc645b57bebee96049f5cb27e7acb40a8a0de01aa509342c6cb8534aaea22633bd93
<3839:3520>	3bff204fc2fd8e502d0f4f285d750d83e51d13ab0db043c68f7267b51cbf8f70ecc40896dc8cbe72
<3519:3200>	f95fc9b807959988abb99cd56f5f37efb6ec4e58d4beb0ff21f679f490f9dd635a10c7b3fc2d5a52
<3199:2880>	56cfb4e62bb695b79d1ba97f4050533dd2c9beab1f1c1653ff6e99ac88a3fde99b4fb1f3f6c82945
<2879:2560>	f7be0efb07b9c6cacb8f9096b308258ef228b01ff7afcba189d2c7e71eed3f207f163b58eec0c02e
<2559:2240>	a8eeb169dd65cf86f319b82e36dac8d8e09077dca3009ac4a7569ffe11a96f735ed0f685e96028ce
<2239:1920>	e8265f8cda9a39540a509936869a2a8267ce3abc560bd80f08cbb0415749facb149c42afe68ebe9b
<1919:1600>	135a6e02c0aa5807906ae384cb74f99f6cb700fa5480d25d52d208fbc295570f642c546e2685255a
<1599:1280>	0fbffd27f056f2a147c33b6f561bb1df9b994e0351d95c8f9b7d104470ff3463815d28a958106069
<1279:960>	9b3324b302d6ac9db6b2a10823b86055e0bacc50f6f847e434c7414e0dcd0d206da50e4d75db65e3
<959:640>	2ee470ad0de1092e04a8812b44a3db0e7a77a9dcc7194fea0da61c433f4435891ee97dbe750c8d16
<639:320>	51543380f0881388b114962019a6962d0d13fc4552e1d6ee5cd6a408eaf6237f5fecd980292cd9c8
<319:0>	4fc29**e524ac56d33248be7cbb1eba5f71f5d872e35e9afa1b54db4d1a4fc09e8057f65fbc3e55fd5**

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Table 119A–6—RS(544,514) codeword B for 400GBASE-R

Index <*i*:*j*>	cx_B<*i*:*j*>
<5439:5120>	a6a9aa6a9aa6a9aa6a9a6499264992649926499264992109a2c8a22acac250b426599565d946559765197
<5119:4800>	6d5b56d5b56d5b56d5b59dbb69dbb6fd936bd8766986849163c6a27c7265713cc5127173faef6bef
<4799:4480>	1fbd07455b4f238f0d79e24405e744e62beda55386bf7c3892ee19da30550caa40d9798a89c6a164
<4479:4160>	621aaec9a824ec9203458118bb4f4402c1635209d83314ca401576ceeba30193ffae9eae9b4feb74
<4159:3840>	940447d476fae46ba3d97965681352e5e1087ded3cc7100984c74c2fbea631da8335e5832cedfce4
<3839:3520>	3f1add1fe9b232567a2041a7782a7aa62a1348f6c874feb76f8145cb65af1ed61554ae7ca3ef4b63
<3519:3200>	999453ac3b093278719a71f4ae09d404c0a9f0014f09f279005a3a7fd289c4eb9b2c98f2c1eb0efd
<3199:2880>	599ce431befd79c6542f4b572684f2a32d19ebcb2267bad81ff1ba2f93274686a4f37361176c78da
<2879:2560>	749aab7ca1415e7ea8bd6b3b4026b13a65af6317dea7dde1e0c8b2eba5412fd1a83da29055292039
<2559:2240>	5c08e6db1b3eee0eeeced94892079d98bee35a6f6410294fe5866667d64598e5f680b711c33c3a08
<2239:1920>	e2fc9d1e0693fd2e999ac9c24fd4dbfcc90338b490c6ac33a5a8111ee5e0a631e0d229f75199927b
<1919:1600>	1d0a36924c20b4f749870eab79f355a4e35a72be70ff713a8b76b5693cddfc9fd73802e6072736df
<1599:1280>	7eadc1281094a1a8b599a41478b9075190836bd54bca458b19862d4729ec9418414e51d233c8fc4c
<1279:960>	f41597ee58eaf5250e435d7710abe4b27f820c4eeffe7b331c27f3d623b942cbe1ac1e46815edb4c
<959:640>	b784b2f192531e6308a04606029071929114e71ab62c36a9d71958111ed9d0d7d45d8e353823a7f4
<639:320>	e53c08de057ad8f1d0bd114438370c05b05422875160140e51c056bc5075f5cffbdbf211044d7a33
<319:0>	efbce**079f39fdd0e6bdd2b8ed1cc0790b29c992d54c852ddd29825193c8caee6f675e83eca4495ad**

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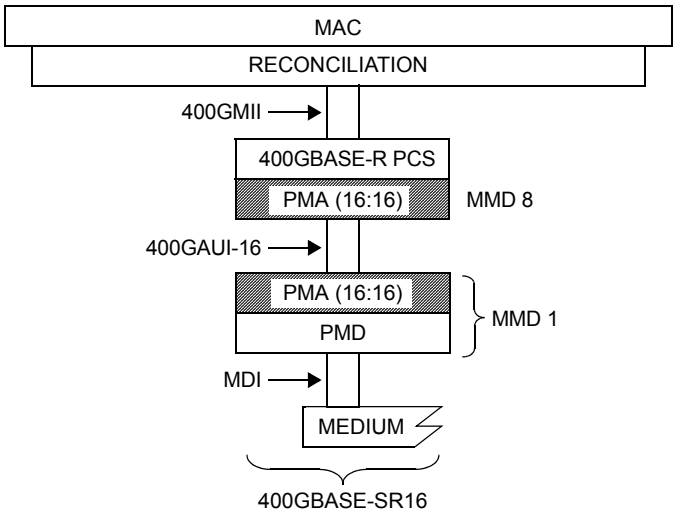
Annex 120A

(informative)

200 Gb/s and 400 Gb/s PMA sublayer partitioning examples

The following subclauses provide various partitioning examples. Partitioning guidelines and MMD numbering conventions are described in 120.1.4.

120A.1 Partitioning example supporting 400GBASE-SR16

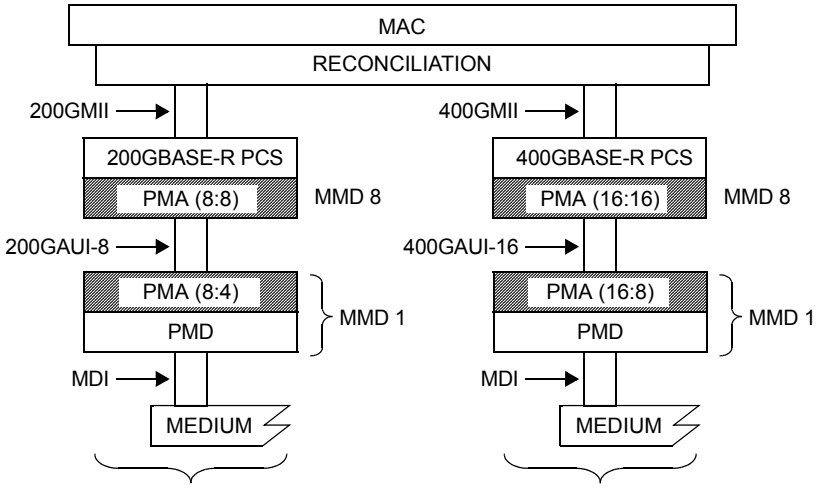


400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120A–1—Example 400GBASE-SR16 PMA layering with single 400GAUI-16 chip-to-module interface

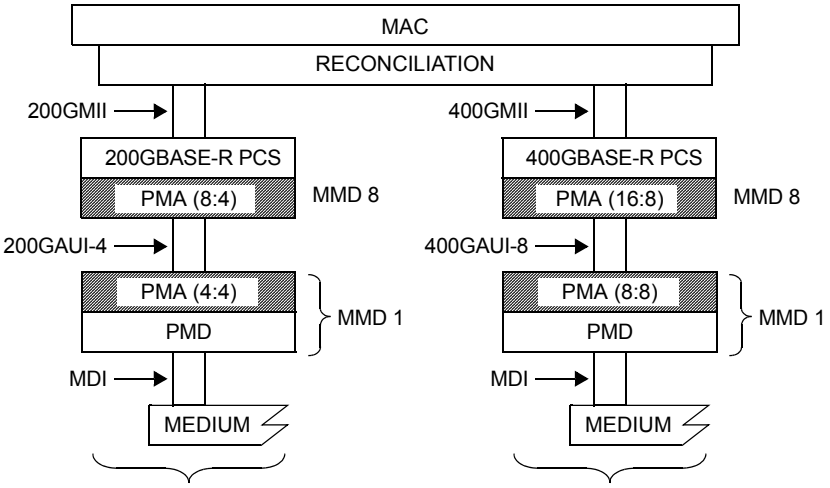
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120A.2 Partitioning examples supporting 200GBASE-DR4/FR4/LR4 and 400GBASE-FR8/LR8



200GAUI = 200 Gb/s ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

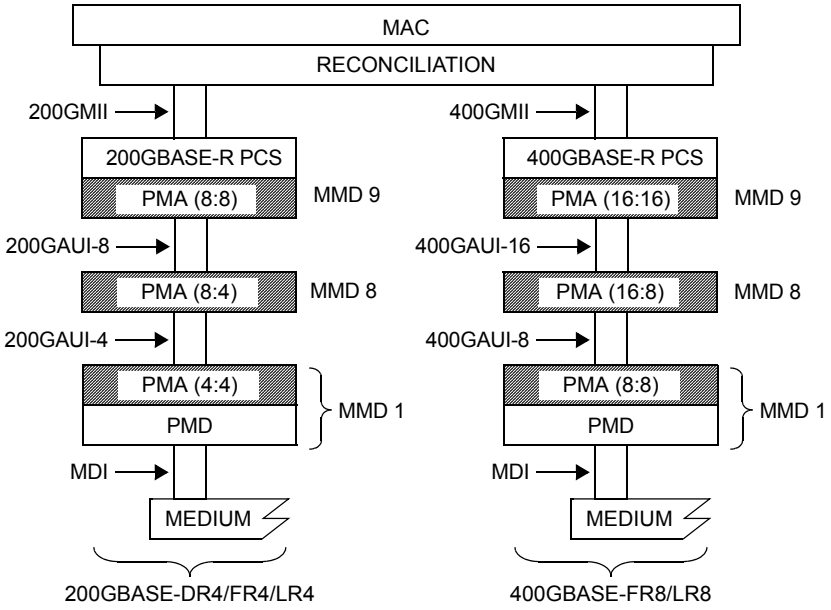
Figure 120A–2—Example 200GBASE-DR4/FR4/LR4 or 400GBASE-FR8/LR8 PMA layering with single 200GAUI-8 or 400GAUI-16 chip-to-module interface



200GAUI = 200 Gb/s ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120A–3—Example 200GBASE-DR4/FR4/LR4 or 400GBASE-FR8/LR8 PMA layering with single 200GAUI-4 or 400GAUI-8 chip-to-module interface

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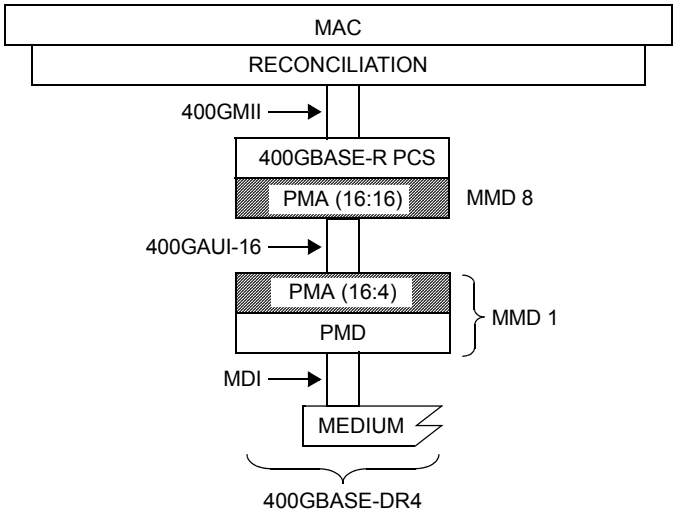


200GAUI = 200 Gb/s ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120A–4—Example 200GBASE-DR4/FR4/LR4 or 400GBASE-FR8/LR8 PMA layering with 200GAUI-8 or 400GAUI-16 chip-to-chip and 200GAUI-4 or 400GAUI-8 chip-to-module interfaces

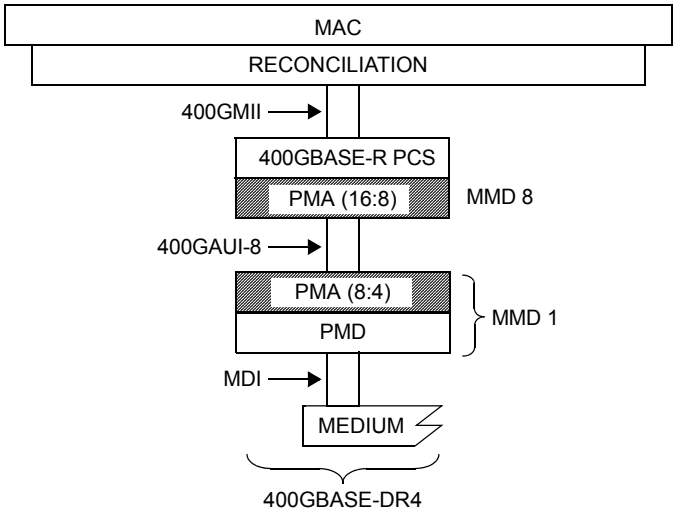
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120A.3 Partitioning examples supporting 400GBASE-DR4



400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120A–5—Example 400GBASE-DR4 PMA layering with single 400GAUI-16 chip-to-module interface

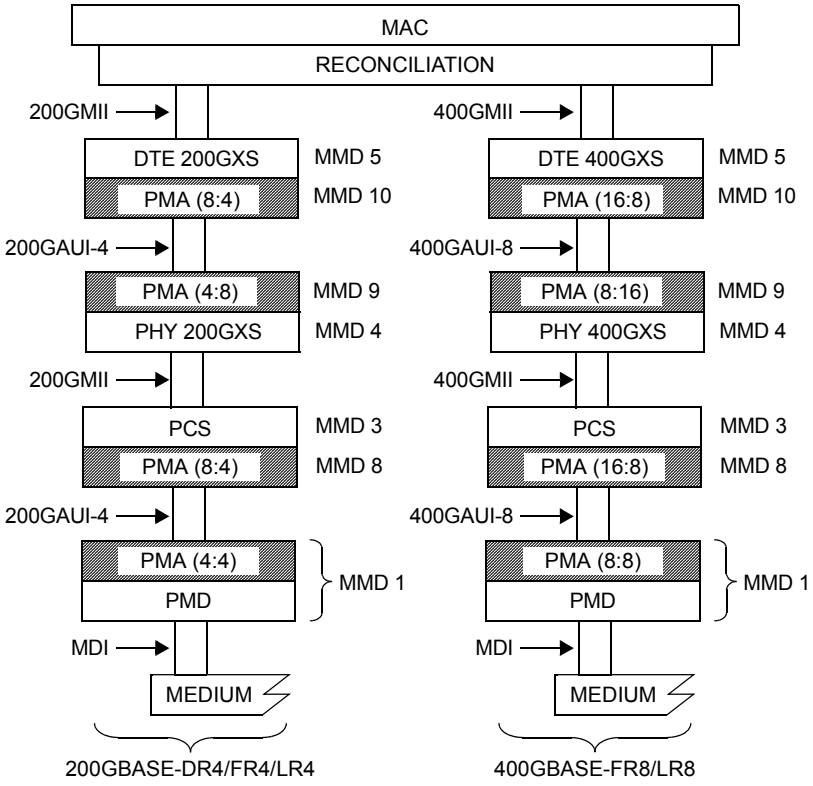


400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120A–6—Example 400GBASE-DR4 PMA layering with single 400GAUI-8 chip-to-module interface

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120A.4 Partitioning example using 200GXS and 400GXS



200GAUI = 200 Gb/s ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
200GXS = 200 Gb/s EXTENDER SUBLAYER
400GAUI = 400 Gb/s ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
400GXS = 400 Gb/s EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
MMD = MDIO MANAGEABLE DEVICE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120A–7—Example 200GBASE-DR4/FR4/LR4 and 400GBASE-FR8/LR8 PMA layering with 200GXS, 400GXS, and two 200GAUI-4, 400GAUI-8 interfaces

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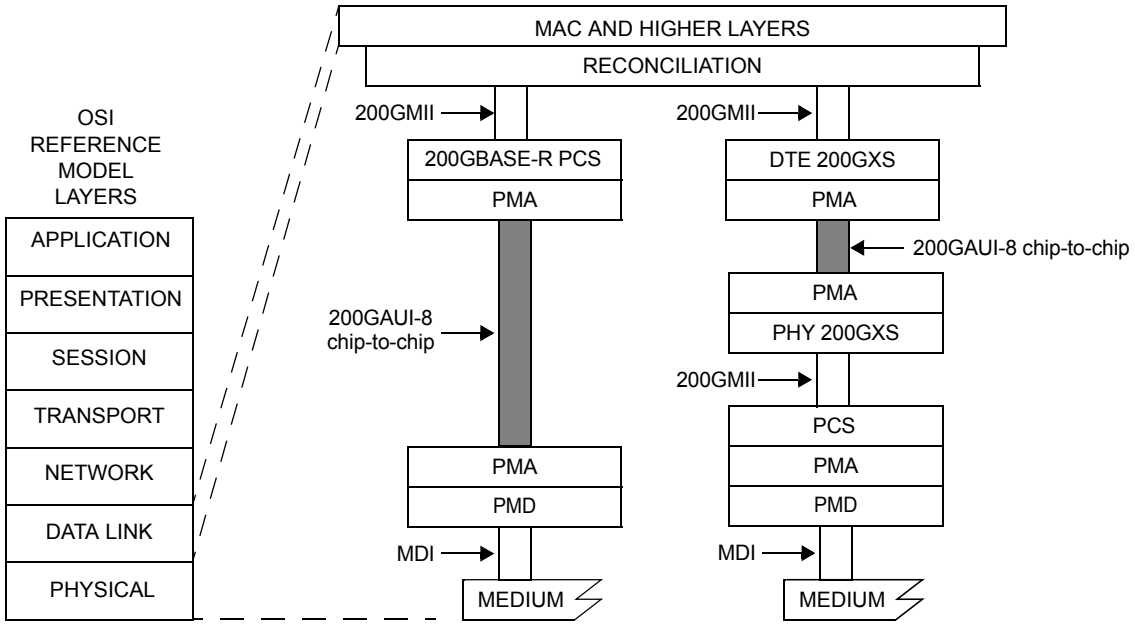
Annex 120B

(normative)

Chip-to-chip 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2C) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2C)

120B.1 Overview

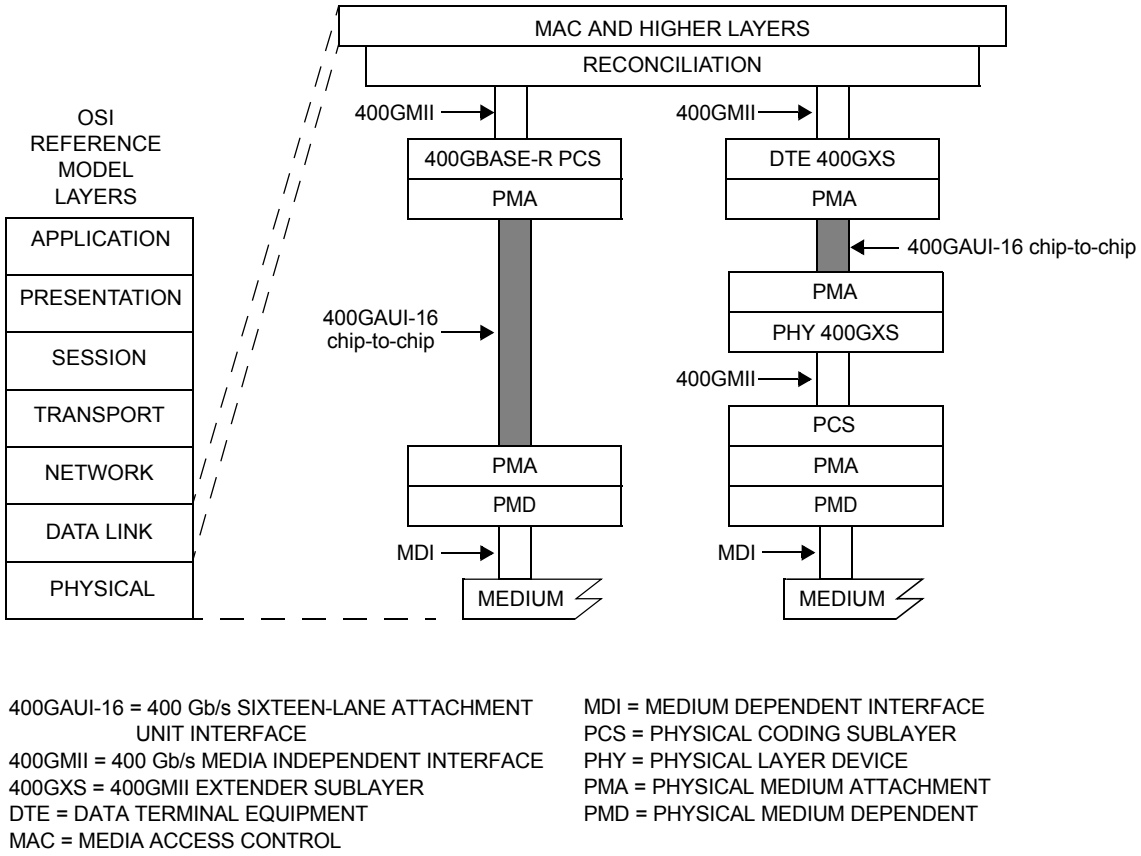
This annex defines the functional and electrical characteristics for the optional chip-to-chip 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2C) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2C). Figure 120B–1 and Figure 120B–2 show example relationships of the 200GAUI-8 and 400GAUI-16 C2C interfaces to the ISO/IEC Open System Interconnection (OSI) reference model, respectively. The 200GAUI-8 and 400GAUI-16 C2C interfaces provide electrical characteristics and associated compliance points, which can optionally be used when designing systems with electrical interconnect of approximately 25 cm in length.



200GAUI-8 = 200 Gb/s EIGHT-LANE ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
200GXS = 200GMII EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120B–1—Example 200GAUI-8 chip-to-chip relationship to the ISO/IEC Open System Interconnection (OSI) reference model and the IEEE 802.3 Ethernet model

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400GAUI-16 = 400 Gb/s SIXTEEN-LANE ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
400GXS = 400GMII EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120B–2—Example 400GAUI-16 chip-to-chip relationship to the ISO/IEC Open System Interconnection (OSI) reference model and the IEEE 802.3 Ethernet model

The 200GAUI-8 C2C link is described in terms of a 200GAUI-8 C2C transmitter, a 200GAUI-8 C2C channel, and a 200GAUI-8 C2C receiver. Figure 120B–3 depicts a typical 200GAUI-8 C2C application, and Equation (83D–1) (illustrated in Figure 83D–3) summarizes the informative differential insertion loss budget associated with the chip-to-chip application. The 200GAUI-8 C2C interface comprises independent data paths in each direction. Each data path contains eight AC-coupled differential lanes using 2-level NRZ (also known as PAM2) signaling, where the highest differential voltage level corresponds to the symbol one and the lowest level corresponds to the symbol zero. The low-frequency 3 dB cutoff of the AC-coupling should be less than 100 kHz.

The 400GAUI-16 C2C link is described in terms of a 400GAUI-16 C2C transmitter, a 400GAUI-16 C2C channel, and a 400GAUI-16 C2C receiver. Figure 120B–4 depicts a typical 400GAUI-16 C2C application, and Equation (83D–1) (illustrated in Figure 83D–3) summarizes the informative differential insertion loss budget associated with the chip-to-chip application. The 400GAUI-16 C2C interface comprises independent data paths in each direction. Each data path contains sixteen AC-coupled differential lanes using 2-level NRZ (also known as PAM2) signaling, where the highest differential voltage level corresponds to the symbol one and the lowest level corresponds to the symbol zero. The low-frequency 3 dB cutoff of the AC-coupling should be less than 100 kHz.

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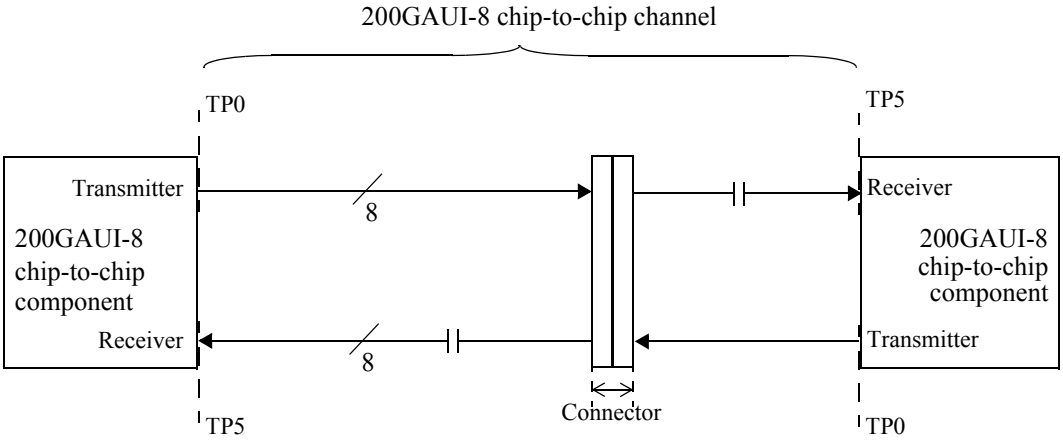


Figure 120B–3—Typical 200GAUI-8 chip-to-chip application

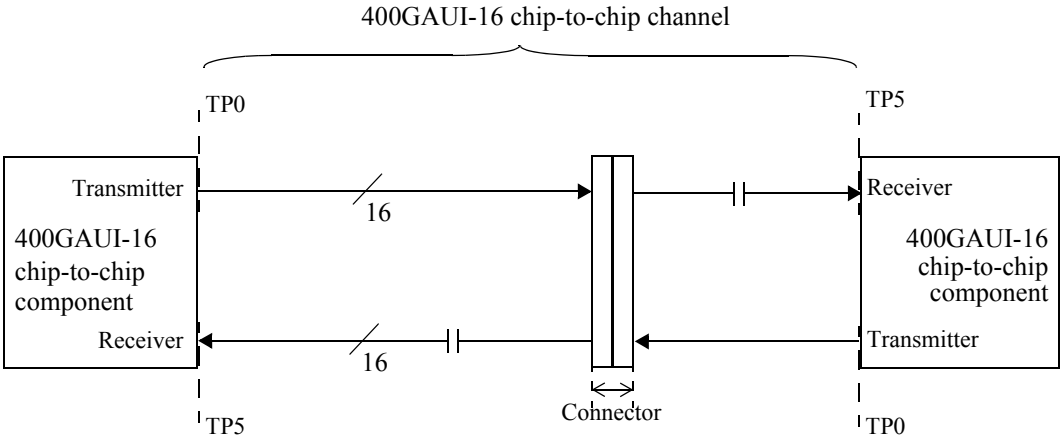


Figure 120B–4—Typical 400GAUI-16 chip-to-chip application

The nominal signaling rate for each lane is 26.5625 GBd. The 200GAUI-8 C2C or 400GAUI-16 C2C transmitter on each end of the link is adjusted to an appropriate setting based on channel knowledge. If implemented, the transmitter equalization feedback mechanism described in 83D.3.3.2 may be used to identify an appropriate setting. The adaptive or adjustable receiver performs the remainder of the equalization.

The normative channel compliance is through 200GAUI-8 C2C or 400GAUI-16 C2C channel operating margin (COM) as described in 120B.4. Actual channel loss could be higher or lower than that given by Equation (83D–1) due to the channel ILD, return loss, and crosstalk.

120B.2 200GAUI-8 and 400GAUI-16 chip-to-chip compliance point definition

The compliance points for the 200GAUI-8 and 400GAUI-16 chip-to-chip interfaces are as defined in 83D.2.

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120B.3 200GAUI-8 and 400GAUI-16 chip-to-chip electrical characteristics

120B.3.1 200GAUI-8 and 400GAUI-16 C2C transmitter characteristics

A 200GAUI-8 or a 400GAUI-16 chip-to-chip transmitter shall meet all specifications in 83D.3.1 with the following exceptions:

- The signaling rate per lane is 26.5625 GBd ± 100 ppm.
- The value for the "Linear fit pulse peak (min)" in Table 83D-1 is $0.75 \times v_f$.
- The value for the "Effective total uncorrelated jitter, peak-to-peak" in Table 83D-1 is 0.19 UI.
- The value of the probability in Table 83D-1 footnote d is 10^{-6}.
- The high-pass filter used for the jitter measurements in 92.8.3.8 has a 3 dB frequency of 4 MHz.

If a Clause 45 MDIO is implemented, *Local_eq_cm1* and *Local_eq_c1* for each lane (0 through 7 for 200GAUI-8 and 0 through 15 for 400GAUI-16) and direction (transmit and receive) are accessible through registers 1.500 through 1.531 (see 45.2.1.116d through 45.2.1.116g).

120B.3.2 200GAUI-8 and 400GAUI-16 C2C receiver characteristics

A 200GAUI-8 or a 400GAUI-16 chip-to-chip receiver shall meet all specifications in 83D.3.3 with the following exceptions:

- The signaling rate per lane is 26.5625 GBd ± 100 ppm.
- The "Bit error ratio" parameter in Table 83D-5 is replaced with "PCS FEC symbol error ratio" with max values of 1.1×10^{-5}.
- The target values for the parameter "COM including effects of broadband noise" in Table 83D-5 are 3 dB.
- The transmitter device package model $S^{(tp)}$ is omitted from Equation (93A-3) in the calculation of COM. The filtered voltage transfer function $H^{(k)}(f)$ calculated in Equation (93A-19) uses the filter $H_t(f)$ defined by Equation (93A-46), where β is 2, T_r is calculated as $T_r = 1.09 \times T_{\mathrm{rm}} - 4.32$ ps, and T_{rm} is the measured 20% to 80% transition time of the signal at TP0a. T_{rm} is measured using the method in 86A.5.3.3, with the exception that the observation filter bandwidth is 33 GHz instead of 12 GHz. T_{rm} is measured with the transmit equalizer turned off (i.e, *Local_eq_cm1* and *Local_eq_c*1 both equal to zero, see 83D.3.1.3).
- Table 83D-5 footnote a is replaced by Table 93-6 footnote a.
- Table 83D-5 footnote b does not apply.
- The "Applied pk-pk sinusoidal jitter" for Test 1 and Test 2 in Table 83D-5 is according to Table 87-13.

If the 200GAUI-8 or 400GAUI-16 chip-to-chip receiver supports the optional transmitter equalization feedback, it shall meet all requirements in 83D.3.3.2. If a Clause 45 MDIO is implemented, the variables *Request_flag*, *Requested_eq_cm1*, *Requested_eq_c1*, *Remote_eq_cm1*, and *Remote_eq_c1* for each lane (0 through 7 for 200GAUI-8 and 0 through 15 for 400GAUI-16) and direction are accessible through registers 1.500 through 1.531 (see 45.2.1.116d through 45.2.1.116g).

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120B.4 200GAUI-8 and 400GAUI-16 chip-to-chip channel characteristics

The 200GAUI-8 or 400GAUI-16 chip-to-chip channel shall meet the requirements specified in 83D.4 with the following exceptions:

— The signaling rate per lane is 26.5625 GBd ± 100 ppm.

— The "Target detector error ratio" (DER_0) value in Table 83D-6 is 10^{-6}.

— The "Normalized DFE coefficient magnitude limit, for n = 1 to N_b" ($b_{max}(n)$) value in Table 83D-6 is 1.

— The Channel Operating Margin (COM) shall be greater than or equal to 3 dB.

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120B.5 Protocol implementation conformance statement (PICS) proforma for Annex 120B, Chip-to-chip 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2C) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2C)[15]

120B.5.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Annex 120B, Chip-to-chip 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2C) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2C), shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

120B.5.2 Identification

120B.5.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

120B.5.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Annex 120B, Chip-to-chip 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2C) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2C)
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

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120B.5.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
NOL	Number of differential AC-coupled lanes	120B.1	Eight independent data paths in each direction for 200GAUI-8 and sixteen independent data paths in each direction for 400GAUI-16	M	Yes [ ]
*CHAN	Channel	120B.4	Items marked with CHAN include channel specifications not applicable to a PMA manufacturer	O	Yes [ ] No [ ]

120B.5.4 PICS proforma tables for Chip-to-chip 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2C) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2C)

120B.5.4.1 Transmitter

Item	Feature	Subclause	Value/Comment	Status	Support
TC1	Signaling rate	120B.3.1	26.5625 GBd ± 100 ppm per lane	M	Yes [ ]
TC2	Peak-to-peak differential output voltage	120B.3.1	1200 mV (max)	M	Yes [ ]
TC3	Peak-to-peak differential output voltage, transmitter disabled	120B.3.1	Less than or equal to 30 mV	M	Yes [ ]
TC4	DC common-mode voltage	120B.3.1	Between 0 V and 1.9 V with respect to signal ground	M	Yes [ ]
TC5	AC common-mode output voltage	120B.3.1	12 mV RMS with respect to signal ground	M	Yes [ ]
TC6	Differential output return loss	120B.3.1	Meets Equation (93–3) constraints	M	Yes [ ]
TC7	Common-mode output return loss	120B.3.1	Meets Equation (93–4) constraints	M	Yes [ ]
TC8	Output waveform	120B.3.1	Meets Table 83D–1 constraints	M	Yes [ ]
TC9	Output jitter	120B.3.1	Meets Table 83D–1 constraints with the exceptions in 120B.3.1	M	Yes [ ]
TC10	Transmit equalization	120B.3.1	Each successive step results in a monotonic change	M	Yes [ ]

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120B.5.4.2 Receiver

Item	Feature	Subclause	Value/Comment	Status	Support
RC1	Differential input return loss	120B.3.2	Meets Equation (93–3) constraints	M	Yes [ ]
RC2	Differential to common-mode input return loss	120B.3.2	Meets Equation (93–5) constraints	M	Yes [ ]
RC3	Interference tolerance	120B.3.2	Satisfy requirements in Table 83D–5 with the exceptions in 120B.3.2	M	Yes [ ]

120B.5.4.3 Channel

Item	Feature	Subclause	Value/Comment	Status	Support
CC1	Channel Operating Margin (COM)	120B.4	Greater than or equal to 3 dB	CHAN:M	Yes [ ]

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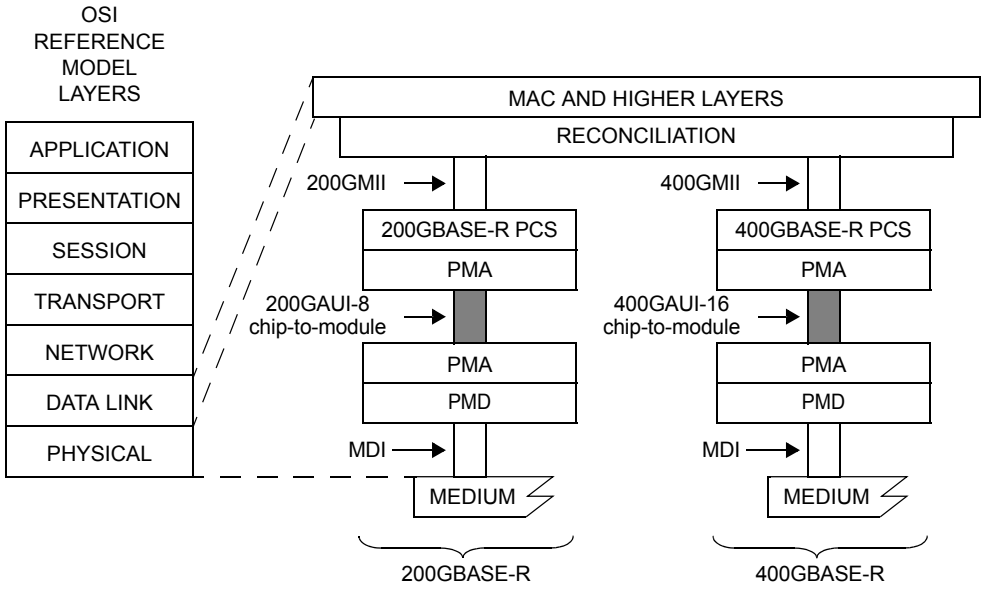
Annex 120C

(normative)

Chip-to-module 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2M) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2M)

120C.1 Overview

This annex defines the functional and electrical characteristics for the optional chip-to-module 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2M) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2M). Figure 120C–1 shows the relationship of the 200GAUI-8 and 400GAUI-16 C2M interfaces to the ISO/IEC Open System Interconnection (OSI) reference model. The chip-to-module interface provides electrical characteristics and associated compliance points, which can optionally be used when designing systems with pluggable module interfaces.



200GAUI-8 = 200 Gb/s EIGHT-LANE ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GAUI-16 = 400 Gb/s SIXTEEN-LANE ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120C–1—Example 200GAUI-8 and 400GAUI-16 chip-to-module relationship to the ISO/IEC Open System Interconnection (OSI) reference model and the IEEE 802.3 Ethernet model

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The sublayers (including the PCS and associated FEC) of each PHY that can optionally include a 200GAUI-8 C2M or 400GAUI-16 C2M are summarized in the tables in 116.1.4 and are specified in the corresponding PMD clause. The positioning of the 200GAUI-8 C2M or 400GAUI-16 C2M relative to other sublayers is shown in 120.1 with further examples in Annex 120A.

The 200GAUI-8 C2M link is described in terms of a host 200GAUI-8 C2M component, a 200GAUI-8 C2M channel with associated insertion loss, and a module 200GAUI-8 C2M component. Figure 120C–2 depicts a typical 200GAUI-8 C2M application and summarizes the differential insertion loss budget associated with the chip-to-module application. The supported insertion loss budget is characterized by Equation (83E–1) and illustrated in Figure 83E–3. The 200GAUI-8 C2M interface comprises independent data paths in each direction. Each data path contains eight differential lanes using 2-level NRZ (also known as PAM2) signaling, where the highest differential voltage level corresponds to the symbol one and the lowest level corresponds to the symbol zero. Each lane is AC-coupled within the module.

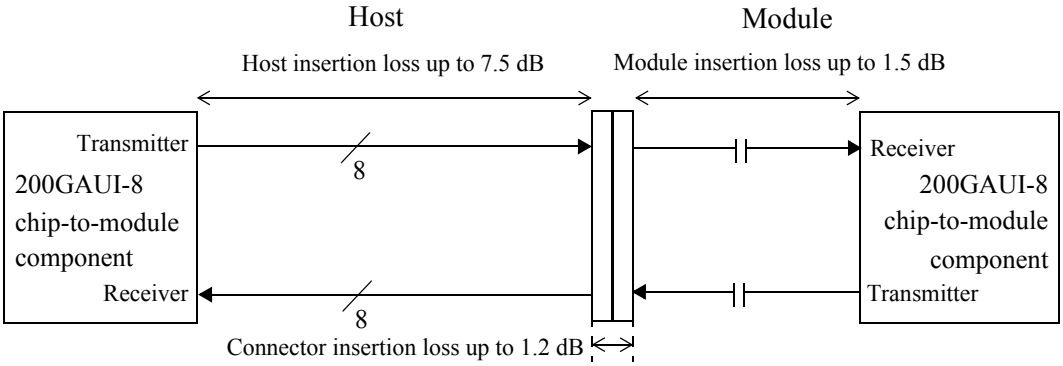


Figure 120C–2—200GAUI-8 chip-to-module insertion loss budget at 13.28 GHz

The 400GAUI-16 C2M link is described in terms of a host 400GAUI-16 C2M component, a 400GAUI-16 C2M channel with associated insertion loss, and a module 400GAUI-16 C2M component. Figure 120C–3 depicts a typical 400GAUI-16 C2M application and summarizes the differential insertion loss budget associated with the chip-to-module application. The supported insertion loss budget is characterized by Equation (83E–1) and illustrated in Figure 83E–3. The 400GAUI-16 C2M interface comprises independent data paths in each direction. Each data path contains sixteen differential lanes using 2-level NRZ (also known as PAM2) signaling, where the highest differential voltage level corresponds to the symbol one and the lowest level corresponds to the symbol zero. Each lane is AC-coupled within the module.

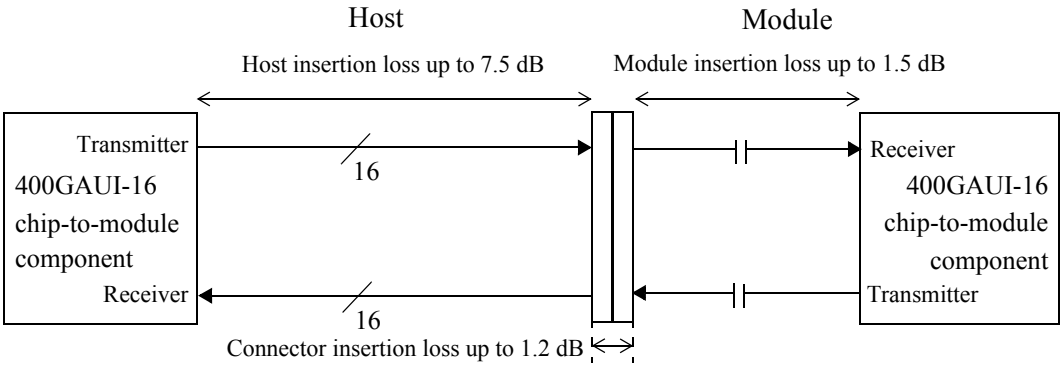


Figure 120C–3—400GAUI-16 chip-to-module insertion loss budget at 13.28 GHz

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The nominal signaling rate for each lane is 26.5625 GBd. The 200GAUI-8 and 400GAUI-16 chip-to-module interfaces are defined using a specification and test methodology that is similar to that used for CEI-28G-VSR defined in OIF-CEI-03.1 [B55].

120C.1.1 Bit error ratio

The bit error ratio (BER) shall be less than 10^{-6}.

120C.2 200GAUI-8 and 400GAUI-16 chip-to-module compliance point definitions

The compliance points for the 200GAUI-8 and 400GAUI-16 chip-to-module interfaces are as defined in 83E.2.

120C.3 200GAUI-8 and 400GAUI-16 chip-to-module electrical characteristics

120C.3.1 200GAUI-8 and 400GAUI-16 C2M host output characteristics

A 200GAUI-8 or 400GAUI-16 C2M host output shall meet all specifications in 83E.3.1 with the following exceptions:

— The signaling rate per lane is 26.5625 GBd ± 100 ppm.

— The clock recovery unit corner frequency is 4 MHz.

If a Clause 45 MDIO is implemented, the variable *Recommended_CTLE_value* is accessible in the module for lanes 0 through 7 for 200GAUI-8 and 0 through 15 for 400GAUI-16 through registers 1.400 to 1.415 (see 45.2.1.116a and 45.2.1.116b).

120C.3.2 200GAUI-8 and 400GAUI-16 C2M module output characteristics

A 200GAUI-8 or a 400GAUI-16 C2M module output shall meet all specifications in 83E.3.2 with the following exceptions:

— The signaling rate per lane is 26.5625 GBd ± 100 ppm.

— The eye height, eye width, and vertical eye closure are as specified in 109B.3.2.1 for a PHY that includes an RS-FEC sublayer.

— The clock recovery unit corner frequency is 4 MHz.

120C.3.3 200GAUI-8 and 400GAUI-16 C2M host input characteristics

A 200GAUI-8 or a 400GAUI-16 C2M host input shall meet all specifications in 83E.3.3 with the following exceptions:

— The signaling rate per lane is 26.5625 GBd ± 100 ppm.

— The BER for the host stressed input test meets the requirement in 120C.1.1 rather than those in 83E.1.1.

— The “Applied pk-pk sinusoidal jitter” in Table 83E-5 is according to Table 87-13.

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120C.3.4 200GAUI-8 and 400GAUI-16 C2M module input characteristics

A 200GAUI-8 or a 400GAUI-16 C2M module input shall meet all specifications in 83E.3.4 with the following exceptions:

— The signaling rate per lane is 26.5625 GBd ± 100 ppm.

— The BER for the module stressed input test meets the requirement in 120C.1.1 rather than those in 83E.1.1.

— The "Applied pk-pk sinusoidal jitter" in Table 83E-8 is according to Table 87-13.

Channel equalization is provided by an equalizer in the module which uses the *Recommended_CTLE_value* provided by the host or an adaptive equalizer in the module which does not use the setting provided by the host.

If a Clause 45 MDIO is implemented, the variable *Recommended_CTLE_value* is accessible in the module for lanes 0 through 7 for 200GAUI-8 and 0 through 15 for 400GAUI-16 through registers 1.400 to 1.415 (see 45.2.1.116a and 45.2.1.116b).

120C.4 200GAUI-8 and 400GAUI-16 C2M measurement methodology

The 200GAUI-8 and 400GAUI-16 C2M measurement methodology is as defined in 83E.4 with the following exceptions:

— The module output eye height, eye width, and vertical eye closure are measured as specified in 109B.3.2.1 for a PHY that includes an RS-FEC sublayer.

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120C.5 Protocol implementation conformance statement (PICS) proforma for Annex 120C, Chip-to-module 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2M) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2M)[16]

120C.5.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Annex 120C, Chip-to-module 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2M) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2M), shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

120C.5.2 Identification

120C.5.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

120C.5.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Annex 120C, Chip-to-module 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2M) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2M)
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

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120C.5.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
NOL	Number of differential AC-coupled lanes	120C.1	Eight independent data paths in each direction for 200GAUI-8 and sixteen independent data paths in each direction for 400GAUI-16	M	Yes [ ]
BER	Meets BER requirement	120C.1.1	See 120C.1.1	M	Yes [ ]
ADE	Adaptive equalizer	120C.3.4	Module receiver does not use *Recommended_CTLE_value*	O	Yes [ ] No [ ]

120C.5.4 PICS proforma tables for Chip-to-module 200 Gb/s eight-lane Attachment Unit Interface (200GAUI-8 C2M) and 400 Gb/s sixteen-lane Attachment Unit Interface (400GAUI-16 C2M)

120C.5.4.1 Host output

Item	Feature	Subclause	Value/Comment	Status	Support
TH1	Signaling rate	120C.3.1	26.5625 GBd ± 100 ppm per lane	M	Yes [ ]
TH2	Peak-to-peak differential output voltage	120C.3.1	900 mV (max)	M	Yes [ ]
TH3	Peak-to-peak differential output voltage, transmitter disabled	120C.3.1	Less than or equal to 35 mV	M	Yes [ ]
TH4	DC common-mode output voltage	120C.3.1	Between –0.3 V and 2.8 V with respect to signal ground	M	Yes [ ]
TH5	AC common-mode output voltage	120C.3.1	17.5 mV RMS with respect to signal ground	M	Yes [ ]
TH6	Differential output return loss	120C.3.1	Meets Equation (83E–2) constraints	M	Yes [ ]
TH7	Reference impedance for output return loss	120C.3.1	100 Ω	M	Yes [ ]
TH8	Common to differential mode conversion	120C.3.1	Meets Equation (83E–3) constraints	M	Yes [ ]
TH9	Differential termination mismatch	120C.3.1	Less than 10%	M	Yes [ ]
TH10	Transition time	120C.3.1	Greater than or equal to 10 ps	M	Yes [ ]
TH11	Eye width	120C.3.1	0.46 UI	M	Yes [ ]
TH12	Eye height A	120C.3.1	95 mV	M	Yes [ ]
TH13	Eye height B	120C.3.1	80 mV	M	Yes [ ]
TH14	Crosstalk source	120C.3.1	Asynchronous crosstalk source using Pattern 5, Pattern 3, or valid 200GBASE-R/400GBASE-R signal	M	Yes [ ]

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120C.5.4.2 Module output

Item	Feature	Subclause	Value/Comment	Status	Support
TM1	Signal rate	120C.3.2	26.5625 GBd ± 100 ppm per lane	M	Yes [ ]
TM2	Peak-to-peak differential output voltage	120C.3.2	900 mV (max)	M	Yes [ ]
TM3	AC common-mode output voltage	120C.3.2	17.5 mV RMS with respect to signal ground	M	Yes [ ]
TM4	Differential output return loss	120C.3.2	Meets Equation (83E–2) constraints	M	Yes [ ]
TM5	Reference impedance for output return loss	120C.3.2	100 Ω	M	Yes [ ]
TM6	Common to differential mode conversion	120C.3.2	Meets Equation (83E–3) constraints	M	Yes [ ]
TM7	Differential termination mismatch	120C.3.2	Less than 10%	M	Yes [ ]
TM8	Transition time	120C.3.2	Greater than or equal to 12 ps	M	Yes [ ]
TM9	Eye width	120C.3.2	0.57 UI	M	Yes [ ]
TM10	Eye height	120C.3.2	228 mV	M	Yes [ ]
TM11	Crosstalk source	120C.3.2	Asynchronous crosstalk source using Pattern 5, Pattern 3, or valid 200GBASE-R/400GBASE-R signal	M	Yes [ ]
TM12	Vertical eye closure	120C.3.2	5.5 dB (max)	M	Yes [ ]

120C.5.4.3 Host input

Item	Feature	Subclause	Value/Comment	Status	Support
RH1	Host input characteristics	120C.3.3	Table 83E–4 with the exceptions in 120C.3.3	M	Yes [ ]

120C.5.4.4 Module input

Item	Feature	Subclause	Value/Comment	Status	Support
RM1	Module input characteristics	120C.3.4	Table 83E–7 with the exceptions in 120C.3.4	M	Yes [ ]
RM2	Module stressed input test BER requirement	120C.3.4	Meet BER requirement of 120C.1.1 with three values of *Recommended_CTLE_value*	M	Yes [ ]

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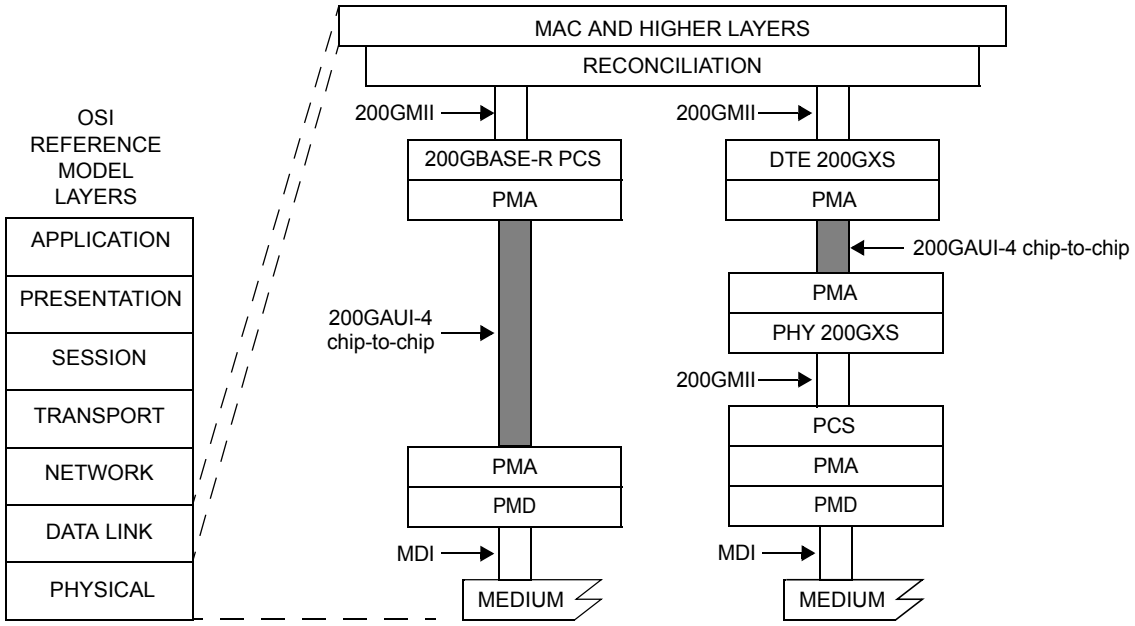
Annex 120D

(normative)

Chip-to-chip 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2C) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2C)

120D.1 Overview

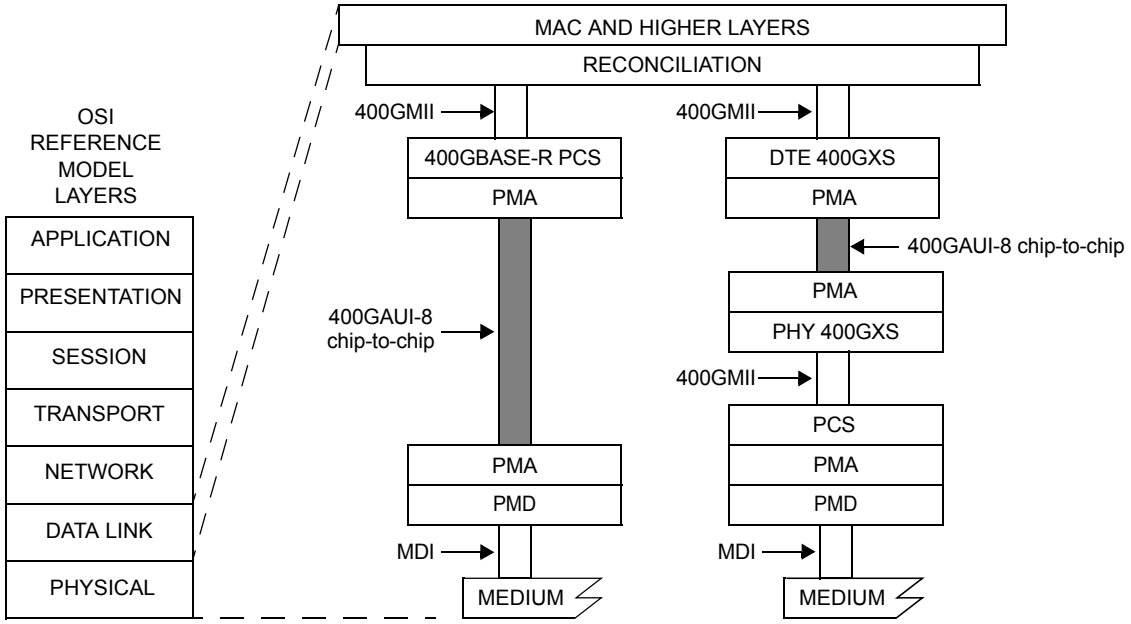
This annex defines the functional and electrical characteristics for the optional chip-to-chip 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2C) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2C). Figure 120D–1 and Figure 120D–2 show example relationships of the 200GAUI-4 and 400GAUI-8 C2C interfaces to the ISO/IEC Open System Interconnection (OSI) reference model, respectively. The 200GAUI-4 and 400GAUI-8 C2C interfaces provide electrical characteristics and associated compliance points, which can optionally be used when designing systems with electrical interconnect of approximately 25 cm in length.



200GAUI-4 = 200 Gb/s FOUR-LANE ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
200GXS = 200GMII EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120D–1—Example 200GAUI-4 chip-to-chip relationship to the ISO/IEC Open System Interconnection (OSI) reference model and the IEEE 802.3 Ethernet model

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400GAUI-8 = 400 Gb/s EIGHT-LANE ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
400GXS = 400GMII EXTENDER SUBLAYER
DTE = DATA TERMINAL EQUIPMENT
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PHY = PHYSICAL LAYER DEVICE
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120D–2—Example 400GAUI-8 chip-to-chip relationship to the ISO/IEC Open System Interconnection (OSI) reference model and the IEEE 802.3 Ethernet model

The 200GAUI-4 C2C link is described in terms of a 200GAUI-4 C2C transmitter, a 200GAUI-4 C2C channel, and a 200GAUI-4 C2C receiver. Figure 120D–3 depicts a typical 200GAUI-4 C2C application, and Equation (120D–1) (illustrated in Figure 120D–5) summarizes the informative differential insertion loss budget associated with the chip-to-chip application. The 200GAUI-4 C2C interface comprises independent data paths in each direction. Each data path contains four AC-coupled differential lanes using PAM4 signaling, where the highest differential level corresponds to the symbol three and the lowest level corresponds to the symbol zero. The low-frequency 3 dB cutoff of the AC-coupling should be less than 100 kHz.

The 400GAUI-8 C2C link is described in terms of a 400GAUI-8 C2C transmitter, a 400GAUI-8 C2C channel, and a 400GAUI-8 C2C receiver. Figure 120D–4 depicts a typical 400GAUI-8 C2C application, and Equation (120D–1) (illustrated in Figure 120D–5) summarizes the informative differential insertion loss budget associated with the chip-to-chip application. The 400GAUI-8 C2C interface comprises independent data paths in each direction. Each data path contains eight AC-coupled differential lanes using PAM4 signaling, where the highest differential level corresponds to the symbol three and the lowest level corresponds to the symbol zero. The low-frequency 3 dB cutoff of the AC-coupling should be less than 100 kHz.

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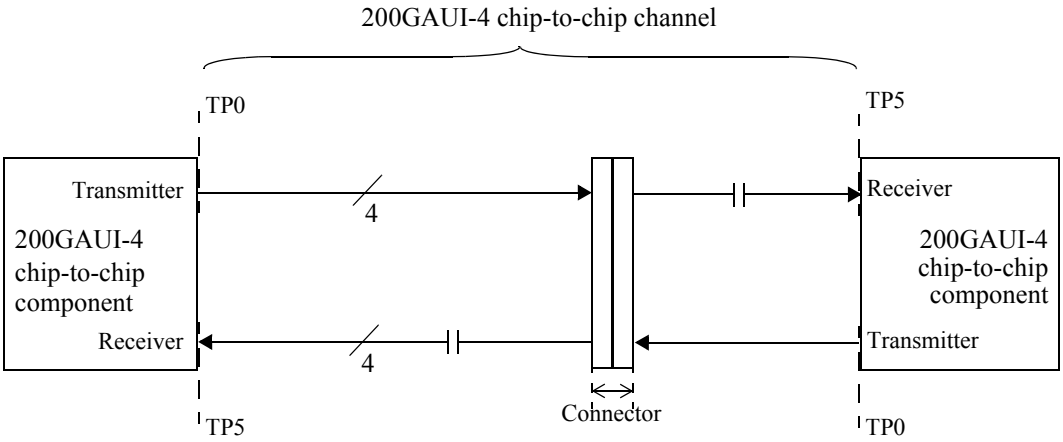


Figure 120D–3—Typical 200GAUI-4 chip-to-chip application

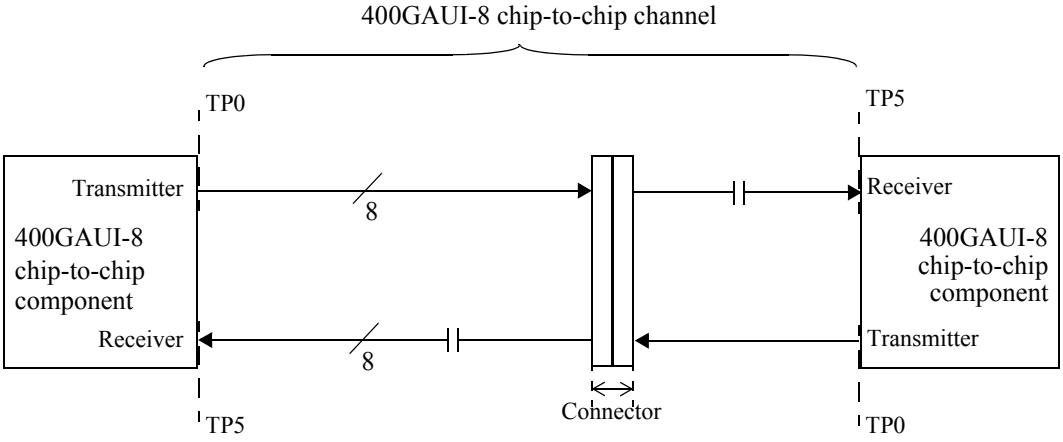


Figure 120D–4—Typical 400GAUI-8 chip-to-chip application

The nominal signaling rate for each lane is 26.5625 GBd. The 200GAUI-4 C2C or 400GAUI-8 C2C transmitter on each end of the link is adjusted to an appropriate setting based on channel knowledge. If implemented, the transmitter equalization feedback mechanism described in 120D.3.2.3 may be used to identify an appropriate setting. The adaptive or adjustable receiver performs the remainder of the equalization.

The channel is normatively defined using channel operating margin (COM) as described in 120D.4. Actual channel loss could be higher or lower than that given by Equation (120D–1) due to the channel ILD, return loss, and crosstalk. Note that for this equation the channel loss at the Nyquist frequency is lower than or equal to 20.457 dB.

$$Insertion_loss(f) \le 1.083 + 2.543\sqrt{f} + 0.761f \quad \text{(dB)}$$
$$\text{for } 0.01 \le f \le 26.5625 \qquad \text{(120D–1)}$$

where

f is the frequency in GHz

Insertion_loss(f) is the informative 200GAUI-4 or 400GAUI-8 chip-to-chip insertion loss

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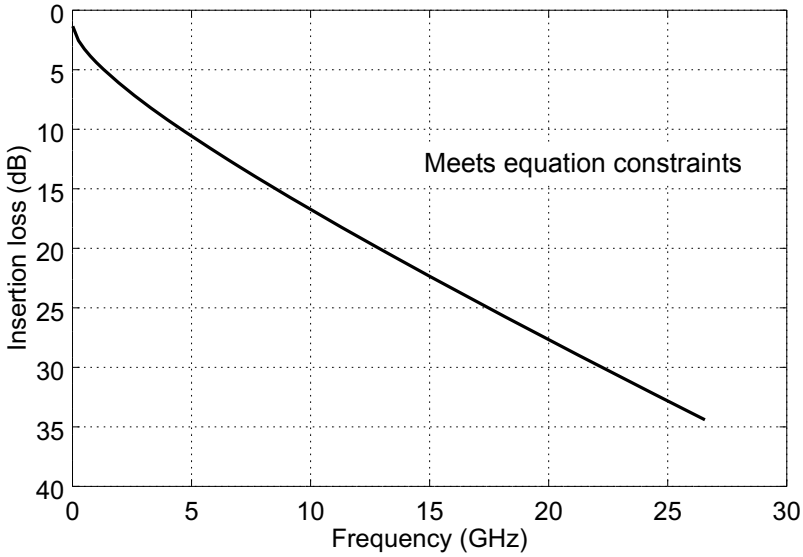


Figure 120D–5—200GAUI-4 or 400GAUI-8 chip-to-chip channel insertion loss

120D.2 200GAUI-4 and 400GAUI-8 chip-to-chip compliance point definition

The electrical characteristics for the 200GAUI-4 and 400GAUI-8 chip-to-chip interfaces are defined at compliance points for the transmitter (TP0a) and receiver (TP5a). The location of TP0a and electrical characteristics of the test fixture used to measure transmitter characteristics are defined in Figure 93-5 and 93.8.1.1, respectively, with the exception that the upper frequency for Equation 93-1 and Equation 93-2 is 26.5625 GHz. The location of TP5a and electrical characteristics of the test fixture used to measure the receiver are defined in Figure 93-10 and 93.8.2.1, respectively, with the exception that the upper frequency for Equation 93-1 and Equation 93-2 is 26.5625 GHz.

120D.3 200GAUI-4 and 400GAUI-8 chip-to-chip electrical characteristics

120D.3.1 200GAUI-4 and 400GAUI-8 C2C transmitter characteristics

A 200GAUI-4 or a 400GAUI-8 chip-to-chip transmitter shall meet the specifications given in Table 120D–1. The transmit output waveform may optionally be manipulated via the feedback mechanism described in 120D.3.2.3.

A test system with a fourth-order Bessel-Thomson low-pass response with 33 GHz 3 dB bandwidth is to be used for all transmitter signal measurements, unless otherwise specified.

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Table 120D–1—200GAUI-4 and 400GAUI-8 C2C transmitter characteristics at TP0a

Parameter	Reference	Value	Units
Signaling rate per lane (range)		26.5625 ± 100 ppm	GBd
Differential peak-to-peak output voltage[a] (max) Transmitter disabled Transmitter enabled	93.8.1.3	 30 1200	 mV mV
Common-mode voltage[a] (max)	93.8.1.3	1.9	V
Common-mode voltage[a] (min)	93.8.1.3	0	V
AC common-mode output voltage[a] (max, RMS)	93.8.1.3	30	mV
Differential output return loss (min)	120D.3.1.1	Equation (120D–2)	dB
Common-mode output return loss (min)	93.8.1.4	Equation 93-4	dB
Output waveform[b] Level separation mismatch ratio R_{LM}(min) Steady state voltage v_f(max) Steady state voltage v_f(min) Linear fit pulse peak (min) Pre-cursor equalization Post-cursor equalization	 120D.3.1.2 120D.3.1.4 120D.3.1.4 120D.3.1.4 120D.3.1.5 120D.3.1.5	 0.95 0.6 0.4 $0.76 \times v_f$ Table 120D–2 Table 120D–3	 — V V V — —
Signal-to-noise-and-distortion ratio SNDR (min)	120D.3.1.6	31.5	dB
Transmitter Output residual ISI SNR_{ISI} (min)	120D.3.1.7	34.8	dB
Output jitter J_{RMS} (max) J4u (max) Even-odd jitter (max)	 120D.3.1.8 120D.3.1.8 120D.3.1.8	 0.023 0.118 0.019	 UI UI UI

[a]Measurement uses the method described in 93.8.1.3 with the exception that the PRBS13Q test pattern is used.
[b]The state of the transmit equalizer is controlled by management interface.

120D.3.1.1 Transmitter differential output return loss

The differential output return loss, in dB, of the transmitter shall meet Equation (120D–2) where f is the frequency in GHz. This requirement applies to all valid output levels. The reference impedance for differential return loss is 100 Ω. The test fixture return loss may be de-embedded from return loss measurements. For the treatment of the insertion loss of the test fixture see 93.8.1.1.

$$RL_d(f) \geq \begin{cases} 14.25 - f & 0.05 \leq f \leq 6 \\ 8.7 - 0.075f & 6 < f \leq 19 \end{cases} \quad \text{dB} \qquad \text{(120D–2)}$$

The differential return loss limit is illustrated by Figure 120D–6.

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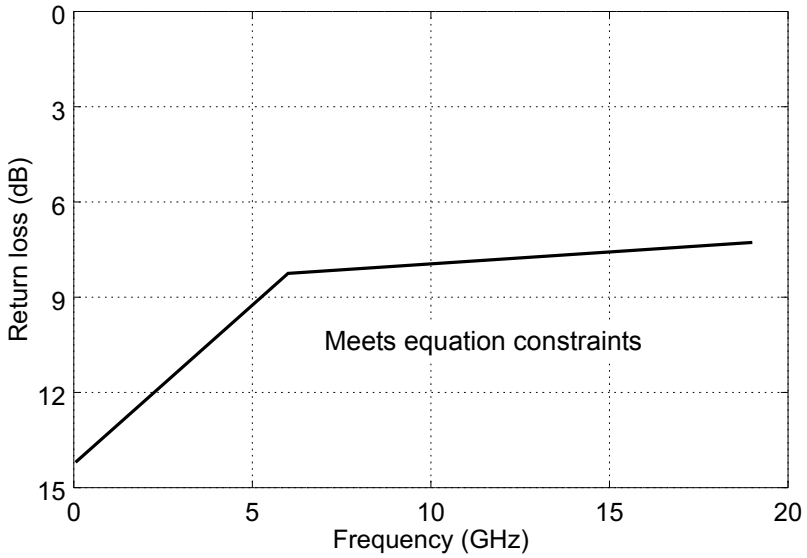


Figure 120D–6—Differential return loss limit

120D.3.1.2 Transmitter linearity

Transmitter linearity is defined as a function of the mean signal level transmitted for each PAM4 symbol level. The mean signal levels of the symbols corresponding to the PAM4 symbol levels 0, 1, 2, and 3 are defined as V_0, V_1, V_2, and V_3 respectively, as defined in 120D.3.1.2.1. The mid-range level V_{mid} is defined by Equation (120D–3). The mean signal levels are then normalized and offset adjusted so that V_{mid} corresponds to 0, V_0 to −1, V_1 to $-ES1$, V_2 to $ES2$, and V_3 to 1. $ES1$ is defined by Equation (120D–4) and $ES2$ is defined by Equation (120D–5).

$$V_{\text{mid}} = \frac{V_0 + V_3}{2} \tag{120D–3}$$

$$ES1 = \frac{V_1 - V_{\text{mid}}}{V_0 - V_{\text{mid}}} \tag{120D–4}$$

$$ES2 = \frac{V_2 - V_{\text{mid}}}{V_3 - V_{\text{mid}}} \tag{120D–5}$$

The level separation mismatch ratio R_{LM} is defined by Equation (120D–6).

$$R_{\text{LM}} = \min((3 \times ES1), (3 \times ES2), (2 - 3 \times ES1), (2 - 3 \times ES2)) \tag{120D–6}$$

120D.3.1.2.1 Measurement of mean signal levels

The signal levels are measured from a waveform captured using the procedure defined in 85.8.3.3.4 while the transmitter is transmitting the PRBS13Q test pattern. The waveform consists of M samples per unit interval and is aligned such that the first M samples of the waveform correspond to the first PAM4 symbol of the test pattern, the second M samples to the second PAM4 symbol, and so on. This allows each sample of the waveform to be associated with a specific PAM4 symbol in the test pattern.

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Denote the number of PAM4 symbols in the test pattern as N. Reduce the captured waveform to N samples by choosing the central sample from each unit interval. The central sample is defined as m^{th} sample in a given unit interval where m is the integer closest to M / 2.

For each PAM4 symbol level x, V_x is the mean value of the waveform samples that correspond to that symbol level.

120D.3.1.3 Linear fit to the measured waveform

The following test procedure shall be followed to determine the linear fit pulse response, linear fit error, and normalized transmitter coefficient values.

For each configuration of the transmit equalizer, capture at least one complete cycle of the PRBS13Q test pattern (120.5.11.2.1) at TP0a per 85.8.3.3.4.

Compute the linear fit pulse response $p(k)$ and linear fit error $e(k)$ from the captured waveform per 85.8.3.3.5 using $N_p = 200$ and $D_p = 2$. The aligned symbols *x(n)* are assigned normalized amplitudes −1, $-ES$, ES, and 1 to represent the PAM4 symbol values 0, 1, 2, and 3 respectively. ES is defined to be $(|ES1| + |ES2|)/2$ where *ES1* and *ES2* are defined in 120D.3.1.2.

Define $r(k)$ to be the linear fit pulse response with *Local_eq_cm1* and *Local_eq_c1* set to zero.

For each configuration of the transmit equalizer, compute the normalized transmit equalizer coefficients, *c(i)*, according to 92.8.3.5.1.

The clock recovery unit (CRU) used in the output waveform measurement has a corner frequency of 4 MHz and a slope of 20 dB/decade

120D.3.1.4 Steady-state voltage and linear fit pulse peak

The linear fit pulse, $p(k)$, is determined according to the linear fit procedure in 120D.3.1.3 with *Local_eq_cm1* and *Local_eq_c1* set to zero. The steady-state voltage v_f is defined to be the sum of the linear fit pulse $p(1)$ through $p(M \times N_v)$ divided by M, determined in step 3 of the linear fit procedure. Here, N_v represents the number of symbols to take into account for the steady-state voltage v_f and has a value of 13.

120D.3.1.5 Transmitter equalization settings

The 200GAUI-4 and 400GAUI-8 chip-to-chip transmitters include programmable equalization to compensate for the frequency-dependent loss of the channel and to facilitate data recovery at the receiver. The functional model for the transmit equalizer is the three tap transversal filter shown in Figure 120D–7. The transmitter output equalization is characterized using the linear fit method described in 120D.3.1.3.

The state of the 200GAUI-4 or 400GAUI-8 C2C transmit output is manipulated via management.

The variable *Local_eq_cm1* controls the weight of the pre-cursor tap $c(-1)$, by changing the ratio $c(-1)/(|c(-1)|+|c(0)|+|c(1)|)$. The valid values of *Local_eq_cm1* and the corresponding ratios are specified in Table 120D–2. The variable *Local_eq_c1* controls the weight of the post-cursor tap $c(1)$, by changing the ratio $c(1)/(|c(-1)|+|c(0)|+|c(1)|)$. The valid values of *Local_eq_c1* and the corresponding ratios are specified in Table 120D–3. *Local_eq_cm1* and *Local_eq_c1* are independent of each other and independent on each lane. Each successive step in *Local_eq_cm1* and *Local_eq_c1* value shall result in a monotonic change in transmitter equalization.

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If a Clause 45 MDIO is implemented, *Local_eq_cm1* and *Local_eq_c1* for each lane (0 through 3 for 200GAUI-4 C2C and 0 through 7 for 400GAUI-8 C2C) and direction (transmit and receive) are accessible through registers 1.500 through 1.507 for receive, and 1.516 through 1.523 for transmit (see 45.2.1.116d through 45.2.1.116g).

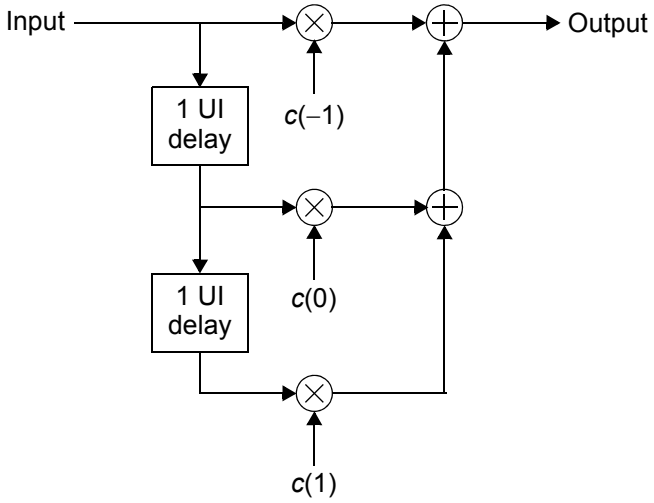


Figure 120D–7—Transmit equalizer functional model

Table 120D–2—Pre-cursor equalization

***Local_eq_cm1* value**	***c*(–1) ratio** $\left(\frac{c(-1)}{\lvert c(-1)\rvert + \lvert c(0)\rvert + \lvert c(1)\rvert}\right)$
0	0 ±0.04
1	–0.05 ±0.04
2	–0.1 ±0.04
3	–0.15 ±0.04

Table 120D–3—Post-cursor equalization

***Local_eq_c1* value**	***c*(1) ratio** $\left(\frac{c(1)}{\lvert c(-1)\rvert + \lvert c(0)\rvert + \lvert c(1)\rvert}\right)$
0	0 ±0.04
1	–0.05 ±0.04
2	–0.1 ±0.04
3	–0.15 ±0.04
4	–0.2 ±0.04
5	–0.25 ±0.04

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120D.3.1.6 Transmitter output noise and distortion

Signal-to-noise and distortion ratio (SNDR) is measured at the transmitter output using the following method, with transmitters on all lanes enabled, with identical transmit equalizer settings, and the lanes not under test transmitting PRBS31Q or a valid 200GBASE-R or 400GBASE-R signal.

Compute the linear fit to the captured waveform and the linear fit pulse response, $p(k)$, and error, $e(k)$, according to 120D.3.1.3. Denote the standard deviation of $e(k)$ as σ_e.

Using the same configuration of the transmitter equalizer, measure the RMS deviation from the mean voltage at a fixed low-slope point in runs of at least 6 consecutive identical PAM4 symbols. PRBS13Q includes such a run for each of the PAM4 levels. The average of the four measurements is denoted as σ_n.

SNDR is defined by Equation (120D–7) where $p_{\max}$ is the maximum value of $p(k)$.

$$SNDR = 10\log_{10}\left(\frac{p_{\max}^2}{\sigma_e^2 + \sigma_n^2}\right) \quad \text{(120D–7)}$$

120D.3.1.7 Transmitter output residual ISI

SNR_{ISI} is defined by Equation (120D–9) computed from $p_{\max}$ and ISI_{cursors} after these have been re-calculated with the continuous time filter described in 93A.1.4.3 using the parameters in Table 120D–8 applied and optimized for maximum SNR_{ISI}. The SNR_{ISI} specification shall be met for all transmit equalization settings.

$$ISI_{\text{cursors}} = [p(t_p + M \times (N_b + 1)), p(t_p + M \times (N_b + 2)), \ldots, p(t_p + M \times (N_p - D_p - 1))] \quad \text{(120D–8)}$$

$$SNR_{\text{ISI}} = 20\log_{10}\left(\frac{p_{\max}}{\sqrt{\sum(ISI_{\text{cursors}}{}^2)}}\right) \quad \text{(120D–9)}$$

ISI_{cursors} are computed from the linear fit pulse response, $p(k)$ in accordance with 120D.3.1.3, using Equation (120D–8), where

t_p is the index of the linear fit pulse where $p(t_p)$ equals $p_{\max}$
M is the oversampling ratio of the measured waveform and linear fit pulse as defined in 85.8.3.3.4
N_p is the linear fit pulse length given in 120D.3.1.3
N_b is given in Table 120D–8

NOTE—The observed SNR_{ISI} can be significantly influenced by the measurement setup, e.g. reflections in cables and connectors. High-precision measurement and careful calibration of the setup are recommended.

120D.3.1.8 Output jitter

J4u, J_{RMS}, and Even-odd jitter are defined by measurements of 12 specific transitions in a PRBS13Q pattern (see 120.5.11.2.1) in order to exclude correlated jitter. The 12 transitions represent all possible combinations of four identical symbols followed by two different identical symbols as shown in Table 120D–4. The sequences are located by the symbol indices given in the table where symbols 1 to 7 are the run of seven 3s.

The threshold used to define each transition is given in Table 120D–4 where V_0, V_1, V_2, and V_3 are as defined in 120D.3.1.2.

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Table 120D–4—PRBS13Q pattern symbols used for jitter measurement

Label	Description	Gray coded PAM4 symbols	Index of first symbol	Index transition begins	Index transition ends	Index of last symbol	Threshold level
REF	Reference for symbol index	3333333	1	—	—	7	—
R03	0 to 3 rise	10000 330	1830	1834	1835	1837	$(V_0+V_3)/2$
F30	3 to 0 fall	23333 001	1269	1273	1274	1276	
R12	1 to 2 rise	0111111 2222221	3638	3644	3645	3651	$(V_1+V_2)/2$
F21	2 to 1 fall	022222 113	1198	1203	1204	1206	
R01	0 to 1 rise	100000 113	6835	6840	6841	6843	$(V_0+V_1)/2$
F10	1 to 0 fall	21111 003	2992	2996	2997	2999	
R23	2 to 3 rise	32222 330	6824	6828	6829	6831	$(V_2+V_3)/2$
F32	3 to 2 fall	033333 2222223	7734	7739	7740	7746	
R02	0 to 2 rise	10000 223	3266	3270	3271	3273	$(V_0+V_2)/2$
F20	2 to 0 fall	122222 0000002	7282	7287	7288	7294	
R13	1 to 3 rise	011111 331	133	138	139	141	$(V_1+V_3)/2$
F31	3 to 1 fall	23333 112	7905	7909	7910	7912	

The jitter is measured with a clock recovery unit (CRU) with a corner frequency of 4 MHz and a slope of 20 dB/decade. Jitter measurements are performed with transmitters on all lanes enabled and using identical transmitter equalizer settings. Transmitters on lanes not under test transmit PRBS31Q, or a valid 200GBASE-R or 400GBASE-R signal. PRBS31Q is described in 120.5.11.2.2.

The J4u, J_{RMS}, and Even-odd jitter specifications shall be met regardless of the transmit equalization setting.

120D.3.1.8.1 J4u and J_{RMS} jitter

For each transition i, $1 \le i \le 12$, of the transitions specified in Table 120D–4, obtain a set $S_i = \{t_i(1), t_i(2), ...\}$ of transition times modulo the period of the pattern. The 12 sets should be of equal size and the size of all sets should be chosen to enable calculation of J4u (as defined below) with sufficient accuracy.

Calculate the average of each set S_i, $T\text{avg}_i$, and subtract it from all elements of that set, to create a set $S0_i = \{t_i(1) - T\text{avg}_i, t_i(2) - T\text{avg}_i, ...\}$.

Combine the sets $S0_i$, i=1 to 12, to create an estimated probability distribution $f_J(t)$.

J4u is defined as the time interval that includes all but 10^{-4} of $f_J(t)$, from the 0.005th to the 99.995th percentile of $f_J(t)$.

J_{RMS} is defined as the standard deviation of $f_J(t)$.

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120D.3.1.8.2 Even-odd Jitter

For one of the 12 specific transitions in PRBS13Q in Table 120D–4:

a) Trigger once in 3 repeats of the PRBS13Q test pattern
   1) Obtain the mean time ($T3$) for this transition in the first PRBS13Q.
   2) Obtain the mean time ($T4$) for the same transition in the second PRBS13Q.
b) The difference between the two means ($T4 - T3$), is the estimated period of the repeating pattern

For each of the 12 specific transitions in PRBS13Q in Table 120D–4:

a1) Trigger once in 2 repeats of the PRBS13Q test pattern.
   1) Obtain the mean time ($T1$) for the specific transition in the first PRBS13Q.
   2) Obtain the mean time ($T2$) for the same transition in the second PRBS13Q.
b1) Calculate even-odd jitter for this transition as $|(T2 - T1) - (T4 - T3)|$

Even-odd jitter EOJ is the maximum of the 12 measurements.

NOTE 1—Both of ($T2 - T1$) and ($T4 - T3$) are about 8191 UI, which is much larger than the EOJ value. Hence, each of $T1$ through $T4$ should have high precision.

NOTE 2—Even-odd jitter has been referred to as *duty cycle distortion* by other Physical Layer specifications for operation over electrical backplane or twinaxial copper cable assemblies (see 72.7.1.9). The term *even-odd jitter* is used here to distinguish it from the duty cycle distortion referred to by Physical Layer specifications for operation over fiber optic cabling.

120D.3.2 200GAUI-4 and 400GAUI-8 C2C receiver characteristics

A 200GAUI-4 or a 400GAUI-8 chip-to-chip receiver shall meet the specifications given in Table 120D–5.

Table 120D–5—200GAUI-4 and 400GAUI-8 C2C receiver characteristics at TP5a

Parameter	Reference	Value	Units
Differential input return loss (min)	120D.3.1.1	Equation (120D–2)	dB
Differential to common mode input return loss	93.8.1.4	Equation (93-5)	dB
Interference tolerance	120D.3.2.1	Table 120D–6	—
Jitter tolerance	120D.3.2.2	Table 120D–7	—

120D.3.2.1 Receiver interference tolerance

Receiver interference tolerance is defined by the procedure in Annex 93C with the exception that transmitter equalization is configured by management (see 120D.3.2.3) to the settings that provide the lowest PCS FEC symbol error ratio. The receiver on each lane shall meet the PCS FEC symbol error ratio requirement with channels matching the Channel Operating Margin (COM) and loss parameters for Test 1 and Test 2 in Table 120D–6. The following additional considerations apply to the interference tolerance test.

a) The test transmitter is constrained such that for any transmitter equalizer setting the differential peak-to-peak voltage (see 93.8.1.3) is less than or equal to 800 mV.
b) The return loss of the test setup in Figure 93C-4 measured at TP5 replica towards TPt meets the requirements of Equation (93-2).
c) The lower frequency bound for the noise spectral density constraints, f_{NSD1}, is 1 GHz.

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d) The transmitter device package model $S^{(tp)}$ is omitted from Equation (93A-3) in the calculation of COM. The filtered voltage transfer function $H^{(k)}(f)$ calculated in Equation (93A-19) uses the filter $H_t(f)$ defined by Equation (93A-46), where β is 2, T_r is calculated as $T_r = 1.09 \times T_{rm} - 4.32$ ps, and T_{rm} is the measured 20% to 80% transition time of the signal at TP0a. T_{rm} is measured using the method in 120E.3.1.5. T_{rm} is measured with *Local_eq_cm1* and *Local_eq_c*1 set to zero (see 120D.3.1.5).

e) For the calculation of test channel COM, the following parameters are based on values measured from the test transmitter. The parameter SNR_{TX} is set to the measured value of SNDR with Np=13, the parameter R_{LM} is set to the measured value of R_{LM}, and the parameters A_{DD} and σ_{RJ} are calculated from the measured values of J4u and J_{RMS} using Equation (120D–10), and Equation (120D–11), respectively.

f) Other COM parameters are set according to the values in Table 120D–8.

g) COM is calculated using both Test 1 and Test 2 device package model transmission line lengths listed in Table 120D–8 on the receiver side. The value of COM is taken as the lower of the two calculated values.

h) The test pattern is the scrambled idle test pattern defined in 119.2.4.9.

i) As an alternative to using the scrambled idle test pattern and measuring FEC symbol error ratio it is permissible to use the PRBS31Q pattern as described in 120.5.11.2.2 and bit error ratio testing. In this case the required bit error ratio is equal to the required FEC symbol error ratio divided by 10. Note that this requirement can be somewhat more stringent than using the scrambled idle test pattern and measuring FEC symbol error ratio, and therefore failing this test requirement with the PRBS31Q pattern does not necessarily imply a failure of the interference tolerance test.

j) A test system with a fourth-order Bessel-Thomson low pass response with 33 GHz 3 dB bandwidth is to be used for measurement of the signal applied by the pattern generator and for measurements of the broadband noise.

$$A_{DD} = \left(\left(\frac{J4u}{2}\right) + Q4\sqrt{\left((Q4^2+1)\times {J_{RMS}}^2 - \left(\frac{J4u}{2}\right)^2\right)}\right)/(Q4^2+1) \quad \text{(120D–10)}$$

$$\sigma_{RJ} = \left(\frac{J4u}{2} - A_{DD}\right)/(Q4) \quad \text{(120D–11)}$$

where $Q4$ is 3.8906.

Table 120D–6—Receiver interference tolerance parameters

Parameter	Test 1 (low loss)			Test 2 (high loss)			Units
	Min	Max	Target	Min	Max	Target	
PCS FEC Symbol error ratio[a]	—	10^{-4}	—	—	10^{-4}	—	—
Insertion loss at 13.2813 GHz[b]	9.5	10.5	—	19.5	20.5	—	dB
RSS_DFE4[c]	0.025	—	—	0.05	—	—	—
COM including effects of broadband noise[d]	—	—	3	—	—	3	dB

[a] The FEC symbol error ratio is measured in step 10 of the receiver interference tolerance method defined in 93C.2
[b] Measured between TPt and TP5 (see Figure 93C-4).
[c] RSS_DFE4 is described in 93A.2.
[d] The COM value is the target for the receiver noise level calibration defined in 93C.2 step 7. The channel noise voltage applied in 93C.2 step 8 should be as close as practical to the value needed to produce the target COM. If higher amplitude values are used, this would demonstrate margin to the specification but this is not required for compliance.

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120D.3.2.2 Receiver jitter tolerance

Receiver jitter tolerance is verified for each pair of jitter frequency and peak-to-peak amplitude values listed in Table 120D–7. The test setup shown in Figure 93-12, or its equivalent, is used. The test channel meets the insertion loss requirement for Test 2 in Table 120D–6. The synthesizer frequency is set to the specified jitter frequency and the synthesizer output amplitude is adjusted until the specified peak-to-peak jitter amplitude for that frequency is measured at TP0a. The test procedure is the same as the one described in 120D.3.2.1, with the following exceptions:

a) No broadband noise is added.

b) The test channel COM, calculated per items d) through g) in 120D.3.2.1, is at least 3 dB.

c) For the COM parameter calibration described in 120D.3.2.1 item e), the test channel transmitter J_{RMS} and J4u values are measured with the jitter frequency and amplitude set according to Case E from Table 120D–7.

d) As an alternative to using the scrambled idle test pattern and measuring FEC symbol error ratio it is permissible to use the PRBS31Q pattern as described in 120.5.11.2.2 and bit error ratio testing. In this case the required bit error ratio is equal to the required FEC symbol error ratio divided by 10. Note that this requirement can be somewhat more stringent than using the scrambled idle test pattern and measuring FEC symbol error ratio, and therefore failing this test requirement with the PRBS31Q pattern does not necessarily imply a failure of the jitter tolerance test.

The receiver under test shall meet the PCS FEC symbol error ratio requirements for each case in Table 120D–7.

Table 120D–7—200GAUI-4 and 400GAUI-8 C2C receiver jitter tolerance parameters

Parameter	Case A	Case B	Case C	Case D	Case E	Units
PCS FEC Symbol error ratio	10^{-4}	10^{-4}	10^{-4}	10^{-4}	10^{-4}	—
Jitter frequency	0.04	1.333	4	12	40	MHz
Jitter amplitude (pk-pk)	5	0.15	0.05	0.05	0.05	UI

120D.3.2.3 Transmitter equalization feedback (optional)

Transmitter equalization feedback is an optional capability for a 200GAUI-4 or a 400GAUI-8 chip-to-chip receiver. If implemented, it shall operate as described in this subclause.

Transmitter equalization feedback is generated for each lane (0 through 3 for 200GAUI-4 C2C and 0 through 7 for 400GAUI-8 C2C) and direction (transmit and receive) independently. The variables that control transmitter equalization feedback are specific for each lane and direction.

A 200GAUI-4 or a 400GAUI-8 chip-to-chip receiver may generate a request to change the transmit equalization coefficients of the remote transmitter to new values by setting the *Request_flag* variable to 1. The variables *Requested_eq_cm1* and *Requested_eq_c1* indicate the requested values of *Local_eq_cm1* and *Local_eq_c1*, respectively, in the remote transmitter (see Table 120D–2 and Table 120D–3). The requested setting may be generated from the remote 200GAUI-4 or 400GAUI-8 chip-to-chip transmitter's equalization setting, which is stored in variables *Remote_eq_cm1* and *Remote_eq_c1*, and from information internal to the receiver, in an implementation specific manner.

When a 200GAUI-4 or a 400GAUI-8 chip-to-chip receiver does not request a change of the remote transmitter's transmit equalization setting, it sets the *Request_flag* variable to 0. A 200GAUI-4 or a

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400GAUI-8 chip-to-chip receiver that does not implement transmitter equalization feedback always sets *Request_flag* to 0.

If a Clause 45 MDIO is implemented, the variables *Request_flag, Requested_eq_cm1, Requested_eq_c1, Remote_eq_cm1,* and *Remote_eq_c1* for each lane and direction are accessible through registers 1.500 through 1.507 for receive, and 1.516 through 1.523 for transmit (see 45.2.1.116d through 45.2.1.116g).

An example of a possible transmitter equalization tuning process using transmitter equalization feedback is provided in 83D.5. Note that 400GAUI-8 C2C uses a different number of lanes and both 200GAUI-4 and 400GAUI-8 use a different register set to those in 83D.5.

120D.4 200GAUI-4 and 400GAUI-8 chip-to-chip channel characteristics

120D.4.1 Channel Operating Margin

The Channel Operating Margin (COM), computed using the procedure in 93A.1, where T_r is 13 ps, β is 2 for $H_t(f)$ in Equation (93A–19), and the parameters in Table 120D–8, shall be greater than or equal to 3 dB. This minimum value allocates margin for practical limitations on the receiver implementation as well as the allowed transmitter equalization coefficients.

Table 120D–8—Channel Operating Margin parameters

Parameter	Symbol	Value	Units
Signaling rate	f_b	26.5625	GBd
Maximum start frequency	f_{min}	0.05	GHz
Maximum frequency step	Δf	0.01	GHz
Device package model Single-ended device capacitance Transmission line length, Test 1 Transmission line length, Test 2 Single-ended package capacitance at package-to-board interface Transmission line characteristic impedance	 C_d z_p z_p C_p Z_c	 1.8×10^{-4} 12 30 1.1×10^{-4} 95	 nF mm mm nF Ω
Single-ended reference resistance	R_o	50	Ω
Single-ended termination resistance	R_d	50	Ω
Receiver 3 dB bandwidth	f_r	$0.75 \times f_b$	GHz
Transmitter equalizer, minimum cursor coefficient	$c(0)$	0.6	—
Transmitter equalizer, pre-cursor coefficient Minimum value Maximum value Step size	$c(-1)$	 –0.15 0 0.05	—
Transmitter equalizer, post-cursor coefficient Minimum value Maximum value Step size	$c(1)$	 –0.25 0 0.05	—

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Table 120D–8—Channel Operating Margin parameters *(continued)*

Parameter	Symbol	Value	Units
Continuous time filter, DC gain Minimum value Maximum value Step size	g_{DC}	 –15 0 1	dB
Continuous time filter, DC gain 2 Minimum value Maximum value Step size	g_{DC2}	 –4 0 1	dB
Continuous time filter, zero frequencies	f_z	f_b/2.5	GHz
Continuous time filter, pole frequencies	f_{p1} f_{p2}	f_b/2.5 $2 \times f_b$	GHz
Continuous time filter, low frequency pole/zero	f_{LF}	f_b/40	GHz
Transmitter differential peak output voltage Victim Far-end aggressor Near-end aggressor	 Av A_{fe} A_{ne}	 0.418 0.418 0.604	 V V V
Number of signal levels	L	4	—
Level separation mismatch ratio	R_{LM}	0.95	
Transmitter signal-to-noise ratio	SNR_{TX}	31	dB
Number of samples per unit interval	M	32	—
Decision feedback equalizer (DFE) length	N_b	10	UI
Normalized DFE coefficient magnitude limit, for $n = 1$ for $n = 2$ to N_b	$b_{max}(n)$	 0.5 0.2	—
Random jitter, RMS	σ_{RJ}	0.01	UI
Dual-Dirac jitter, peak	A_{DD}	0.02	UI
One-sided noise spectral density	η_o	2.6×10^{-8}	V^2/GHz
Target detector error ratio	DER_0	10^{-5}	—

120D.4.2 Channel return loss

The differential return loss of the channel shall meet Equation (120D–12) if the computed value of the Channel Operating Margin (COM) is less than or equal to 4 dB. For COM greater than 4 dB it is recommended that the differential return loss of the channel meet Equation (120D–12).

$$RL_d(f) \geq \begin{cases} 12 & 0.05 \leq f \leq f_b/4 \\ 12 - 15\log_{10}(4f/f_b) & f_b/4 < f \leq f_b \end{cases} \quad \text{dB} \qquad \text{(120D–12)}$$

where

f is the frequency in GHz

f_b is the signaling rate (26.5625) in GHz

$RL(f)$ is the return loss at frequency f

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The differential return loss limit is illustrated in Figure 120D–8.

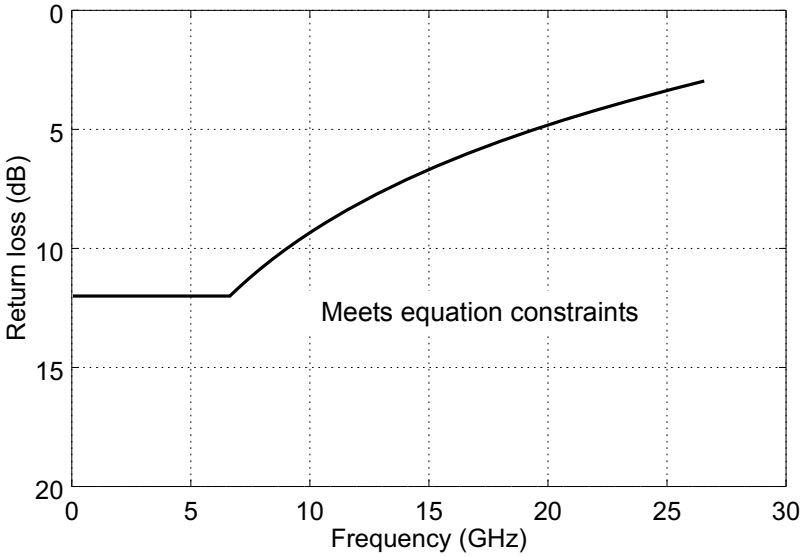


Figure 120D–8—Channel differential return loss limit

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120D.5 Protocol implementation conformance statement (PICS) proforma for Annex 120D, Chip-to-chip 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2C) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2C)[17]

120D.5.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Annex 120D, Chip-to-chip 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2C) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2C), shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

120D.5.2 Identification

120D.5.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

120D.5.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Annex 120D, Chip-to-chip 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2C) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2C)
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[17] *Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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120D.5.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
NOL	Number of differential AC-coupled lanes	120D.1	Four independent data paths in each direction for 200GAUI-4 C2C and eight independent data paths in each direction for 400GAUI-8 C2C	M	Yes [ ]
*CHAN	Channel	120D.4	Items marked with CHAN include channel specifications not applicable to a PMA manufacturer	O	Yes [ ] No [ ]

120D.5.4 PICS proforma tables for Chip-to-chip 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2C) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2C)

120D.5.4.1 Transmitter

Item	Feature	Subclause	Value/Comment	Status	Support
TC1	Signaling rate	120D.3.1	26.5625 GBd ± 100 ppm per lane	M	Yes [ ]
TC2	Peak-to-peak differential output voltage	120D.3.1	1200 mV (max)	M	Yes [ ]
TC3	Peak-to-peak differential output voltage, transmitter disabled	120D.3.1	Less than or equal to 30 mV	M	Yes [ ]
TC4	DC common-mode voltage	120D.3.1	Between 0 V and 1.9 V with respect to signal ground	M	Yes [ ]
TC5	AC common-mode output voltage	120D.3.1	30 mV RMS with respect to signal ground	M	Yes [ ]
TC6	Differential output return loss	120D.3.1	Meets Equation (120D–2) constraints	M	Yes [ ]
TC7	Common-mode output return loss	120D.3.1	Meets Equation (93-4) constraints	M	Yes [ ]
TC8	Output waveform	120D.3.1	Meets Table 120D–1 constraints	M	Yes [ ]
TC9	Output jitter	120D.3.1	Meets Table 120D–1 constraints	M	Yes [ ]
TC10	Transmit equalization	120D.3.1.5	Each successive step results in a monotonic change	M	Yes [ ]
TC11	Transmitter Output residual ISI	120D.3.1	Meets Table 120D–1 constraints	M	Yes [ ]
TC12	Signal-to-noise-and-distortion ratio	120D.3.1	Meets Table 120D–1 constraints	M	Yes [ ]

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120D.5.4.2 Receiver

Item	Feature	Subclause	Value/Comment	Status	Support
RC1	Differential input return loss	120D.3.2	Meets Equation (120D–2) constraints	M	Yes [ ]
RC2	Differential to common mode input return loss	120D.3.2	Meets Equation (93-5) constraints	M	Yes [ ]
RC3	Interference tolerance	120D.3.2.1	Satisfy requirements in Table 120D–6	M	Yes [ ]
RC4	Jitter tolerance	120D.3.2.2	Satisfy requirements in Table 120D–7	M	Yes [ ]

120D.5.4.3 Channel

Item	Feature	Subclause	Value/Comment	Status	Support
CC1	Channel Operating Margin (COM)	120D.4.1	Greater than or equal to 3 dB	CHAN: M	Yes [ ]
CC2	Return loss for channels with COM less than or equal to 4 dB	120D.4.2	Meets Equation (120D–12) constraints	CHAN: M	Yes [ ]

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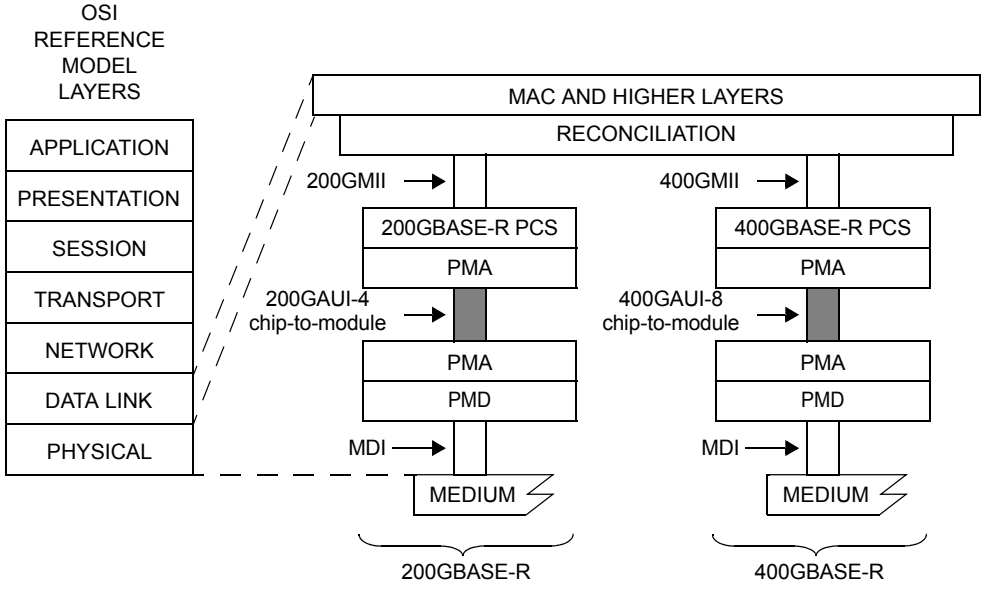
Annex 120E

(normative)

Chip-to-module 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2M) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2M)

120E.1 Overview

This annex defines the functional and electrical characteristics for the optional chip-to-module 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2M) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2M). Figure 120E–1 shows the relationship of the 200GAUI-4 and 400GAUI-8 C2M interfaces to the ISO/IEC Open System Interconnection (OSI) reference model. The chip-to-module interface provides electrical characteristics and associated compliance points, which can optionally be used when designing systems with pluggable module interfaces.



200GAUI-4 = 200 Gb/s FOUR-LANE ATTACHMENT UNIT INTERFACE
200GMII = 200 Gb/s MEDIA INDEPENDENT INTERFACE
400GAUI-8 = 400 Gb/s EIGHT-LANE ATTACHMENT UNIT INTERFACE
400GMII = 400 Gb/s MEDIA INDEPENDENT INTERFACE
MAC = MEDIA ACCESS CONTROL
MDI = MEDIUM DEPENDENT INTERFACE
PCS = PHYSICAL CODING SUBLAYER
PMA = PHYSICAL MEDIUM ATTACHMENT
PMD = PHYSICAL MEDIUM DEPENDENT

Figure 120E–1—Example 200GAUI-4 and 400GAUI-8 chip-to-module relationship to the ISO/IEC Open System Interconnection (OSI) reference model and the IEEE 802.3 Ethernet model

The sublayers (including the PCS and associated FEC) for each PHY that can optionally include a 200GAUI-4 C2M or 400GAUI-8 C2M are summarized in the tables in 116.1.4 and are specified in the corresponding PMD clause. The positioning of the 200GAUI-4 C2M or 400GAUI-8 C2M relative to other sublayers is shown in 120.1 with further examples in Annex 120A.

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The 200GAUI-4 C2M link is described in terms of a host 200GAUI-4 C2M component, a 200GAUI-4 C2M channel with associated insertion loss, and a module 200GAUI-4 C2M component. Figure 120E–2 depicts a typical 200GAUI-4 C2M application and summarizes the differential insertion loss budget associated with the chip-to-module application. The supported insertion loss budget is characterized by Equation (120E–1) and illustrated in Figure 120E–4. The 200GAUI-4 C2M interface comprises independent data paths in each direction. Each data path contains four differential lanes using PAM4 signaling, where the highest differential level corresponds to the symbol three and the lowest level corresponds to the symbol zero. Each lane is AC-coupled within the module.

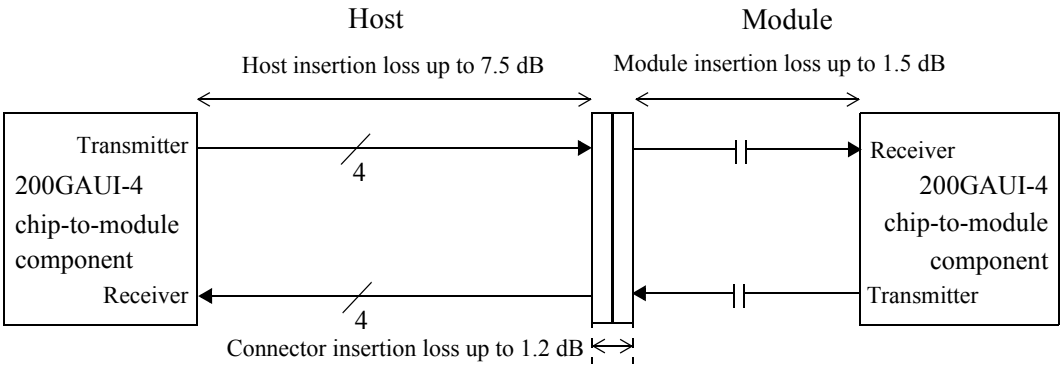


Figure 120E–2—200GAUI-4 chip-to-module insertion loss budget at 13.28 GHz

The 400GAUI-8 C2M link is described in terms of a host 400GAUI-8 C2M component, a 400GAUI-8 C2M channel with associated insertion loss, and a module 400GAUI-8 C2M component. Figure 120E–3 depicts a typical 400GAUI-8 C2M application and summarizes the differential insertion loss budget associated with the chip-to-module application. The recommended insertion loss budget is characterized by Equation (120E–1) and illustrated in Figure 120E–4. The 400GAUI-8 C2M interface comprises independent data paths in each direction. Each data path contains eight differential lanes using PAM4 signaling, where the highest differential level corresponds to the symbol three and the lowest level corresponds to the symbol zero. Each lane is AC-coupled within the module.

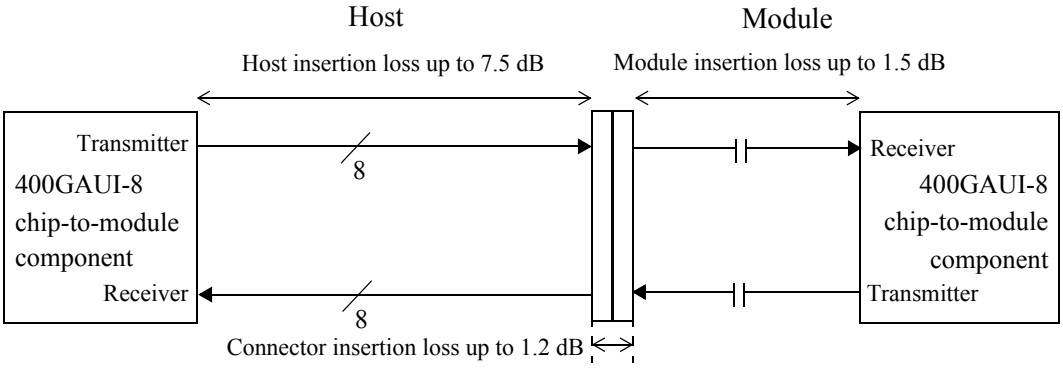


Figure 120E–3—400GAUI-8 chip-to-module insertion loss budget at 13.28 GHz

The nominal signaling rate for each lane is 26.5625 GBd. The chip-to-module interface is defined using a specification and test methodology that is similar to that used for CEI-56G-VSR-PAM defined in OIF-CEI-04.0 [B55a].

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$$Insertion_loss(f) \le \left\{ \begin{array}{cc} 0.05 + 1.8\sqrt{f} + 0.2705f & 0.01 \le f \le 13.28 \\ -4.0096 + 1.07f & 13.28 < f \le 26.5625 \end{array} \right\} (\text{dB}) \qquad (120E–1)$$

where

f is the frequency in GHz

Insertion_loss(f) is the 200GAUI-4 or 400GAUI-8 chip-to-module insertion loss

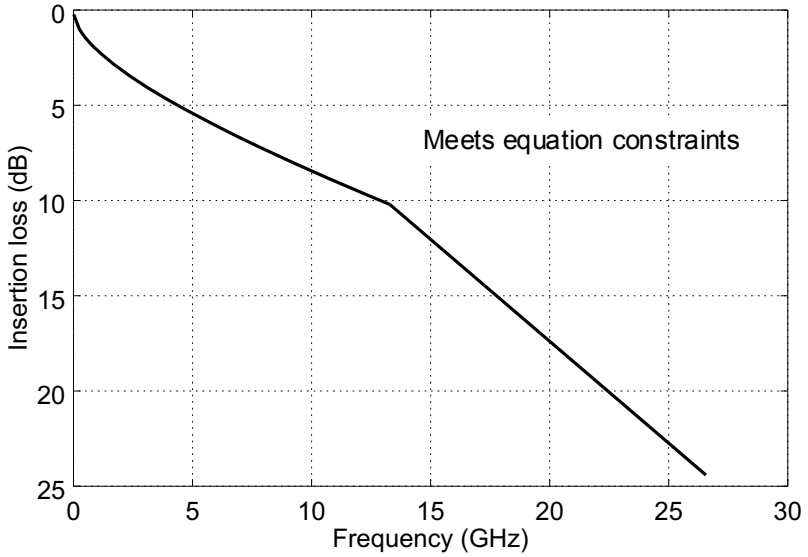


Figure 120E–4—Recommended 200GAUI-4 or 400GAUI-8 chip-to-module channel insertion loss

120E.1.1 Bit error ratio

The bit error ratio (BER) when processed according to Clause 120 shall be less than 10^{-5}.

120E.2 200GAUI-4 and 400GAUI-8 chip-to-module compliance point definitions

The electrical characteristics for the 200GAUI-4 and 400GAUI-8 chip-to-module interfaces are defined at compliance points for the host and module, respectively. Reference test fixtures, called compliance boards, are used to access the electrical specification parameters. Figure 120E–5 depicts the location of compliance points when measuring host 200GAUI-4 or 400GAUI-8 C2M compliance. The output of the Host Compliance Board (HCB) is used to verify the host electrical output signal at TP1a. Similarly, the input of the HCB at TP4a is used to verify the host input compliance.

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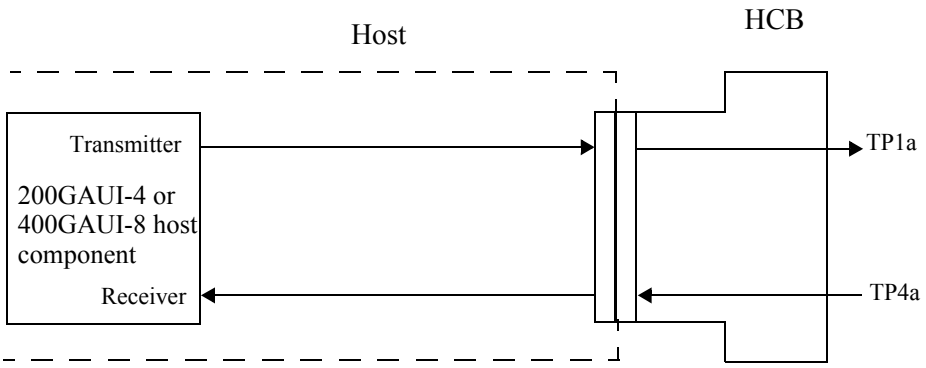


Figure 120E–5—Host 200GAUI-4 or 400GAUI-8 C2M compliance points

Figure 120E–6 depicts the location of compliance points when measuring module 200GAUI-4 or 400GAUI-8 C2M compliance. The output of the Module Compliance Board (MCB) is used to verify the module electrical output signal at TP4. Similarly, the input of the MCB at TP1 is used to verify the module input compliance. Additional details on the requirements for the MCB and HCB are given in 120E.4.1.

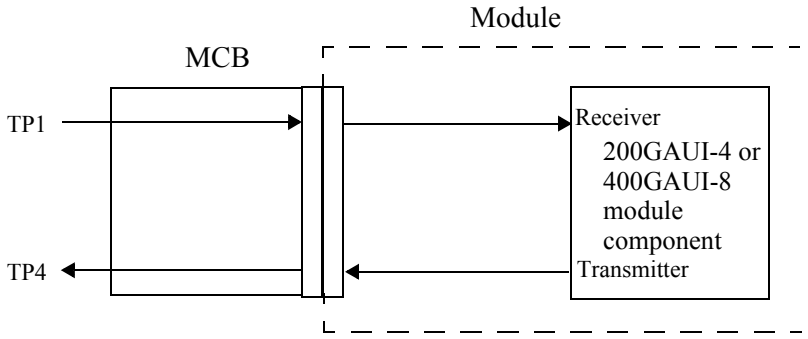


Figure 120E–6—Module 200GAUI-4 or 400GAUI-8 C2M compliance points

120E.3 200GAUI-4 and 400GAUI-8 chip-to-module electrical characteristics

120E.3.1 200GAUI-4 and 400GAUI-8 C2M host output characteristics

A 200GAUI-4 or a 400GAUI-8 C2M host output shall meet the specifications given in Table 120E–1.

A test system with a fourth-order Bessel-Thomson low-pass response with 33 GHz 3 dB bandwidth is to be used for all output signal measurements, unless otherwise specified.

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Table 120E–1—200GAUI-4 and 400GAUI-8 C2M host output characteristics (at TP1a)

Parameter	Reference	Value	Units
Signaling rate per lane (range)	120E.3.1.1	26.5625 ± 100 ppm	GBd
DC common-mode output voltage (max)	120E.3.1.2	2.8	V
DC common-mode output voltage (min)	120E.3.1.2	–0.3	V
Single-ended output voltage (max)	120E.3.1.2	3.3	V
Single-ended output voltage (min)	120E.3.1.2	–0.4	V
AC common-mode output voltage (max, RMS)	120E.3.1.2	17.5	mV
Differential peak-to-peak output voltage (max) Transmitter disabled Transmitter enabled	120E.3.1.2	 35 880	mV
ESMW (Eye symmetry mask width)	120E.4.2	0.22	UI
Eye height, differential (min)	120E.4.2	32	mV
Differential output return loss (min)	83E.3.1.3	Equation (83E-2)	dB
Common to differential mode conversion return loss (min)	83E.3.1.3	Equation (83E-3)	dB
Differential termination mismatch (max)	120E.3.1.4	10	%
Transition time (min, 20% to 80%)	120E.3.1.5	10	ps

120E.3.1.1 Signaling rate and range

The 200GAUI-4 and 400GAUI-8 C2M signaling rate is 26.5625 GBd ± 100 ppm per lane. This translates to a nominal unit interval of 37.647059 ps.

120E.3.1.2 Signal levels

The differential output voltage v_{di} is defined to be the difference between the single-ended output voltages, SL*i*<p> minus SL*i*<n>. The common-mode voltage v_{cmi} is defined to be one half of the sum of SL*i*<p> and SL*i*<n>. These definitions are illustrated by Figure 120E–7.

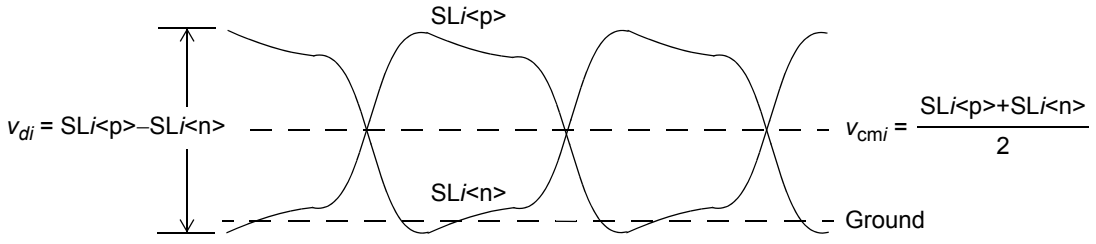


Figure 120E–7—Voltage definitions

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The DC common-mode output voltage and AC common-mode output voltage are defined with respect to signal ground.

Unless otherwise noted, differential and common-mode signal voltages are measured with a PRBS13Q test pattern.

120E.3.1.3 Output return loss

Differential output return loss is defined in 83E.3.1.3.

120E.3.1.4 Differential termination mismatch

Differential termination mismatch is defined in 86A.5.3.2.

120E.3.1.5 Transition time

In this annex, transition times are specified for transitions between three consecutive "zero" symbols and three consecutive "three" symbols, or vice versa. The specified times are between the crossings of 20% and 80% levels of the signal.

The test pattern used is PRBS13Q, the transitions within sequences of three zeros and three threes, and three threes and three zeros, respectively, are measured. These are PAM4 symbols 1820 to 1825 and 2086 to 2091, respectively, where symbols 1 to 7 are the run of seven threes. The 0% level and the 100% level are defined as the average signal within windows from –1.5 UI to –1 UI and from 1.5 UI to 2 UI relative to the edge.

The waveform is observed through a fourth-order Bessel-Thomson low-pass response with a 3 dB bandwidth of 33 GHz.

NOTE—This definition is not the same as the rise and fall times typically reported by an oscilloscope from an eye diagram, which take all the edges into account.

120E.3.1.6 Host output eye width and eye height

Figure 120E–8 depicts an example host output eye width and eye height test configuration. Host output eye width and eye height are measured at TP1a using compliance boards defined in 120E.4.1. The host output eye is measured using a reference receiver with a continuous time linear equalizer (CTLE) defined in 120E.3.1.7. Eye width and eye height measurement methodology is described in 120E.4.2. All counter-propagating signals shall be asynchronous to the co-propagating signals using the PRBS13Q or PRBS31Q pattern, or a valid 200GBASE-R or 400GBASE-R signal. PRBS13Q is described in 120.5.11.2.1. PRBS31Q is described in 120.5.11.2.2. For the case where PRBS13Q or PRBS31Q are used with a common clock, there is at least 31 UI delay between the patterns on one lane and any other lane, so that the symbols on each lane are not correlated. The crosstalk generator is calibrated at TP4 (without the use of a CTLE) with target differential peak-to-peak amplitude of 900 mV and slew time of 12 ps between ± 0.27 V.

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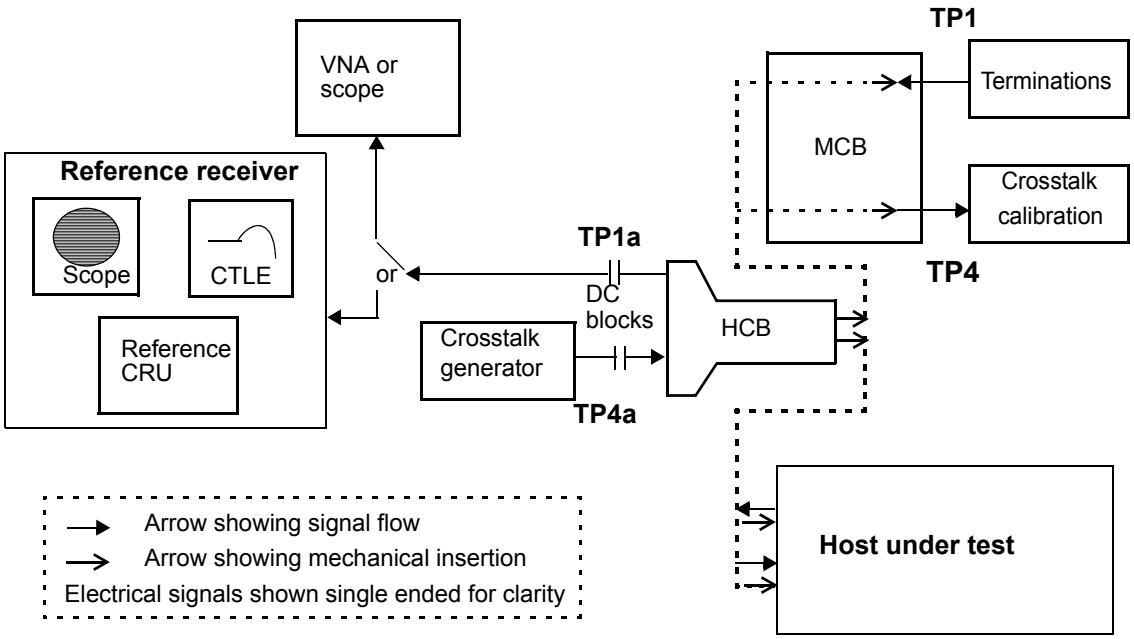


Figure 120E–8—Example host output test configuration

120E.3.1.7 Reference receiver for eye width and eye height evaluation

The reference receiver is used to measure eye width and eye height. The reference receiver includes a selectable continuous time linear equalizer (CTLE), which is described by Equation (120E–2) with coefficients given in Table 120E–2 and illustrated in Figure 120E–9. The equalizer may be implemented in software; however, the measured signal is not averaged.

$$H(f) = \frac{GP_1P_2P_{\mathrm{LF}}}{Z_1Z_{\mathrm{LF}}} \times \frac{\mathrm{j}2\pi f + Z_1}{(\mathrm{j}2\pi f + P_1)(\mathrm{j}2\pi f + P_2)} \times \frac{\mathrm{j}2\pi f + Z_{\mathrm{LF}}}{\mathrm{j}2\pi f + P_{\mathrm{LF}}} \quad \text{(120E–2)}$$

where

$H(f)$	is the CTLE transfer function
G	is the CTLE gain
P_1, P_2	are the CTLE poles in Grad/s
Z_1	is the CTLE zero in Grad/s
P_{LF}	is the low frequency CTLE pole in Grad/s
Z_{LF}	is the low frequency CTLE zero in Grad/s
j	is the square root of –1
f	is the frequency in GHz

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Table 120E–2—Reference CTLE coefficients

Peaking (dB)	G	$\frac{P_1}{2\pi}$	$\frac{P_2}{2\pi}$	$\frac{Z_1}{2\pi}$	$\frac{P_{LF}}{2\pi}$	$\frac{Z_{LF}}{2\pi}$
1	0.891251	26.5625	14.1	9.463748	1.2	1.2
1.5	0.841395	26.5625	14.1	9.248465	1.2	1.15
2	0.794328	26.5625	14.1	9.069645	1.2	1.1
2.5	0.749894	26.5625	14.1	8.640319	1.2	1.075
3	0.707946	26.5625	14.1	8.255665	1.2	1.05
3.5	0.668344	26.5625	14.1	7.906766	1.2	1.025
4	0.630957	26.5625	14.1	7.58765	1.2	1
4.5	0.595662	26.5625	14.1	7.076858	1.2	1
5	0.562341	26.5625	14.1	6.614781	1.2	1
5.5	0.530884	26.5625	14.1	6.193091	1.2	1
6	0.501187	26.5625	14.1	5.805801	1.2	1
6.5	0.473151	26.5625	14.1	5.448395	1.2	1
7	0.446684	26.5625	14.1	5.117337	1.2	1
7.5	0.421697	26.5625	14.1	4.809777	1.2	1
8	0.398107	26.5625	14.1	4.523367	1.2	1
8.5	0.375837	26.5625	14.1	4.256129	1.2	1
9	0.354813	26.5625	14.1	4.006377	1.2	1

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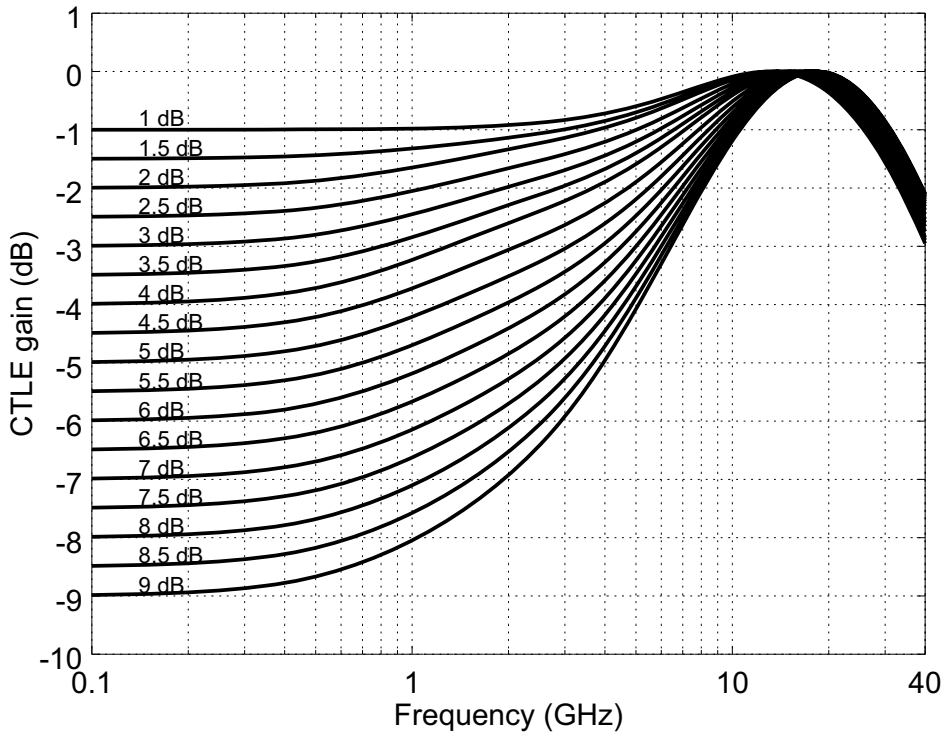


Figure 120E–9—Reference continuous time linear equalizer (CTLE) characteristic

120E.3.2 200GAUI-4 and 400GAUI-8 C2M module output characteristics

A 200GAUI-4 or a 400GAUI-8 C2M module output shall meet the specifications given in Table 120E–3. The low-frequency 3 dB cutoff of the output AC-coupling within the module should be less than 100 kHz. A test system with a fourth-order Bessel-Thomson low-pass response with 33 GHz 3 dB bandwidth is to be used for all output signal measurements, unless otherwise specified.

Table 120E–3—200GAUI-4 and 400GAUI-8 C2M module output characteristics (at TP4)

Parameter	Reference	Value	Units
Signaling rate per lane (range)	120E.3.1.1	26.5625 ± 100 ppm	GBd
AC common-mode output voltage (max, RMS)	120E.3.1.2	17.5	mV
Differential peak-to-peak output voltage (max)	120E.3.1.2	900	mV
Near-end ESMW (Eye symmetry mask width)	120E.4.2	0.265	UI
Near-end Eye height, differential (min)	120E.4.2	70	mV
Far-end ESMW (Eye symmetry mask width)	120E.4.2	0.2	UI
Far-end Eye height, differential (min)	120E.4.2	30	mV
Far-end pre-cursor ISI ratio	120E.3.2.1.2	–4.5 to +2.5	%

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Table 120E–3—200GAUI-4 and 400GAUI-8 C2M module output characteristics (at TP4) *(continued)*

Parameter	Reference	Value	Units
Differential output return loss (min)	83E.3.1.3	Equation (83E-2)	dB
Common to differential mode conversion return loss (min)	83E.3.1.3	Equation (83E-3)	dB
Differential termination mismatch (max)	120E.3.1.4	10	%
Transition time (min, 20% to 80%)	120E.3.1.5	9.5	ps
DC common mode voltage (min)[a]	120E.3.1.2	–350	mV
DC common mode voltage (max)[a]	120E.3.1.2	2850	mV

[a] DC common mode voltage is generated by the host. Specification includes effects of ground offset voltage.

120E.3.2.1 Module output eye width, eye height, and pre-cursor ISI ratio

Figure 120E–10 depicts an example module output eye width and eye height test configuration. Module output eye width and eye height are measured at TP4 using compliance boards defined in 120E.4.1. The module output eye is measured using a reference receiver with a continuous time linear equalizer (CTLE) defined in 120E.3.2.1.1. Eye width and eye height measurement methodology is described in 120E.4.2. All counter-propagating signals shall be asynchronous to the co-propagating signals using the PRBS13Q or PRBS31Q pattern, or a valid 200GBASE-R or 400GBASE-R signal. PRBS13Q is described in 120.5.11.2.1 and PRBS31Q is described in 120.5.11.2.2. For the case where PRBS13Q or PRBS31Q are used with a common clock, there is at least 31 UI delay between the patterns on one lane and any other lane, so that the symbols on each lane are not correlated. The crosstalk generator is calibrated at TP1a (without the use of a CTLE) with target differential peak-to-peak amplitude of 880 mV and target transition time of 19 ps.

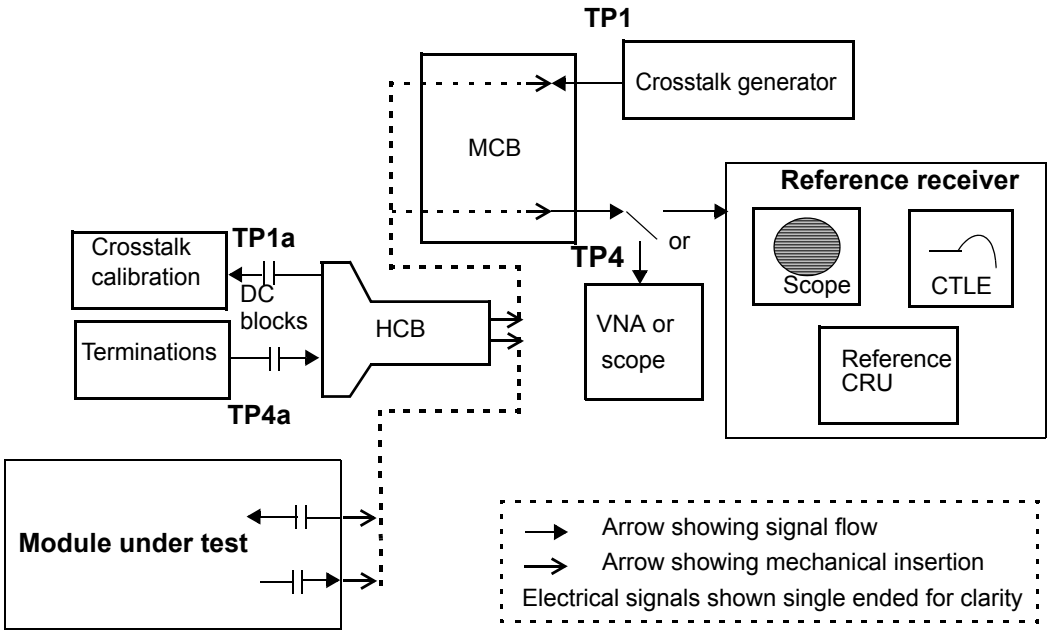


Figure 120E–10—Example module output test configuration

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120E.3.2.1.1 Reference receiver for module output evaluation

A reference receiver is used to measure module eye width and eye height. The reference receiver includes a selectable continuous time linear equalizer (CTLE), which is defined in 120E.3.1.7. The equalizer may be implemented in software; however, the measured signal is not averaged.

The near-end eye is measured with the reference receiver. Equalizer settings corresponding to less than or equal to 3 dB of peaking from Table 120E–2 may be used to meet the output eye width and height requirements.

For the far-end eye, the signal measured at TP4 is first convolved with a loss channel (~6.4 dB loss at Nyquist) that represents the worst case channel loss. The loss channel is the host trace defined in 92.10.7.1.1 with Z_p = 151 mm. The reference receiver is then used to measure the eye width and height. Any of the equalizer settings from Table 120E–2 may be used.

120E.3.2.1.2 Far-end pre-cursor ISI ratio

Capture the PRBS13Q waveform corresponding to the far-end eye (see 120E.3.2.1.1) and calculate the linear fit pulse using the procedure defined in 120D.3.1.3. Any setting of the reference CTLE for which the far-end eye width and height satisfy the limits in Table 120E–3, may be used.

The peak amplitude of the linear fit pulse is p_{max}. The pre-cursor ISI p_{pre} is the value of the linear fit pulse 1 UI prior to the time of the pulse peak. The pre-cursor ISI ratio is p_{pre} / p_{max}.

120E.3.3 200GAUI-4 and 400GAUI-8 C2M host input characteristics

A 200GAUI-4 or a 400GAUI-8 C2M host input shall meet the specifications given in Table 120E–4.

Table 120E–4—200GAUI-4 and 400GAUI-8 C2M host input characteristics

Parameter	Reference	Test point	Value	Units
Signaling rate, per lane (range)	120E.3.1.1	TP4a	26.5625 ± 100 ppm	GBd
Differential pk-pk input voltage tolerance (min)	120E.3.1.2	TP4	900	mV
Differential input return loss (min)	83E.3.3.1	TP4a	Equation (83E-5)	dB
Differential to common mode input return loss (min)	83E.3.3.1	TP4a	Equation (83E-6)	dB
Host stressed input test[a]	120E.3.3.2	TP4	See 120E.3.3.2	
Differential termination mismatch (max)	120E.3.1.4	TP4a	10	%
Common-mode voltage[b] Min Max	120E.3.1.2	TP4a	 –0.3 2.8	V

[a]Meets BER specified in 120E.1.1.
[b]Generated by host, referred to host ground.

120E.3.3.1 Input return loss

Differential input return loss is defined in 83E.3.3.1.

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120E.3.3.2 Host stressed input test

The host stressed input tolerance is measured using the procedure defined in 120E.3.3.2.1. The input shall satisfy the input tolerance with the parameters in Table 120E–5.

Table 120E–5—Host stressed input parameters

Parameter	Value
Far-end ESMW (Eye symmetry mask width)	0.2 UI
Far-end Eye width	0.2 UI
Applied pk-pk sinusoidal jitter	Table 120E–6
Far-end Eye height	30 mV

120E.3.3.2.1 Host stressed input test procedure

The host stressed input test is summarized in Figure 120E–11. The stressed signal is applied at TP4a and is calibrated at TP4. A reference CRU with a corner frequency of 4 MHz and slope of 20 dB/decade is used to calibrate the stressed signal using the PRBS13Q Pattern (see 120.5.11.2.1). The reference receiver includes a selectable CTLE as specified in 120E.3.2.1.1. The stressed signal is generated by adding sinusoidal jitter, random jitter, and bounded uncorrelated jitter to a clean pattern. The frequency and peak-to-peak amplitude of the sinusoidal jitter used for the host stressed input test is given in Table 120E–6.

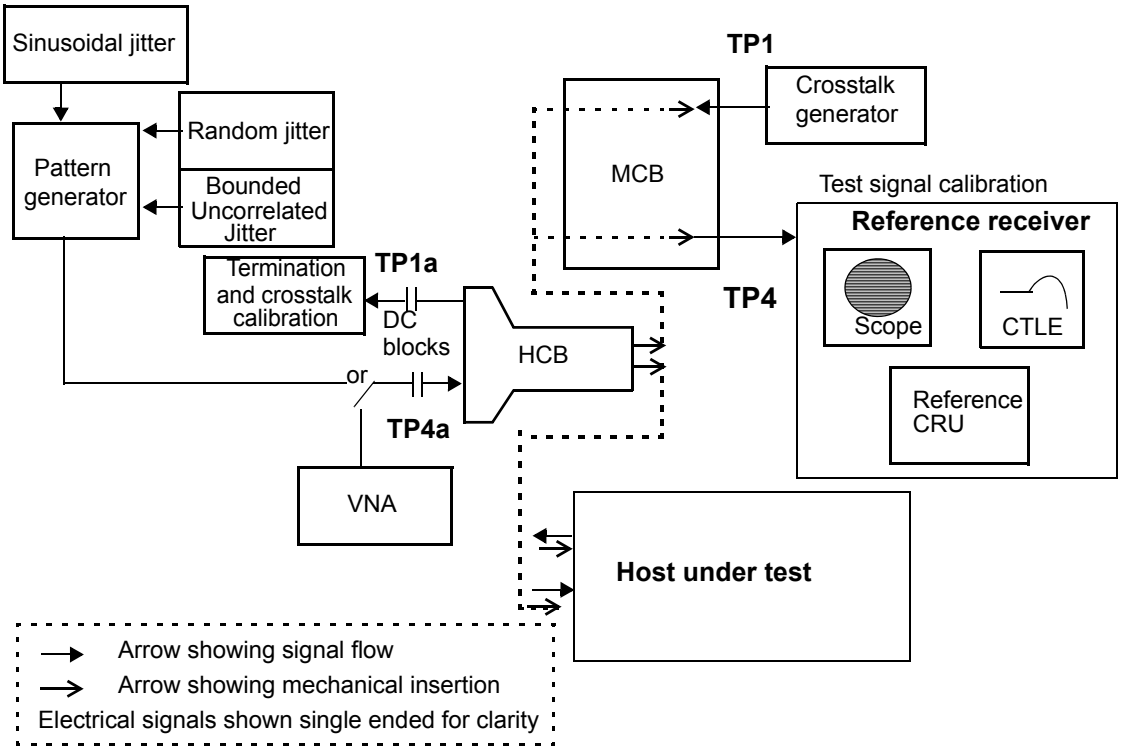


Figure 120E–11—Example host stressed input test

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Table 120E–6—Applied sinusoidal jitter

Parameter	Case A	Case B	Case C	Case D	Case E	Units
Jitter frequency	0.04	1.333	4	12	40	MHz
Jitter amplitude	5	0.15	0.05	0.05	0.05	UI

Bounded uncorrelated jitter provides a source of bounded high probability jitter uncorrelated with the signal stream. This jitter stress source may not be present in all stressed pattern generators or bit error ratio testers. It can be generated by driving the pattern generator external jitter modulation input with a filtered PRBS pattern. The PRBS pattern length should be between PRBS7 and PRBS9 with a signaling rate approximately 1/10 of the stressed pattern signaling rate (e.g., 2.65625 GBd). The clock source for the PRBS generator is asynchronous to the pattern generator clock source to assure non-correlation of the jitter. The low-pass filter that operates on the PRBS pattern to generate the bounded uncorrelated jitter should exhibit 20 dB/decade roll-off with a –3 dB corner frequency between 150 MHz and 300 MHz. This value must also be below the upper frequency limit of the pattern generator external modulator input. Random jitter and bounded uncorrelated jitter are added such that the output of the pattern generator approximates the 200GAUI-4 and 400GAUI-8 C2C output jitter profile given in Table 120D–1.

The counter propagating crosstalk channels during calibration of the stressed signal are asynchronous with target amplitude of 880 mV peak-to-peak differential and 20% to 80% target transition time of 19 ps as measured at TP1a (without the use of a CTLE). The crosstalk signal transition time is calibrated with PRBS13Q. The pattern may be changed to a valid 400GBASE-R signal for amplitude calibration and the stressed input test. The PRBS13Q pattern is described in 120.5.11.2.1. For the case where the PRBS13Q pattern is used with a common clock, there is at least 31 UI delay between the PRBS13Q patterns on one lane and any other lane, so that the symbols on each lane are not correlated. Any one of these patterns is sufficient as a crosstalk aggressor with all lanes active during the stressed input test.

The far-end eye height and eye width, measured to a probability of 10^{-5}, are then measured at TP4 with the reference receiver defined in 120E.3.2.1.1 using the measurement methodology given in 120E.4.2. Note that the reference receiver for far-end eye height and width measurements includes a loss channel. Random jitter and the pattern generator output levels are adjusted (without exceeding the differential pk-pk input voltage tolerance specification as shown in Table 120E–4) to result in the eye height for all three eyes and eye width for the smallest eye given in Table 120E–5 with the setting of the CTLE that maximizes the product of eye height and eye width.

The far-end pre-cursor ISI ratio is measured using the method defined in 120E.3.2.1.2 and it shall meet the specification in Table 120E–3. Pre-emphasis capability is likely to be required in the pattern generator to meet this requirement.

The pattern is then changed to Pattern 5, Pattern 3, or a valid 200GBASE-R/400GBASE-R signal for the input test which is conducted by inserting the HCB into the host under test. Patterns 3 and 5 are described in Table 124–9.

If the test is performed with pattern 3, the host BER may be calculated using the host's PMA test pattern checker (see 120.5.11.2.2). If the test is performed with pattern 5 or a valid 200GBASE-R/400GBASE-R signal, the BER may be calculated using the host's PCS Reed-Solomon decoder error counters (see 119.3.1), as the number of FEC symbol errors divided by the number of received bits. The number of received bits may be estimated based on the test time.

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120E.3.4 200GAUI-4 and 400GAUI-8 C2M module input characteristics

A 200GAUI-4 or a 400GAUI-8 C2M module input shall meet the specifications given in Table 120E–7. Channel equalization is provided by an adaptive equalizer in the module.

Table 120E–7—200GAUI-4 and 400GAUI-8 C2M module input characteristics

Parameter	Reference	Test point	Value	Units
Signaling rate per lane (range)	120E.3.1.1	TP1	26.5625 ± 100 ppm	GBd
Differential pk-pk input voltage tolerance (min)[a]	120E.3.1.2	TP1a	900	mV
Differential input return loss (min)	83E.3.3.1	TP1	Equation (83E-5)	dB
Differential to common mode input return loss (min)	83E.3.3.1	TP1	Equation (83E-6)	dB
Differential termination mismatch (max)	120E.3.1.4	TP1	10	%
Module stressed input test[b]	120E.3.4.1	TP1a	See 120E.3.4.1	
Single-ended voltage tolerance range (min)	120E.3.1.2	TP1a	–0.4 to 3.3	V
DC common mode voltage (min)[c]	120E.3.1.2	TP1	–350	mV
DC common mode voltage (max)[c]	120E.3.1.2	TP1	2850	mV

[a] With the exception to 120E.3.1.2 that the pattern is PRBS31Q or scrambled idle.
[b] Meets BER specified in 120E.1.1.
[c] DC common mode voltage generated by the host. Specification includes effects of ground offset voltage.

120E.3.4.1 Module stressed input test

The module stressed input tolerance is measured using the procedure defined in 120E.3.4.1.1. The input shall satisfy the input tolerance with the parameters in Table 120E–8.

Table 120E–8—Module stressed input parameters

Parameter	Value
ESMW (Eye symmetry mask width)	0.22 UI
Eye width	0.22 UI
Applied pk-pk sinusoidal jitter	Table 120E–6
Eye height	32 mV

120E.3.4.1.1 Module stressed input test procedure

The module stressed input test is summarized in Figure 120E–12. The stressed signal is applied at TP1 and is calibrated at TP1a. A reference CRU with a corner frequency of 4 MHz and slope of 20 dB/decade is used to calibrate the stressed signal using a PRBS13Q pattern. The reference receiver includes a selectable CTLE as specified in 120E.3.1.7. The stressed signal is generated by adding sinusoidal jitter, random jitter, and

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bounded uncorrelated jitter to a clean pattern, followed by frequency-dependent attenuation. The frequency-dependent attenuator represents the host channel and may be implemented with PCB traces. The amount of applied peak-to-peak sinusoidal jitter used for the module stressed input test is given in Table 120E–6. Bounded uncorrelated jitter provides a source of bounded high probability jitter uncorrelated with the signal stream. This jitter stress source may not be present in all stressed pattern generators or bit error ratio testers. It can be generated by driving the pattern generator external jitter modulation input with a filtered PRBS pattern. The PRBS pattern length should be between PRBS7 and PRBS9 with a signaling rate approximately 1/10 of the stressed pattern signaling rate (e.g., 2.65625 GBd). The clock source for the PRBS generator is asynchronous to the pattern generator clock source to assure non-correlation of the jitter. The low-pass filter that operates on the PRBS pattern to generate the bounded uncorrelated jitter should exhibit 20 dB/decade roll-off with a –3 dB corner frequency between 150 MHz and 300 MHz. This value must also be below the upper frequency limit of the pattern generator external modulator input. Random jitter and bounded uncorrelated jitter are added such that the output of the pattern generator approximates the 200GAUI-4 and 400GAUI-8 C2C output jitter profile given in Table 120D–1. The target pattern generator 20% to 80% transition time at the input to the test channel in the module stressed input test is 10.5 ps. The return loss of the test system as measured at TP1 meets the specification given in Equation (83E-2).

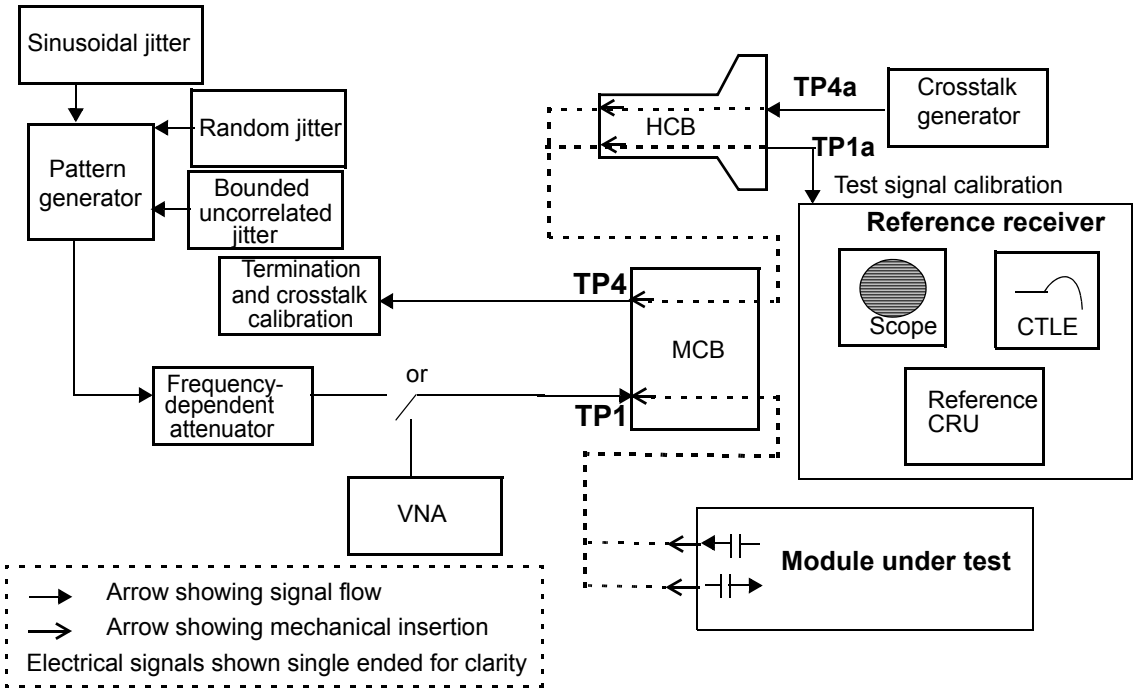


Figure 120E–12—Example module stressed input test

For the high loss case, pre-emphasis capability is likely to be required in the pattern generator to meet the TP1a eye height and eye width specifications.

The counter propagating crosstalk channels during calibration of the stressed signal are asynchronous with target amplitude of 900 mV peak-to-peak differential and target slew time between ± 270 mV of 12 ps as measured at TP4 (without the use of a CTLE). The crosstalk signal transition time is calibrated with a PRBS13Q pattern. The pattern may be changed to a valid 200GBASE-R/400GBASE-R signal for amplitude calibration and the stressed input test. The PRBS13Q pattern is described in 120.5.11.2.1. For the case where the PRBS13Q pattern is used with a common clock, there is at least 31 UI delay between the PRBS13Q patterns on one lane and any other lane, so that the symbols on each lane are not correlated. Any one of these patterns is sufficient as a crosstalk aggressor with all lanes being active during the stressed input test.

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Two levels of frequency-dependent attenuation are used for the module stressed input test: high loss and low loss. For the high loss case, frequency-dependent attenuation is added such that the loss at 13.28 GHz from the output of the pattern generator to TP1a is 12.2 dB. The 12.2 dB loss represents 10.5 dB channel loss with an additional allowance for host transmitter package loss. Eye height and eye width are then measured at TP1a based on the eye measurement methodology given in 120E.4.2. Random jitter and the pattern generator output levels are adjusted (without exceeding the differential pk-pk input voltage tolerance specification as shown in Table 120E–7) to result in the eye height for all three eyes and eye width for the smallest eye given in Table 120E–8 using the reference receiver with the setting of the CTLE that maximizes the product of eye height and eye width. This CTLE setting has to be greater than or equal to 7 dB. For the low loss case, discrete frequency-dependent attenuation is removed such that from the output of the pattern generator to TP1a comprises the mated HCB/MCB pair as described in 120E.4.1. Eye height and eye width at TP1a are then adjusted in the same way as described for the high loss case except that the restriction that the CTLE setting has to be greater than or equal to 7 dB does not apply. The pattern is then changed to Pattern 5, Pattern 3, or a valid 200GBASE-R/400GBASE-R signal for the input test, which is conducted by inserting the module into the MCB. Patterns 3 and 5 are described in Table 124–9.

If the test is performed with pattern 3, the module bit errors may be counted using a PMA test pattern checker (see 120.5.11.2.2) if this option is implemented in the module.

If the test is performed with pattern 5 or a valid 200GBASE-R/400GBASE-R signal, the module BER may be calculated by placing the module under test into local loopback (see 120.5.9) and feeding the module output into a compliant host or its equivalent. The BER may then be calculated using the host's PCS Reed-Solomon decoder error counters (see 119.3.1), as the number of FEC symbol errors divided by the number of received bits.

Methods of extracting the received bit pattern and counting errors other than the ones described above may be used if they generate equivalent results.

The number of received bits may be estimated based on the test time.

The module 200GAUI-4 or 400GAUI-8 C2M receiver under test shall meet the BER requirement as described in 120E.1.1 for both the high loss test and low loss test.

120E.4 200GAUI-4 and 400GAUI-8 C2M measurement methodology

This subclause describes common measurement tools and methodologies to be used for the 200GAUI-4 and 400GAUI-8 chip-to-module interface. Details of HCB and MCB characteristics are given in 120E.4.1 and details of the eye diagram measurement methodology are given in 120E.4.2.

120E.4.1 HCB/MCB characteristics

HCB characteristics are described in 92.11.1 where the HCB performs the equivalent function as the TP2 or TP3 test fixture with the exception that the upper frequency of 25 GHz is replaced with 26.5625 GHz. The MCB characteristics are described in 92.11.2 where the MCB performs the equivalent functionality as the cable assembly test fixture with the exception that the upper frequency of 25 GHz is replaced with 26.5625 GHz.

The mated compliance board characteristics are described in 92.11.3 where the MCB and HCB perform the equivalent functionality as the cable assembly test fixtures with the exceptions that the upper frequency of 25 GHz is replaced with 26.5625 GHz, MDNEXT shall be less than 1.5 mV RMS, MDFEXT shall be less than 4.2 mV RMS, ICN shall be less than 4.4 mV RMS, FOM_{ILD} limit is 0.1 dB, and the reference insertion loss is given in Equation (120E–3) for 0.01 GHz $\leq f \leq$ 25 GHz.

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$$IL(f) = 0.471\sqrt{f} + 0.1194f + 0.002f^2 \quad \text{(dB)} \qquad \text{(120E–3)}$$

where

f is the frequency in GHz

IL(f) is the mated compliance board reference insertion loss

120E.4.2 Eye width and eye height measurement method

Eye diagrams in 200GAUI-4 and 400GAUI-8 chip-to-module are measured using a reference receiver. The reference receiver includes a fourth-order Bessel-Thomson low-pass filter response with 33 GHz 3 dB bandwidth, and a selectable continuous time linear equalizer (CTLE) to measure eye height and width. The pattern used for output eye diagram measurements is PRBS13Q. Unless specified otherwise the probabilities are relative to the number of PAM4 symbols measured. The following procedure should be used to obtain the eye height and eye width parameters, as illustrated by Figure 120E–13:

1) Capture the PRBS13Q using a clock recovery unit with a corner frequency of 4 MHz and slope of 20 dB/decade. The capture includes a minimum of 3 samples per symbol, or equivalent. Collect sufficient samples equivalent to at least 1.2 million PAM4 symbols to allow for construction of a normalized cumulative distribution function (CDF) to a probability of 10^{-5} without extrapolation.
2) Apply the reference receiver including the appropriate CTLE to the captured signal. For module compliance, any single CTLE setting as described in 120E.3.2.1.1 that meets both eye width and eye height requirements is acceptable. For host compliance, any single CTLE setting as described in 120E.3.1.7 that meets both eye width and eye height requirements is acceptable.
3) Use the differential equalized signal from step 2) to construct CDF of the signal threshold crossing for both the left edge (MID0CDFL) and right edge (MID0CDFR) of the middle eye, as a function of the distance from the center of the eye. Calculate the time center of the middle eye width (TCmid) as the mid-point in time between MID0CDFR and MID0CDFL with a value of 10^{-3}. MID0CDFL and MID0CDFR are calculated as the cumulative sum of histograms of the zero crossing samples at the left and right edges of the middle eye normalized by the total number of sampled symbols. The MID0CDFL and MID0CDFR are equivalent to bath tub curves where the BER is plotted versus sampling time.
4) Use the differential equalized signal from step 2) to construct the CDF of the signal voltage of the middle eye within 0.025 UI of time TCmid, for both +1/3 (MIDCDF1) and –1/3 (MIDCDF0), as a function of the distance from the center of the eye. Calculate the eye height (Vmid) as the difference in voltage between MIDCDF1 and MIDCDF0 with a value of 10^{-5}. Calculate the voltage center (VCmid) of the middle eye as the mid-point in voltage between MIDCDF1 and MIDCDF0 with a value of 10^{-5}. MIDCDF0 and MIDCDF1 are calculated as the cumulative sum of histograms of the voltage at the top and bottom of the eye normalized by the total number of sampled bits.
5) Using the method in step 4), construct CDFs of signal voltages of the upper eye within 0.025 UI of time TCmid, and calculate the 10^{-5} upper eye height (Vupp). Calculate the voltage center (VCupp) of the upper eye as the mid-point in voltage between UPPCDF1 and UPPCDF0 with a value of 10^{-5}.
6) Using the method in step 4), construct CDFs of signal voltages of the lower eye within 0.025 UI of time TCmid, and calculate the 10^{-5} lower eye height (Vlow). Calculate the voltage center (VClow) of the lower eye as the mid-point in voltage between LOWCDF1 and LOWCDF0 with a value of 10^{-5}.
7) The eye height is defined as the minimum of Vmid, Vupp, and Vlow.
8) Use the differential equalized signal from step 2) to construct new CDFs of the signal for both the left edge (MIDCDFL) and right edge (MIDCDFR) of the middle eye at VCmid, as a function of the distance from the center of the eye. Calculate the middle eye width (Hmid) as the difference in time between MIDCDFR and MIDCDFL with a value of 10^{-5}.

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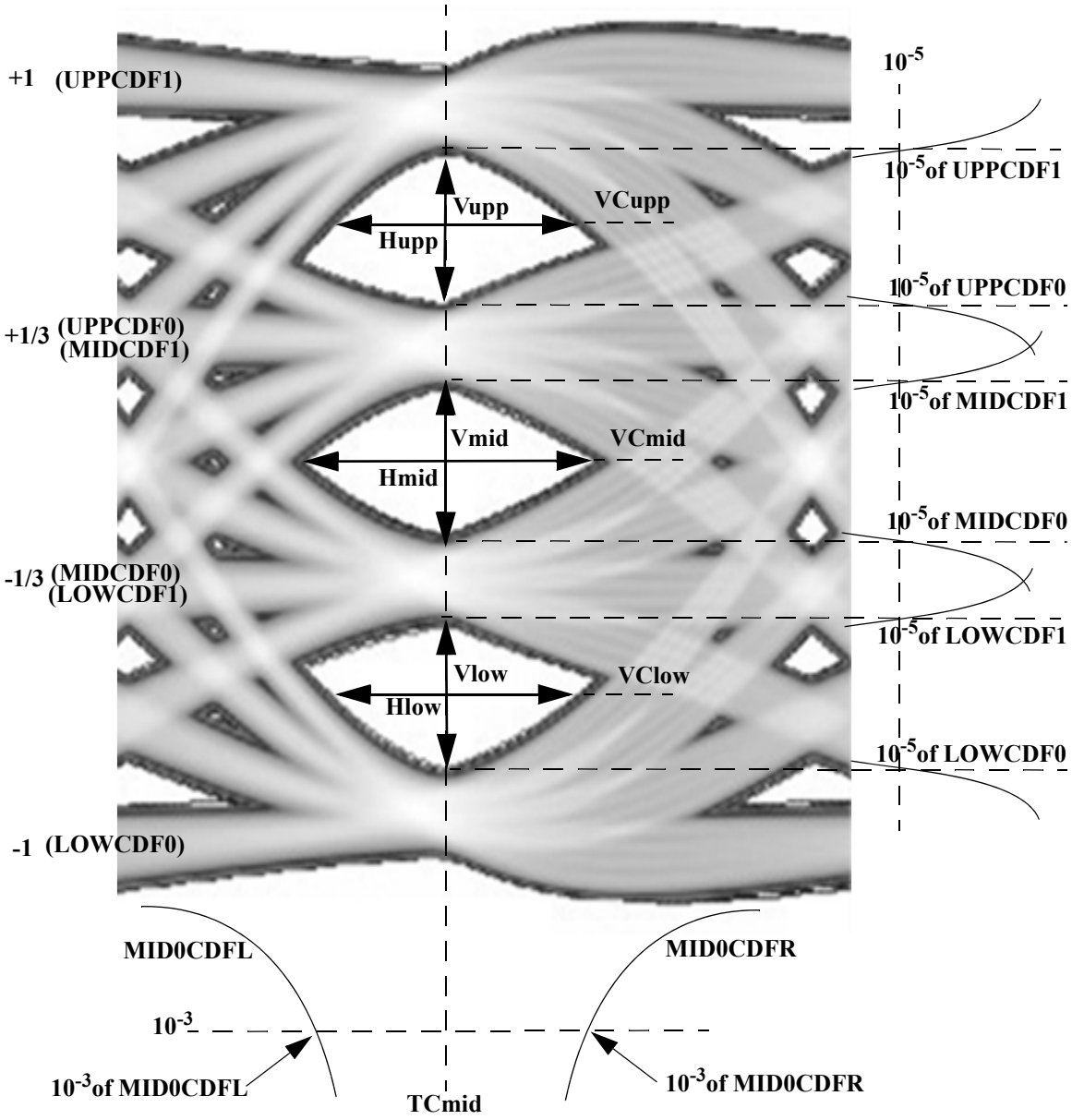


Figure 120E–13—PAM4 eye measurements

9) Using the method in step 8), calculate the 10^{-5} upper eye width (Hupp) sliced at VCupp.
10) Using the method in step 8), calculate the 10^{-5} lower eye width (Hlow) sliced at VClow.
11) The eye width is defined as the minimum of Hmid, Hupp, and Hlow.
12) Check the upper and lower eye alignment to the middle eye. Apply an ESMW wide mask centered on TCmid, as illustrated by Figure 120E–14. The 10^{-5} horizontal openings of the middle eye at VCmid, the upper eye at VCupp, and of the lower eye at VClow must all extend beyond this mask.

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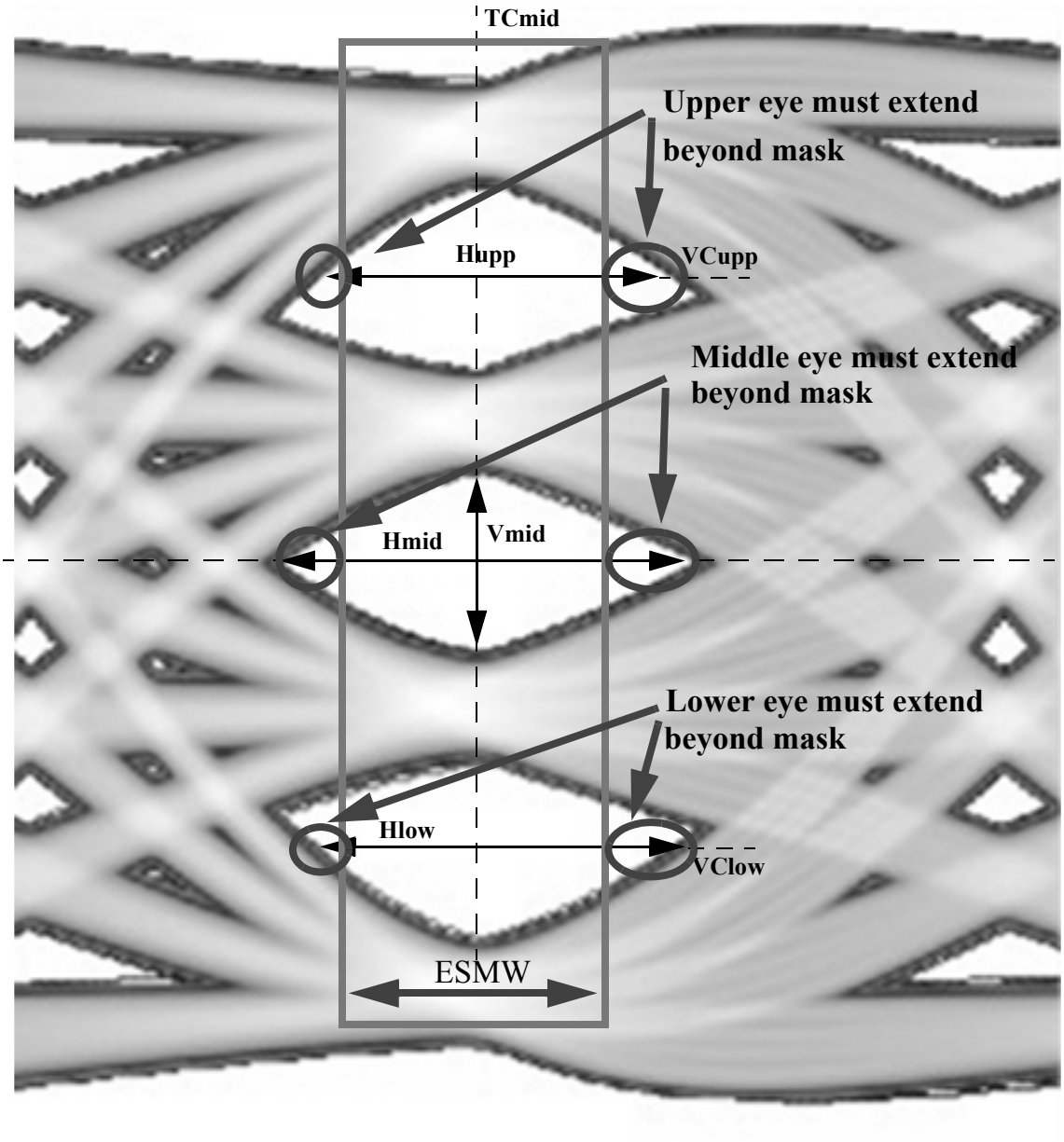


Figure 120E–14—PAM4 upper, middle, and lower eye mask

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120E.5 Protocol implementation conformance statement (PICS) proforma for Annex 120E, Chip-to-module 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2M) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2M)[18]

120E.5.1 Introduction

The supplier of a protocol implementation that is claimed to conform to Annex 120E, Chip-to-module 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2M) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2M), shall complete the following protocol implementation conformance statement (PICS) proforma.

A detailed description of the symbols used in the PICS proforma, along with instructions for completing the PICS proforma, can be found in Clause 21.

120E.5.2 Identification

120E.5.2.1 Implementation identification

Supplier[1]	
Contact point for inquiries about the PICS[1]	
Implementation Name(s) and Version(s)[1,3]	
Other information necessary for full identification—e.g., name(s) and version(s) for machines and/or operating systems; System Name(s)[2]	
NOTE 1—Required for all implementations. NOTE 2—May be completed as appropriate in meeting the requirements for the identification. NOTE 3—The terms Name and Version should be interpreted appropriately to correspond with a supplier's terminology (e.g., Type, Series, Model).	

120E.5.2.2 Protocol summary

Identification of protocol standard	IEEE Std 802.3bs-2017, Annex 120E, Chip-to-module 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2M) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2M)
Identification of amendments and corrigenda to this PICS proforma that have been completed as part of this PICS	
Have any Exception items been required? No [ ] Yes [ ] (See Clause 21; the answer Yes means that the implementation does not conform to IEEE Std 802.3bs-2017.)	

Date of Statement	

[18] *Copyright release for PICS proformas:* Users of this standard may freely reproduce the PICS proforma in this subclause so that it can be used for its intended purpose and may further publish the completed PICS.

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120E.5.3 Major capabilities/options

Item	Feature	Subclause	Value/Comment	Status	Support
NOL	Number of differential AC-coupled lanes	120E.1	Four independent data paths in each direction for 200GAUI-4 C2M and eight independent data paths in each direction for 400GAUI-8 C2M	M	Yes [ ]
BER	Meets BER requirement	120E.1.1	See 120E.1.1	M	Yes [ ]
ADE	Adaptive Equalizer	120E.3.4	See 120E.3.4	M	Yes [ ]

120E.5.4 PICS proforma tables for Chip-to-module 200 Gb/s four-lane Attachment Unit Interface (200GAUI-4 C2M) and 400 Gb/s eight-lane Attachment Unit Interface (400GAUI-8 C2M)

120E.5.4.1 Host output

Item	Feature	Subclause	Value/Comment	Status	Support
TH1	Signaling rate	120E.3.1.1	26.5625 GBd ± 100 ppm per lane	M	Yes [ ]
TH2	DC common-mode output voltage	120E.3.1	Between –0.3 V and 2.8 V with respect to signal ground	M	Yes [ ]
TH3	Peak-to-peak differential output voltage, transmitter disabled	120E.3.1.2	Less than or equal to 35 mV	M	Yes [ ]
TH4	Peak-to-peak differential output voltage	120E.3.1.2	880 mV (max)	M	Yes [ ]
TH5	AC common-mode output voltage	120E.3.1	17.5 mV RMS with respect to signal ground	M	Yes [ ]
TH6	Eye width, all 3 eyes	120E.3.1	0.22 UI	M	Yes [ ]
TH7	Eye height, all 3 eyes	120E.3.1	32 mV	M	Yes [ ]
TH8	Differential output return loss	120E.3.1.3	Meets Equation (83E-2) constraints	M	Yes [ ]
TH9	Common to differential mode conversion	120E.3.1.3	Meets Equation (83E-3) constraints	M	Yes [ ]
TH10	Differential termination mismatch	120E.3.1	Less than 10%	M	Yes [ ]
TH11	Transition time	120E.3.1.5	Greater than or equal to 10 ps	M	Yes [ ]
TH12	Reference impedance for output return loss	120E.3.1.3	100 Ω	M	Yes [ ]
TH13	Crosstalk source	120E.3.1.6	Asynchronous crosstalk source using PRBS13Q or PRBS31Q pattern, or valid 200GBASE-R/400GBASE-R signal	M	Yes [ ]

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120E.5.4.2 Module output

Item	Feature	Subclause	Value/Comment	Status	Support
TM1	Signal rate	120E.3.1.1	26.5625 GBd ± 100 ppm per lane	M	Yes [ ]
TM2	AC common-mode output voltage	120E.3.1	17.5 mV RMS with respect to signal ground	M	Yes [ ]
TM3	Peak-to-peak differential output voltage	120E.3.1.2	900 mV (max)	M	Yes [ ]
TM4	Near-end eye width, all 3 eyes	120E.3.2.1	0.265 UI	M	Yes [ ]
TM5	Near-end eye height, all 3 eyes	120E.3.2.1	70 mV	M	Yes [ ]
TM6	Far-end eye width, all 3 eyes	120E.3.2.1	0.2 UI	M	Yes [ ]
TM7	Far-end eye height, all 3 eyes	120E.3.2.1	30 mV	M	Yes [ ]
TM8	Far-end pre-cursor ISI ratio	120E.3.2.1.2	–4.5% to +2.5%	M	Yes [ ]
TM9	Differential output return loss	120E.3.1.3	Meets Equation (83E-2) constraints	M	Yes [ ]
TM10	Common to differential mode conversion	120E.3.1.3	Meets Equation (83E-3) constraints	M	Yes [ ]
TM11	Differential termination mismatch	120E.3.1	Less than 10%	M	Yes [ ]
TM12	Transition time	120E.3.1.5	Greater than or equal to 9.5 ps	M	Yes [ ]
TM13	Reference impedance for output return loss	120E.3.1.3	100 Ω	M	Yes [ ]
TM14	Crosstalk source	120E.3.2.1	Asynchronous crosstalk source using PRBS13Q or PRBS31Q pattern, or valid 200GBASE-R/400GBASE-R signal	M	Yes [ ]

120E.5.4.3 Host input

Item	Feature	Subclause	Value/Comment	Status	Support
RH1	Host input characteristics	120E.3.3	Table 120E–4	M	Yes [ ]

120E.5.4.4 Module input

Item	Feature	Subclause	Value/Comment	Status	Support
RM1	Module input characteristics	120E.3.4	Table 120E–7	M	Yes [ ]

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